

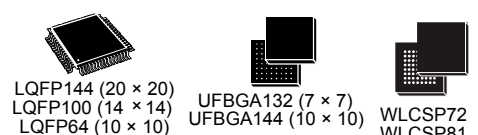


THE DATASHEET OF STM32L476QGI6

Ultra-low-power Arm[®] Cortex[®]-M4 32-bit MCU+FPU, 100DMIPS, up to 1MB Flash, 128 KB SRAM, USB OTG FS, LCD, ext. SMPS

Datasheet - production data

Features

- Ultra-low-power with FlexPowerControl
 - 1.71 V to 3.6 V power supply
 - -40 °C to 85/105/125 °C temperature range
 - 300 nA in V_{BAT} mode: supply for RTC and 32x32-bit backup registers
 - 30 nA Shutdown mode (5 wakeup pins)
 - 120 nA Standby mode (5 wakeup pins)
 - 420 nA Standby mode with RTC
 - 1.1 µA Stop 2 mode, 1.4 µA with RTC
 - 100 µA/MHz run mode (LDO Mode)
 - 39 µA/MHz run mode (@3.3 V SMPS Mode)
 - Batch acquisition mode (BAM)
 - 4 µs wakeup from Stop mode
 - Brown out reset (BOR)
 - Interconnect matrix
 - Core: Arm[®] 32-bit Cortex[®]-M4 CPU with FPU, Adaptive real-time accelerator (ART Accelerator™) allowing 0-wait-state execution from Flash memory, frequency up to 80 MHz, MPU, 100DMIPS and DSP instructions
 - Performance benchmark
 - 1.25 DMIPS/MHz (Drystone 2.1)
 - 273.55 CoreMark[®] (3.42 CoreMark/MHz @ 80 MHz)
 - Energy benchmark
 - 294 ULPMark™ CP score
 - 106 ULPMark™ PP score
 - Clock Sources
 - 4 to 48 MHz crystal oscillator
 - 32 kHz crystal oscillator for RTC (LSE)
 - Internal 16 MHz factory-trimmed RC (±1%)
 - Internal low-power 32 kHz RC (±5%)
 - Internal multispeed 100 kHz to 48 MHz oscillator, auto-trimmed by LSE (better than ±0.25 % accuracy)
 - 3 PLLs for system clock, USB, audio, ADC
- 

LQFP144 (20 × 20)
LQFP100 (14 × 14)
LQFP64 (10 × 10)
UFBGA132 (7 × 7)
UFBGA144 (10 × 10)
WLCSP72
WLCSP81

 - Up to 114 fast I/Os, most 5 V-tolerant, up to 14 I/Os with independent supply down to 1.08 V
 - RTC with HW calendar, alarms and calibration
 - LCD 8× 40 or 4× 44 with step-up converter
 - Up to 24 capacitive sensing channels: support touchkey, linear and rotary touch sensors
 - 16x timers: 2x 16-bit advanced motor-control, 2x 32-bit and 5x 16-bit general purpose, 2x 16-bit basic, 2x low-power 16-bit timers (available in Stop mode), 2x watchdogs, SysTick timer
 - Memories
 - Up to 1 MB Flash, 2 banks read-while-write, proprietary code readout protection
 - Up to 128 KB of SRAM including 32 KB with hardware parity check
 - External memory interface for static memories supporting SRAM, PSRAM, NOR and NAND memories
 - Quad SPI memory interface
 - 4x digital filters for sigma delta modulator
 - Rich analog peripherals (independent supply)
 - 3x 12-bit ADC 5 Msps, up to 16-bit with hardware oversampling, 200 µA/Msps
 - 2x 12-bit DAC output channels, low-power sample and hold
 - 2x operational amplifiers with built-in PGA
 - 2x ultra-low-power comparators
 - 20x communication interfaces
 - USB OTG 2.0 full-speed, LPM and BCD
 - 2x SAs (serial audio interface)
 - 3x I2C FM+(1 Mbit/s), SMBus/PMBus
 - 5x USARTs (ISO 7816, LIN, IrDA, modem)
 - 1x LPUART (Stop 2 wake-up)

- 3x SPIs (and 1x Quad SPI)
- CAN (2.0B Active) and SDMMC interface
- SWPMI single wire protocol master I/F
- IRTIM (Infrared interface)
- 14-channel DMA controller
- True random number generator
- CRC calculation unit, 96-bit unique ID
- Development support: serial wire debug (SWD), JTAG, Embedded Trace Macrocell™
- All packages are ECOPACK2® compliant

Table 1. Device summary

Reference	Part numbers
STM32L476xx	STM32L476RG, STM32L476JG, STM32L476MG, STM32L476ME, STM32L476VG, STM32L476QG, STM32L476ZG, STM32L476RE, STM32L476JE, STM32L476VE, STM32L476QE, STM32L476ZE, STM32L476RC, STM32L476VC

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1 Introduction

This datasheet provides the ordering information and mechanical device characteristics of the STM32L476xx microcontrollers.

This document should be read in conjunction with the STM32L4x6 reference manual (RM0351). The reference manual is available from the STMicroelectronics website www.st.com.

For information on the Arm^{®(a)} Cortex[®]-M4 core, please refer to the Cortex[®]-M4 Technical Reference Manual, available from the www.arm.com website.



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2 Description

The STM32L476xx devices are the ultra-low-power microcontrollers based on the high-performance Arm® Cortex®-M4 32-bit RISC core operating at a frequency of up to 80 MHz. The Cortex-M4 core features a Floating point unit (FPU) single precision which supports all Arm® single-precision data-processing instructions and data types. It also implements a full set of DSP instructions and a memory protection unit (MPU) which enhances application security.

The STM32L476xx devices embed high-speed memories (Flash memory up to 1 Mbyte, up to 128 Kbyte of SRAM), a flexible external memory controller (FSMC) for static memories (for devices with packages of 100 pins and more), a Quad SPI flash memories interface (available on all packages) and an extensive range of enhanced I/Os and peripherals connected to two APB buses, two AHB buses and a 32-bit multi-AHB bus matrix.

The STM32L476xx devices embed several protection mechanisms for embedded Flash memory and SRAM: readout protection, write protection, proprietary code readout protection and Firewall.

The devices offer up to three fast 12-bit ADCs (5 Msps), two comparators, two operational amplifiers, two DAC channels, an internal voltage reference buffer, a low-power RTC, two general-purpose 32-bit timer, two 16-bit PWM timers dedicated to motor control, seven general-purpose 16-bit timers, and two 16-bit low-power timers. The devices support four digital filters for external sigma delta modulators (DFSDM).

In addition, up to 24 capacitive sensing channels are available. The devices also embed an integrated LCD driver 8x40 or 4x44, with internal step-up converter.

They also feature standard and advanced communication interfaces.

- Three I2Cs
- Three SPIs
- Three USARTs, two UARTs and one Low-Power UART.
- Two SAs (Serial Audio Interfaces)
- One SDMMC
- One CAN
- One USB OTG full-speed
- One SWPMI (Single Wire Protocol Master Interface)

The STM32L476xx operates in the -40 to +85 °C (+105 °C junction), -40 to +105 °C (+125 °C junction) and -40 to +125 °C (+130 °C junction) temperature ranges from a 1.71 to 3.6 V V_{DD} power supply when using internal LDO regulator and a 1.05 to 1.32V V_{DD12} power supply when using external SMPS supply. A comprehensive set of power-saving modes allows the design of low-power applications.

Some independent power supplies are supported: analog independent supply input for ADC, DAC, OPAMPs and comparators, 3.3 V dedicated supply input for USB and up to 14 I/Os can be supplied independently down to 1.08V. A VBAT input allows to backup the RTC and backup registers. Dedicated V_{DD12} power supplies can be used to bypass the internal LDO regulator when connected to an external SMPS.

The STM32L476xx family offers six packages from 64-pin to 144-pin packages.

Table 2. STM32L476xx family device features and peripheral counts

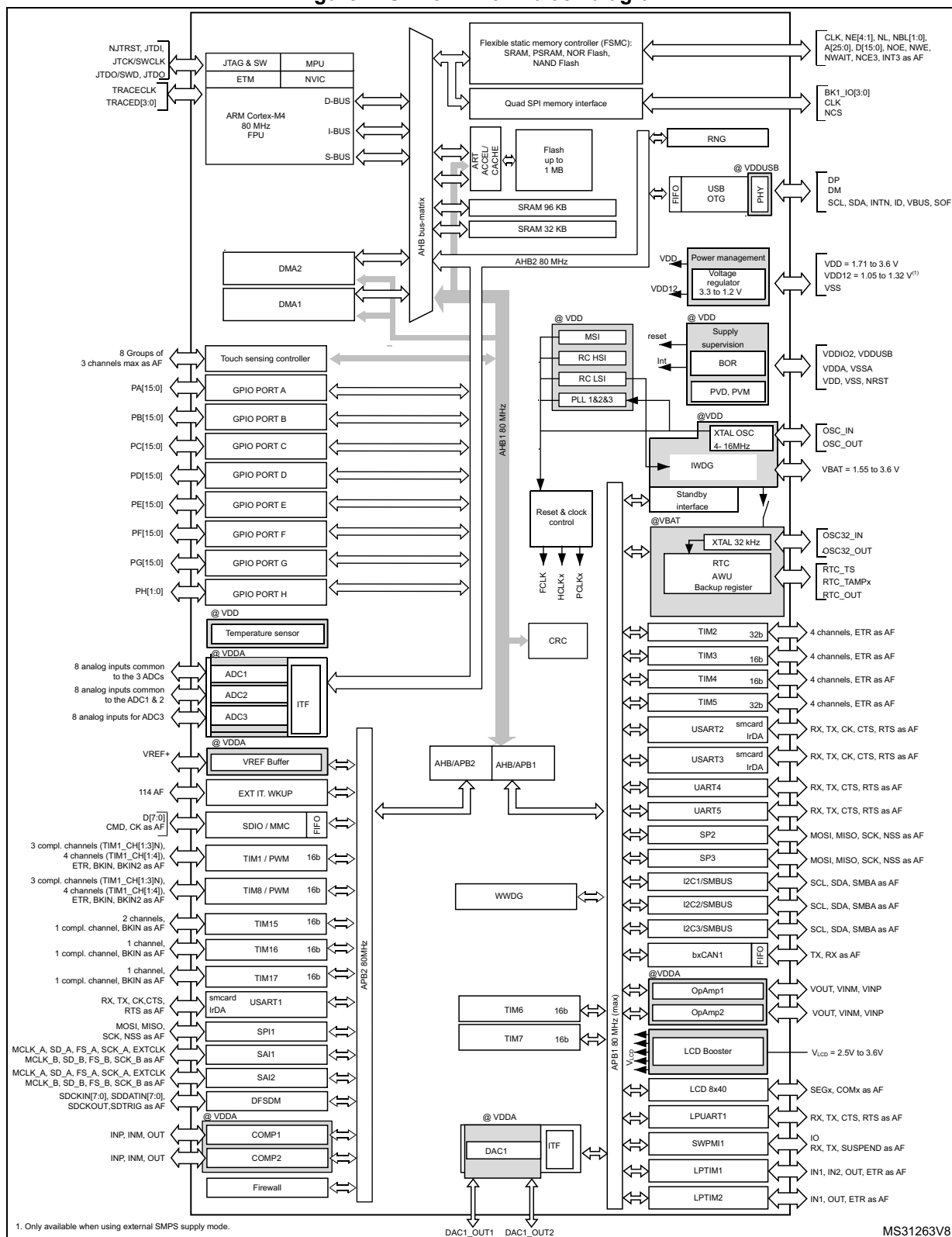
Peripheral		STM32 L476Zx		STM32 L476Qx		STM32 L476Vx			STM32 L476Mx		STM32 L476Jx		STM32 L476Rx		
Flash memory		512K B	1MB	512K B	1MB	256K B	512K B	1MB	512K B	1MB	512K B	1MB	256K B	512K B	1MB
SRAM		128KB													
External memory controller for static memories		Yes		Yes		Yes ⁽¹⁾			No		No		No		
Quad SPI		Yes													
Timers	Advanced control	2 (16-bit)													
	General purpose	5 (16-bit) 2 (32-bit)													
	Basic	2 (16-bit)													
	Low -power	2 (16-bit)													
	SysTick timer	1													
	Watchdog timers (independent, window)	2													
Comm. interfaces	SPI	3													
	I ² C	3													
	USART	3													
	UART	2													
	LPUART	1													
	SAI	2													
	CAN	1													
	USB OTG FS	Yes													
	SDMMC	Yes													
	SWPMI	Yes													
Digital filters for sigma-delta modulators		Yes (4 filters)													
Number of channels		8													
RTC		Yes													
Tamper pins		3							2		2		2		
LCD COM x SEG		Yes 8x40 or 4x44		Yes 8x40 or 4x44		Yes 8x40 or 4x44			Yes 8x30 or 4x32		Yes 8x28 or 4x32		Yes 8x28 or 4x32		

Table 2. STM32L476xx family device features and peripheral counts (continued)

Peripheral	STM32 L476Zx	STM32 L476Qx	STM32 L476Vx	STM32 L476Mx	STM32 L476Jx	STM32 L476Rx
Random generator	Yes					
GPIOs ⁽²⁾	114	109	82	65	57	51
Wakeup pins	5	5	5	4	4	4
Nb of I/Os down to 1.08 V	14	14	0	6	6	0
Capacitive sensing Number of channels	24	24	21	12	12	12
12-bit ADCs Number of channels	3 24	3 19	3 16	3 16	3 16	3 16
12-bit DAC channels	2					
Internal voltage reference buffer	Yes					No
Analog comparator	2					
Operational amplifiers	2					
Max. CPU frequency	80 MHz					
Operating voltage (V _{DD})	1.71 to 3.6 V					
Operating voltage (V _{DD12})	1.05 to 1.32 V					
Operating temperature	Ambient operating temperature: -40 to 85 °C / -40 to 105 °C / -40 to 125 °C Junction temperature: -40 to 105 °C / -40 to 125 °C / -40 to 130 °C					
Packages	LQFP144 UFBGA144	UFBGA 132	LQFP100	WLCSP81	WLCSP72	LQFP64

1. For the LQFP100 package, only FMC Bank1 is available. Bank1 can only support a multiplexed NOR/PSRAM memory using the NE1 Chip Select.
2. In case external SMPS package type is used, 2 GPIO's are replaced by VDD12 pins to connect the SMPS power supplies hence reducing the number of available GPIO's by 2.

Figure 1. STM32L476xx block diagram



Note: AF: alternate function on I/O pins.

3 Functional overview

3.1 Arm® Cortex®-M4 core with FPU

The Arm® Cortex®-M4 with FPU processor is the latest generation of Arm® processors for embedded systems. It was developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced response to interrupts.

The Arm® Cortex®-M4 with FPU 32-bit RISC processor features exceptional code-efficiency, delivering the high-performance expected from an Arm® core in the memory size usually associated with 8- and 16-bit devices.

The processor supports a set of DSP instructions which allow efficient signal processing and complex algorithm execution.

Its single precision FPU speeds up software development by using metalanguage development tools, while avoiding saturation.

With its embedded Arm® core, the STM32L476xx family is compatible with all Arm® tools and software.

Figure 1 shows the general block diagram of the STM32L476xx family devices.

3.2 Adaptive real-time memory accelerator (ART Accelerator™)

The ART Accelerator™ is a memory accelerator which is optimized for STM32 industry-standard Arm® Cortex®-M4 processors. It balances the inherent performance advantage of the Arm® Cortex®-M4 over Flash memory technologies, which normally requires the processor to wait for the Flash memory at higher frequencies.

To release the processor near 100 DMIPS performance at 80MHz, the accelerator implements an instruction prefetch queue and branch cache, which increases program execution speed from the 64-bit Flash memory. Based on CoreMark benchmark, the performance achieved thanks to the ART accelerator is equivalent to 0 wait state program execution from Flash memory at a CPU frequency up to 80 MHz.

3.3 Memory protection unit

The memory protection unit (MPU) is used to manage the CPU accesses to memory to prevent one task to accidentally corrupt the memory or resources used by any other active task. This memory area is organized into up to 8 protected areas that can in turn be divided up into 8 subareas. The protection area sizes are between 32 bytes and the whole 4 gigabytes of addressable memory.

The MPU is especially helpful for applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting, based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.

3.4 Embedded Flash memory

STM32L476xx devices feature up to 1 Mbyte of embedded Flash memory available for storing programs and data. The Flash memory is divided into two banks allowing read-while-write operations. This feature allows to perform a read operation from one bank while an erase or program operation is performed to the other bank. The dual bank boot is also supported. Each bank contains 256 pages of 2 Kbyte.

Flexible protections can be configured thanks to option bytes:

- Readout protection (RDP) to protect the whole memory. Three levels are available:
 - Level 0: no readout protection
 - Level 1: memory readout protection: the Flash memory cannot be read from or written to if either debug features are connected, boot in RAM or bootloader is selected
 - Level 2: chip readout protection: debug features (Cortex-M4 JTAG and serial wire), boot in RAM and bootloader selection are disabled (JTAG fuse). This selection is irreversible.

Table 3. Access status versus readout protection level and execution modes

Area	Protection level	User execution			Debug, boot from RAM or boot from system memory (loader)		
		Read	Write	Erase	Read	Write	Erase
Main memory	1	Yes	Yes	Yes	No	No	No
	2	Yes	Yes	Yes	N/A	N/A	N/A
System memory	1	Yes	No	No	Yes	No	No
	2	Yes	No	No	N/A	N/A	N/A
Option bytes	1	Yes	Yes	Yes	Yes	Yes	Yes
	2	Yes	No	No	N/A	N/A	N/A
Backup registers	1	Yes	Yes	N/A ⁽¹⁾	No	No	N/A ⁽¹⁾
	2	Yes	Yes	N/A	N/A	N/A	N/A
SRAM2	1	Yes	Yes	Yes ⁽¹⁾	No	No	No ⁽¹⁾
	2	Yes	Yes	Yes	N/A	N/A	N/A

1. Erased when RDP change from Level 1 to Level 0.

- Write protection (WRP): the protected area is protected against erasing and programming. Two areas per bank can be selected, with 2-Kbyte granularity.
- Proprietary code readout protection (PCROP): a part of the flash memory can be protected against read and write from third parties. The protected area is execute-only: it can only be reached by the STM32 CPU, as an instruction code, while all other accesses (DMA, debug and CPU data read, write and erase) are strictly prohibited. One area per bank can be selected, with 64-bit granularity. An additional option bit (PCROP_RDP) allows to select if the PCROP area is erased or not when the RDP protection is changed from Level 1 to Level 0.

