



**THE DATASHEET OF  
ACS722LLCTR-10AB-T**

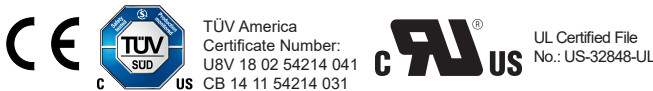


## High Accuracy, Galvanically Isolated Current Sensor IC with Small Footprint SOIC8 Package

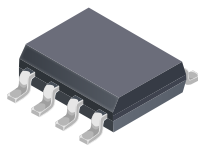
### FEATURES AND BENEFITS

- Patented integrated digital temperature compensation circuitry allows for near closed loop accuracy over temperature in an open loop sensor
- UL60950-1 (ed. 2) certified
  - Dielectric Strength Voltage = 2.4 kVrms
  - Basic Isolation Working Voltage = 420 Vpk/297 Vrms
- Industry-leading noise performance with greatly improved bandwidth through proprietary amplifier and filter design techniques
- Pin-selectable band width: 80 kHz for high bandwidth applications or 20 kHz for low noise performance
- 0.65 mΩ primary conductor resistance for low power loss and high inrush current withstand capability
- Small footprint, low-profile SOIC8 package suitable for space-constrained applications
- Integrated shield virtually eliminates capacitive coupling from current conductor to die, greatly suppressing output noise due to high dv/dt transients
- 3 to 3.6 V, single supply operation

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**Package: 8-pin SOIC (suffix LC)**



*Not to scale*

### DESCRIPTION

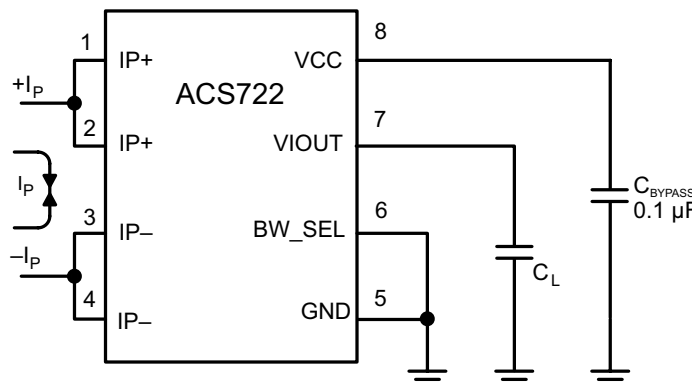
The Allegro™ ACS722 current sensor IC is an economical and precise solution for AC or DC current sensing in industrial, commercial, and communications systems. The small package is ideal for space constrained applications while also saving costs due to reduced board area. Typical applications include motor control, load detection and management, switched-mode power supplies, and overcurrent fault protection.

The device consists of a precise, low-offset, linear Hall sensor circuit with a copper conduction path located near the surface of the die. Applied current flowing through this copper conduction path generates a magnetic field which is sensed by the integrated Hall IC and converted into a proportional voltage. Device accuracy is optimized through the close proximity of the magnetic field to the Hall transducer. A precise, proportional voltage is provided by the low-offset, chopper-stabilized BiCMOS Hall IC, which includes Allegro's patented digital temperature compensation, resulting in extremely accurate performance over temperature. The output of the device has a positive slope when an increasing current flows through the primary copper conduction path (from pins 1 and 2, to pins 3 and 4), which is the path used for current sensing. The internal resistance of this conductive path is 0.65 mΩ typical, providing low power loss.

The terminals of the conductive path are electrically isolated from the sensor leads (pins 5 through 8). This allows the ACS722 current sensor IC to be used in high-side current sense applications without the use of high-side differential amplifiers or other costly isolation techniques.

The ACS722 is provided in a small, low profile surface mount SOIC8 package. The leadframe is plated with 100% matte tin,

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**Typical Application**

The ACS722 outputs an analog signal,  $V_{IOUT}$ , that changes, proportionally, with the bidirectional AC or DC primary sensed current,  $I_P$ , within the specified measurement range. The BW\_SEL pin can be used to select one of the two bandwidths to optimize the noise performance. Grounding the BW\_SEL pin puts the part in the high bandwidth (80 kHz) mode.

### FEATURES AND BENEFITS (continued)

- Output voltage proportional to AC or DC current
- Factory-trimmed sensitivity and quiescent output voltage for improved accuracy
- Chopper stabilization results in extremely stable quiescent output voltage
- Nearly zero magnetic hysteresis
- Ratiometric output from supply voltage

### DESCRIPTION (continued)

which is compatible with standard lead (Pb) free printed circuit board assembly processes. Internally, the flip-chip device is considered Pb-free. However, the solder bump connections are available in a Pb-free or high-temperature Pb-based option. Part numbers followed by -S are manufactured with tin-silver-based solder bumps, making these parts Pb-free compliant without the use of RoHS exemptions. Part numbers followed by -T are manufactured with Pb-based solder bumps using allowed RoHS exemptions. The device is fully calibrated prior to shipment from the factory.

### SELECTION GUIDE

Part Number	$I_{PR}$ (A)	Sens(Typ) at $V_{CC} = 3.3$ V (mV/A)	$T_A$ (°C)	Packing
<b>-S VARIANT [1]</b>				
ACS722LLCTR-05AB-S	±5	264	-40 to 150	Tape and Reel, 3000 pieces per reel
ACS722LLCTR-10AU-S	10			
ACS722LLCTR-10AB-S	±10	132		
ACS722LLCTR-20AU-S	20			
ACS722LLCTR-20AB-S	±20	66		
ACS722LLCTR-40AU-S	40			
ACS722LLCTR-40AB-S	±40	33		
<b>-T VARIANT [2]</b>				
ACS722LLCTR-05AB-T	±5	264	-40 to 150	Tape and Reel, 3000 pieces per reel
ACS722LLCTR-10AU-T	10			
ACS722LLCTR-10AB-T	±10	132		
ACS722LLCTR-20AU-T	20			
ACS722LLCTR-20AB-T	±20	66		
ACS722LLCTR-40AU-T	40			
ACS722LLCTR-40AB-T	±40	33		

[1] -S denotes the lead-free construction with tin-silver-based solder bumps.

[2] -T denotes Pb-contained construction with Pb-based solder bumps. Operating performance of -T and -S devices are identical. -T devices are RoHS compliant using allowed exemptions provided in Annex III and IV of Directive 2011/65/EU [Exemptions 7(a), 15, 15(a), as applicable].

### SPECIFICATIONS

#### ABSOLUTE MAXIMUM RATINGS

Characteristic	Symbol	Notes	Rating	Units
Supply Voltage	$V_{CC}$		6	V
Reverse Supply Voltage	$V_{RCC}$		-0.1	V
Output Voltage	$V_{IOUT}$		25	V
Reverse Output Voltage	$V_{RIOUT}$		-0.1	V
Maximum Continuous Current	$I_{CMAX}$	$T_A = 25^\circ\text{C}$	65	A
Operating Ambient Temperature	$T_A$	Range L	-40 to 150	$^\circ\text{C}$
Junction Temperature	$T_J(\text{max})$		165	$^\circ\text{C}$
Storage Temperature	$T_{stg}$		-65 to 165	$^\circ\text{C}$

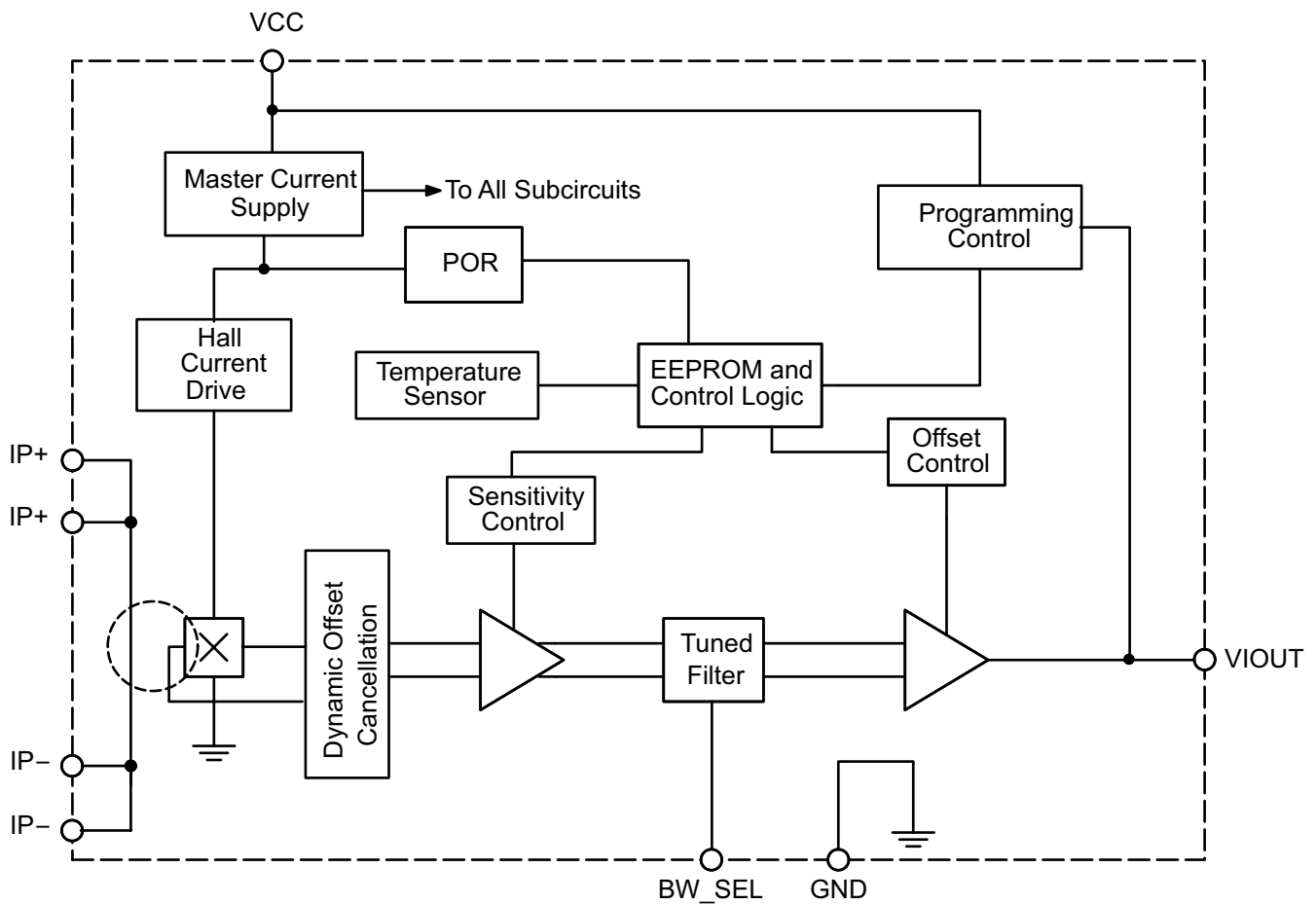
#### ISOLATION CHARACTERISTICS

Characteristic	Symbol	Notes	Rating	Unit
Dielectric Strength Test Voltage	$V_{ISO}$	Agency type-tested for 60 seconds per UL 60950-1 (edition. 2). Production tested at $V_{ISO}$ for 1 second, in accordance with UL 60950-1 (edition. 2).	2400	$V_{RMS}$
Working Voltage for Basic Isolation	$V_{WVBI}$	Maximum approved working voltage for basic (single) isolation according UL 60950-1 (edition 2).	420	$V_{PK}$ or VDC
			297	$V_{RMS}$
Clearance	$D_{cl}$	Minimum distance through air from IP leads to signal leads.	3.9	mm
Creepage	$D_{cr}$	Minimum distance along package body from IP leads to signal leads.	3.9	mm

#### THERMAL CHARACTERISTICS

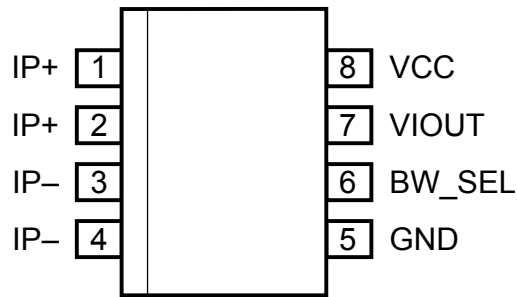
Characteristic	Symbol	Test Conditions*	Value	Units
Package Thermal Resistance (Junction to Ambient)	$R_{\theta JA}$	Mounted on the Allegro 85-0593 evaluation board with 400 mm <sup>2</sup> of 4 oz. copper on each side, connected to pins 1 and 2, and to pins 3 and 4, with thermal vias connecting the layers. Performance values include the power consumed by the PCB.	23	$^\circ\text{C}/\text{W}$
Package Thermal Resistance (Junction to Lead)	$R_{\theta JL}$	Mounted on the Allegro ASEK 722 evaluation board.	5	$^\circ\text{C}/\text{W}$

\*Additional thermal information available on the Allegro website.



Functional Block Diagram

### PINOUT DIAGRAM AND TERMINAL LIST



Pinout Diagram

#### TERMINAL LIST TABLE

Number	Name	Description
1, 2	IP+	Terminals for current being sensed; fused internally
3, 4	IP-	Terminals for current being sensed; fused internally
5	GND	Signal ground terminal
6	BW_SEL	Terminal for selecting 20 kHz or 80 kHz bandwidth
7	VIOUT	Analog output signal
8	VCC	Device power supply terminal

**COMMON ELECTRICAL CHARACTERISTICS [1]:** Valid through the full range of  $T_A = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ,  
and at  $V_{CC} = 3.3\text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
Supply Voltage	$V_{CC}$		3	3.3	3.6	V
Supply Current	$I_{CC}$	$V_{CC}$ within $V_{CC}(\text{min})$ and $V_{CC}(\text{max})$	–	9	12	mA
Output Capacitance Load	$C_L$	VIOOUT to GND	–	–	10	nF
Output Resistive Load	$R_L$	VIOOUT to GND	4.7	–	–	k $\Omega$
Primary Conductor Resistance	$R_{IP}$	$T_A = 25^\circ\text{C}$	–	0.65	–	m $\Omega$
Magnetic Coupling Factor	$C_F$		–	10	–	G/A
Rise Time	$t_r$	$I_P = I_P(\text{max})$ , $T_A = 25^\circ\text{C}$ , $C_L = 1\text{ nF}$ , BW_SEL tied to GND	–	4	–	$\mu\text{s}$
		$I_P = I_P(\text{max})$ , $T_A = 25^\circ\text{C}$ , $C_L = 1\text{ nF}$ , BW_SEL tied to VCC	–	17.5	–	$\mu\text{s}$
Propagation Delay	$t_{pd}$	$I_P = I_P(\text{max})$ , $T_A = 25^\circ\text{C}$ , $C_L = 1\text{ nF}$ , BW_SEL tied to GND	–	1	–	$\mu\text{s}$
		$I_P = I_P(\text{max})$ , $T_A = 25^\circ\text{C}$ , $C_L = 1\text{ nF}$ , BW_SEL tied to VCC	–	5	–	$\mu\text{s}$
Response Time	$t_{\text{RESPONSE}}$	$I_P = I_P(\text{max})$ , $T_A = 25^\circ\text{C}$ , $C_L = 1\text{ nF}$ , BW_SEL tied to GND	–	5	–	$\mu\text{s}$
		$I_P = I_P(\text{max})$ , $T_A = 25^\circ\text{C}$ , $C_L = 1\text{ nF}$ , BW_SEL tied to VCC	–	22.5	–	$\mu\text{s}$
Internal Bandwidth	BW <sub>i</sub>	Small signal –3 dB; $C_L = 1\text{ nF}$ , BW_SEL tied to GND	–	80	–	kHz
		Small signal –3 dB; $C_L = 1\text{ nF}$ , BW_SEL tied to VCC	–	20	–	kHz
Noise Density	$I_{ND}$	Input referenced noise density; $T_A = 25^\circ\text{C}$ , $C_L = 1\text{ nF}$	–	150	–	$\mu\text{A}_{(\text{rms})}/\sqrt{\text{Hz}}$
Noise	$I_N$	Input referenced noise; BW <sub>i</sub> = 80 kHz, $T_A = 25^\circ\text{C}$ , $C_L = 1\text{ nF}$	–	42	–	$\text{mA}_{(\text{rms})}$
		Input referenced noise; BW <sub>i</sub> = 20 kHz, $T_A = 25^\circ\text{C}$ , $C_L = 1\text{ nF}$	–	21	–	$\text{mA}_{(\text{rms})}$
Nonlinearity	$E_{LIN}$	Through full range of $I_P$	–	$\pm 1$	–	%
Saturation Voltage [2]	$V_{OH}$	$R_L = 4.7\text{ k}\Omega$ , $T_A = 25^\circ\text{C}$	$V_{CC} - 0.33$	–	–	V
	$V_{OL}$	$R_L = 4.7\text{ k}\Omega$ , $T_A = 25^\circ\text{C}$	–	–	0.33	V
Power-On Time	$t_{PO}$	Output reaches 90% of steady-state level, $T_A = 25^\circ\text{C}$ , $I_P = I_{PR}(\text{max})$ applied	–	64	–	$\mu\text{s}$

[1] Device may be operated at higher primary current levels,  $I_P$ , ambient temperatures,  $T_A$ , and internal leadframe temperatures, provided the Maximum Junction Temperature,  $T_J(\text{max})$ , is not exceeded.

[2] The sensor IC will continue to respond to current beyond the range of  $I_P$  until the high or low saturation voltage; however, the nonlinearity in this region will be worse than through the rest of the measurement range.

**xLLCTR-5AB PERFORMANCE CHARACTERISTICS:  $T_A$  Range L, valid at  $T_A = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ , unless otherwise specified**

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		-5	-	5	A
Sensitivity	Sens	$I_{PR}(\text{min}) < I_P < I_{PR}(\text{max})$	-	264	-	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional; $I_P = 0\text{ A}$	-	$V_{CC} \times 0.5$	-	V
<b>ACCURACY PERFORMANCE</b>						
Sensitivity Error	$E_{\text{sens}}$	$T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$ ; measured at $I_P = I_{PR}(\text{max})$	-2	-	2	%
		$T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$ ; ; measured at $I_P = I_{PR}(\text{max})$	-	$\pm 2.5$	-	%
Offset Voltage [1]	$V_{OE}$	$I_P = 0\text{ A}$ ; $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	-15	-	15	mV
		$I_P = 0\text{ A}$ ; $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	$\pm 20$	-	mV
Total Output Error [2]	$E_{TOT}$	$I_P = I_{PR}(\text{max})$ , $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	-2.5	-	2.5	%
		$I_P = I_{PR}(\text{max})$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	-	$\pm 3$	-	%
<b>LIFETIME DRIFT CHARACTERISTICS</b>						
Sensitivity Error Lifetime Drift	$E_{\text{sens\_drift}}$		-	$\pm 2$	-	%
Total Output Error Lifetime Drift	$E_{\text{tot\_drift}}$		-	$\pm 2$	-	%

[1] Offset Voltage does not incorporate any error due to external magnetic fields. See section: Impact of External Magnetic Fields.

[2] Percentage of  $I_P$ , with  $I_P = I_{PR}(\text{max})$ .

**xLLCTR-10AU PERFORMANCE CHARACTERISTICS:  $T_A$  Range L, valid at  $T_A = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ , unless otherwise specified**

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		0	–	10	A
Sensitivity	Sens	$I_{PR}(\text{min}) < I_P < I_{PR}(\text{max})$	–	264	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Unidirectional; $I_P = 0\text{ A}$	–	$V_{CC} \times 0.1$	–	V
<b>ACCURACY PERFORMANCE</b>						
Sensitivity Error	$E_{\text{sens}}$	$T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$ ; measured at $I_P = I_{PR}(\text{max})$	–2	–	2	%
		$T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$ ; ; measured at $I_P = I_{PR}(\text{max})$	–	$\pm 2.5$	–	%
Offset Voltage [1]	$V_{OE}$	$I_P = 0\text{ A}$ ; $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–15	–	15	mV
		$I_P = 0\text{ A}$ ; $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 20$	–	mV
Total Output Error [2]	$E_{TOT}$	$I_P = I_{PR}(\text{max})$ , $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–2.5	–	2.5	%
		$I_P = I_{PR}(\text{max})$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 3$	–	%
<b>LIFETIME DRIFT CHARACTERISTICS</b>						
Sensitivity Error Lifetime Drift	$E_{\text{sens\_drift}}$		–	$\pm 2$	–	%
Total Output Error Lifetime Drift	$E_{\text{tot\_drift}}$		–	$\pm 2$	–	%

[1] Offset Voltage does not incorporate any error due to external magnetic fields. See section: Impact of External Magnetic Fields.

[2] Percentage of  $I_P$ , with  $I_P = I_{PR}(\text{max})$ .

**xLLCTR-10AB PERFORMANCE CHARACTERISTICS:  $T_A$  Range L, valid at  $T_A = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ , unless otherwise specified**

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		–10	–	10	A
Sensitivity	Sens	$I_{PR}(\text{min}) < I_P < I_{PR}(\text{max})$	–	132	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional; $I_P = 0\text{ A}$	–	$V_{CC} \times 0.5$	–	V
<b>ACCURACY PERFORMANCE</b>						
Sensitivity Error	$E_{\text{sens}}$	$T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$ ; measured at $I_P = I_{PR}(\text{max})$	–1.5	–	1.5	%
		$T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$ ; ; measured at $I_P = I_{PR}(\text{max})$	–	$\pm 2$	–	%
Offset Voltage [1]	$V_{OE}$	$I_P = 0\text{ A}$ ; $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–10	–	10	mV
		$I_P = 0\text{ A}$ ; $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 15$	–	mV
Total Output Error [2]	$E_{TOT}$	$I_P = I_{PR}(\text{max})$ , $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–2	–	2	%
		$I_P = I_{PR}(\text{max})$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 3$	–	%
<b>LIFETIME DRIFT CHARACTERISTICS</b>						
Sensitivity Error Lifetime Drift	$E_{\text{sens\_drift}}$		–	$\pm 2$	–	%
Total Output Error Lifetime Drift	$E_{\text{tot\_drift}}$		–	$\pm 2$	–	%

[1] Offset Voltage does not incorporate any error due to external magnetic fields. See section: Impact of External Magnetic Fields.

[2] Percentage of  $I_P$ , with  $I_P = I_{PR}(\text{max})$ .