The whole non-volatile memory embeds the error correction code (ECC) feature supporting:

- single error detection and correction
- double error detection.
- The address of the ECC fail can be read in the ECC register

3.5 Embedded SRAM

STM32L476xx devices feature up to 128 Kbyte of embedded SRAM. This SRAM is split into two blocks:

- 96 Kbyte mapped at address 0x2000 0000 (SRAM1)
- 32 Kbyte located at address 0x1000 0000 with hardware parity check (SRAM2).

This block is accessed through the ICode/DCode buses for maximum performance. These 32 Kbyte SRAM can also be retained in Standby mode.

The SRAM2 can be write-protected with 1 Kbyte granularity.

The memory can be accessed in read/write at CPU clock speed with 0 wait states.

3.6 Firewall

The device embeds a Firewall which protects code sensitive and secure data from any access performed by a code executed outside of the protected areas.

Each illegal access generates a reset which kills immediately the detected intrusion.

The Firewall main features are the following:

- Three segments can be protected and defined thanks to the Firewall registers:
 - Code segment (located in Flash or SRAM1 if defined as executable protected area)
 - Non-volatile data segment (located in Flash)
 - Volatile data segment (located in SRAM1)
- The start address and the length of each segments are configurable:
 - Code segment: up to 1024 Kbyte with granularity of 256 bytes
 - Non-volatile data segment: up to 1024 Kbyte with granularity of 256 bytes
 - Volatile data segment: up to 96 Kbyte with a granularity of 64 bytes
- Specific mechanism implemented to open the Firewall to get access to the protected areas (call gate entry sequence)
- Volatile data segment can be shared or not with the non-protected code
- Volatile data segment can be executed or not depending on the Firewall configuration

The Flash readout protection must be set to level 2 in order to reach the expected level of protection.

3.7 Boot modes

At startup, BOOT0 pin and BOOT1 option bit are used to select one of three boot options:

- Boot from user Flash
- Boot from system memory
- Boot from embedded SRAM

The boot loader is located in system memory. It is used to reprogram the Flash memory by using USART, I2C, SPI, CAN or USB OTG FS in Device mode through DFU (device firmware upgrade).

3.8 Cyclic redundancy check calculation unit (CRC)

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code using a configurable generator polynomial value and size.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a signature of the software during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

3.9 Power supply management

3.9.1 Power supply schemes

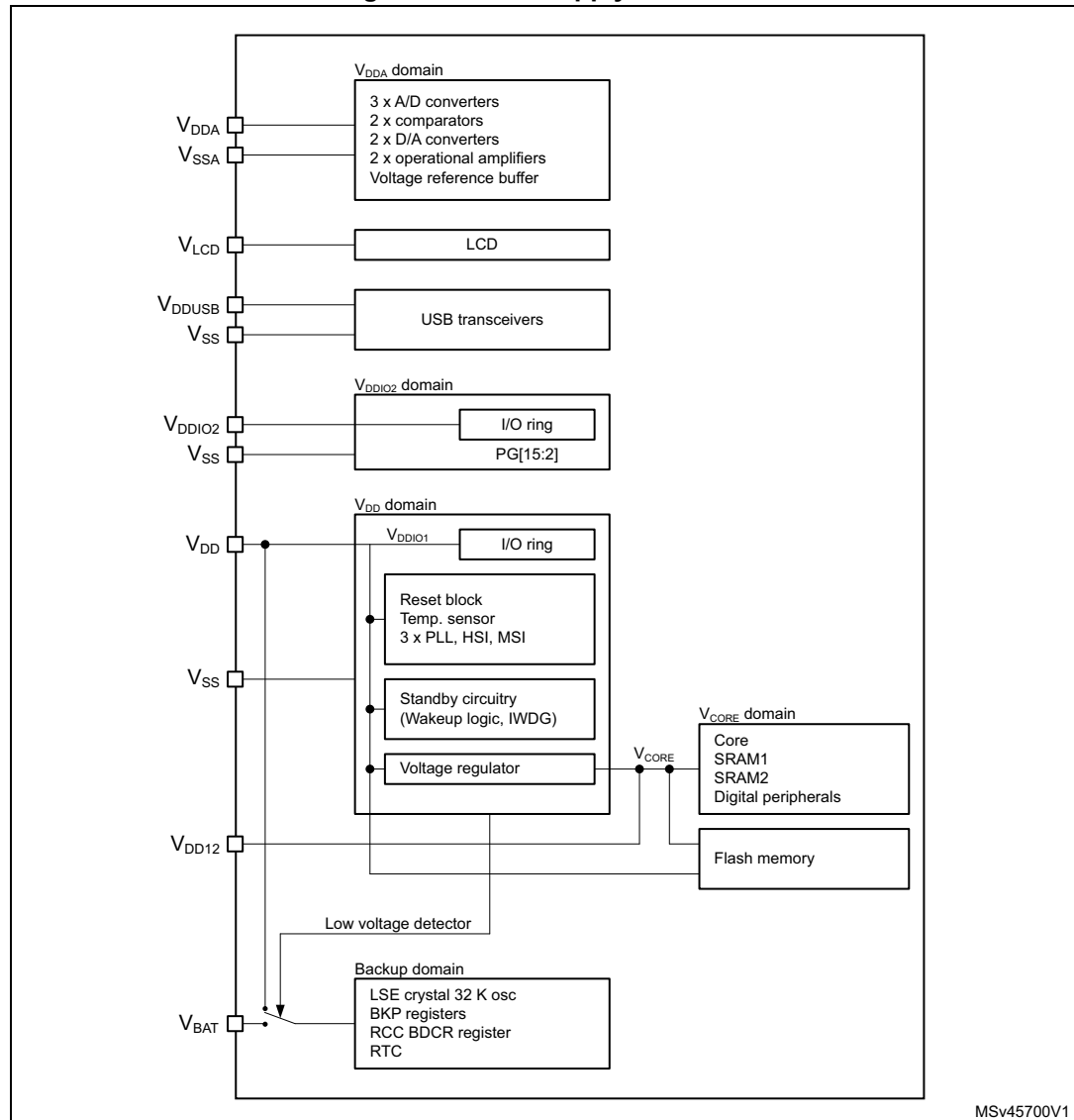
- $V_{DD} = 1.71$ to 3.6 V: external power supply for I/Os (V_{DDIO1}), the internal regulator and the system analog such as reset, power management and internal clocks. It is provided externally through VDD pins.
- $V_{DD12} = 1.05$ to 1.32 V: external power supply bypassing internal regulator when connected to an external SMPS. It is provided externally through VDD12 pins and only available on packages with the external SMPS supply option. VDD12 does not require any external decoupling capacitance and cannot support any external load.
- $V_{DDA} = 1.62$ V (ADCs/COMPAs) / 1.8 (DAC/OPAMPs) to 3.6 V: external analog power supply for ADCs, DAC, OPAMPs, Comparators and Voltage reference buffer. The V_{DDA} voltage level is independent from the V_{DD} voltage.
- $V_{DDUSB} = 3.0$ to 3.6 V: external independent power supply for USB transceivers. The V_{DDUSB} voltage level is independent from the V_{DD} voltage.
- $V_{DDIO2} = 1.08$ to 3.6 V: external power supply for 14 I/Os (PG[15:2]). The V_{DDIO2} voltage level is independent from the V_{DD} voltage.
- $V_{LCD} = 2.5$ to 3.6 V: the LCD controller can be powered either externally through VLCD pin, or internally from an internal voltage generated by the embedded step-up converter.
- $V_{BAT} = 1.55$ to 3.6 V: power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when V_{DD} is not present.

Note: When the functions supplied by V_{DDA} , V_{DDUSB} or V_{DDIO2} are not used, these supplies should preferably be shorted to V_{DD} .

Note: If these supplies are tied to ground, the I/Os supplied by these power supplies are not 5 V tolerant (refer to [Table 20: Voltage characteristics](#)).

Note: V_{DDIOx} is the I/Os general purpose digital functions supply. V_{DDIOx} represents V_{DDIO1} or V_{DDIO2} , with $V_{DDIO1} = V_{DD}$. V_{DDIO2} supply voltage level is independent from V_{DDIO1} .

Figure 2. Power supply overview

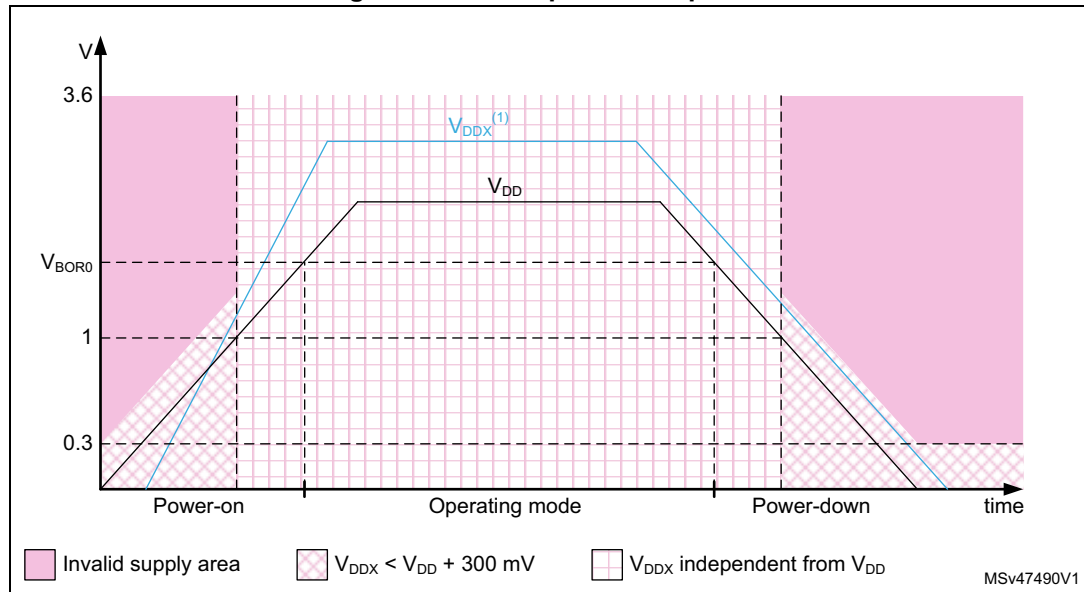


During power-up and power-down phases, the following power sequence requirements must be respected:

- When V_{DD} is below 1 V, other power supplies (V_{DDA} , V_{DDUSB} , V_{DDIO2} , V_{LCD}) must remain below $V_{DD} + 300$ mV.
- When V_{DD} is above 1 V, all power supplies are independent.

During the power-down phase, V_{DD} can temporarily become lower than other supplies only if the energy provided to the MCU remains below 1 mJ; this allows external decoupling capacitors to be discharged with different time constants during the power-down transient phase.

Figure 3. Power-up/down sequence



1. V_{DDX} refers to any power supply among V_{DDA} , V_{DDUSB} , V_{DDIO2} , V_{LCD} .

3.9.2 Power supply supervisor

The device has an integrated ultra-low-power brown-out reset (BOR) active in all modes except Shutdown and ensuring proper operation after power-on and during power down. The device remains in reset mode when the monitored supply voltage V_{DD} is below a specified threshold, without the need for an external reset circuit.

The lowest BOR level is 1.71V at power on, and other higher thresholds can be selected through option bytes. The device features an embedded programmable voltage detector (PVD) that monitors the V_{DD} power supply and compares it to the VPVD threshold. An interrupt can be generated when V_{DD} drops below the VPVD threshold and/or when V_{DD} is higher than the VPVD threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

In addition, the device embeds a Peripheral Voltage Monitor which compares the independent supply voltages V_{DDA} , V_{DDUSB} , V_{DDIO2} with a fixed threshold in order to ensure that the peripheral is in its functional supply range.

3.9.3 Voltage regulator

Two embedded linear voltage regulators supply most of the digital circuitries: the main regulator (MR) and the low-power regulator (LPR).

- The MR is used in the Run and Sleep modes and in the Stop 0 mode.
- The LPR is used in Low-Power Run, Low-Power Sleep, Stop 1 and Stop 2 modes. It is also used to supply the 32 Kbyte SRAM2 in Standby with SRAM2 retention.
- Both regulators are in power-down in Standby and Shutdown modes: the regulator output is in high impedance, and the kernel circuitry is powered down thus inducing zero consumption.

The ultralow-power STM32L476xx supports dynamic voltage scaling to optimize its power consumption in run mode. The voltage from the Main Regulator that supplies the logic (V_{CORE}) can be adjusted according to the system's maximum operating frequency.

There are two power consumption ranges:

- Range 1 with the CPU running at up to 80 MHz.
- Range 2 with a maximum CPU frequency of 26 MHz. All peripheral clocks are also limited to 26 MHz.

The V_{CORE} can be supplied by the low-power regulator, the main regulator being switched off. The system is then in Low-power run mode.

- Low-power run mode with the CPU running at up to 2 MHz. Peripherals with independent clock can be clocked by HSI16.

When the MR is in use, the STM32L476xx with the external SMPS option allows to force an external V_{CORE} supply on the VDD12 supply pins.

When V_{DD12} is forced by an external source and is higher than the output of the internal LDO, the current is taken from this external supply and the overall power efficiency is significantly improved if using an external step down DC/DC converter.

3.9.4 Low-power modes

The ultra-low-power STM32L476xx supports seven low-power modes to achieve the best compromise between low-power consumption, short startup time, available peripherals and available wakeup sources.

Table 4. STM32L476xx modes overview

Mode	Regulator (1)	CPU	Flash	SRAM	Clocks	DMA & Peripherals(2)	Wakeup source
Run	MR range 1	Yes	ON(4)	ON	Any	All	N/A
	SMPS range 2 High						
	MR range2					All except OTG_FS, RNG	
	SMPS range 2 Low						
LPRun	LPR	Yes	ON(4)	ON	Any except PLL	All except OTG_FS, RNG	N/A
Sleep	MR range 1	No	ON(4)	ON(7)	Any	All	Any interrupt or event
	SMPS range 2 High						
	MR range2					All except OTG_FS, RNG	
	SMPS range 2 Low						
LPSleep	LPR	No	ON(4)	ON(7)	Any except PLL	All except OTG_FS, RNG	Any interrupt or event

Table 4. STM32L476xx modes overview (continued)

Mode	Regulator ⁽¹⁾	CPU	Flash	SRAM	Clocks	DMA & Peripherals ⁽²⁾	Wakeup source
Stop 0	Range 1 ⁽⁸⁾	No	Off	ON	LSE LSI	BOR, PVD, PVM RTC, LCD, IWDG COMPx (x=1,2) DAC1 OPAMPx (x=1,2) USARTx (x=1...5) ⁽⁹⁾ LPUART1 ⁽⁹⁾ I2Cx (x=1...3) ⁽¹⁰⁾ LPTIMx (x=1,2) *** All other peripherals are frozen.	Reset pin, all I/Os BOR, PVD, PVM RTC, LCD, IWDG COMPx (x=1,2) USARTx (x=1...5) ⁽⁹⁾ LPUART1 ⁽⁹⁾ I2Cx (x=1...3) ⁽¹⁰⁾ LPTIMx (x=1,2) OTG_FS ⁽¹¹⁾ SWPMI1 ⁽¹²⁾
	Range 2 ⁽⁸⁾						
Stop 1	LPR	No	Off	ON	LSE LSI	BOR, PVD, PVM RTC, LCD, IWDG COMPx (x=1,2) DAC1 OPAMPx (x=1,2) USARTx (x=1...5) ⁽⁹⁾ LPUART1 ⁽⁹⁾ I2Cx (x=1...3) ⁽¹⁰⁾ LPTIMx (x=1,2) *** All other peripherals are frozen.	Reset pin, all I/Os BOR, PVD, PVM RTC, LCD, IWDG COMPx (x=1,2) USARTx (x=1...5) ⁽⁹⁾ LPUART1 ⁽⁹⁾ I2Cx (x=1...3) ⁽¹⁰⁾ LPTIMx (x=1,2) OTG_FS ⁽¹¹⁾ SWPMI1 ⁽¹²⁾

Table 4. STM32L476xx modes overview (continued)

Mode	Regulator ⁽¹⁾	CPU	Flash	SRAM	Clocks	DMA & Peripherals ⁽²⁾	Wakeup source
Stop 2	LPR	No	Off	ON	LSE LSI	BOR, PVD, PVM RTC, LCD, IWDG COMPx (x=1..2) I2C3 ⁽¹⁰⁾ LPUART1 ⁽⁹⁾ LPTIM1 *** All other peripherals are frozen.	Reset pin, all I/Os BOR, PVD, PVM RTC, LCD, IWDG COMPx (x=1..2) I2C3 ⁽¹⁰⁾ LPUART1 ⁽⁹⁾ LPTIM1
Standby	LPR	Powered Off	Off	SRAM2 ON	LSE LSI	BOR, RTC, IWDG *** All other peripherals are powered off. *** I/O configuration can be floating, pull-up or pull-down	Reset pin 5 I/Os (WKUPx) ⁽¹³⁾ BOR, RTC, IWDG
	OFF			Powered Off			
Shutdown	OFF	Powered Off	Off	Powered Off	LSE	RTC *** All other peripherals are powered off. *** I/O configuration can be floating, pull-up or pull-down ⁽¹⁴⁾	Reset pin 5 I/Os (WKUPx) ⁽¹³⁾ RTC

1. LPR means Main regulator is OFF and Low-power regulator is ON.

2. All peripherals can be active or clock gated to save power consumption.

3. Typical current at $V_{DD} = 1.8\text{ V}$, 25°C . Consumptions values provided running from SRAM, Flash memory Off, 80 MHz in Range 1, 2 LPRun/LPSleep.

4. The Flash memory can be put in power-down and its clock can be gated off when executing from SRAM.

5. Theoretical value based on $V_{DD} = 3.3\text{ V}$, DC/DC Efficiency of 85%, $V_{CORE} = 1.10\text{ V}$

6. Theoretical value based on $V_{DD} = 3.3\text{ V}$, DC/DC Efficiency of 85%, $V_{CORE} = 1.05\text{ V}$

7. The SRAM1 and SRAM2 clocks can be gated on or off independently.

8. SMPS mode can be used in STOP0 Mode, but no significant power gain can be expected.
9. U(S)ART and LPUART reception is functional in Stop mode, and generates a wakeup interrupt on Start, address match or received data match.
10. I2C address detection is functional in Stop mode, and generates a wakeup interrupt in case of address match.
11. OTG_FS wakeup by resume from suspend and attach detection protocol event.
12. SWPMI1 wakeup by resume from suspend.
13. The I/Os with wakeup from Standby/Shutdown capability are: PA0, PC13, PE6, PA2, PC5.
14. I/Os can be configured with internal pull-up, pull-down or floating in Shutdown mode but the configuration is lost when exiting the Shutdown mode.