**xLLCTR-20AU PERFORMANCE CHARACTERISTICS:  $T_A$  Range L, valid at  $T_A = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ , unless otherwise specified**

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		0	–	20	A
Sensitivity	Sens	$I_{PR}(\text{min}) < I_P < I_{PR}(\text{max})$	–	132	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Unidirectional; $I_P = 0\text{ A}$	–	$V_{CC} \times 0.1$	–	V
<b>ACCURACY PERFORMANCE</b>						
Sensitivity Error	$E_{\text{sens}}$	$T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$ ; measured at $I_P = I_{PR}(\text{max})$	–1.5	–	1.5	%
		$T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$ ; ; measured at $I_P = I_{PR}(\text{max})$	–	$\pm 2$	–	%
Offset Voltage [1]	$V_{OE}$	$I_P = 0\text{ A}$ ; $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–10	–	10	mV
		$I_P = 0\text{ A}$ ; $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 15$	–	mV
Total Output Error [2]	$E_{TOT}$	$I_P = I_{PR}(\text{max})$ , $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–2	–	2	%
		$I_P = I_{PR}(\text{max})$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 3$	–	%
<b>LIFETIME DRIFT CHARACTERISTICS</b>						
Sensitivity Error Lifetime Drift	$E_{\text{sens\_drift}}$		–	$\pm 2$	–	%
Total Output Error Lifetime Drift	$E_{\text{tot\_drift}}$		–	$\pm 2$	–	%

[1] Offset Voltage does not incorporate any error due to external magnetic fields. See section: Impact of External Magnetic Fields.

[2] Percentage of  $I_P$ , with  $I_P = I_{PR}(\text{max})$ .

**xLLCTR-20AB PERFORMANCE CHARACTERISTICS:  $T_A$  Range L, valid at  $T_A = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ , unless otherwise specified**

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		–20	–	20	A
Sensitivity	Sens	$I_{PR}(\text{min}) < I_P < I_{PR}(\text{max})$	–	66	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional; $I_P = 0\text{ A}$	–	$V_{CC} \times 0.5$	–	V
<b>ACCURACY PERFORMANCE</b>						
Sensitivity Error	$E_{\text{sens}}$	$T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$ ; measured at $I_P = I_{PR}(\text{max})$	–1.5	–	1.5	%
		$T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$ ; ; measured at $I_P = I_{PR}(\text{max})$	–	$\pm 2$	–	%
Offset Voltage [1]	$V_{OE}$	$I_P = 0\text{ A}$ ; $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–10	–	10	mV
		$I_P = 0\text{ A}$ ; $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 15$	–	mV
Total Output Error [2]	$E_{TOT}$	$I_P = I_{PR}(\text{max})$ , $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–2	–	2	%
		$I_P = I_{PR}(\text{max})$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 3$	–	%
<b>LIFETIME DRIFT CHARACTERISTICS</b>						
Sensitivity Error Lifetime Drift	$E_{\text{sens\_drift}}$		–	$\pm 2$	–	%
Total Output Error Lifetime Drift	$E_{\text{tot\_drift}}$		–	$\pm 2$	–	%

[1] Offset Voltage does not incorporate any error due to external magnetic fields. See section: Impact of External Magnetic Fields.

[2] Percentage of  $I_P$ , with  $I_P = I_{PR}(\text{max})$ .

**xLLCTR-40AU PERFORMANCE CHARACTERISTICS:  $T_A$  Range L, valid at  $T_A = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ , unless otherwise specified**

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		0	–	40	A
Sensitivity	Sens	$I_{PR}(\text{min}) < I_P < I_{PR}(\text{max})$	–	66	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Unidirectional; $I_P = 0\text{ A}$	–	$V_{CC} \times 0.1$	–	V
<b>ACCURACY PERFORMANCE</b>						
Sensitivity Error	$E_{\text{sens}}$	$T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$ ; measured at $I_P = I_{PR}(\text{max})$	–1.5	–	1.5	%
		$T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$ ; ; measured at $I_P = I_{PR}(\text{max})$	–	$\pm 2$	–	%
Offset Voltage [1]	$V_{OE}$	$I_P = 0\text{ A}$ ; $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–10	–	10	mV
		$I_P = 0\text{ A}$ ; $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 15$	–	mV
Total Output Error [2]	$E_{TOT}$	$I_P = I_{PR}(\text{max})$ , $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–2	–	2	%
		$I_P = I_{PR}(\text{max})$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 3$	–	%
<b>LIFETIME DRIFT CHARACTERISTICS</b>						
Sensitivity Error Lifetime Drift	$E_{\text{sens\_drift}}$		–	$\pm 2$	–	%
Total Output Error Lifetime Drift	$E_{\text{tot\_drift}}$		–	$\pm 2$	–	%

[1] Offset Voltage does not incorporate any error due to external magnetic fields. See section: Impact of External Magnetic Fields.

[2] Percentage of  $I_P$ , with  $I_P = I_{PR}(\text{max})$ .

**xLLCTR-40AB PERFORMANCE CHARACTERISTICS:  $T_A$  Range L, valid at  $T_A = -40^\circ\text{C}$  to  $150^\circ\text{C}$ ,  $V_{CC} = 3.3\text{ V}$ , unless otherwise specified**

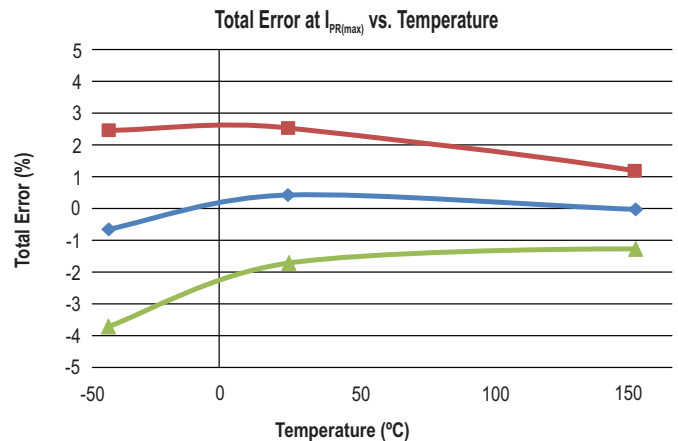
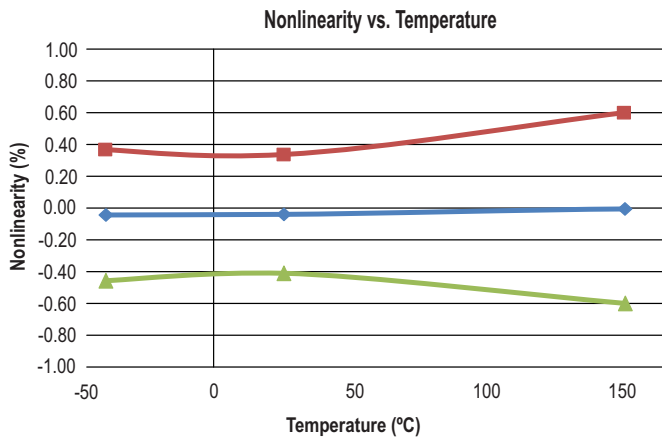
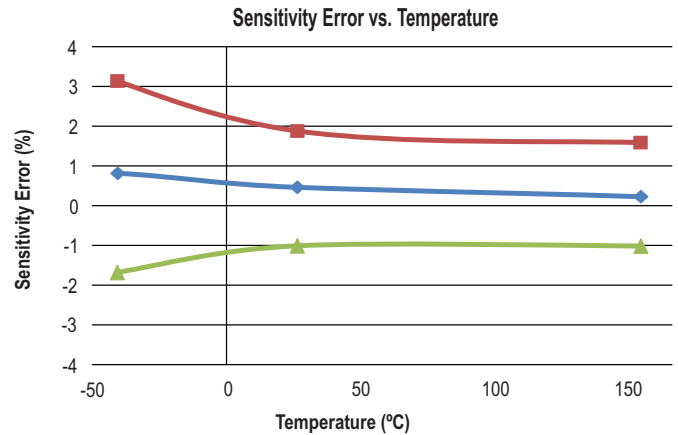
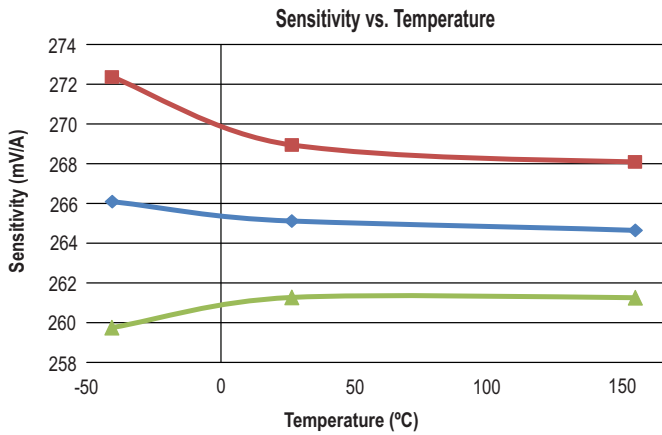
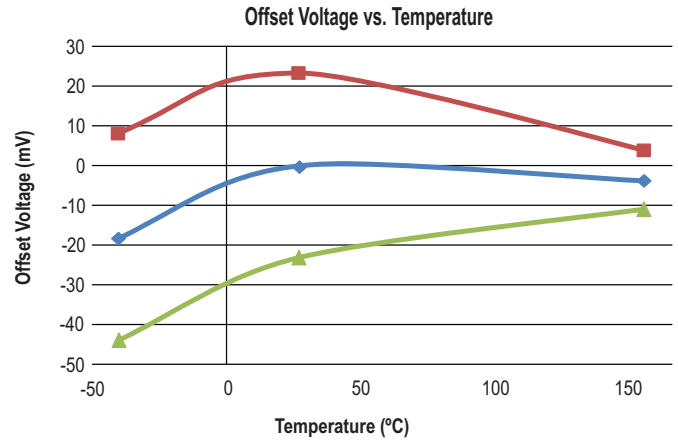
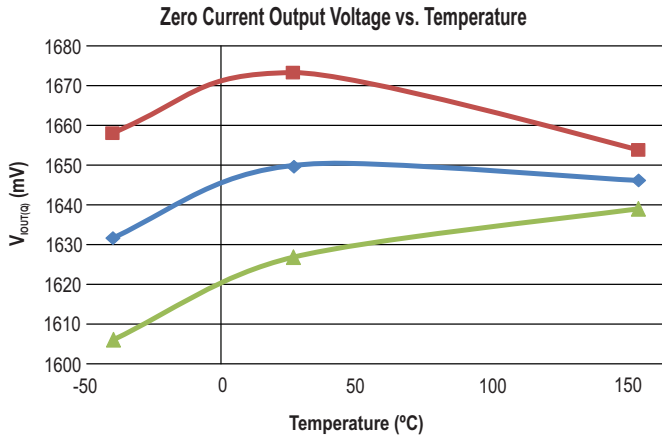
Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Units
<b>NOMINAL PERFORMANCE</b>						
Current Sensing Range	$I_{PR}$		–40	–	40	A
Sensitivity	Sens	$I_{PR}(\text{min}) < I_P < I_{PR}(\text{max})$	–	33	–	mV/A
Zero Current Output Voltage	$V_{IOUT(Q)}$	Bidirectional; $I_P = 0\text{ A}$	–	$V_{CC} \times 0.5$	–	V
<b>ACCURACY PERFORMANCE</b>						
Sensitivity Error	$E_{\text{sens}}$	$T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$ ; measured at $I_P = I_{PR}(\text{max})$	–1.5	–	1.5	%
		$T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$ ; ; measured at $I_P = I_{PR}(\text{max})$	–	$\pm 2$	–	%
Offset Voltage [1]	$V_{OE}$	$I_P = 0\text{ A}$ ; $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–10	–	10	mV
		$I_P = 0\text{ A}$ ; $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 15$	–	mV
Total Output Error [2]	$E_{TOT}$	$I_P = I_{PR}(\text{max})$ , $T_A = 25^\circ\text{C}$ to $150^\circ\text{C}$	–2	–	2	%
		$I_P = I_{PR}(\text{max})$ , $T_A = -40^\circ\text{C}$ to $25^\circ\text{C}$	–	$\pm 3$	–	%
<b>LIFETIME DRIFT CHARACTERISTICS</b>						
Sensitivity Error Lifetime Drift	$E_{\text{sens\_drift}}$		–	$\pm 2$	–	%
Total Output Error Lifetime Drift	$E_{\text{tot\_drift}}$		–	$\pm 2$	–	%

[1] Offset Voltage does not incorporate any error due to external magnetic fields. See section: Impact of External Magnetic Fields.

[2] Percentage of  $I_P$ , with  $I_P = I_{PR}(\text{max})$ .

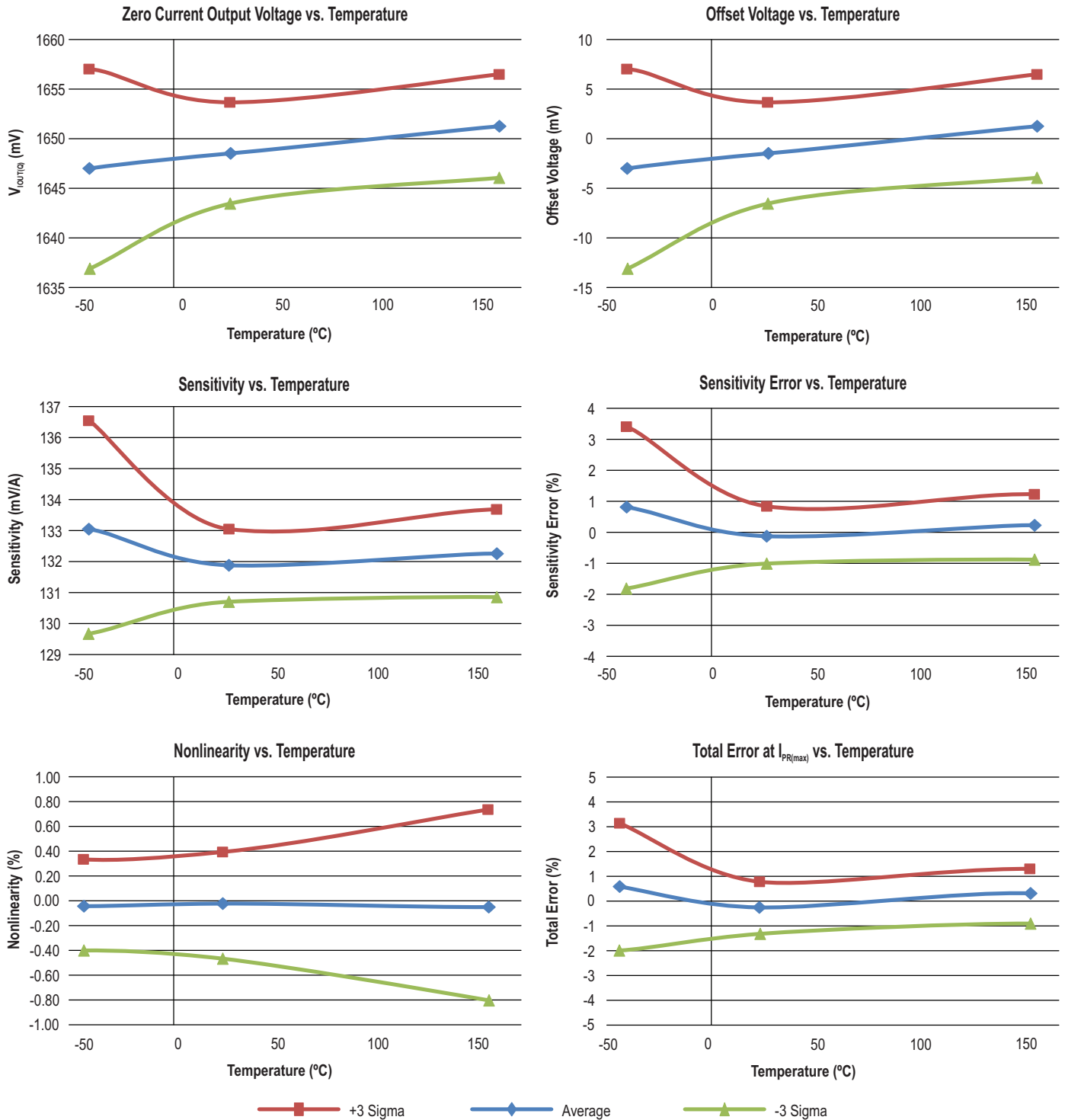
### CHARACTERISTIC PERFORMANCE

#### xLLCTR-5AB Key Parameters

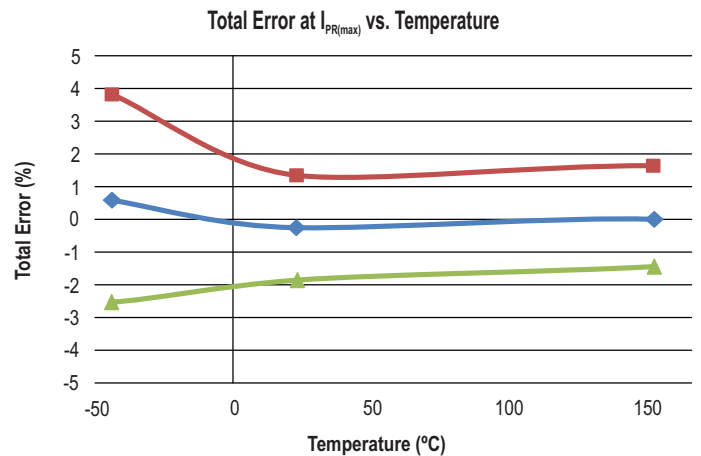
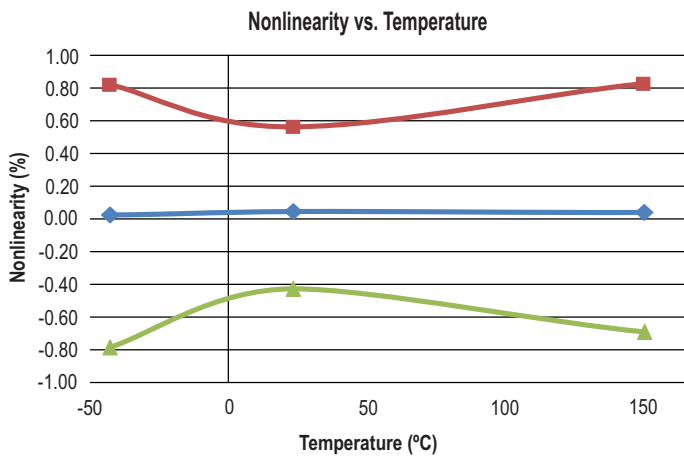
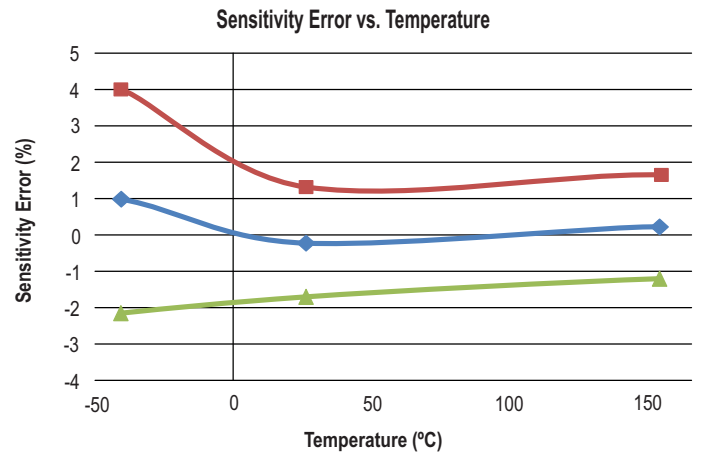
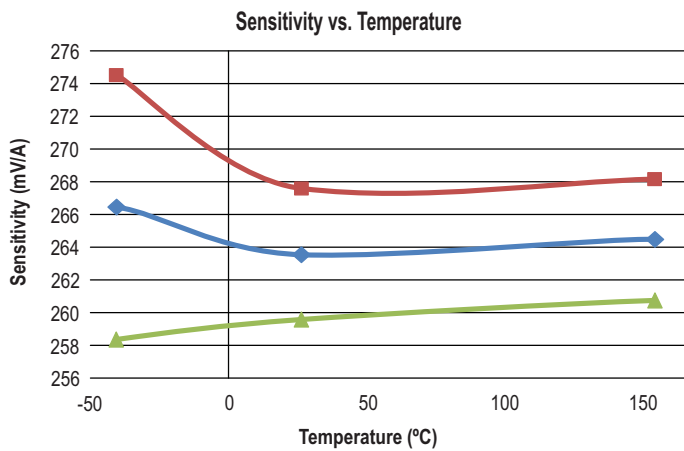
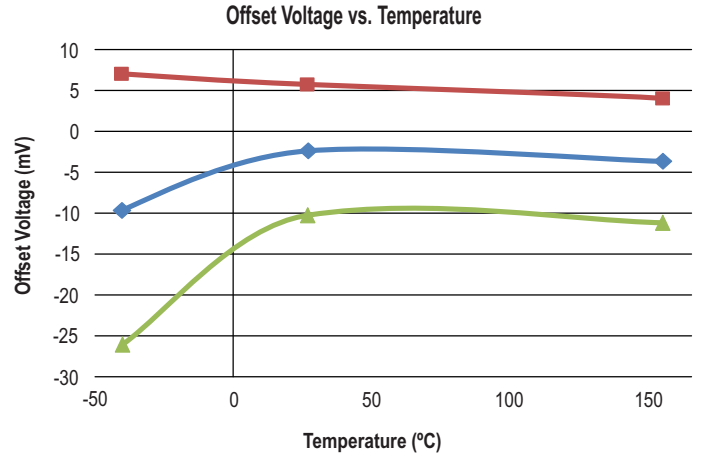
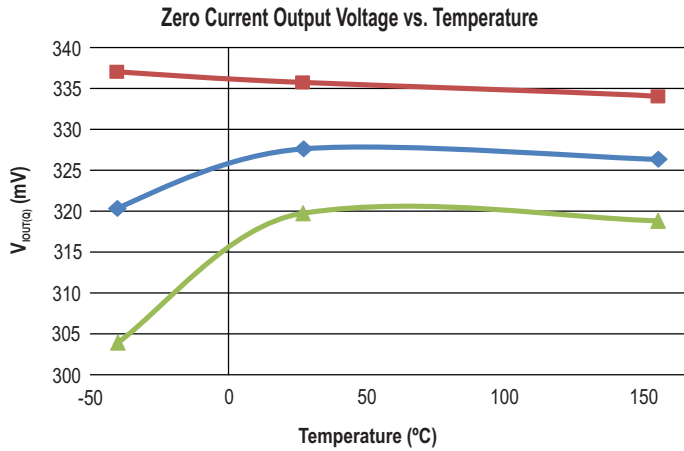