By default, the microcontroller is in Run mode after a system or a power Reset. It is up to the user to select one of the low-power modes described below:

- **Sleep mode**

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

- **Low-power run mode**

This mode is achieved with V_{CORE} supplied by the low-power regulator to minimize the regulator's operating current. The code can be executed from SRAM or from Flash, and the CPU frequency is limited to 2 MHz. The peripherals with independent clock can be clocked by HSI16.

- **Low-power sleep mode**

This mode is entered from the low-power run mode. Only the CPU clock is stopped. When wakeup is triggered by an event or an interrupt, the system reverts to the low-power run mode.

- **Stop 0, Stop 1 and Stop 2 modes**

Stop mode achieves the lowest power consumption while retaining the content of SRAM and registers. All clocks in the V_{CORE} domain are stopped, the PLL, the MSI RC, the HSI16 RC and the HSE crystal oscillators are disabled. The LSE or LSI is still running.

The RTC can remain active (Stop mode with RTC, Stop mode without RTC).

Some peripherals with wakeup capability can enable the HSI16 RC during Stop mode to detect their wakeup condition.

Three Stop modes are available: Stop 0, Stop 1 and Stop 2 modes. In Stop 2 mode, most of the V_{CORE} domain is put in a lower leakage mode.

Stop 1 offers the largest number of active peripherals and wakeup sources, a smaller wakeup time but a higher consumption than Stop 2. In Stop 0 mode, the main regulator remains ON, allowing a very fast wakeup time but with much higher consumption.

The system clock when exiting from Stop 0, Stop 1 or Stop 2 modes can be either MSI up to 48 MHz or HSI16, depending on software configuration.

- **Standby mode**

The Standby mode is used to achieve the lowest power consumption with BOR. The internal regulator is switched off so that the V_{CORE} domain is powered off. The PLL, the MSI RC, the HSI16 RC and the HSE crystal oscillators are also switched off.

The RTC can remain active (Standby mode with RTC, Standby mode without RTC).

The brown-out reset (BOR) always remains active in Standby mode.

The state of each I/O during standby mode can be selected by software: I/O with internal pull-up, internal pull-down or floating.

After entering Standby mode, SRAM1 and register contents are lost except for registers in the Backup domain and Standby circuitry. Optionally, SRAM2 can be retained in Standby mode, supplied by the low-power Regulator (Standby with SRAM2 retention mode).

The device exits Standby mode when an external reset (NRST pin), an IWDG reset, WKUP pin event (configurable rising or falling edge), or an RTC event occurs (alarm, periodic wakeup, timestamp, tamper) or a failure is detected on LSE (CSS on LSE).

The system clock after wakeup is MSI up to 8 MHz.

- **Shutdown mode**

The Shutdown mode allows to achieve the lowest power consumption. The internal regulator is switched off so that the V_{CORE} domain is powered off. The PLL, the HSI16, the MSI, the LSI and the HSE oscillators are also switched off.

The RTC can remain active (Shutdown mode with RTC, Shutdown mode without RTC).

The BOR is not available in Shutdown mode. No power voltage monitoring is possible in this mode, therefore the switch to Backup domain is not supported.

SRAM1, SRAM2 and register contents are lost except for registers in the Backup domain.

The device exits Shutdown mode when an external reset (NRST pin), a WKUP pin event (configurable rising or falling edge), or an RTC event occurs (alarm, periodic wakeup, timestamp, tamper).

The system clock after wakeup is MSI at 4 MHz.

Table 5. Functionalities depending on the working mode⁽¹⁾

Peripheral	Run	Sleep	Low-power run	Low-power sleep	Stop 0/1		Stop 2		Standby		Shutdown		VBAT
					-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	
CPU	Y	-	Y	-	-	-	-	-	-	-	-	-	-
Flash memory (up to 1 MB)	O ⁽²⁾	O ⁽²⁾	O ⁽²⁾	O ⁽²⁾	-	-	-	-	-	-	-	-	-
SRAM1 (up to 96 KB)	Y	Y ⁽³⁾	Y	Y ⁽³⁾	Y	-	Y	-	-	-	-	-	-
SRAM2 (32 KB)	Y	Y ⁽³⁾	Y	Y ⁽³⁾	Y	-	Y	-	O ⁽⁴⁾	-	-	-	-
FSMC	O	O	O	O	-	-	-	-	-	-	-	-	-
Quad SPI	O	O	O	O	-	-	-	-	-	-	-	-	-
Backup Registers	Y	Y	Y	Y	Y	-	Y	-	Y	-	Y	-	Y
Brown-out reset (BOR)	Y	Y	Y	Y	Y	Y	Y	Y	Y	Y	-	-	-
Programmable Voltage Detector (PVD)	O	O	O	O	O	O	O	O	-	-	-	-	-
Peripheral Voltage Monitor (PVMx; x=1,2,3,4)	O	O	O	O	O	O	O	O	-	-	-	-	-
DMA	O	O	O	O	-	-	-	-	-	-	-	-	-
High Speed Internal (HSI16)	O	O	O	O	(5)	-	(5)	-	-	-	-	-	-
High Speed External (HSE)	O	O	O	O	-	-	-	-	-	-	-	-	-
Low Speed Internal (LSI)	O	O	O	O	O	-	O	-	O	-	-	-	-
Low Speed External (LSE)	O	O	O	O	O	-	O	-	O	-	O	-	O
Multi-Speed Internal (MSI)	O	O	O	O	-	-	-	-	-	-	-	-	-
Clock Security System (CSS)	O	O	O	O	-	-	-	-	-	-	-	-	-
Clock Security System on LSE	O	O	O	O	O	O	O	O	O	O	-	-	-
RTC / Auto wakeup	O	O	O	O	O	O	O	O	O	O	O	O	O
Number of RTC Tamper pins	3	3	3	3	3	O	3	O	3	O	3	O	3

Table 5. Functionalities depending on the working mode⁽¹⁾ (continued)

Peripheral	Run	Sleep	Low-power run	Low-power sleep	Stop 0/1		Stop 2		Standby		Shutdown		VBAT
					-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	
LCD	O	O	O	O	O	O	O	O	-	-	-	-	-
USB OTG FS	O ⁽⁸⁾	O ⁽⁸⁾	-	-	-	O	-	-	-	-	-	-	-
USARTx (x=1,2,3,4,5)	O	O	O	O	O ⁽⁶⁾	O ⁽⁶⁾	-	-	-	-	-	-	-
Low-power UART (LPUART)	O	O	O	O	O ⁽⁶⁾	O ⁽⁶⁾	O ⁽⁶⁾	O ⁽⁶⁾	-	-	-	-	-
I2Cx (x=1,2)	O	O	O	O	O ⁽⁷⁾	O ⁽⁷⁾	-	-	-	-	-	-	-
I2C3	O	O	O	O	O ⁽⁷⁾	O ⁽⁷⁾	O ⁽⁷⁾	O ⁽⁷⁾	-	-	-	-	-
SPIx (x=1,2,3)	O	O	O	O	-	-	-	-	-	-	-	-	-
CAN	O	O	O	O	-	-	-	-	-	-	-	-	-
SDMMC1	O	O	O	O	-	-	-	-	-	-	-	-	-
SWPMI1	O	O	O	O	-	O	-	-	-	-	-	-	-
SAIx (x=1,2)	O	O	O	O	-	-	-	-	-	-	-	-	-
DFSDM1	O	O	O	O	-	-	-	-	-	-	-	-	-
ADCx (x=1,2,3)	O	O	O	O	-	-	-	-	-	-	-	-	-
DAC1	O	O	O	O	O	-	-	-	-	-	-	-	-
VREFBUF	O	O	O	O	O	-	-	-	-	-	-	-	-
OPAMPx (x=1,2)	O	O	O	O	O	-	-	-	-	-	-	-	-
COMPx (x=1,2)	O	O	O	O	O	O	O	O	-	-	-	-	-
Temperature sensor	O	O	O	O	-	-	-	-	-	-	-	-	-
Timers (TIMx)	O	O	O	O	-	-	-	-	-	-	-	-	-
Low-power timer 1 (LPTIM1)	O	O	O	O	O	O	O	O	-	-	-	-	-
Low-power timer 2 (LPTIM2)	O	O	O	O	O	O	-	-	-	-	-	-	-
Independent watchdog (IWDG)	O	O	O	O	O	O	O	O	O	O	-	-	-
Window watchdog (WWDG)	O	O	O	O	-	-	-	-	-	-	-	-	-
SysTick timer	O	O	O	O	-	-	-	-	-	-	-	-	-
Touch sensing controller (TSC)	O	O	O	O	-	-	-	-	-	-	-	-	-

Table 5. Functionalities depending on the working mode⁽¹⁾ (continued)

Peripheral	Run	Sleep	Low-power run	Low-power sleep	Stop 0/1		Stop 2		Standby		Shutdown		VBAT
					-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	-	Wakeup capability	
Random number generator (RNG)	O ⁽⁸⁾	O ⁽⁸⁾	-	-	-	-	-	-	-	-	-	-	-
CRC calculation unit	O	O	O	O	-	-	-	-	-	-	-	-	-
GPIOs	O	O	O	O	O	O	O	O	⁽⁹⁾ 5 pins ⁽¹⁰⁾	⁽¹¹⁾ 5 pins ⁽¹⁰⁾	-	-	-

- Legend: Y = Yes (Enable). O = Optional (Disable by default. Can be enabled by software). - = Not available.
- The Flash can be configured in power-down mode. By default, it is not in power-down mode.
- The SRAM clock can be gated on or off.
- SRAM2 content is preserved when the bit RRS is set in PWR_CR3 register.
- Some peripherals with wakeup from Stop capability can request HSI16 to be enabled. In this case, HSI16 is woken up by the peripheral, and only feeds the peripheral which requested it. HSI16 is automatically put off when the peripheral does not need it anymore.
- UART and LPUART reception is functional in Stop mode, and generates a wakeup interrupt on Start, address match or received frame event.
- I2C address detection is functional in Stop mode, and generates a wakeup interrupt in case of address match.
- Voltage scaling Range 1 only.
- I/Os can be configured with internal pull-up, pull-down or floating in Standby mode.
- The I/Os with wakeup from Standby/Shutdown capability are: PA0, PC13, PE6, PA2, PC5.
- I/Os can be configured with internal pull-up, pull-down or floating in Shutdown mode but the configuration is lost when exiting the Shutdown mode.

3.9.5 Reset mode

In order to improve the consumption under reset, the I/Os state under and after reset is “analog state” (the I/O schmitt trigger is disable). In addition, the internal reset pull-up is deactivated when the reset source is internal.

3.9.6 VBAT operation

The VBAT pin allows to power the device VBAT domain from an external battery, an external supercapacitor, or from V_{DD} when no external battery and an external supercapacitor are present. The VBAT pin supplies the RTC with LSE and the backup registers. Three anti-tamper detection pins are available in VBAT mode.

VBAT operation is automatically activated when V_{DD} is not present.

An internal VBAT battery charging circuit is embedded and can be activated when V_{DD} is present.

Note: When the microcontroller is supplied from VBAT, external interrupts and RTC alarm/events do not exit it from VBAT operation.

3.10 Interconnect matrix

Several peripherals have direct connections between them. This allows autonomous communication between peripherals, saving CPU resources thus power supply consumption. In addition, these hardware connections allow fast and predictable latency.

Depending on peripherals, these interconnections can operate in Run, Sleep, low-power run and sleep, Stop 0, Stop 1 and Stop 2 modes.

Table 6. STM32L476xx peripherals interconnect matrix

Interconnect source	Interconnect destination	Interconnect action	Run	Sleep	Low-power run	Low-power sleep	Stop 0 / Stop 1	Stop 2
TIMx	TIMx	Timers synchronization or chaining	Y	Y	Y	Y	-	-
	ADCx DAC1 DFSDM1	Conversion triggers	Y	Y	Y	Y	-	-
	DMA	Memory to memory transfer trigger	Y	Y	Y	Y	-	-
	COMPx	Comparator output blanking	Y	Y	Y	Y	-	-
TIM16/TIM17	IRTIM	Infrared interface output generation	Y	Y	Y	Y	-	-
COMPx	TIM1, 8 TIM2, 3	Timer input channel, trigger, break from analog signals comparison	Y	Y	Y	Y	-	-
	LPTIMERx	Low-power timer triggered by analog signals comparison	Y	Y	Y	Y	Y ⁽¹⁾	-
ADCx	TIM1, 8	Timer triggered by analog watchdog	Y	Y	Y	Y	-	-
RTC	TIM16	Timer input channel from RTC events	Y	Y	Y	Y	-	-
	LPTIMERx	Low-power timer triggered by RTC alarms or tampers	Y	Y	Y	Y	Y ⁽¹⁾	-
All clocks sources (internal and external)	TIM2 TIM15, 16, 17	Clock source used as input channel for RC measurement and trimming	Y	Y	Y	Y	-	-
USB	TIM2	Timer triggered by USB SOF	Y	Y	-	-	-	-
CSS CPU (hard fault) RAM (parity error) Flash memory (ECC error) COMPx PVD DFSDM1 (analog watchdog, short circuit detection)	TIM1,8 TIM15,16,17	Timer break	Y	Y	Y	Y	-	-

Table 6. STM32L476xx peripherals interconnect matrix (continued)

Interconnect source	Interconnect destination	Interconnect action	Run	Sleep	Low-power run	Low-power sleep	Stop 0 / Stop 1	Stop 2
GPIO	TIMx	External trigger	Y	Y	Y	Y	-	-
	LPTIMERx	External trigger	Y	Y	Y	Y	Y	Y ⁽¹⁾
	ADCx DAC1 DFSDM1	Conversion external trigger	Y	Y	Y	Y	-	-

1. LPTIM1 only.

3.11 Clocks and startup

The clock controller (see [Figure 4](#)) distributes the clocks coming from different oscillators to the core and the peripherals. It also manages clock gating for low-power modes and ensures clock robustness. It features:

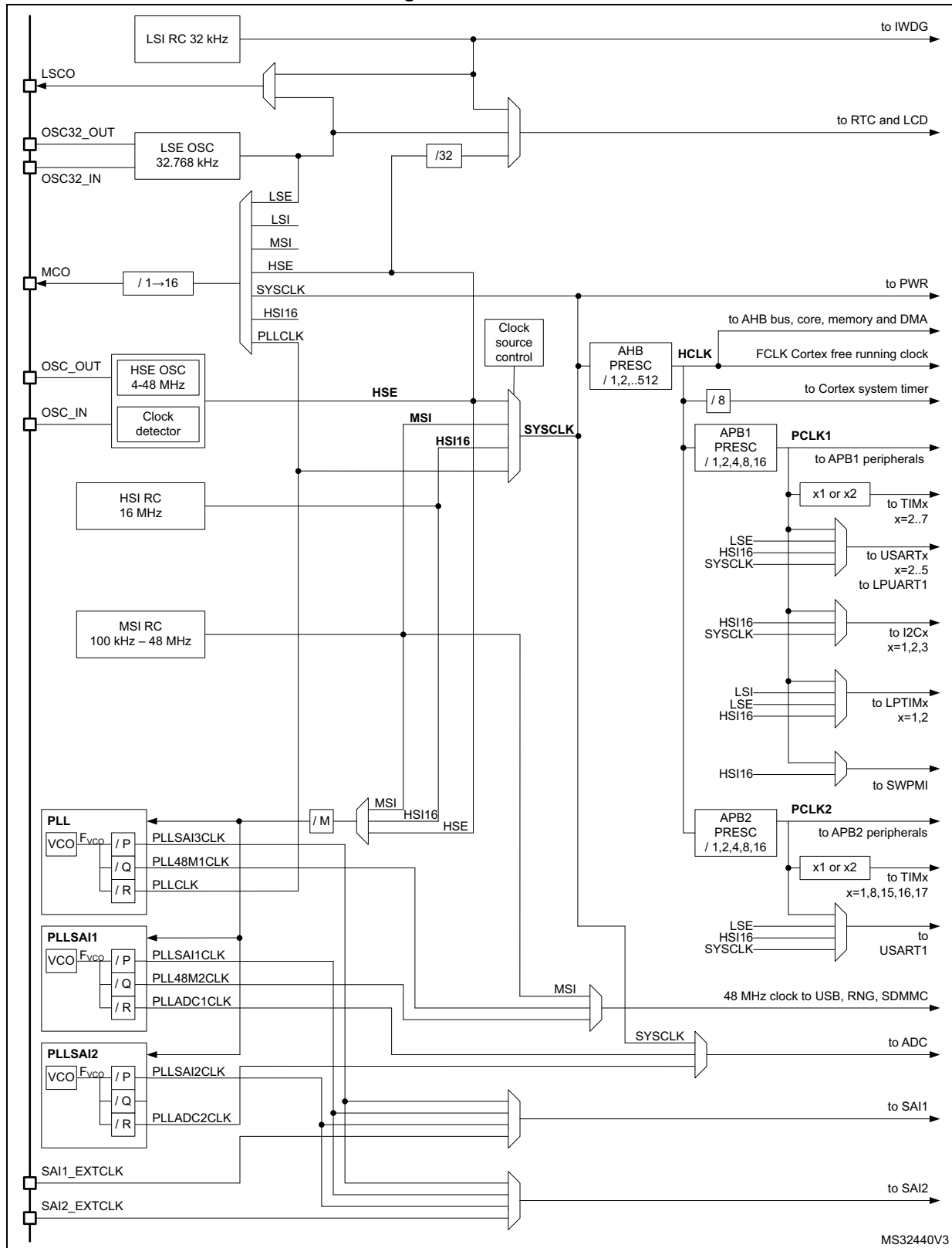
- **Clock prescaler:** to get the best trade-off between speed and current consumption, the clock frequency to the CPU and peripherals can be adjusted by a programmable prescaler
- **Safe clock switching:** clock sources can be changed safely on the fly in run mode through a configuration register.
- **Clock management:** to reduce power consumption, the clock controller can stop the clock to the core, individual peripherals or memory.
- **System clock source:** four different clock sources can be used to drive the master clock SYSCLK:
 - 4-48 MHz high-speed external crystal or ceramic resonator (HSE), that can supply a PLL. The HSE can also be configured in bypass mode for an external clock.
 - 16 MHz high-speed internal RC oscillator (HSI16), trimmable by software, that can supply a PLL
 - Multispeed internal RC oscillator (MSI), trimmable by software, able to generate 12 frequencies from 100 kHz to 48 MHz. When a 32.768 kHz clock source is available in the system (LSE), the MSI frequency can be automatically trimmed by hardware to reach better than $\pm 0.25\%$ accuracy. In this mode the MSI can feed the USB device, saving the need of an external high-speed crystal (HSE). The MSI can supply a PLL.
 - System PLL which can be fed by HSE, HSI16 or MSI, with a maximum frequency at 80 MHz.
- **Auxiliary clock source:** two ultralow-power clock sources that can be used to drive the LCD controller and the real-time clock:
 - 32.768 kHz low-speed external crystal (LSE), supporting four drive capability modes. The LSE can also be configured in bypass mode for an external clock.
 - 32 kHz low-speed internal RC (LSI), also used to drive the independent watchdog. The LSI clock accuracy is $\pm 5\%$ accuracy.
- **Peripheral clock sources:** Several peripherals (USB, SDMMC, RNG, SAI, USARTs, I2Cs, LPTimers, ADC, SWPMI) have their own independent clock whatever the system clock. Three PLLs, each having three independent outputs allowing the highest flexibility, can generate independent clocks for the ADC, the USB/SDMMC/RNG and the two SAIs.
- **Startup clock:** after reset, the microcontroller restarts by default with an internal 4 MHz clock (MSI). The prescaler ratio and clock source can be changed by the application program as soon as the code execution starts.
- **Clock security system (CSS):** this feature can be enabled by software. If a HSE clock failure occurs, the master clock is automatically switched to HSI16 and a software

interrupt is generated if enabled. LSE failure can also be detected and generated an interrupt.