■ +3 Sigma     
 ◆ Average     
 ▲ -3 Sigma

### xLLCTR-10AB Key Parameters

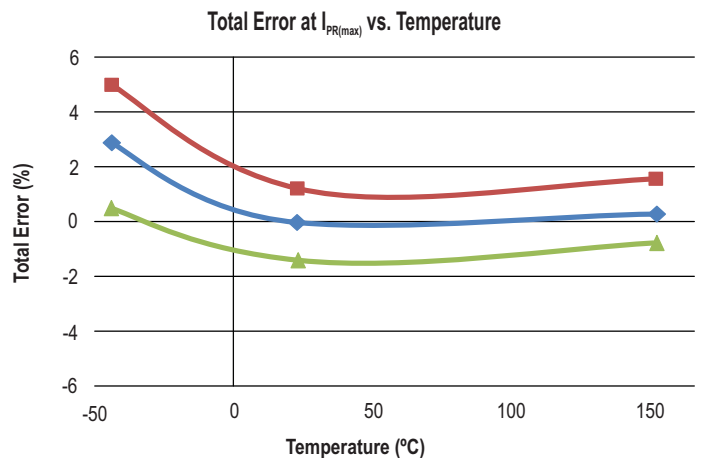
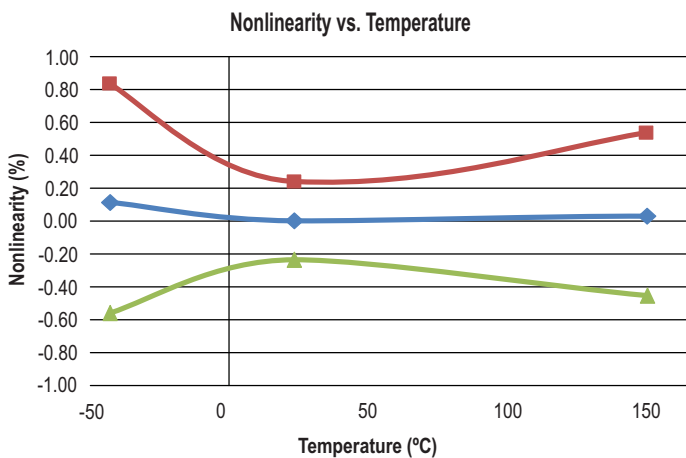
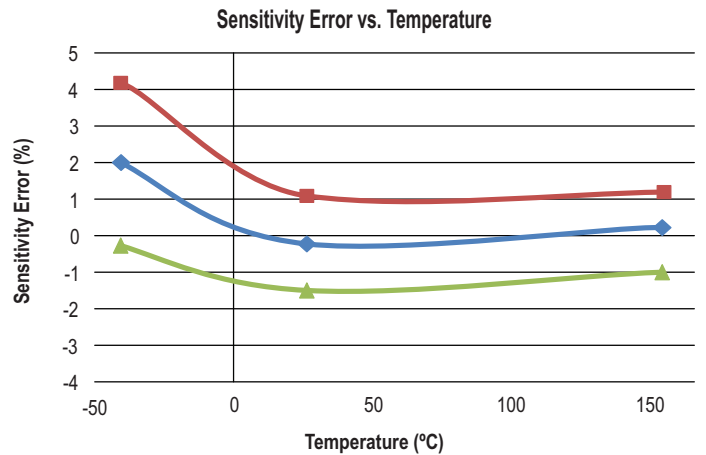
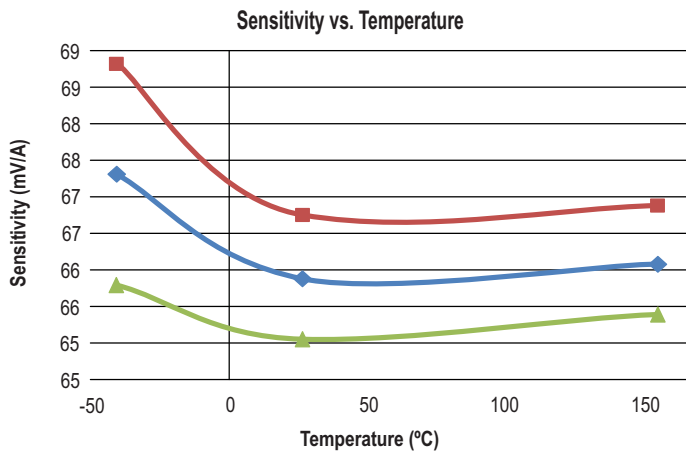
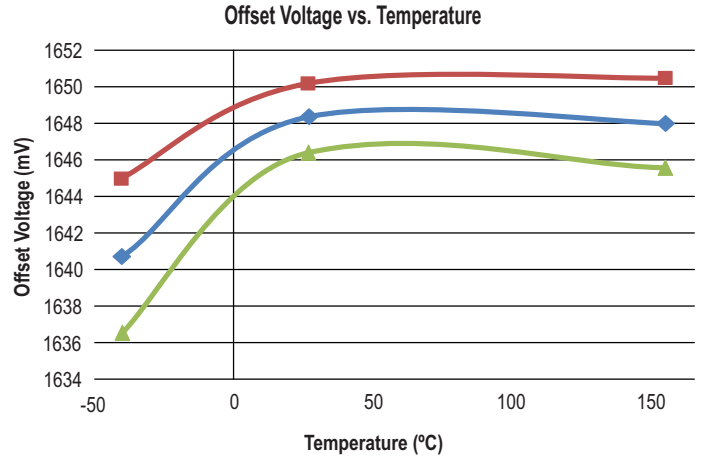
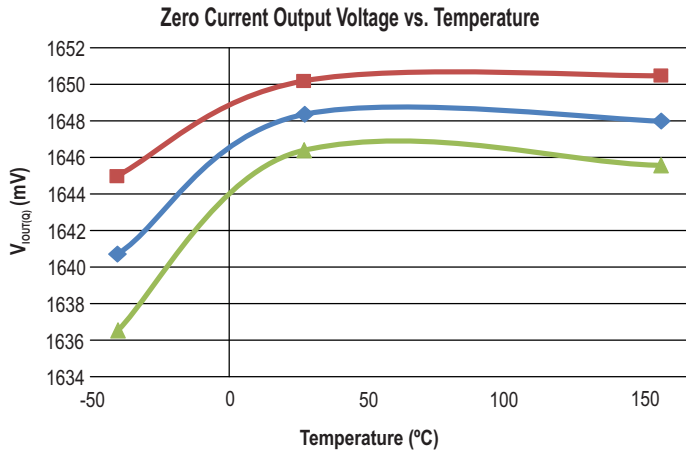


### xLLCTR-10AU Key Parameters



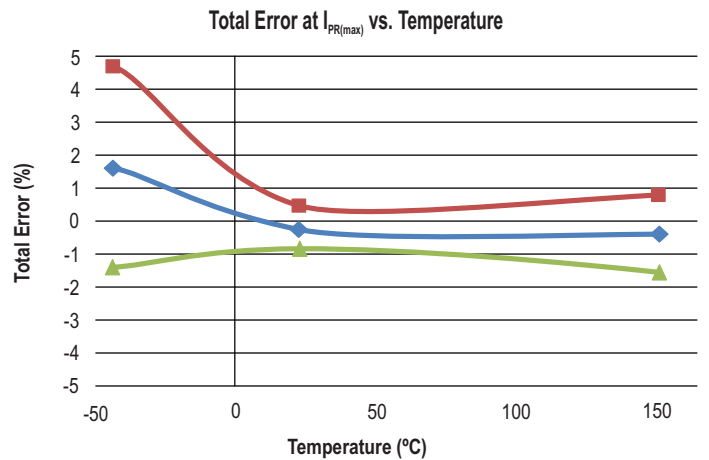
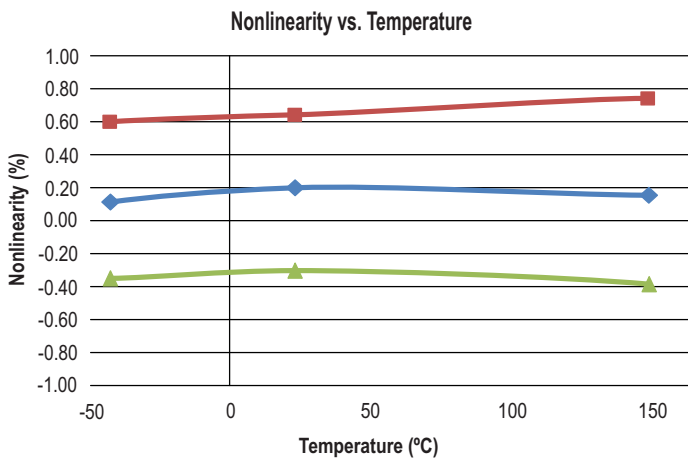
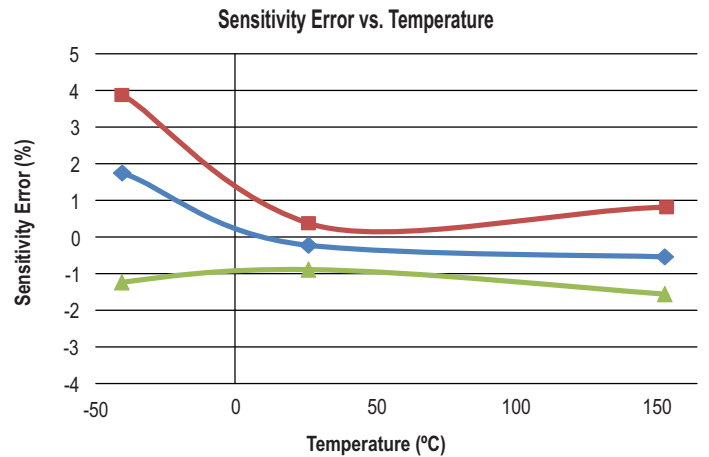
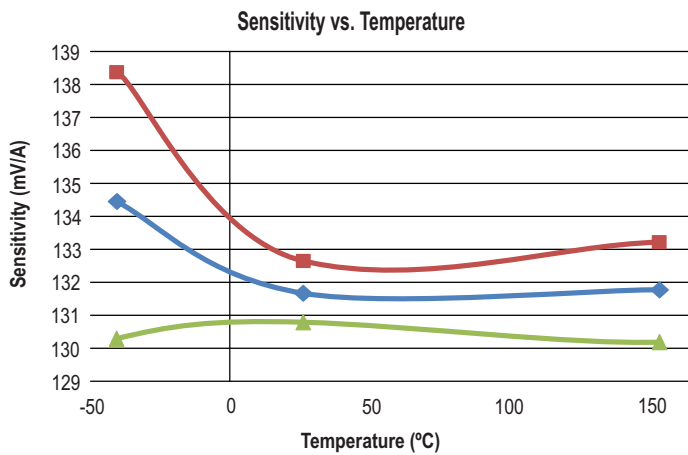
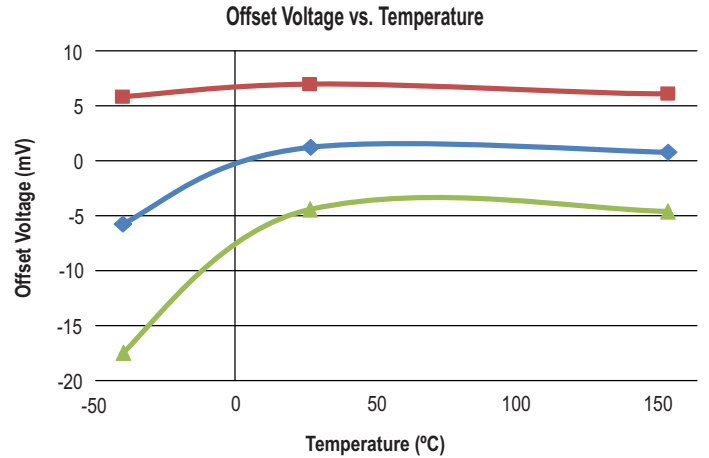
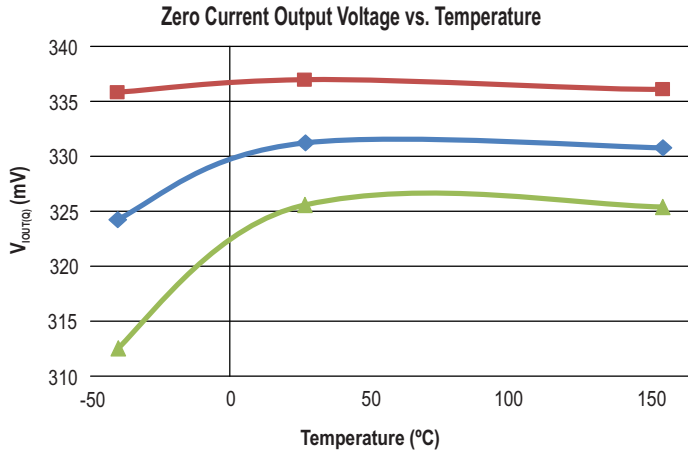
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### xLLCTR-20AB Key Parameters



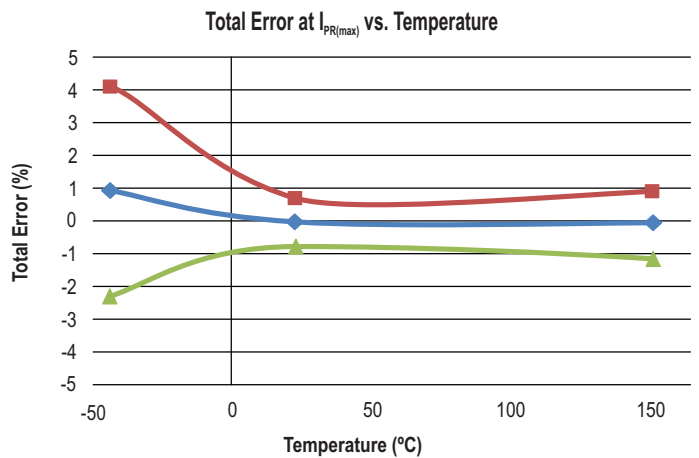
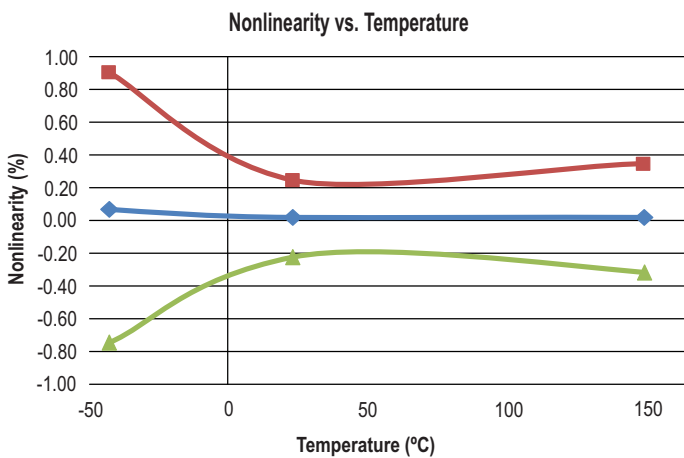
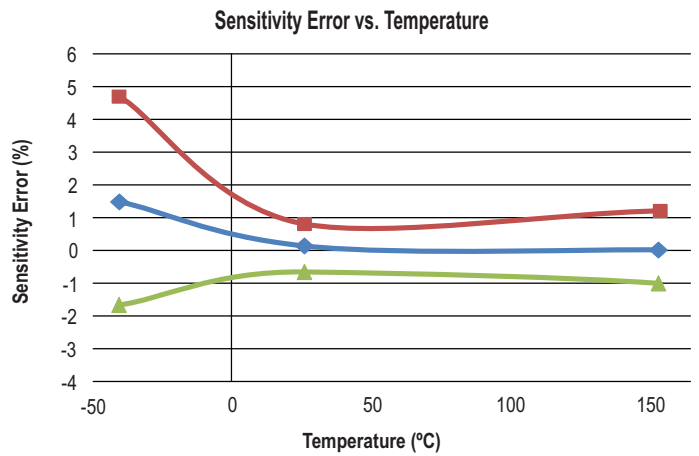
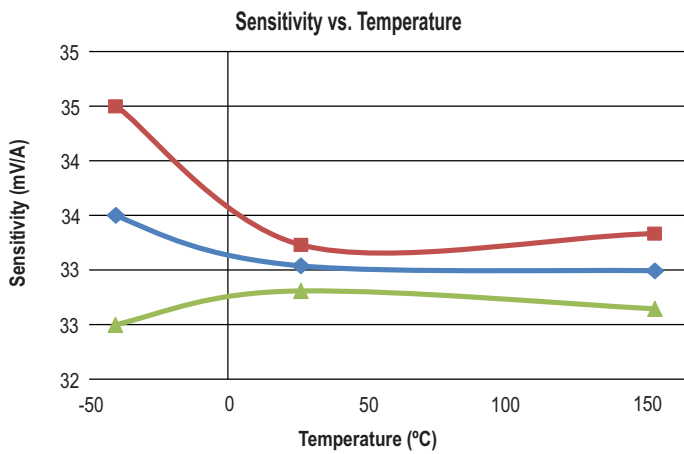
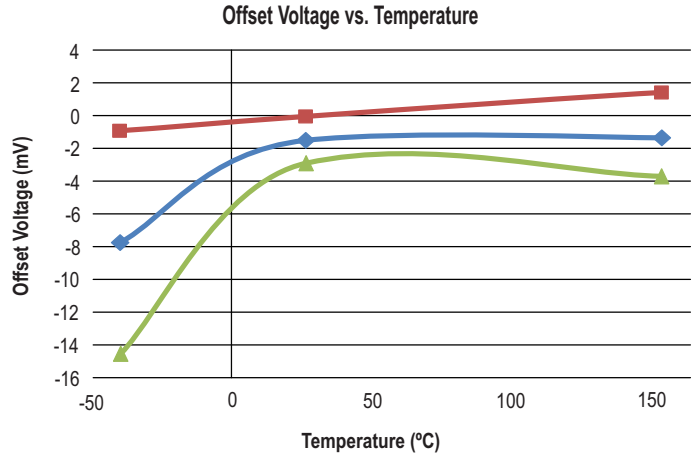
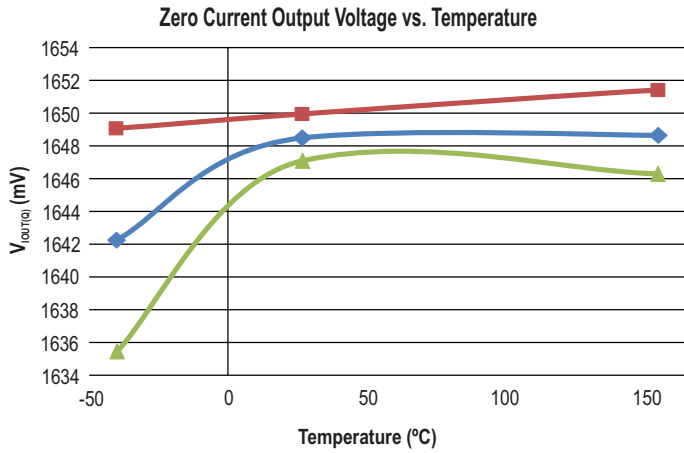
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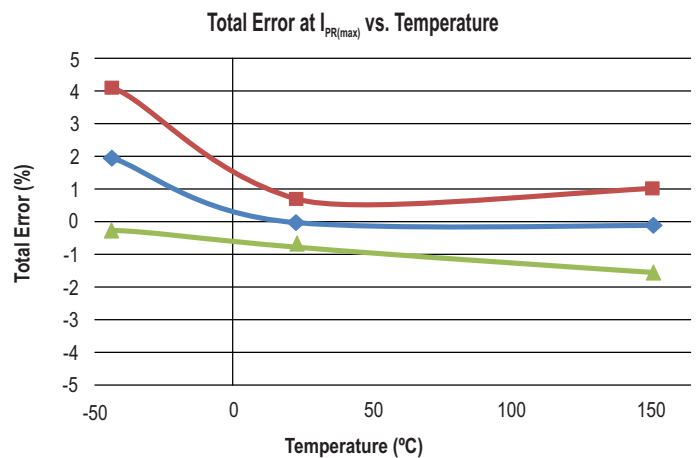
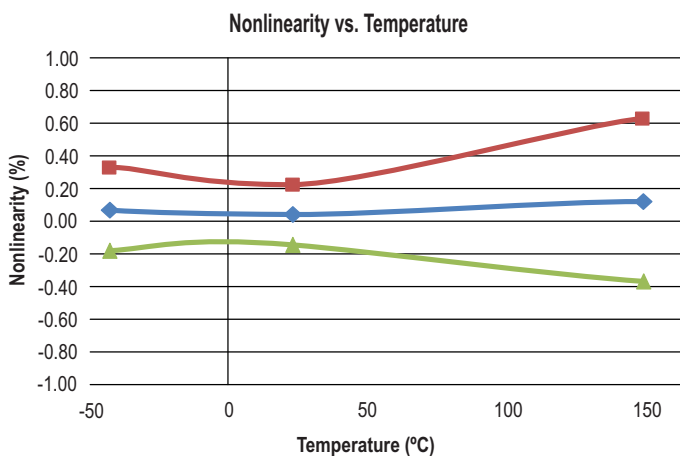
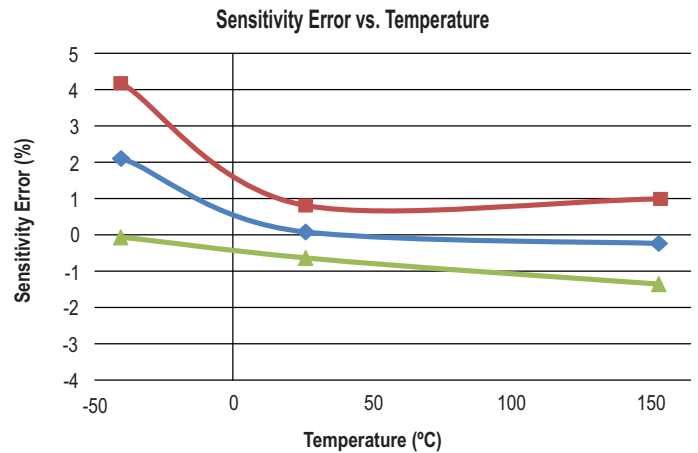
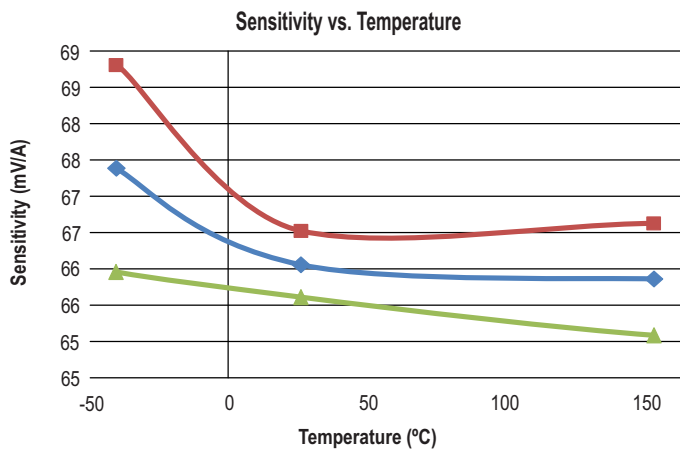
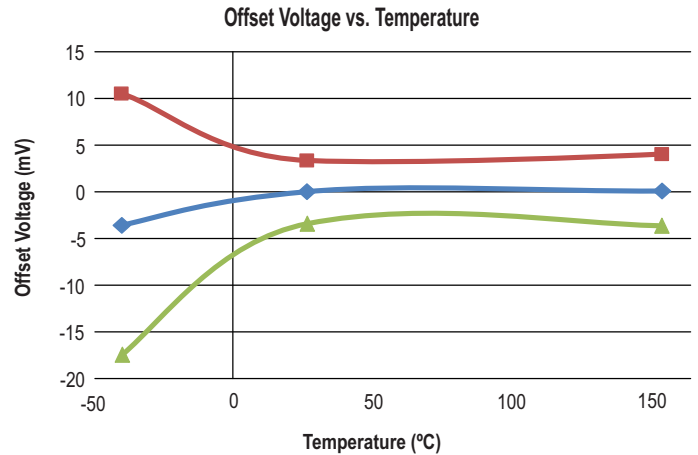
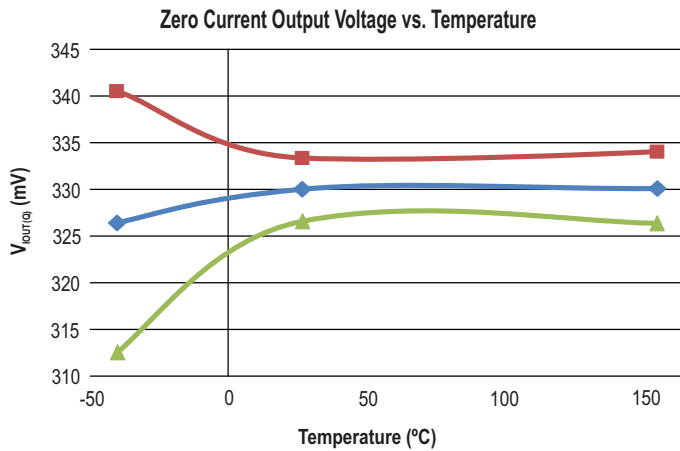
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 ▲ -3 Sigma