- Clock-out capability:
 - **MCO: microcontroller clock output:** it outputs one of the internal clocks for external use by the application. Low frequency clocks (LSI, LSE) are available down to Stop 1 low power state.
 - **LSCO: low speed clock output:** it outputs LSI or LSE in all low-power modes down to Standby mode. LSE can also be output on LSCO in Shutdown mode. LSCO is not available in VBAT mode.

Several prescalers allow to configure the AHB frequency, the high speed APB (APB2) and the low speed APB (APB1) domains. The maximum frequency of the AHB and the APB domains is 80 MHz.

Figure 4. Clock tree



MS32440V3

3.12 General-purpose inputs/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain), as input (with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. Fast I/O toggling can be achieved thanks to their mapping on the AHB2 bus.

The I/Os alternate function configuration can be locked if needed following a specific sequence in order to avoid spurious writing to the I/Os registers.

3.13 Direct memory access controller (DMA)

The device embeds 2 DMAs. Refer to [Table 7: DMA implementation](#) for the features implementation.

Direct memory access (DMA) is used in order to provide high-speed data transfer between peripherals and memory as well as memory to memory. Data can be quickly moved by DMA without any CPU actions. This keeps CPU resources free for other operations.

The two DMA controllers have 14 channels in total, each dedicated to managing memory access requests from one or more peripherals. Each has an arbiter for handling the priority between DMA requests.

The DMA supports:

- 14 independently configurable channels (requests)
- Each channel is connected to dedicated hardware DMA requests, software trigger is also supported on each channel. This configuration is done by software.
- Priorities between requests from channels of one DMA are software programmable (4 levels consisting of very high, high, medium, low) or hardware in case of equality (request 1 has priority over request 2, etc.)
- Independent source and destination transfer size (byte, half word, word), emulating packing and unpacking. Source/destination addresses must be aligned on the data size.
- Support for circular buffer management
- 3 event flags (DMA Half Transfer, DMA Transfer complete and DMA Transfer Error) logically ORed together in a single interrupt request for each channel
- Memory-to-memory transfer
- Peripheral-to-memory and memory-to-peripheral, and peripheral-to-peripheral transfers
- Access to Flash, SRAM, APB and AHB peripherals as source and destination
- Programmable number of data to be transferred: up to 65536.

Table 7. DMA implementation

DMA features	DMA1	DMA2
Number of regular channels	7	7

3.14 Interrupts and events

3.14.1 Nested vectored interrupt controller (NVIC)

The devices embed a nested vectored interrupt controller able to manage 16 priority levels, and handle up to 81 maskable interrupt channels plus the 16 interrupt lines of the Cortex[®]-M4.

The NVIC benefits are the following:

- Closely coupled NVIC gives low latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Allows early processing of interrupts
- Processing of late arriving higher priority interrupts
- Support for tail chaining
- Processor state automatically saved on interrupt entry, and restored on interrupt exit, with no instruction overhead

The NVIC hardware block provides flexible interrupt management features with minimal interrupt latency.

3.14.2 Extended interrupt/event controller (EXTI)

The extended interrupt/event controller consists of 40 edge detector lines used to generate interrupt/event requests and wake-up the system from Stop mode. Each external line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The internal lines are connected to peripherals with wakeup from Stop mode capability. The EXTI can detect an external line with a pulse width shorter than the internal clock period. Up to 114 GPIOs can be connected to the 16 external interrupt lines.

3.15 Analog to digital converter (ADC)

The device embeds 3 successive approximation analog-to-digital converters with the following features:

- 12-bit native resolution, with built-in calibration
- 5.33 Msps maximum conversion rate with full resolution
 - Down to 18.75 ns sampling time
 - Increased conversion rate for lower resolution (up to 8.88 Msps for 6-bit resolution)
- Up to 24 external channels, some of them shared between ADC1 and ADC2, or ADC1, ADC2 and ADC3.
- 5 internal channels: internal reference voltage, temperature sensor, VBAT/3, DAC1_OUT1 and DAC1_OUT2.
- One external reference pin is available on some package, allowing the input voltage range to be independent from the power supply
- Single-ended and differential mode inputs
- Low-power design
 - Capable of low-current operation at low conversion rate (consumption decreases linearly with speed)
 - Dual clock domain architecture: ADC speed independent from CPU frequency
- Highly versatile digital interface
 - Single-shot or continuous/discontinuous sequencer-based scan mode: 2 groups of analog signals conversions can be programmed to differentiate background and high-priority real-time conversions
 - Handles two ADC converters for dual mode operation (simultaneous or interleaved sampling modes)
 - Each ADC support multiple trigger inputs for synchronization with on-chip timers and external signals
 - Results stored into 3 data register or in RAM with DMA controller support
 - Data pre-processing: left/right alignment and per channel offset compensation
 - Built-in oversampling unit for enhanced SNR
 - Channel-wise programmable sampling time
 - Three analog watchdog for automatic voltage monitoring, generating interrupts and trigger for selected timers
 - Hardware assistant to prepare the context of the injected channels to allow fast context switching

3.15.1 Temperature sensor

The temperature sensor (TS) generates a voltage V_{TS} that varies linearly with temperature.

The temperature sensor is internally connected to the ADC1_IN17 and ADC3_IN17 input channels which is used to convert the sensor output voltage into a digital value.

The sensor provides good linearity but it has to be calibrated to obtain good overall accuracy of the temperature measurement. As the offset of the temperature sensor varies from chip to chip due to process variation, the uncalibrated internal temperature sensor is suitable for applications that detect temperature changes only.

To improve the accuracy of the temperature sensor measurement, each device is individually factory-calibrated by ST. The temperature sensor factory calibration data are stored by ST in the system memory area, accessible in read-only mode.

Table 8. Temperature sensor calibration values

Calibration value name	Description	Memory address
TS_CAL1	TS ADC raw data acquired at a temperature of 30 °C (± 5 °C), $V_{DDA} = V_{REF+} = 3.0$ V (± 10 mV)	0x1FFF 75A8 - 0x1FFF 75A9
TS_CAL2	TS ADC raw data acquired at a temperature of 110 °C (± 5 °C), $V_{DDA} = V_{REF+} = 3.0$ V (± 10 mV)	0x1FFF 75CA - 0x1FFF 75CB

3.15.2 Internal voltage reference (V_{REFINT})

The internal voltage reference (V_{REFINT}) provides a stable (bandgap) voltage output for the ADC and Comparators. V_{REFINT} is internally connected to the ADC1_IN0 input channel. The precise voltage of V_{REFINT} is individually measured for each part by ST during production test and stored in the system memory area. It is accessible in read-only mode.

Table 9. Internal voltage reference calibration values

Calibration value name	Description	Memory address
V_{REFINT}	Raw data acquired at a temperature of 30 °C (± 5 °C), $V_{DDA} = V_{REF+} = 3.0$ V (± 10 mV)	0x1FFF 75AA - 0x1FFF 75AB

3.15.3 V_{BAT} battery voltage monitoring

This embedded hardware feature allows the application to measure the V_{BAT} battery voltage using the internal ADC channel ADC1_IN18 or ADC3_IN18. As the V_{BAT} voltage may be higher than V_{DDA} , and thus outside the ADC input range, the V_{BAT} pin is internally connected to a bridge divider by 3. As a consequence, the converted digital value is one third the V_{BAT} voltage.

3.16 Digital to analog converter (DAC)

Two 12-bit buffered DAC channels can be used to convert digital signals into analog voltage signal outputs. The chosen design structure is composed of integrated resistor strings and an amplifier in inverting configuration.

This digital interface supports the following features:

- Up to two DAC output channels
- 8-bit or 12-bit output mode
- Buffer offset calibration (factory and user trimming)
- Left or right data alignment in 12-bit mode
- Synchronized update capability
- Noise-wave generation
- Triangular-wave generation
- Dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- External triggers for conversion
- Sample and hold low-power mode, with internal or external capacitor

The DAC channels are triggered through the timer update outputs that are also connected to different DMA channels.

3.17 Voltage reference buffer (VREFBUF)

The STM32L476xx devices embed an voltage reference buffer which can be used as voltage reference for ADCs, DAC and also as voltage reference for external components through the VREF+ pin.

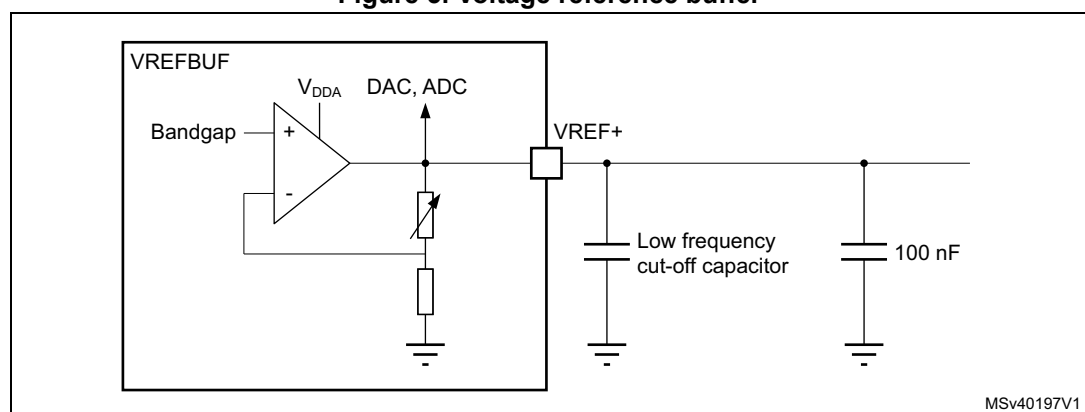
The internal voltage reference buffer supports two voltages:

- 2.048 V
- 2.5 V

An external voltage reference can be provided through the VREF+ pin when the internal voltage reference buffer is off.

The VREF+ pin is double-bonded with VDDA on some packages. In these packages the internal voltage reference buffer is not available.

Figure 5. Voltage reference buffer



3.18 Comparators (COMP)

The STM32L476xx devices embed two rail-to-rail comparators with programmable reference voltage (internal or external), hysteresis and speed (low speed for low-power) and with selectable output polarity.

The reference voltage can be one of the following:

- External I/O
- DAC output channels
- Internal reference voltage or submultiple (1/4, 1/2, 3/4).

All comparators can wake up from Stop mode, generate interrupts and breaks for the timers and can be also combined into a window comparator.

3.19 Operational amplifier (OPAMP)

The STM32L476xx embeds two operational amplifiers with external or internal follower routing and PGA capability.

The operational amplifier features:

- Low input bias current
- Low offset voltage
- Low-power mode
- Rail-to-rail input

3.20 Touch sensing controller (TSC)

The touch sensing controller provides a simple solution for adding capacitive sensing functionality to any application. Capacitive sensing technology is able to detect finger presence near an electrode which is protected from direct touch by a dielectric (glass, plastic, ...). The capacitive variation introduced by the finger (or any conductive object) is measured using a proven implementation based on a surface charge transfer acquisition principle.

The touch sensing controller is fully supported by the STMTouch touch sensing firmware library which is free to use and allows touch sensing functionality to be implemented reliably in the end application.

The main features of the touch sensing controller are the following:

- Proven and robust surface charge transfer acquisition principle
- Supports up to 24 capacitive sensing channels
- Up to 3 capacitive sensing channels can be acquired in parallel offering a very good response time
- Spread spectrum feature to improve system robustness in noisy environments
- Full hardware management of the charge transfer acquisition sequence
- Programmable charge transfer frequency
- Programmable sampling capacitor I/O pin
- Programmable channel I/O pin
- Programmable max count value to avoid long acquisition when a channel is faulty
- Dedicated end of acquisition and max count error flags with interrupt capability
- One sampling capacitor for up to 3 capacitive sensing channels to reduce the system components
- Compatible with proximity, touchkey, linear and rotary touch sensor implementation
- Designed to operate with STMTouch touch sensing firmware library

Note: The number of capacitive sensing channels is dependent on the size of the packages and subject to I/O availability.

3.21 Liquid crystal display controller (LCD)

The LCD drives up to 8 common terminals and 44 segment terminals to drive up to 320 pixels.

- Internal step-up converter to guarantee functionality and contrast control irrespective of V_{DD} . This converter can be deactivated, in which case the VLCD pin is used to provide the voltage to the LCD
- Supports static, 1/2, 1/3, 1/4 and 1/8 duty
- Supports static, 1/2, 1/3 and 1/4 bias
- Phase inversion to reduce power consumption and EMI
- Integrated voltage output buffers for higher LCD driving capability
- Up to 8 pixels can be programmed to blink
- Unneeded segments and common pins can be used as general I/O pins
- LCD RAM can be updated at any time owing to a double-buffer
- The LCD controller can operate in Stop mode

3.22 Digital filter for Sigma-Delta Modulators (DFSDM)

The device embeds one DFSDM with 4 digital filters modules and 8 external input serial channels (transceivers) or alternately 8 internal parallel inputs support.

The DFSDM peripheral is dedicated to interface the external $\Sigma\Delta$ modulators to microcontroller and then to perform digital filtering of the received data streams (which represent analog value on $\Sigma\Delta$ modulators inputs). DFSDM can also interface PDM (Pulse Density Modulation) microphones and perform PDM to PCM conversion and filtering in

hardware. DFSDM features optional parallel data stream inputs from microcontrollers memory (through DMA/CPU transfers into DFSDM).

DFSDM transceivers support several serial interface formats (to support various $\Sigma\Delta$ modulators). DFSDM digital filter modules perform digital processing according user selected filter parameters with up to 24-bit final ADC resolution.

The DFSDM peripheral supports:

- 8 multiplexed input digital serial channels:
 - configurable SPI interface to connect various SD modulator(s)
 - configurable Manchester coded 1 wire interface support
 - PDM (Pulse Density Modulation) microphone input support
 - maximum input clock frequency up to 20 MHz (10 MHz for Manchester coding)
 - clock output for SD modulator(s): 0..20 MHz
- alternative inputs from 8 internal digital parallel channels (up to 16 bit input resolution):
 - internal sources: device memory data streams (DMA)
- 4 digital filter modules with adjustable digital signal processing:
 - Sinc^x filter: filter order/type (1..5), oversampling ratio (up to 1..1024)
 - integrator: oversampling ratio (1..256)
- up to 24-bit output data resolution, signed output data format
- automatic data offset correction (offset stored in register by user)
- continuous or single conversion
- start-of-conversion triggered by:
 - software trigger
 - internal timers
 - external events
 - start-of-conversion synchronously with first digital filter module (DFSDM1_FLT0)
- analog watchdog feature:
 - low value and high value data threshold registers
 - dedicated configurable Sincx digital filter (order = 1..3, oversampling ratio = 1..32)
 - input from final output data or from selected input digital serial channels
 - continuous monitoring independently from standard conversion
- short circuit detector to detect saturated analog input values (bottom and top range):
 - up to 8-bit counter to detect 1..256 consecutive 0's or 1's on serial data stream
 - monitoring continuously each input serial channel
- break signal generation on analog watchdog event or on short circuit detector event
- extremes detector:
 - storage of minimum and maximum values of final conversion data
 - refreshed by software
- DMA capability to read the final conversion data
- interrupts: end of conversion, overrun, analog watchdog, short circuit, input serial channel clock absence
- “regular” or “injected” conversions:
 - “regular” conversions can be requested at any time or even in continuous mode

without having any impact on the timing of “injected” conversions

- “injected” conversions for precise timing and with high conversion priority

Table 10. DFSDM1 implementation

DFSDM features	DFSDM1
Number of channels	8
Number of filters	4
Input from internal ADC	-
Supported trigger sources	10
Pulses skipper	-
ID registers support	-

3.23 Random number generator (RNG)

All devices embed an RNG that delivers 32-bit random numbers generated by an integrated analog circuit.

3.24 Timers and watchdogs

The STM32L476xx includes two advanced control timers, up to nine general-purpose timers, two basic timers, two low-power timers, two watchdog timers and a SysTick timer. The table below compares the features of the advanced control, general purpose and basic timers.

Table 11. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
Advanced control	TIM1, TIM8	16-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	3
General-purpose	TIM2, TIM5	32-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	No
General-purpose	TIM3, TIM4	16-bit	Up, down, Up/down	Any integer between 1 and 65536	Yes	4	No
General-purpose	TIM15	16-bit	Up	Any integer between 1 and 65536	Yes	2	1

Table 11. Timer feature comparison (continued)

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary outputs
General-purpose	TIM16, TIM17	16-bit	Up	Any integer between 1 and 65536	Yes	1	1
Basic	TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No

3.24.1 Advanced-control timer (TIM1, TIM8)

The advanced-control timer can each be seen as a three-phase PWM multiplexed on 6 channels. They have complementary PWM outputs with programmable inserted dead-times. They can also be seen as complete general-purpose timers. The 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge or center-aligned modes) with full modulation capability (0-100%)
- One-pulse mode output

In debug mode, the advanced-control timer counter can be frozen and the PWM outputs disabled to turn off any power switches driven by these outputs.

Many features are shared with those of the general-purpose TIMx timers (described in [Section 3.24.2](#)) using the same architecture, so the advanced-control timers can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

3.24.2 General-purpose timers (TIM2, TIM3, TIM4, TIM5, TIM15, TIM16, TIM17)

There are up to seven synchronizable general-purpose timers embedded in the STM32L476xx (see [Table 11](#) for differences). Each general-purpose timer can be used to generate PWM outputs, or act as a simple time base.

- TIM2, TIM3, TIM4 and TIM5

They are full-featured general-purpose timers:

- TIM2 and TIM5 have a 32-bit auto-reload up/downcounter and 32-bit prescaler
- TIM3 and TIM4 have 16-bit auto-reload up/downcounter and 16-bit prescaler.

These timers feature 4 independent channels for input capture/output compare, PWM or one-pulse mode output. They can work together, or with the other general-purpose timers via the Timer Link feature for synchronization or event chaining.

The counters can be frozen in debug mode.

All have independent DMA request generation and support quadrature encoders.

- TIM15, 16 and 17

They are general-purpose timers with mid-range features:

They have 16-bit auto-reload upcounters and 16-bit prescalers.

- TIM15 has 2 channels and 1 complementary channel
- TIM16 and TIM17 have 1 channel and 1 complementary channel

All channels can be used for input capture/output compare, PWM or one-pulse mode output.

The timers can work together via the Timer Link feature for synchronization or event chaining. The timers have independent DMA request generation.

The counters can be frozen in debug mode.

3.24.3 Basic timers (TIM6 and TIM7)

The basic timers are mainly used for DAC trigger generation. They can also be used as generic 16-bit timebases.

3.24.4 Low-power timer (LPTIM1 and LPTIM2)

The devices embed two low-power timers. These timers have an independent clock and are running in Stop mode if they are clocked by LSE, LSI or an external clock. They are able to wakeup the system from Stop mode.

LPTIM1 is active in Stop 0, Stop 1 and Stop 2 modes.

LPTIM2 is active in Stop 0 and Stop 1 mode.

This low-power timer supports the following features:

- 16-bit up counter with 16-bit autoreload register
- 16-bit compare register
- Configurable output: pulse, PWM
- Continuous/ one shot mode
- Selectable software/hardware input trigger
- Selectable clock source
 - Internal clock sources: LSE, LSI, HSI16 or APB clock
 - External clock source over LPTIM input (working even with no internal clock source running, used by pulse counter application).
- Programmable digital glitch filter
- Encoder mode (LPTIM1 only)

3.24.5 Infrared interface (IRTIM)

The STM32L476xx includes one infrared interface (IRTIM). It can be used with an infrared LED to perform remote control functions. It uses TIM16 and TIM17 output channels to generate output signal waveforms on IR_OUT pin.

3.24.6 Independent watchdog (IWDG)

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 32 kHz internal RC (LSI) and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free running timer for application timeout management. It is hardware or software configurable through the option bytes. The counter can be frozen in debug mode.

3.24.7 System window watchdog (WWDG)

The window watchdog is based on a 7-bit downcounter that can be set as free running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

3.24.8 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard down counter. It features:

- A 24-bit down counter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0.
- Programmable clock source

3.25 Real-time clock (RTC) and backup registers

The RTC is an independent BCD timer/counter. It supports the following features:

- Calendar with subsecond, seconds, minutes, hours (12 or 24 format), week day, date, month, year, in BCD (binary-coded decimal) format.
- Automatic correction for 28, 29 (leap year), 30, and 31 days of the month.
- Two programmable alarms.
- On-the-fly correction from 1 to 32767 RTC clock pulses. This can be used to synchronize it with a master clock.
- Reference clock detection: a more precise second source clock (50 or 60 Hz) can be used to enhance the calendar precision.
- Digital calibration circuit with 0.95 ppm resolution, to compensate for quartz crystal inaccuracy.
- Three anti-tamper detection pins with programmable filter.
- Timestamp feature which can be used to save the calendar content. This function can be triggered by an event on the timestamp pin, or by a tamper event, or by a switch to VBAT mode.
- 17-bit auto-reload wakeup timer (WUT) for periodic events with programmable resolution and period.

The RTC and the 32 backup registers are supplied through a switch that takes power either from the V_{DD} supply when present or from the VBAT pin.

The backup registers are 32-bit registers used to store 128 bytes of user application data when V_{DD} power is not present. They are not reset by a system or power reset, or when the device wakes up from Standby or Shutdown mode.

The RTC clock sources can be:

- A 32.768 kHz external crystal (LSE)
- An external resonator or oscillator (LSE)
- The internal low power RC oscillator (LSI, with typical frequency of 32 kHz)
- The high-speed external clock (HSE) divided by 32.

The RTC is functional in VBAT mode and in all low-power modes when it is clocked by the LSE. When clocked by the LSI, the RTC is not functional in VBAT mode, but is functional in all low-power modes except Shutdown mode.

All RTC events (Alarm, WakeUp Timer, Timestamp or Tamper) can generate an interrupt and wakeup the device from the low-power modes.

3.26 Inter-integrated circuit interface (I²C)

The device embeds three I2C. Refer to [Table 12: I2C implementation](#) for the features implementation.

The I²C bus interface handles communications between the microcontroller and the serial I²C bus. It controls all I²C bus-specific sequencing, protocol, arbitration and timing.

The I2C peripheral supports:

- I²C-bus specification and user manual rev. 5 compatibility:
 - Slave and master modes, multimaster capability
 - Standard-mode (Sm), with a bitrate up to 100 kbit/s
 - Fast-mode (Fm), with a bitrate up to 400 kbit/s
 - Fast-mode Plus (Fm+), with a bitrate up to 1 Mbit/s and 20 mA output drive I/Os
 - 7-bit and 10-bit addressing mode, multiple 7-bit slave addresses
 - Programmable setup and hold times
 - Optional clock stretching
- System Management Bus (SMBus) specification rev 2.0 compatibility:
 - Hardware PEC (Packet Error Checking) generation and verification with ACK control
 - Address resolution protocol (ARP) support
 - SMBus alert
- Power System Management Protocol (PMBusTM) specification rev 1.1 compatibility
- Independent clock: a choice of independent clock sources allowing the I2C communication speed to be independent from the PCLK reprogramming. Refer to [Figure 4: Clock tree](#).
- Wakeup from Stop mode on address match
- Programmable analog and digital noise filters
- 1-byte buffer with DMA capability

Table 12. I2C implementation

I2C features ⁽¹⁾	I2C1	I2C2	I2C3
Standard-mode (up to 100 kbit/s)	X	X	X
Fast-mode (up to 400 kbit/s)	X	X	X
Fast-mode Plus with 20mA output drive I/Os (up to 1 Mbit/s)	X	X	X
Programmable analog and digital noise filters	X	X	X
SMBus/PMBus hardware support	X	X	X
Independent clock	X	X	X
Wakeup from Stop 0 / Stop 1 mode on address match	X	X	X
Wakeup from Stop 2 mode on address match	-	-	X

1. X: supported

3.27 Universal synchronous/asynchronous receiver transmitter (USART)

The STM32L476xx devices have three embedded universal synchronous receiver transmitters (USART1, USART2 and USART3) and two universal asynchronous receiver transmitters (UART4, UART5).

These interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode and have LIN Master/Slave capability. They provide hardware management of the CTS and RTS signals, and RS485 Driver Enable. They are able to communicate at speeds of up to 10Mbit/s.

USART1, USART2 and USART3 also provide Smart Card mode (ISO 7816 compliant) and SPI-like communication capability.

All USART have a clock domain independent from the CPU clock, allowing the USARTx (x=1,2,3,4,5) to wake up the MCU from Stop mode using baudrates up to 204 Kbaud. The wake up events from Stop mode are programmable and can be:

- Start bit detection
- Any received data frame
- A specific programmed data frame

All USART interfaces can be served by the DMA controller.

Table 13. STM32L476xx USART/UART/LPUART features

USART modes/features ⁽¹⁾	USART1	USART2	USART3	UART4	UART5	LPUART1
Hardware flow control for modem	X	X	X	X	X	X
Continuous communication using DMA	X	X	X	X	X	X
Multiprocessor communication	X	X	X	X	X	X
Synchronous mode	X	X	X	-	-	-
Smartcard mode	X	X	X	-	-	-
Single-wire half-duplex communication	X	X	X	X	X	X
IrDA SIR ENDEC block	X	X	X	X	X	-
LIN mode	X	X	X	X	X	-
Dual clock domain	X	X	X	X	X	X
Wakeup from Stop 0 / Stop 1 modes	X	X	X	X	X	X
Wakeup from Stop 2 mode	-	-	-	-	-	X
Receiver timeout interrupt	X	X	X	X	X	-
Modbus communication	X	X	X	X	X	-
Auto baud rate detection	X (4 modes)					-
Driver Enable	X	X	X	X	X	X
LPUART/USART data length	7, 8 and 9 bits					

1. X = supported.

3.28 Low-power universal asynchronous receiver transmitter (LPUART)

The device embeds one Low-Power UART. The LPUART supports asynchronous serial communication with minimum power consumption. It supports half duplex single wire communication and modem operations (CTS/RTS). It allows multiprocessor communication.

The LPUART has a clock domain independent from the CPU clock, and can wakeup the system from Stop mode using baudrates up to 220 Kbaud. The wake up events from Stop mode are programmable and can be:

- Start bit detection
- Any received data frame
- A specific programmed data frame

Only a 32.768 kHz clock (LSE) is needed to allow LPUART communication up to 9600 baud. Therefore, even in Stop mode, the LPUART can wait for an incoming frame while having an extremely low energy consumption. Higher speed clock can be used to reach higher baudrates.

LPUART interface can be served by the DMA controller.

3.29 Serial peripheral interface (SPI)

Three SPI interfaces allow communication up to 40 Mbits/s in master and up to 24 Mbits/s slave modes, in half-duplex, full-duplex and simplex modes. The 3-bit prescaler gives 8 master mode frequencies and the frame size is configurable from 4 bits to 16 bits. The SPI interfaces support NSS pulse mode, TI mode and Hardware CRC calculation.

All SPI interfaces can be served by the DMA controller.

3.30 Serial audio interfaces (SAI)

The device embeds 2 SAI. Refer to [Table 14: SAI implementation](#) for the features implementation. The SAI bus interface handles communications between the microcontroller and the serial audio protocol.

The SAI peripheral supports:

- Two independent audio sub-blocks which can be transmitters or receivers with their respective FIFO.
- 8-word integrated FIFOs for each audio sub-block.
- Synchronous or asynchronous mode between the audio sub-blocks.
- Master or slave configuration independent for both audio sub-blocks.
- Clock generator for each audio block to target independent audio frequency sampling when both audio sub-blocks are configured in master mode.
- Data size configurable: 8-, 10-, 16-, 20-, 24-, 32-bit.
- Peripheral with large configurability and flexibility allowing to target as example the following audio protocol: I2S, LSB or MSB-justified, PCM/DSP, TDM, AC'97 and SPDIF out.
- Up to 16 slots available with configurable size and with the possibility to select which ones are active in the audio frame.
- Number of bits by frame may be configurable.
- Frame synchronization active level configurable (offset, bit length, level).
- First active bit position in the slot is configurable.
- LSB first or MSB first for data transfer.
- Mute mode.
- Stereo/Mono audio frame capability.
- Communication clock strobing edge configurable (SCK).
- Error flags with associated interrupts if enabled respectively.
 - Overrun and underrun detection.
 - Anticipated frame synchronization signal detection in slave mode.
 - Late frame synchronization signal detection in slave mode.
 - Codec not ready for the AC'97 mode in reception.
- Interruption sources when enabled:
 - Errors.
 - FIFO requests.
- DMA interface with 2 dedicated channels to handle access to the dedicated integrated FIFO of each SAI audio sub-block.

Table 14. SAI implementation

SAI features ⁽¹⁾	SAI1	SAI2
I2S, LSB or MSB-justified, PCM/DSP, TDM, AC'97	X	X
Mute mode	X	X
Stereo/Mono audio frame capability.	X	X
16 slots	X	X
Data size configurable: 8-, 10-, 16-, 20-, 24-, 32-bit	X	X
FIFO Size	X (8 Word)	X (8 Word)
SPDIF	X	X

1. X: supported

3.31 Single wire protocol master interface (SWPMI)

The Single wire protocol master interface (SWPMI) is the master interface corresponding to the Contactless Frontend (CLF) defined in the ETSI TS 102 613 technical specification. The main features are:

- full-duplex communication mode
- automatic SWP bus state management (active, suspend, resume)
- configurable bitrate up to 2 Mbit/s
- automatic SOF, EOF and CRC handling

SWPMI can be served by the DMA controller.

3.32 Controller area network (CAN)

The CAN is compliant with specifications 2.0A and B (active) with a bit rate up to 1 Mbit/s. It can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers. It has three transmit mailboxes, two receive FIFOs with 3 stages and 14 scalable filter banks.

The CAN peripheral supports:

- Supports CAN protocol version 2.0 A, B Active
- Bit rates up to 1 Mbit/s
- Transmission
 - Three transmit mailboxes
 - Configurable transmit priority
- Reception
 - Two receive FIFOs with three stages
 - 14 Scalable filter banks
 - Identifier list feature
 - Configurable FIFO overrun
- Time-triggered communication option
 - Disable automatic retransmission mode
 - 16-bit free running timer
 - Time Stamp sent in last two data bytes
- Management
 - Maskable interrupts
 - Software-efficient mailbox mapping at a unique address space

3.33 Secure digital input/output and MultiMediaCards Interface (SDMMC)

The card host interface (SDMMC) provides an interface between the APB peripheral bus and MultiMediaCards (MMCs), SD memory cards and SDIO cards.

The SDMMC features include the following:

- Full compliance with MultiMediaCard System Specification Version 4.2. Card support for three different databus modes: 1-bit (default), 4-bit and 8-bit
- Full compatibility with previous versions of MultiMediaCards (forward compatibility)
- Full compliance with SD Memory Card Specifications Version 2.0
- Full compliance with SD I/O Card Specification Version 2.0: card support for two different databus modes: 1-bit (default) and 4-bit
- Data transfer up to 48 MHz for the 8 bit mode
- Data write and read with DMA capability

3.34 Universal serial bus on-the-go full-speed (OTG_FS)

The devices embed an USB OTG full-speed device/host/OTG peripheral with integrated transceivers. The USB OTG FS peripheral is compliant with the USB 2.0 specification and with the OTG 2.0 specification. It has software-configurable endpoint setting and supports suspend/resume. The USB OTG controller requires a dedicated 48 MHz clock that can be provided by the internal multispeed oscillator (MSI) automatically trimmed by 32.768 kHz external oscillator (LSE). This allows to use the USB device without external high speed crystal (HSE).

The major features are:

- Combined Rx and Tx FIFO size of 1.25 KB with dynamic FIFO sizing
- Supports the session request protocol (SRP) and host negotiation protocol (HNP)
- 1 bidirectional control endpoint + 5 IN endpoints + 5 OUT endpoints
- 12 host channels with periodic OUT support
- HNP/SNP/IP inside (no need for any external resistor)
- USB 2.0 LPM (Link Power Management) support
- Battery Charging Specification Revision 1.2 support
- Internal FS OTG PHY support

For OTG/Host modes, a power switch is needed in case bus-powered devices are connected.

3.35 Flexible static memory controller (FSMC)

The Flexible static memory controller (FSMC) includes two memory controllers:

- The NOR/PSRAM memory controller
- The NAND/memory controller

This memory controller is also named Flexible memory controller (FMC).

The main features of the FMC controller are the following:

- Interface with static-memory mapped devices including:
 - Static random access memory (SRAM)
 - NOR Flash memory/OneNAND Flash memory
 - PSRAM (4 memory banks)
 - NAND Flash memory with ECC hardware to check up to 8 Kbyte of data
- 8-,16- bit data bus width
- Independent Chip Select control for each memory bank
- Independent configuration for each memory bank
- Write FIFO
- The Maximum FMC_CLK frequency for synchronous accesses is HCLK/2.

LCD parallel interface

The FMC can be configured to interface seamlessly with most graphic LCD controllers. It supports the Intel 8080 and Motorola 6800 modes, and is flexible enough to adapt to specific LCD interfaces. This LCD parallel interface capability makes it easy to build cost effective graphic applications using LCD modules with embedded controllers or high performance solutions using external controllers with dedicated acceleration.

3.36 Quad SPI memory interface (QUADSPI)

The Quad SPI is a specialized communication interface targeting single, dual or quad SPI flash memories. It can operate in any of the three following modes:

- Indirect mode: all the operations are performed using the QUADSPI registers
- Status polling mode: the external flash status register is periodically read and an interrupt can be generated in case of flag setting
- Memory-mapped mode: the external Flash is memory mapped and is seen by the system as if it were an internal memory

The Quad SPI interface supports:

- Three functional modes: indirect, status-polling, and memory-mapped
- SDR and DDR support
- Fully programmable opcode for both indirect and memory mapped mode
- Fully programmable frame format for both indirect and memory mapped mode
- Each of the 5 following phases can be configured independently (enable, length, single/dual/quad communication)
 - Instruction phase
 - Address phase
 - Alternate bytes phase
 - Dummy cycles phase
 - Data phase
- Integrated FIFO for reception and transmission
- 8, 16, and 32-bit data accesses are allowed
- DMA channel for indirect mode operations
- Programmable masking for external flash flag management
- Timeout management
- Interrupt generation on FIFO threshold, timeout, status match, operation complete, and access error

3.37 Development support

3.37.1 Serial wire JTAG debug port (SWJ-DP)

The Arm® SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target.