### xLLCTR-40AB Key Parameters



■ +3 Sigma     
 ◆ Average     
 ▲ -3 Sigma

### xLLCTR-40AU Key Parameters



■ +3 Sigma     
 ◆ Average     
 ▲ -3 Sigma

## DEFINITIONS OF ACCURACY CHARACTERISTICS

### Sensitivity (Sens)

The change in sensor IC output in response to a 1A change through the primary conductor. The sensitivity is the product of the magnetic circuit sensitivity (G/A) (1 G = 0.1 mT) and the linear IC amplifier gain (mV/G). The linear IC amplifier gain is programmed at the factory to optimize the sensitivity (mV/A) for the full-scale current of the device.

### Nonlinearity ( $E_{LIN}$ )

The nonlinearity is a measure of how linear the output of the sensor IC is over the full current measurement range. The nonlinearity is calculated as:

$$E_{LIN} = \left\{ 1 - \left[ \frac{V_{IOUT}(I_{PR(max)}) - V_{IOUT(Q)}}{2 \times V_{IOUT}(I_{PR(max)/2}) - V_{IOUT(Q)}} \right] \right\} \times 100 (\%)$$

where  $V_{IOUT}(I_{PR(max)})$  is the output of the sensor IC with the maximum measurement current flowing through it and  $V_{IOUT}(I_{PR(max)/2})$  is the output of the sensor IC with half of the maximum measurement current flowing through it.

### Zero Current Output Voltage ( $V_{IOUT(Q)}$ )

The output of the sensor when the primary current is zero. For a unipolar supply voltage, it nominally remains at  $0.5 \times V_{CC}$  for a bidirectional device and  $0.1 \times V_{CC}$  for a unidirectional device. For example, in the case of a bidirectional output device,  $V_{CC} = 3.3 \text{ V}$  translates into  $V_{IOUT(Q)} = 1.65 \text{ V}$ . Variation in  $V_{IOUT(Q)}$  can be attributed to the resolution of the Allegro linear IC quiescent voltage trim and thermal drift.

### Offset Voltage ( $V_{OE}$ )

The deviation of the device output from its ideal quiescent value of  $0.5 \times V_{CC}$  (bidirectional) or  $0.1 \times V_{CC}$  (unidirectional) due to nonmagnetic causes. To convert this voltage to amperes, divide by the device sensitivity, Sens.

### Total Output Error ( $E_{TOT}$ )

The difference between the current measurement from the sensor IC and the actual current ( $I_P$ ), relative to the actual current. This is equivalent to the difference between the ideal output voltage and the actual output voltage, divided by the ideal sensitivity, relative to the current flowing through the primary conduction path:

$$E_{TOT}(I_P) = \frac{V_{IOUT\_ideal}(I_P) - V_{IOUT}(I_P)}{\text{Sens}_{ideal}(I_P) \times I_P} \times 100 (\%)$$

The Total Output Error incorporates all sources of error and is a function of  $I_P$ . At relatively high currents,  $E_{TOT}$  will be mostly

due to sensitivity error, and at relatively low currents,  $E_{TOT}$  will be mostly due to Offset Voltage ( $V_{OE}$ ). In fact, at  $I_P = 0$ ,  $E_{TOT}$  approaches infinity due to the offset. This is illustrated in Figures 1 and 2. Figure 1 shows a distribution of output voltages versus  $I_P$  at 25°C and across temperature. Figure 2 shows the corresponding  $E_{TOT}$  versus  $I_P$ .

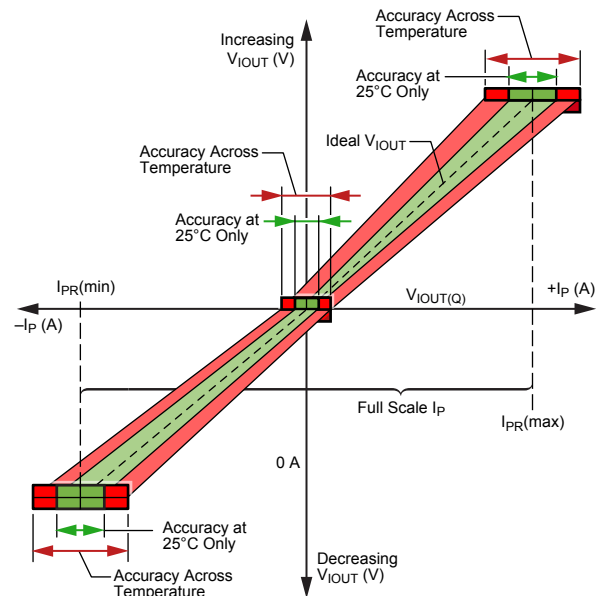


Figure 1: Output Voltage versus Sensed Current

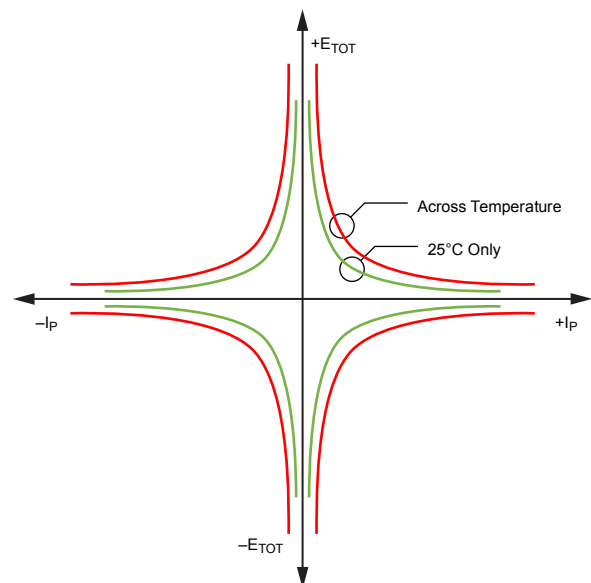


Figure 2: Total Output Error versus Sensed Current

### APPLICATION INFORMATION

#### Impact of External Magnetic Fields

The ACS722 works by sensing the magnetic field created by the current flowing through the package. However, the sensor cannot differentiate between fields created by the current flow and external magnetic fields. This means that external magnetic fields can cause errors in the output of the sensor. Magnetic fields which are perpendicular to the surface of the package affect the output of the sensor, as it only senses fields in that one plane. The error in Amperes can be quantified as:

$$Error(B) = \frac{B}{C_F}$$

where B is the strength of the external field perpendicular to the

surface of the package in Gauss, and  $C_F$  is the coupling factor in G/A. Then, multiplying by the sensitivity of the part, Sens, gives the error in mV.

For example, an external field of 1 Gauss will result in around 0.1 A of error. If the ACS722LLCTR-10AB, which has a nominal sensitivity of 132 mV/A, is being used, that equates to 13.2 mV of error on the output of the sensor.

External Field (Gauss)	Error (A)	Error (mV)			
		5AB	10AB	20AB	40AB
0.5	0.05	13.2	6.6	3.3	1.65
1	0.1	26.4	13.2	6.6	3.3
2	0.2	52.8	26.4	13.2	6.6

### Thermal Rise vs. Primary Current

Self-heating due to the flow of current should be considered during the design of any current sensing system. The sensor, printed circuit board (PCB), and contacts to the PCB will generate heat as current moves through the system.

The thermal response is highly dependent on PCB layout, copper thickness, cooling techniques, and the profile of the injected current. The current profile includes peak current, current “on-time”, and duty cycle. While the data presented in this section was collected with direct current (DC), these numbers may be used to approximate thermal response for both AC signals and current pulses.