Debug is performed using 2 pins only instead of 5 required by the JTAG (JTAG pins could be re-use as GPIO with alternate function): the JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

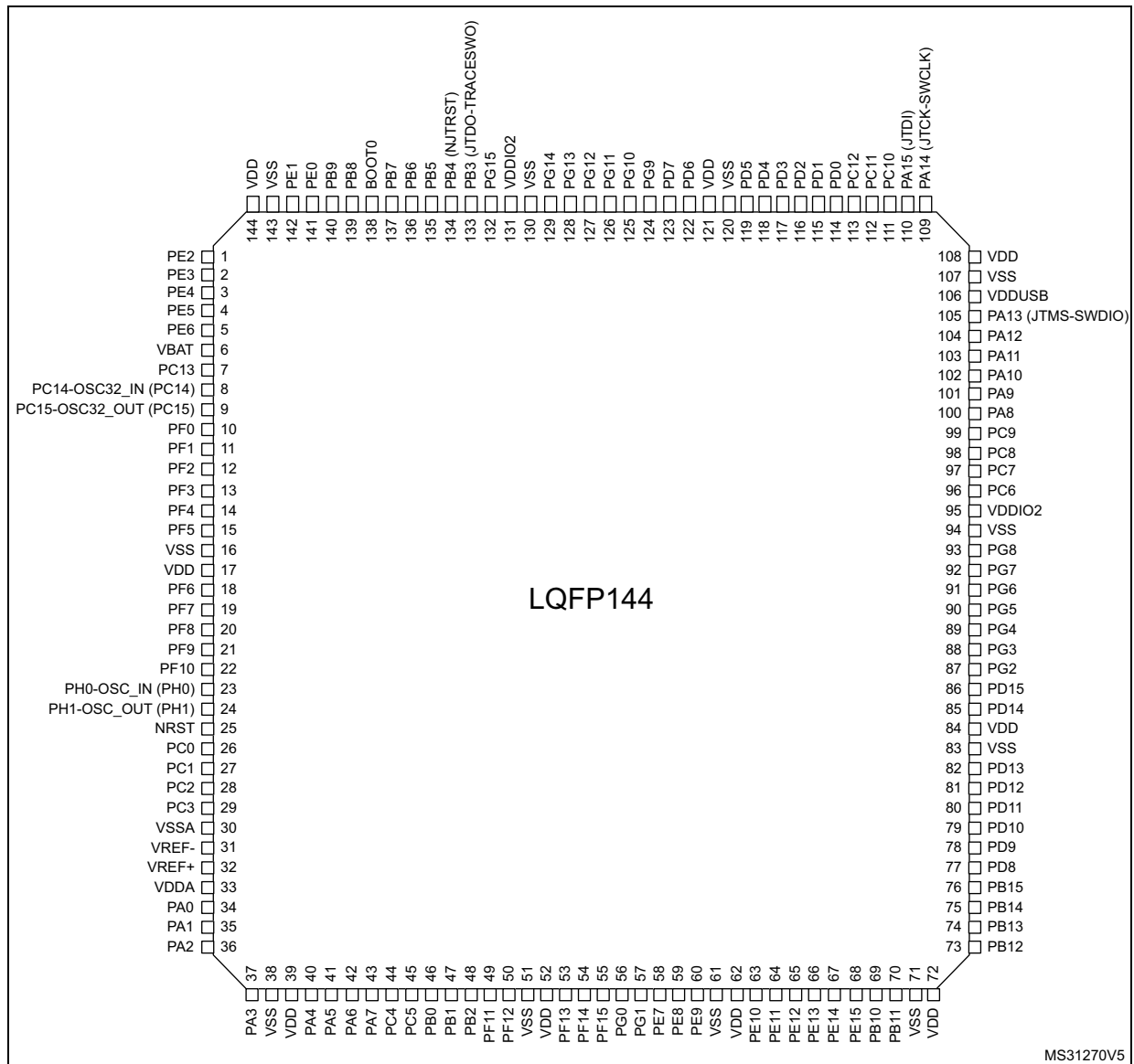
3.37.2 Embedded Trace Macrocell™

The Arm® Embedded Trace Macrocell™ provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32L476xx through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. Real-time instruction and data flow activity be recorded and then formatted for display on the host computer that runs the debugger software. TPA hardware is commercially available from common development tool vendors.

The Embedded Trace Macrocell™ operates with third party debugger software tools.

4 Pinouts and pin description

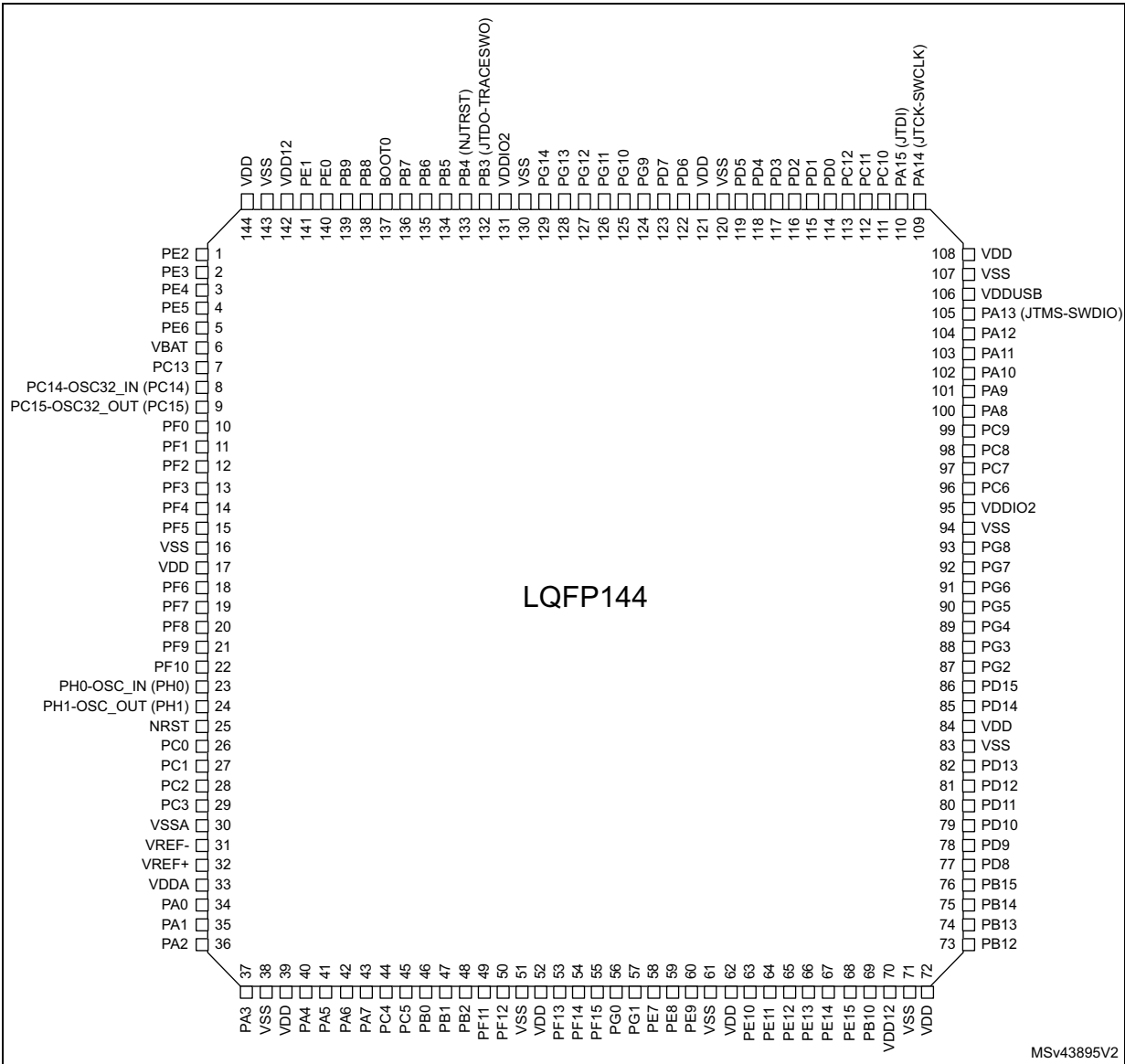
Figure 6. STM32L476Zx LQFP144 pinout⁽¹⁾



MS31270V5

1. The above figure shows the package top view.

Figure 7. STM32L476Zx, external SMPS device, LQFP144 pinout⁽¹⁾



MSv43895V2

1. The above figure shows the package top view.

Figure 8. STM32L476Zx UFBGA144 ballout⁽¹⁾

	1	2	3	4	5	6	7	8	9	10	11	12
A	VSS	PE0	PB8	BOOT0	PB7	PG14	PG12	PD7	PD6	PD1	PD0	VSS
B	VBAT	PE4	PE3	PE1	PB6	PG15	PG11	PD5	PC12	PC10	PA12	PA11
C	PC15- OSC32_OUT	PE5	PE2	PB9	PB5	PB3	PG9	PD4	PC11	PA14	PA13	PA10
D	PF4	PC14- OSC32_IN	PE6	PC13	PB4	PG13	PG10	PD3	PD2	PA15	PA9	PA8
E	PF6	PF1	PF0	PF2	VSS	VDDIO2	VDD	VSS	VDDUSB	PC6	PC9	PC8
F	PF8	PF7	PF5	PF3	VDD	VSS	VSS	VDDIO2	PG7	PG6	PG8	PC7
G	PH1- OSC_OUT	PH0-OSC_IN	PF10	PF9	VDD	VSS	VSS	VDD	PG4	PD13	PG3	PG5
H	PC2	PC0	PC1	NRST	VSS	VDD	VDD	VSS	PD12	PD11	PD14	PG2
J	VSSA	VREF-	PA0	PC3	PC4	PF11	PG1	PE9	PB13	PB14	PD10	PD15
K	VREF+	VDDA	PA1	PA6	PB2	PF12	PG0	PE11	PB11	PB12	PD8	PD9
L	OPAMP1 _VINM	PA2	PA4	OPAMP2 _VINM	PB0	PF13	PE8	PE12	PE13	PE14	PB10	PB15
M	VSS	PA3	PA5	PA7	PC5	PB1	PF14	PE7	PF15	PE10	PE15	VSS

MSv50902V1

1. The above figure shows the package top view.

Figure 9. STM32L476Qx UFBGA132 ballout⁽¹⁾

	1	2	3	4	5	6	7	8	9	10	11	12				
A	PE3	PE1	PB8	BOOT0	PD7	PD5	PB4	PB3	PA15	PA14	PA13	PA12				
B	PE4	PE2	PB9	PB7	PB6	PD6	PD4	PD3	PD1	PC12	PC10	PA11				
C	PC13	PE5	PE0	VDD	PB5	PG14	PG13	PD2	PD0	PC11	VDDUSB	PA10				
D	PC14- OSC32_IN	PE6	VSS	PF2	PF1	PF0	PG12	PG10	PG9	PA9	PA8	PC9				
E	PC15- OSC32_OUT	VBAT	VSS	PF3	<table><tr><td>VSS</td><td>VSS</td></tr><tr><td>VDD</td><td>VDDIO2</td></tr></table>				VSS	VSS	VDD	VDDIO2	PG5	PC8	PC7	PC6
VSS	VSS															
VDD	VDDIO2															
F	PH0-OSC_IN	VSS	PF4	PF5	PG3	PG4	VSS	VSS								
G	PH1- OSC_OUT	VDD	PG11	PG6					PG1	PG2	VDD	VDD				
H	PC0	NRST	VDD	PG7					PG0	PD15	PD14	PD13				
J	VSSA/VREF-	PC1	PC2	PA4	PA7	PG8	PF12	PF14	PF15	PD12	PD11	PD10				
K	PG15	PC3	PA2	PA5	PC4	PF11	PF13	PD9	PD8	PB15	PB14	PB13				
L	VREF+	PA0	PA3	PA6	PC5	PB2	PE8	PE10	PE12	PB10	PB11	PB12				
M	VDDA	PA1	OPAMP1_ VINM	OPAMP2_ VINM	PB0	PB1	PE7	PE9	PE11	PE13	PE14	PE15				

MSV35003V7

MSv35003V7

1. The above figure shows the package top view.

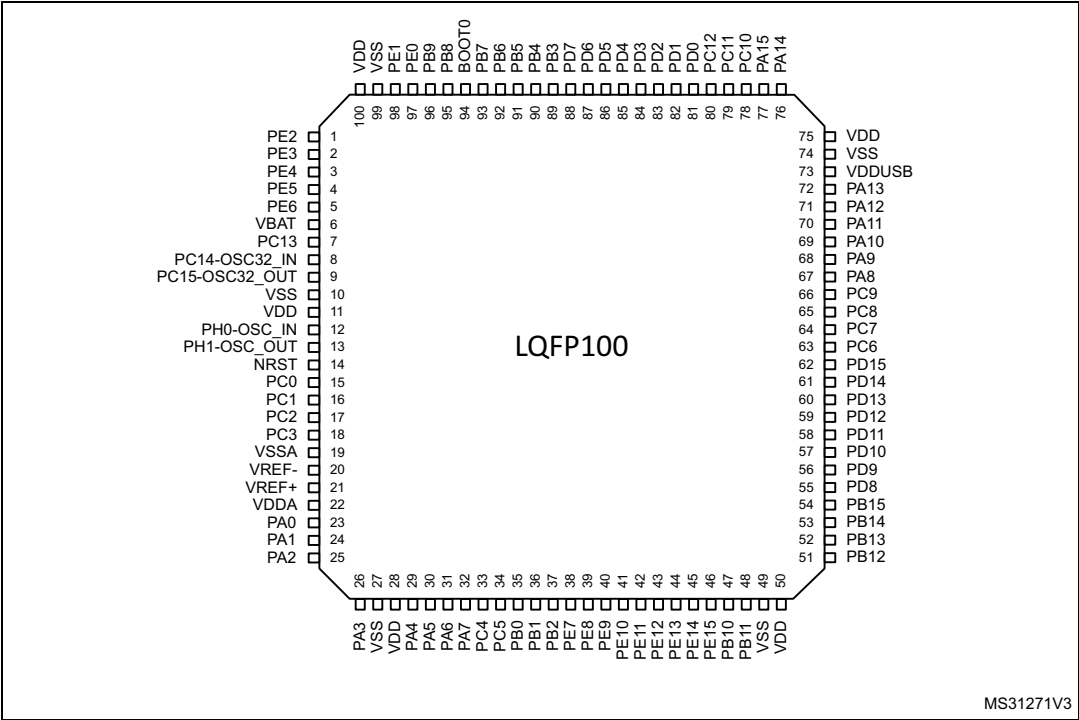
Figure 10. STM32L476Qx, external SMPS device, UFBGA132 ballout

	1	2	3	4	5	6	7	8	9	10	11	12				
A	PE3	PE1	PB8	BOOT0	PD7	PD5	PB4	PB3	PA15	PA14	PA13	PA12				
B	PE4	PE2	PB9	PB7	PB6	PD6	PD4	PD3	PD1	PC12	PC10	PA11				
C	PC13	PE5	PE0	VDD	PB5	VDD12	PG13	PD2	PD0	PC11	VDDUSB	PA10				
D	PC14-OSC32_IN	PE6	VSS	PF2	PF1	PF0	PG12	PG10	PG9	PA9	PA8	PC9				
E	PC15-OSC32_OUT	VBAT	VSS	PF3	<table><tr><td>VSS</td><td>VSS</td></tr><tr><td>VDD</td><td>VDDIO2</td></tr></table>				VSS	VSS	VDD	VDDIO2	PG5	PC8	PC7	PC6
VSS	VSS															
VDD	VDDIO2															
F	PH0-OSC_IN	VSS	PF4	PF5	PG3	PG4	VSS	VSS								
G	PH1-OSC_OUT	VDD	PG11	PG6					PG1	PG2	VDD	VDD				
H	PC0	NRST	VDD	PG7					PG0	PD15	PD14	PD13				
J	VSSA/VREF-	PC1	PC2	PA4	PA7	PG8	PF12	PF14	PF15	PD12	PD11	PD10				
K	PG15	PC3	PA2	PA5	PC4	PF11	PF13	PD9	PD8	PB15	PB14	PB13				
L	VREF+	PA0	PA3	PA6	PC5	PB2	PE8	PE10	PE12	PB10	VDD12	PB12				
M	VDDA	PA1	OPAMP1_VINM	OPAMP2_VINM	PB0	PB1	PE7	PE9	PE11	PE13	PE14	PE15				

MSv47486V1

1. The above figure shows the package top view.

Figure 11. STM32L476Vx LQFP100 pinout⁽¹⁾



1. The above figure shows the package top view.

Figure 12. STM32L476Mx WLCSP81 ballout⁽¹⁾

	1	2	3	4	5	6	7	8	9
A	VDDUSB	PA15	PD2	PG9	PG14	PB3	PB7	VSS	VDD
B	VSS	PA14	PC12	PG10	PG13	VDDIO2	PB6	PC13	VBAT
C	PA12	PA13	PC11	PG11	PG12	PB4	PB5	PC15-OSC32_OUT	PC14-OSC32_IN
D	PA11	PA10	PC10	PD5	PD6	PD7	BOOT0	PH1-OSC_OUT	PH0-OSC_IN
E	PC9	PA8	PA9	VDD	PD4	PE7	PB8	PB9	NRST
F	PC7	PC8	PC6	PD9	PD8	PE8	PC2	PC1	PC0
G	PB15	PB14	PB11	PA1	PA4	PA2	PC3	VREF+	VSSA/VREF-
H	PB12	PB13	PB10	PA7	PA6	PA5	PA3	PA0	VDDA
J	VDD	VSS	PB2	PB1	PB0	PC5	PC4	VDD	VSS

MSv38020V3

1. The above figure shows the package top view.

Figure 13. STM32L476Jx WLCSP72 ballout⁽¹⁾

	1	2	3	4	5	6	7	8	9
A	VDDUSB	PA15	PD2	PG9	PG14	PB3	PB7	VSS	VDD
B	VSS	PA14	PC12	PG10	PG13	VDDIO2	PB6	PC13	VBAT
C	PA12	PA13	PC11	PG11	PG12	PB4	PB5	PC15-OSC32_OUT	PC14-OSC32_IN
D	PA11	PA10	PC10	WLCSP72			BOOT0	PH1-OSC_OUT	PH0-OSC_IN
E	PC9	PA8	PA9				PB8	PB9	NRST
F	PC7	PC8	PC6				PC2	PC1	PC0
G	PB15	PB14	PB11	PA1	PA4	PA2	PC3	VREF+	VSSA/VREF-
H	PB12	PB13	PB10	PA7	PA6	PA5	PA3	PA0	VDDA
J	VDD	VSS	PB2	PB1	PB0	PC5	PC4	VDD	VSS

MSv35083V7

1. The above figure shows the package top view.

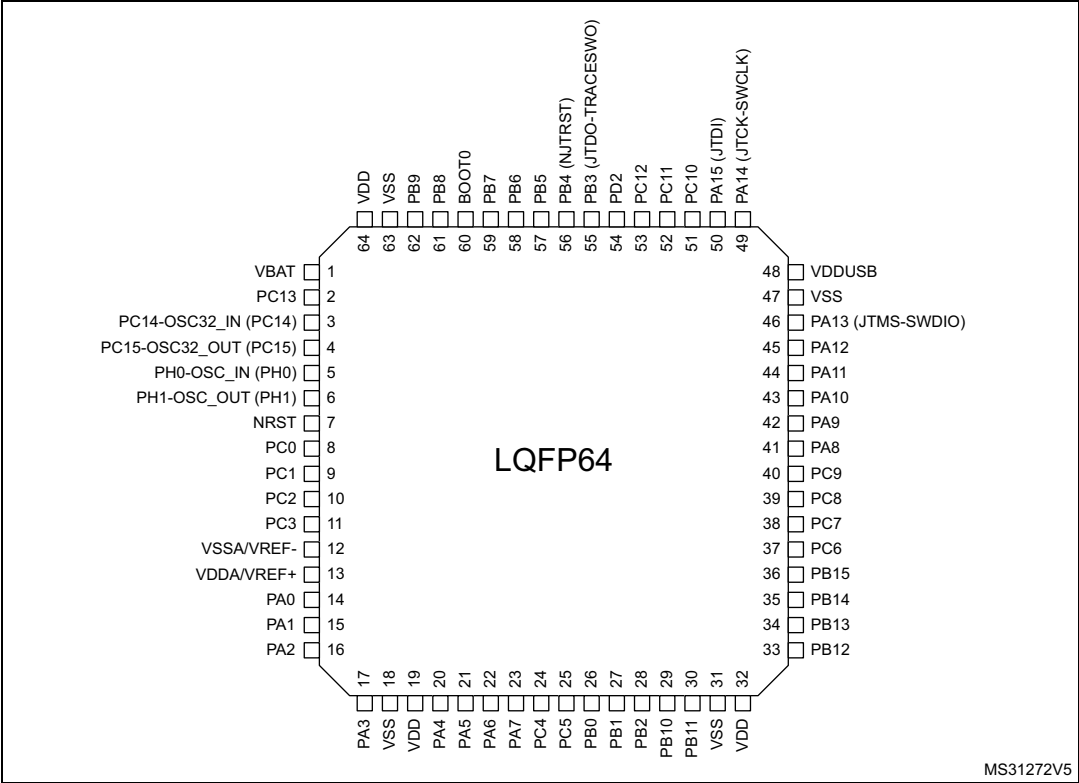
Figure 14. STM32L476Jx, external SMPS device, WLCSP72 ballout⁽¹⁾