The plot in Figure 3 shows the measured rise in steady-state die temperature of the ACS722 versus continuous current at an ambient temperature,  $T_A$ , of 25 °C. The thermal offset curves may be directly applied to other values of  $T_A$ . Conversely, Figure 4 shows the maximum continuous current at a given  $T_A$ . Surges beyond the maximum current listed in Figure 4 are allowed given the maximum junction temperature,  $T_{J(MAX)}$  (165°C), is not exceeded.

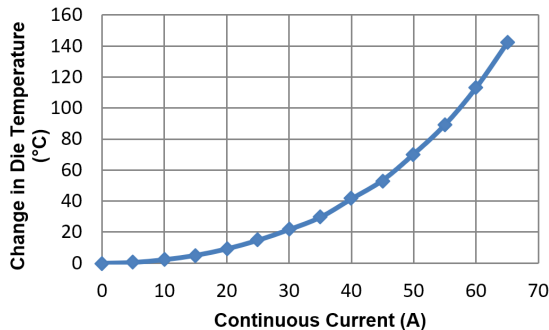


Figure 3: Self-Heating in the LC2 Package Due to Current Flow

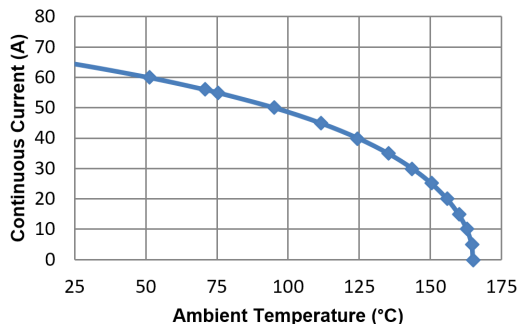


Figure 4: Maximum Continuous Current at a Given  $T_A$

The thermal capacity of the ACS722 should be verified by the end user in the application’s specific conditions. The maximum junction temperature,  $T_{J(MAX)}$  (165°C), should not be exceeded. Further information on this application testing is available in the [DC and Transient Current Capability application note](#) on the Allegro website.

### ASEK722 Evaluation Board Layout

Thermal data shown in Figure 3 was collected using the ASEK722 Evaluation Board (TED-85-0593-002). This board includes 1388 mm<sup>2</sup> of 4 oz. copper (0.1388) connected to pins 1 and 2, and to pins 3 and 4, with thermal vias connecting the layers. Top and bottom layers of the PCB are shown below in Figure 5.

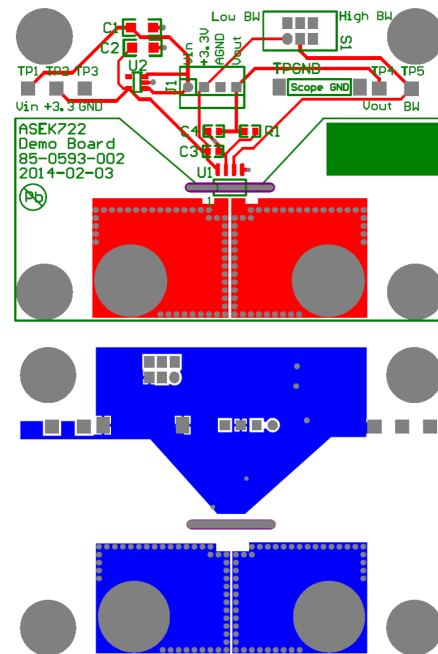


Figure 5: Top and Bottom Layers for ASEK722 Evaluation Board

Gerber files for the ASEK722 evaluation board are available for download from the Allegro website. See the technical documents section of the [ACS722 device webpage](#).

## DEFINITIONS OF DYNAMIC RESPONSE CHARACTERISTICS

### Power-On Time ( $t_{PO}$ )

When the supply is ramped to its operating voltage, the device requires a finite time to power its internal components before responding to an input magnetic field.

Power-On Time,  $t_{PO}$ , is defined as the time it takes for the output voltage to settle within  $\pm 10\%$  of its steady state value under an applied magnetic field, after the power supply has reached its minimum specified operating voltage,  $V_{CC(min)}$ , as shown in the chart at right.

### Rise Time ( $t_r$ )

The time interval between a) when the sensor IC reaches 10% of its full scale value, and b) when it reaches 90% of its full scale value. The rise time to a step response is used to derive the bandwidth of the current sensor IC, in which  $f(-3 \text{ dB}) = 0.35/t_r$ . Both  $t_r$  and  $t_{RESPONSE}$  are detrimentally affected by eddy current losses observed in the conductive IC ground plane.

### Propagation Delay ( $t_{pd}$ )

The propagation delay is measured as the time interval a) when the primary current signal reaches 20% of its final value, and b) when the device reaches 20% of its output corresponding to the applied current.

### Response Time ( $t_{RESPONSE}$ )

The time interval between a) when the primary current signal reaches 90% of its final value, and b) when the device reaches 90% of its output corresponding to the applied current.

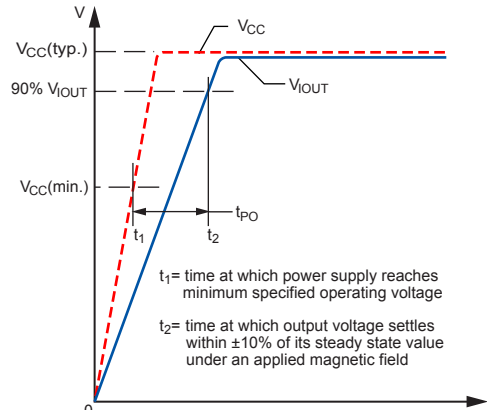


Figure 6: Power-On Time ( $t_{PO}$ )

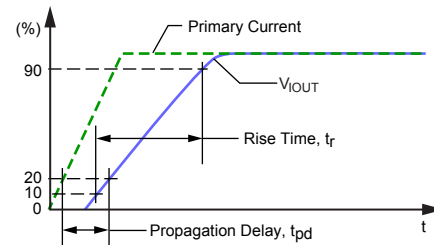


Figure 7: Rise Time ( $t_r$ ) and Propagation Delay ( $t_{pd}$ )

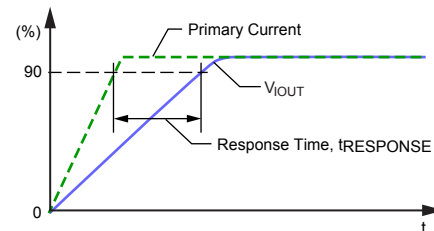


Figure 8: Response Time ( $t_{RESPONSE}$ )

### PACKAGE OUTLINE DRAWING

#### For Reference Only – Not for Tooling Use

(Reference Allegro DWG-0000385, Rev. 2 or JEDEC MS-012AA)

Dimensions in millimeters – NOT TO SCALE

Dimensions exclusive of mold flash, gate burrs, and dambar protrusions  
Exact case and lead configuration at supplier discretion within limits shown

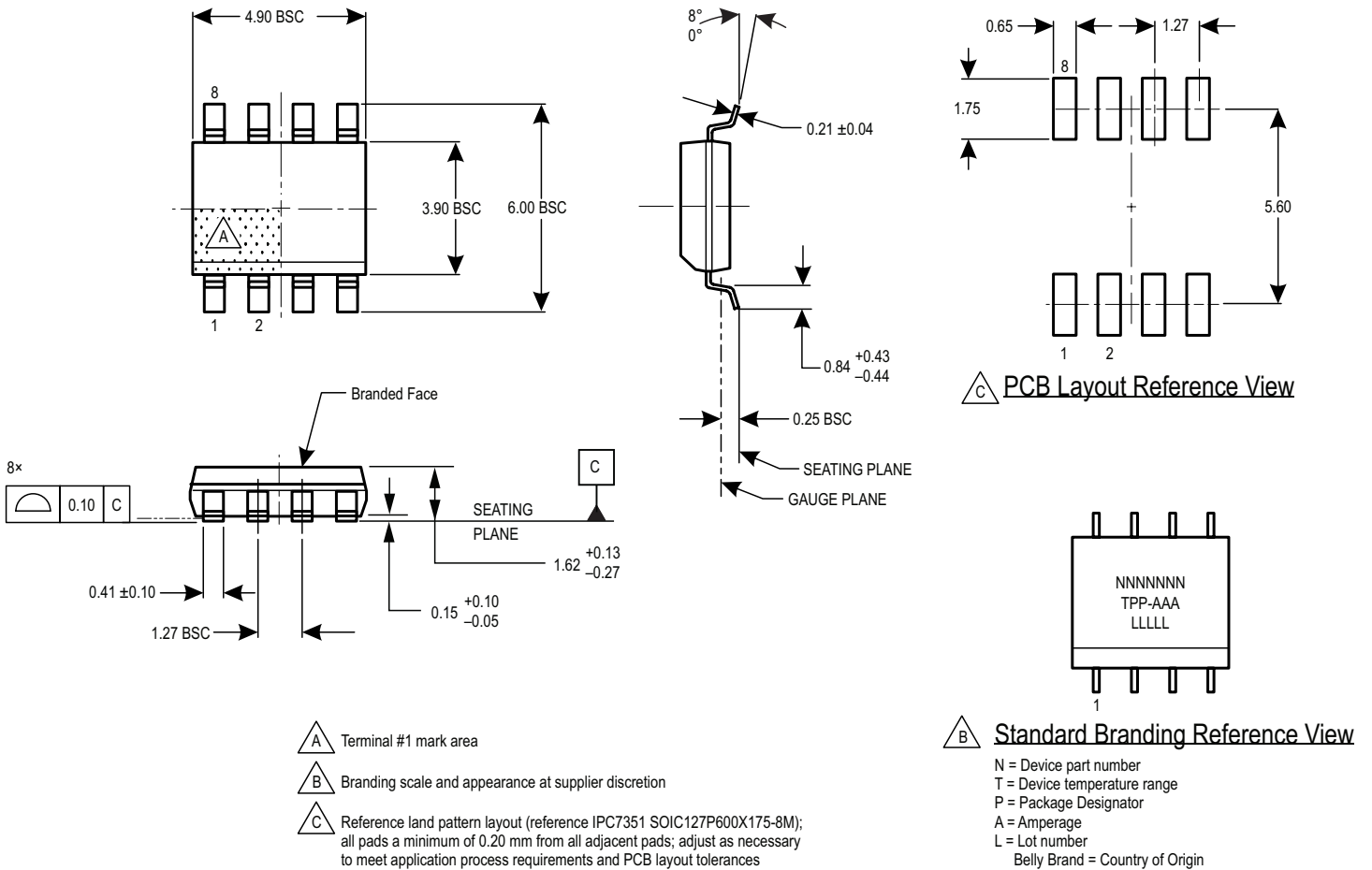


Figure 9: Package LC, 8-pin SOICN

### REVISION HISTORY

Number	Date	Description
–	June 10, 2014	Initial release.
1	October 29, 2014	Added Magnetic Coupling Factor characteristic and Error Due to External Magnetic Fields section
2	April 29, 2015	Added Characteristic Performance graphs
3	December 11, 2018	Updated certificate numbers
4	June 3, 2019	Updated TUV certificate mark
5	September 3, 2019	Added Maximum Continuous Current to Absolute Maximum Ratings table (page 3) and thermal data section (page 20)
6	September 9, 2021	Updated package drawing (page 22)
7	November 8, 2023	Added -S variants (pages 1-2)

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