	1	2	3	4	5	6	7	8	9
A	VDDUSB	PC10	PD2	PG9	PG14	PB3	BOOT0	VSS	VDD
B	VSS	PA14	PC12	PG10	PG13	VDDIO2	PB7	VDD12	VBAT
C	PA12	PA13	PA15	PG12	PB4	PB8	PC13	PC15-OSC32_OUT	PC14-OSC32_IN
D	PA11	PA10	PC11	WLCSP72			PB9	PH1-OSC_OUT	PH0-OSC_IN
E	PC9	PA8	PA9				PB5	PB6	NRST
F	VDD	PC7	PC8				PC2	PC1	PC0
G	PB15	PC6	PB14	PA2	PA0	PA1	PC3	VREF+	VSSA/VREF-
H	PB12	PB13	PB11	PA7	PA5	PA4	PA3	VDD	VDDA
J	VDD12	VSS	PB10	PB0	PB1	PB2	PC4	PA6	VSS

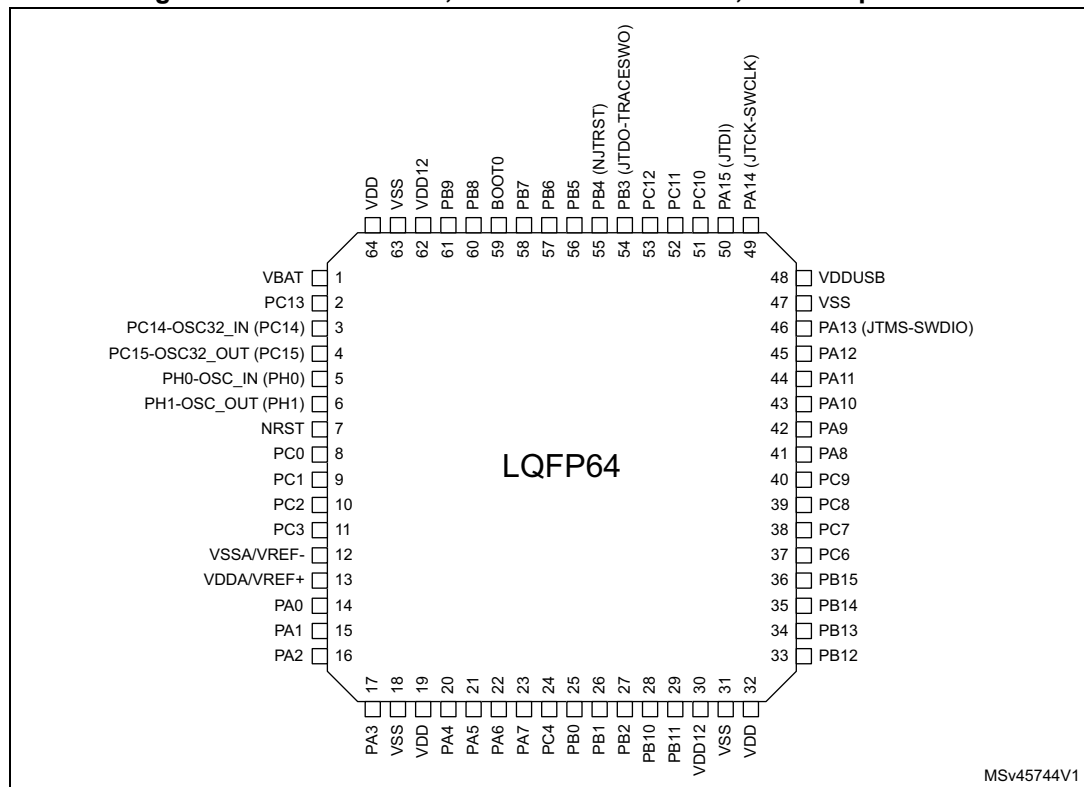
MSv43896V1

1. The above figure shows the package top view.

Figure 15. STM32L476Rx LQFP64 pinout⁽¹⁾



1. The above figure shows the package top view.

Figure 16. STM32L476Rx, external SMPS device, LQFP64 pinout⁽¹⁾

1. The above figure shows the package top view.

Table 15. Legend/abbreviations used in the pinout table

Name	Abbreviation	Definition
Pin name	Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name	
Pin type	S	Supply pin
	I	Input only pin
	I/O	Input / output pin
I/O structure	FT	5 V tolerant I/O
	TT	3.6 V tolerant I/O
	B	Dedicated BOOT0 pin
	RST	Bidirectional reset pin with embedded weak pull-up resistor
	Option for TT or FT I/Os	
	$_f^{(1)}$	I/O, Fm+ capable
	$_l^{(2)}$	I/O, with LCD function supplied by V_{LCD}
	$_u^{(3)}$	I/O, with USB function supplied by V_{DDUSB}
	$_a^{(4)}$	I/O, with Analog switch function supplied by V_{DDA}
	$_s^{(5)}$	I/O supplied only by V_{DDIO2}

Table 15. Legend/abbreviations used in the pinout table (continued)

Name		Abbreviation	Definition
Notes		Unless otherwise specified by a note, all I/Os are set as analog inputs during and after reset.	
Pin functions	Alternate functions	Functions selected through GPIOx_AFR registers	
	Additional functions	Functions directly selected/enabled through peripheral registers	

1. The related I/O structures in [Table 16](#) are: FT_f, FT_fa, FT_fl, FT fla.
2. The related I/O structures in [Table 16](#) are: FT_l, FT_fl, FT_lu.
3. The related I/O structures in [Table 16](#) are: FT_u, FT_lu.
4. The related I/O structures in [Table 16](#) are: FT_a, FT_la, FT_fa, FT fla, TT_a, TT_la.
5. The related I/O structures in [Table 16](#) are: FT_s, FT_fs.

Table 16. STM32L476xx pin definitions

Pin Number										Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs				
2	2	B8	C7	B8	7	C1	C1	7	D4	I/O	FT	(1) (2)	EVENTO
1	1	B9	B9	B9	6	E2	E2	6	B1	S	-	-	-
-	-	-	-	-	5	D2	D2	5	D3	I/O	FT	-	TRACED3, TIM FMC_A22, SAI1 EVENTO
-	-	-	-	-	4	C2	C2	4	C2	I/O	FT	-	TRACED2, TIM DFSDM1_C0 TSC_G7_IO4; F SAI1_SCH1 EVENTO
-	-	-	-	-	3	B1	B1	3	B2	I/O	FT	-	TRACED1, TIM DFSDM1_D0 TSC_G7_IO3; F SAI1_FS_A, EV
-	-	-	-	-	2	A1	A1	2	B3	I/O	FT_I	-	TRACED0, TIM TSC_G7_ LCD_SEG39, F SAI1_SD_B, EV
-	-	-	-	-	1	B2	B2	1	C3	I/O	FT_I	-	TRACECK, TIM TSC_G7_ LCD_SEG38, F SAI1_MCL1 EVENTO



Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs				
3	3	C9	C9	C9	8	D1	D1	8	8	I/O	FT	(1) (2)	EVENTO
4	4	C8	C8	C8	9	E1	E1	9	C1	I/O	FT	(1) (2)	EVENTO
-	-	-	-	-	-	D6	D6	10	E3	I/O	FT_f	-	I2C2_SDA, FT EVENTO
-	-	-	-	-	-	D5	D5	11	E2	I/O	FT_f	-	I2C2_SCL, FT EVENTO
-	-	-	-	-	-	D4	D4	12	E4	I/O	FT	-	I2C2_SMBA, FT EVENTO
-	-	-	-	-	-	E4	E4	13	F4	I/O	FT_a	-	FMC_A3, EVE
-	-	-	-	-	-	F3	F3	14	D1	I/O	FT_a	-	FMC_A4, EVE
-	-	-	-	-	-	F4	F4	15	F3	I/O	FT_a	-	FMC_A5, EVE
-	-	-	-	-	10	F2	F2	16	F6	S	-	-	-
-	-	-	-	-	11	G2	G2	17	G5	S	-	-	-
-	-	-	-	-	-	-	-	18	E1	I/O	FT_a	-	TIM5_ETR, TIM SAI1_SD_B, EV
-	-	-	-	-	-	-	-	19	F2	I/O	FT_a	-	TIM5_CH SAI1_MCL EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WLSP72	WLSP72_SMPs	WLCSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs				
5	-	-	-	-	-	-	-	20	20	I/O	FT_a	-	TIM5_CH3, SA11 EVENTO
6	-	-	-	-	-	-	-	21	21	I/O	FT_a	-	TIM5_CH4, SA11 TIM15_CH1, EV
7	-	-	-	-	-	-	-	22	22	I/O	FT_a	-	TIM15_CH2, EV
5	5	D9	D9	D9	12	F1	F1	23	23	I/O	FT	-	EVENTO
6	6	D8	D8	D8	13	G1	G1	24	24	I/O	FT	-	EVENTO
7	7	E9	E9	E9	14	H2	H2	25	25	I/O	RST	-	-
8	8	F9	F9	F9	15	H1	H1	26	26	I/O	FT fla	-	LPTIM1_IN1, I2 DFSDM1_D1 LPUART1 LCD_SEG LPTIM2_IN1, EV
9	9	F8	F8	F8	16	J2	J2	27	27	I/O	FT fla	-	LPTIM1_OUT, I2 DFSDM1_C LPUART1 LCD_SEG EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number											Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144					
10	10	F7	F7	F7	17	J3	J3	28	28	H1	PC2	I/O	FT_la	-	LPTIM1_I SPI2_MIS DFSDM1_C LCD_SEG EVENTO
11	11	G7	G7	G7	18	K2	K2	29	29	J4	PC3	I/O	FT_a	-	LPTIM1_E SPI2_MOSI, LCD SAI1_SD LPTIM2_E EVENTO
-	-	-	-	-	19	-	-	30	30	J1	VSSA	S	-	-	-
-	-	-	-	-	20	-	-	31	31	J2	VREF-	S	-	-	-
12	12	G9	G9	G9	-	J1	J1	-	-	-	VSSA/ VREF-	S	-	-	-
-	-	G8	G8	G8	21	L1	L1	32	32	K1	VREF+	S	-	-	-
-	-	H9	H9	H9	22	M1	M1	33	33	K2	VDDA	S	-	-	-
13	13	-	-	-	-	-	-	-	-	-	VDDA/ VREF+	S	-	-	-
14	14	H8	G5	H8	23	L2	L2	34	34	J3	PA0	I/O	FT_a	-	TIM2_CH1, TIM8_ET USART2_C UART4_- SAI1_EXT TIM2_ETR, EV

Table 16. STM32L476xx pin definitions (continued)

[illegible]

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate function	
LQFP64	LQFP64_SMPs	WLCSP72	WLCSP72_SMPs	WLCSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs						
21	21	H6		H5	H6	30	K4	K4	41	M3	PA5	I/O	TT_a	-	TIM2_CH1, TIM8_CH1N, SPI1_LPTIM2_EVENTO
22	22	H5		J8	H5	31	L4	L4	42	K4	PA6	I/O	FT_la	-	TIM1_BKIN, TIM8_BKIN, SPI1_USART3_CQUADSPI_BK_LCD_SEGMENT1_TIM1_BKIN_CTIM8_BKIN_CTIM16_CH1, EV
-	-	-		-	-	-	M4	M4	-	L4	OPAMP2_VINM	I	TT	-	-
23	23	H4		H4	H4	32	J5	J5	43	M4	PA7	I/O	FT_la	(3)	TIM1_CH1N, TIM8_CH1N, SPI1_USART3_CQUADSPI_BK_LCD_SEGMENT4, TIM16_CH1, EV
24	24	J7		J7	J7	33	K5	K5	44	J5	PC4	I/O	FT_la	-	USART3_LCD_SEGMENT1_EVENTO
25	-	J6		-	J6	34	L5	L5	45	M5	PC5	I/O	FT_la	-	USART3_LCD_SEGMENT1_EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WLSP72	WLSP72_SMPs	WLCSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144			
26	25	J5	J4	J5	35	M5	M5	46	46	L5	I/O	-	TIM1_CH2N, TIM8_CH2N, USART3_QUADSPI_BK, LCD_SEG6, LPTCOMP1_EVENTO
27	26	J4	J5	J4	36	M6	M6	47	47	M6	I/O	-	TIM1_CH3N, TIM8_CH3N, DFSDM1_DTOUT, USART3_RT, QUADSPI_BK, LCD_SEG6, LPTCOMP1_EVENTO
28	27	J3	J6	J3	37	L6	L6	48	48	K5	I/O	-	RTC_OUT, LPTIM2, I2C3_SM, DFSDM1_C, EVENTO
-	-	-	-	-	-	K6	K6	49	49	J6	I/O	-	EVENTO
-	-	-	-	-	-	J7	J7	50	50	K6	I/O	-	FMC_A6, EVENTO
-	-	-	-	-	-	-	-	51	51	G6	S	-	-
-	-	-	-	-	-	-	-	52	52	H6	S	-	-
-	-	-	-	-	-	K7	K7	53	53	L6	I/O	-	DFSDM1_DTOUT, FMC_A7, EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144				
-	-	-	-	-	-	J8	J8	54	54	M7	I/O	FT	-	DFSDM1_C TSC_G8_IO1, I EVENTO
-	-	-	-	-	-	J9	J9	55	55	M9	I/O	FT	-	TSC_G8_IO2, I EVENTO
-	-	-	-	-	-	H9	H9	56	56	K7	I/O	FT	-	TSC_G8_IO3, F EVENTO
-	-	-	-	-	-	G9	G9	57	57	J7	I/O	FT	-	TSC_G8_IO4, F EVENTO
-	-	-	-	E6	38	M7	M7	58	58	M8	I/O	FT	-	TIM1_ET DFSDM1_D FMC_D4, SAI EVENTO
-	-	-	-	F6	39	L7	L7	59	59	L7	I/O	FT	-	TIM1_CH DFSDM1_C FMC_D5, SAI1 EVENTO
-	-	-	-	-	40	M8	M8	60	60	J8	I/O	FT	-	TIM1_CH DFSDM1_C FMC_D6, SAI EVENTO
-	-	-	-	-	-	F6	F6	61	61	G7	S	-	-	-
-	-	-	-	-	-	G6	G6	62	62	E7	S	-	-	-



Table 16. STM32L476xx pin definitions (continued)

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Table 16. STM32L476xx pin definitions (continued)

[illegible]

Alternate

TIM1_	TSC_GG	TIM1_CH1N
BKIN	LCD_S	SP12_
TIM1_BKIN	WPM11_IO,	DFSDM11
2C2_SMBAA	M15_BKIN,	USART
DFSDM1_		LPUART
USART		TSC_GG
LPUART1		LCD_S
		SWPM
		SAI2_S
		TIM15_
		EVEN

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
	LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144		
	35	35	G2	G3	G2	G2	53	K11	K11	75	J10	-	TIM1_CH2N, I2C2_MISO, DFSDM1_DATA0, USART3_RTSC, TSC_G1, LCD_SEGMENT0, SWPMI1_SAI2_MCLK, TIM15_CH1, EV
	36	36	G1	G1	G1	54	K10	K10	76	76	L12	-	RTC_RELOAD, TIM1_CH2N, SPI2_MISO, DFSDM1_DATA0, TSC_G1, LCD_SEGMENT0, SWPMI1_SAI2_SD_A, TIM15_EVENTO
	-	-	-	-	F5	55	K9	K9	77	77	K11	-	USART3_SEGMENT0, F, EVENTO
	-	-	-	-	F4	56	K8	K8	78	78	K12	-	USART3_SEGMENT0, F, SAI2_MCLK, EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WLSP72	WLSP72_SMPs	WLCSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144			
-	-	-	-	-	57	J12	J12	J12	79	J11	I/O	-	USART3_TSC_G6_LCD_SEG30, FSAI2_SCK_EVENTO
-	-	-	-	-	58	J11	J11	J11	80	H10	I/O	-	USART3_TSC_G6_LCD_SEG31, FSAI2_SD_LPTIM2_EVENTO
-	-	-	-	-	59	J10	J10	J10	81	H9	I/O	-	TIM4_CH_USART3_RT_TSC_G6_LCD_SEG32, FSAI2_FS_A, LPTIM2_EVENTO
-	-	-	-	-	60	H12	H12	H12	82	G10	I/O	-	TIM4_CH_USART3_RT_TSC_G6_LCD_SEG33, FSAI2_FS_A, LPTIM2_EVENTO
-	-	-	-	-	-	-	-	83	83	E5	S	-	-
-	-	-	-	-	-	-	-	84	84	F5	S	-	-
-	-	-	-	-	61	H11	H11	H11	85	H11	I/O	-	TIM4_CH3, LCD_FMC_D0, EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs				
-	-	-	-	-	62	H10	H10	86	J12	I/O	FT_I	-	TIM4_CH4, LCD_FMC_D1, EVENTO
-	-	-	-	-	-	G10	G10	87	H12	I/O	FT_s	-	SPI1_SCK, FMC_SAI2_SCK, EVENTO
-	-	-	-	-	-	F9	F9	88	G11	I/O	FT_s	-	SPI1_MISO, FMC_SAI2_FS_B, EVENTO
-	-	-	-	-	-	F10	F10	89	G9	I/O	FT_s	-	SPI1_MOSI, FMC_SAI2_MCL, EVENTO
-	-	-	-	-	-	E9	E9	90	G12	I/O	FT_s	-	SPI1_NS, LPUART1_FMC_A15, SAI2_EVENTO
-	-	-	-	-	-	G4	G4	91	F10	I/O	FT_s	-	I2C3_SM, LPUART1_R, EVENTO
-	-	-	-	-	-	H4	H4	92	F9	I/O	FT_fs	-	I2C3_SC, LPUART1_TX, FMC_SAI2_EVENTO
-	-	-	-	-	-	J6	J6	93	F11	I/O	FT_fs	-	I2C3_SD, LPUART1, EVENTO
-	-	-	-	-	-	-	-	94	M12	S	-	-	-
-	-	-	-	-	-	-	-	95	F8	S	-	-	-

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
	LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144		
	37	37	F3	G2	F3	63	E12	E12	96	96	E10	-	TIM3_CH1, TIM8_CH1 DFSDM1_OUT TSC_G4 LCD_SEG1 SDMMC1 SAI2_MCLK EVENTO
	38	38	F1	F2	F1	64	E11	E11	97	97	F12	-	TIM3_CH2, TIM8_CH2 DFSDM1_DATA0 TSC_G4 LCD_SEG1 SDMMC1 SAI2_MCLK EVENTO
	39	39	F2	F3	F2	65	E10	E10	98	98	E12	-	TIM3_CH3, TIM8_CH3 TSC_G4 LCD_SEG1 SDMMC1 EVENTO
	40	40	E1	E1	E1	66	D12	D12	99	99	E11	-	TIM8_BKIN2, TIM8_CH2 TIM8_CH1 TSC_G4 OTG_FS_N LCD_SEG1 SDMMC1 SAI2_EXTCLK TIM8_BKIN2 EVENTO



Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
	LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144		
41	41	E2	E3	E2	E2	67	D11	D11	100	100	D12	-	MCO, TIM1_USART1_OTG_FS_LCD_CO LPTIM2_EVENTO
42	42	E3	E3	E3	E3	68	D10	D10	101	101	D11	-	TIM1_CH2, USART1_LCD_CO TIM15_BKIN, EVENTO
43	43	D2	D2	D2	D2	69	C12	C12	102	102	C12	-	TIM1_CH3, USART1_OTG_FS_LCD_CO TIM17_BKIN, EVENTO
44	44	D1	D1	D1	D1	70	B12	B12	103	103	B12	-	TIM1_CH4, USART1_CAN1_RX, OTG_TIM1_BKIN2_EVENTO
45	45	C1	C1	C1	C1	71	A12	A12	104	104	B11	-	TIM1_ET USART1_RT CAN1_TX, OTG_EVENTO
46	46	C2	C2	C2	C2	72	A11	A11	105	105	C11	(4)	JTMS-SWDIO, OTG_FS_N EVENTO
47	47	B1	B1	B1	B1	-	-	-	-	-	E8	-	-

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WLSP72	WLSP72_SMPs	WLCSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144			
48	48	A1	A1	A1	73	C11	C11	106	106	E9	S	-	-
-	-	-	-	-	74	F11	F11	107	107	H8	S	-	-
-	-	-	-	-	75	G11	G11	108	108	H7	S	-	-
49	49	B2	B2	B2	76	A10	A10	109	109	C10	I/O	(4)	JTCK-SWCLK EVENTO
50	50	A2	C3	A2	77	A9	A9	110	110	D10	I/O	(4)	JTDI, TIM2_ETR, SPI3_NSS, UART4_RTSP, TSC_G3_LCD_SEG, SAI2_FS_B, EV
51	51	D3	A2	D3	78	B11	B11	111	111	B10	I/O	-	SPI3_SCK, USART4_TX, TSC_G3_LCD_COM4/LCD_SEG, SDMMC1_SAI2_SCK, EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number											Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WLCSP72	WLCSP72_SMPs	WLCSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144					
52	52	C3	D3	C3	79	C10	C10	112	112	C9	PC11	I/O	FT_I	-	SPI3_MISO USART3_ UART4_F TSC_G3_ LCD_COM5/LCD_SE /LCD_SE SDMMC1_ SAI2_MCL_ EVENTO
53	53	B3	B3	B3	80	B10	B10	113	113	B9	PC12	I/O	FT_I	-	SPI3_MOSI USART3_ UART5_ TSC_G3_ LCD_COM6/LCD_SE /LCD_SE SDMMC1_ SAI2_SD_B, EVENTO
-	-	-	-	-	81	C9	C9	114	114	A11	PD0	I/O	FT	-	SPI2_NSS DFSDM1_D CAN1_RX, F EVENTO
-	-	-	-	-	82	B9	B9	115	115	A10	PD1	I/O	FT	-	SPI2_SC DFSDM1_C CAN1_TX, F EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WLSP72	WLSP72_SMPs	WLCSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs					
54	-	A3	A3	A3	83	C8	C8	116	116	D9	I/O	FT_I	-	TIM3_ET USART3_RT UART5_RX, TS LCD_COM7/LCD /LCD_SEC SDMMC1_C EVENTO
-	-	-	-	-	84	B8	B8	117	117	D8	I/O	FT	-	SPI2_MI9 DFSDM1_D USART2_C FMC_CLK, EV
-	-	-	-	E5	85	B7	B7	118	118	C8	I/O	FT	-	SPI2_MC DFSDM1_C USART2_RT FMC_NOE, EV
-	-	-	-	D4	86	A6	A6	119	119	B8	I/O	FT	-	USART2_TX, FMC EVENTO
-	-	-	-	-	-	-	-	120	120	A1	S	-	-	-
-	-	-	-	E4	-	-	-	121	121	-	S	-	-	-
-	-	-	-	D5	87	B6	B6	122	122	A9	I/O	FT	-	DFSDM1_D USART2 FMC_NW SAI1_SD_A, EV
-	-	-	-	D6	88	A5	A5	123	123	A8	I/O	FT	-	DFSDM1_C USART2_CK, FMC EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number											Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144					
-	-	A4	A4	A4	-	D9	D9	D9	124	C7	PG9	I/O	FT_s	-	SPI3_SCK, USAA FMC_NCE/FM SAI2_SCK TIM15_CH EVENTO
-	-	B4	B4	B4	-	D8	D8	D8	125	D7	PG10	I/O	FT_s	-	LPTIM1_I SPI3_MIS USART1_RX, F SAI2_FS_A, TIM EVENTO
-	-	C4	-	C4	-	G3	G3	G3	126	B7	PG11	I/O	FT_s	-	LPTIM1_I SPI3_MC USART1_C SAI2_MCL TIM15_CH2, EV
-	-	C5	C4	C5	-	D7	D7	D7	127	A7	PG12	I/O	FT_s	-	LPTIM1_ETR, S USART1_RT FMC_NE4, SAA EVENTO
-	-	B5	B5	B5	-	C7	C7	C7	128	D6	PG13	I/O	FT_fs	-	I2C1_SDA, USAA FMC_A24, EV
-	-	A5	A5	A5	-	C6	-	-	129	A6	PG14	I/O	FT_fs	-	I2C1_SCL, FM EVENTO
-	-	-	-	-	-	F7	F7	F7	130	A12	VSS	S	-	-	
-	-	B6	B6	B6	-	G7	G7	G7	131	E6	VDDIO2	S	-	-	

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin type	I/O structure	Notes	Alternate function
	LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144		
	55	54	A6	A6	A6	89	A8	A8	133	132	B6	-	LPTIM1_C I2C1_SMBA, EV
	56	55	C6	C5	C6	90	A7	A7	134	133	C6	(4)	JTDO-TRACE TIM2_CH2, SPI3_SC USART1_RT LCD_SEE SAI1_SCK EVENTO
	57	56	C7	E7	C7	91	C5	C5	135	134	C5	(4)	NJTRST, TIM SPI1_MISO, SPI USART1_C USART5_RT TSC_G2_ LCD_SEE SAI1_MCL TIM17_BKIN, EV
												-	LPTIM1_IN1, TIM I2C1_SMBA, SPI SPI3_MC USART11 USART5_C TSC_G2_ LCD_SEE COMP2_C SAI1_SD TIM16_BKIN, EV

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate function	
	LQFP64	LQFP64_SMPs	WL CSP72	WL CSP72_SMPs	WL CSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144			LPTIM1_ENCODER TIM4_CH1, TIM4_CH2 I2C1_SCL, I2C1_SDA DFSDM1_DATA0 USART1_TX, USART1_RX TSC_G2_GPIO TIM8_BKIN2, TIM8_BKIN1 SAI1_FS, SAI1_MCLK TIM16_CH1, TIM16_CH2 EVENTO	
58	57	B7	E8	B7	B7	92	B5	B5	136	135	B5	PB6	I/O	FT_fa	-
59	58	A7	B7	A7	A7	93	B4	B4	137	136	A5	PB7	I/O	FT fla	-
60	59	D7	A7	D7	D7	94	A4	A4	138	137	A4	BOOT0	I	-	-
61	60	E7	C6	E7	E7	95	A3	A3	139	138	A3	PB8	I/O	FT_fla	TIM4_CH3, I2C1_SCL DFSDM1_DATA0 CAN1_RX, LCD_SEGMENT1 SDMMC1_DATA0 SAI1_MCLK, SAI1_FS TIM16_CH1, EVENTO

Table 16. STM32L476xx pin definitions (continued)

Pin Number										Pin name (function after reset)	Pin type	I/O structure	Notes	Alternate function
LQFP64	LQFP64_SMPs	WLCSP72	WLCSP72_SMPs	WLCSP81	LQFP100	UFBGA132	UFBGA132_SMPs	LQFP144	LQFP144_SMPs	UFBGA144				
62	61	E8	D7	E8	96	B3	B3	140	139	C4	PB9	I/O FT_fl	-	IR_OUT, TIM4_I2C1_SDA, SPI2_C1_TX, LCD0_CAN1_TX, LCD0_SDMMC1_SAI1_FS_A, TIM4_EVENTO
-	-	-	-	-	97	C3	C3	141	140	A2	PE0	I/O FT_I	-	TIM4_ETR, LCD0_FMC_NBL0, TIM4_EVENTO
-	-	-	-	-	98	A2	A2	142	141	B4	PE1	I/O FT_I	-	LCD_SEGMENT0_FMC_NBL1, TIM4_EVENTO
-	62	-	J1	-	-	-	C6	-	142	-	VDD12	S	-	-
63	63	A8	A8	A8	99	D3	D3	143	143	M1	VSS	S	-	-
64	64	A9	A9	A9	100	C4	C4	144	144	-	VDD	S	-	-

1. PC13, PC14 and PC15 are supplied through the power switch. Since the switch only sinks a limited amount of current, PC15 in output mode is limited:
 - The speed should not exceed 2 MHz with a maximum load of 30 pF
 - These GPIOs must not be used as current sources (e.g. to drive an LED).
2. After a Backup domain power-up, PC13, PC14 and PC15 operate as GPIOs. Their function then depends on the core reset by the system reset. For details on how to manage these GPIOs, refer to the Backup domain and RTC register manual.
3. OPAMPx_VINM pins are not available as additional functions on pins PA1 and PA7 on UFBGA packages. On UFBGA packages, dedicated pins.
4. After reset, these pins are configured as JTAG/SW debug alternate functions, and the internal pull-up on PA15, PA16 on PA14 pin are activated.

Table 17. Alternate function AF0 to AF7⁽¹⁾

Port	AF0	AF1	AF2	AF3	AF4	AF5
	SYS_AF	TIM1/TIM2/ TIM5/TIM8/ LPTIM1	TIM1/TIM2/ TIM3/TIM4/ TIM5	TIM8	I2C1/I2C2/I2C3	SPI1/SPI2
Port A	PA0	TIM2_CH1	TIM5_CH1	TIM8_ETR	-	-
	PA1	TIM2_CH2	TIM5_CH2	-	-	-
	PA2	TIM2_CH3	TIM5_CH3	-	-	-
	PA3	TIM2_CH4	TIM5_CH4	-	-	-
	PA4	-	-	-	-	SPI1_NSS
	PA5	TIM2_CH1	TIM2_ETR	TIM8_CH1N	-	SPI1_SCK
	PA6	TIM1_BKIN	TIM3_CH1	TIM8_BKIN	-	SPI1_MISO
	PA7	TIM1_CH1N	TIM3_CH2	TIM8_CH1N	-	SPI1_MOSI
	PA8	MCO	TIM1_CH1	-	-	-
	PA9	-	TIM1_CH2	-	-	-
	PA10	-	TIM1_CH3	-	-	-
	PA11	-	TIM1_CH4	TIM1_BKIN2	-	-
	PA12	-	TIM1_ETR	-	-	-
	PA13	JTMS-SWDIO	IR_OUT	-	-	-
	PA14	JTCK-SWCLK	-	-	-	-
	PA15	JTDI	TIM2_ETR	-	-	SPI1_NSS

Table 17. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5
	SYS_AF	TIM1/TIM2/ TIM5/TIM8/ LPTIM1	TIM1/TIM2/ TIM3/TIM4/ TIM5	TIM8	I2C1/I2C2/I2C3	SPI1/SPI2
Port B	PB0	-	TIM1_CH2N	TIM3_CH3	TIM8_CH2N	-
	PB1	-	TIM1_CH3N	TIM3_CH4	TIM8_CH3N	-
	PB2	RTC_OUT	LPTIM1_OUT	-	-	I2C3_SMBA
	PB3	JTDO- TRACESWO	TIM2_CH2	-	-	SPI1_SCK
	PB4	NJTRST	-	TIM3_CH1	-	SPI1_MISO
	PB5	-	LPTIM1_IN1	TIM3_CH2	-	SPI1_MOSI
	PB6	-	LPTIM1_ETR	TIM4_CH1	TIM8_BKIN2	-
	PB7	-	LPTIM1_IN2	TIM4_CH2	TIM8_BKIN	-
	PB8	-	-	TIM4_CH3	I2C1_SCL	-
	PB9	-	IR_OUT	TIM4_CH4	I2C1_SDA	SPI2_NSS
	PB10	-	TIM2_CH3	-	I2C2_SCL	SPI2_SCK
	PB11	-	TIM2_CH4	-	I2C2_SDA	-
	PB12	-	TIM1_BKIN	-	TIM1_BKIN_ COMP2	SPI2_NSS
	PB13	-	TIM1_CH1N	-	I2C2_SCL	SPI2_SCK
	PB14	-	TIM1_CH2N	-	I2C2_SDA	SPI2_MISO
	PB15	RTC_REFIN	TIM1_CH3N	-	-	SPI2_MOSI

Table 17. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5
	SYS_AF	TIM1/TIM2/ TIM5/TIM8/ LPTIM1	TIM1/TIM2/ TIM3/TIM4/ TIM5	TIM8	I2C1/I2C2/I2C3	SPI1/SPI2
Port C	PC0	-	LPTIM1_IN1	-	I2C3_SCL	-
	PC1	-	LPTIM1_OUT	-	I2C3_SDA	-
	PC2	-	LPTIM1_IN2	-	-	SPI2_MISO
	PC3	-	LPTIM1_ETR	-	-	SPI2_MOSI
	PC4	-	-	-	-	-
	PC5	-	-	-	-	-
	PC6	-	-	TIM3_CH1	TIM8_CH1	-
	PC7	-	-	TIM3_CH2	TIM8_CH2	-
	PC8	-	-	TIM3_CH3	TIM8_CH3	-
	PC9	-	TIM8_BKIN2	TIM3_CH4	TIM8_CH4	-
	PC10	-	-	-	-	-
	PC11	-	-	-	-	-
	PC12	-	-	-	-	-
	PC13	-	-	-	-	-
	PC14	-	-	-	-	-
	PC15	-	-	-	-	-

Table 17. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5
	SYS_AF	TIM1/TIM2/ TIM5/TIM8/ LPTIM1	TIM1/TIM2/ TIM3/TIM4/ TIM5	TIM8	I2C1/I2C2/I2C3	SPI1/SPI2
Port D	PD0	-	-	-	-	SPI2_NSS
	PD1	-	-	-	-	SPI2_SCK
	PD2	-	TIM3_ETR	-	-	-
	PD3	-	-	-	-	SPI2_MISO
	PD4	-	-	-	-	SPI2_MOSI
	PD5	-	-	-	-	-
	PD6	-	-	-	-	-
	PD7	-	-	-	-	-
	PD8	-	-	-	-	-
	PD9	-	-	-	-	-
	PD10	-	-	-	-	-
	PD11	-	-	-	-	-
	PD12	-	TIM4_CH1	-	-	-
	PD13	-	TIM4_CH2	-	-	-
	PD14	-	TIM4_CH3	-	-	-
	PD15	-	TIM4_CH4	-	-	-

Table 17. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5
	SYS_AF	TIM1/TIM2/ TIM5/TIM8/ LPTIM1	TIM1/TIM2/ TIM3/TIM4/ TIM5	TIM8	I2C1/I2C2/I2C3	SPI1/SPI2
Port E	PE0	-	TIM4_ETR	-	-	-
	PE1	-	-	-	-	-
	PE2	TRACECK	TIM3_ETR	-	-	-
	PE3	TRACED0	TIM3_CH1	-	-	-
	PE4	TRACED1	TIM3_CH2	-	-	-
	PE5	TRACED2	TIM3_CH3	-	-	-
	PE6	TRACED3	TIM3_CH4	-	-	-
	PE7	-	TIM1_ETR	-	-	-
	PE8	-	TIM1_CH1N	-	-	-
	PE9	-	TIM1_CH1	-	-	-
	PE10	-	TIM1_CH2N	-	-	-
	PE11	-	TIM1_CH2	-	-	-
	PE12	-	TIM1_CH3N	-	-	SPI1_NSS
	PE13	-	TIM1_CH3	-	-	SPI1_SCK
	PE14	-	TIM1_CH4	TIM1_BKIN2_ COMP2	-	SPI1_MISO
	PE15	-	TIM1_BKIN	TIM1_BKIN_ COMP1	-	SPI1_MOSI



Table 17. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5
	SYS_AF	TIM1/TIM2/ TIM5/TIM8/ LPTIM1	TIM1/TIM2/ TIM3/TIM4/ TIM5	TIM8	I2C1/I2C2/I2C3	SPI1/SPI2
Port F	PF0	-	-	-	I2C2_SDA	-
	PF1	-	-	-	I2C2_SCL	-
	PF2	-	-	-	I2C2_SMBA	-
	PF3	-	-	-	-	-
	PF4	-	-	-	-	-
	PF5	-	-	-	-	-
	PF6	-	TIM5_ETR	-	-	-
	PF7	-	TIM5_CH1	-	-	-
	PF8	-	TIM5_CH2	-	-	-
	PF9	-	TIM5_CH3	-	-	-
	PF10	-	TIM5_CH4	-	-	-
	PF11	-	-	-	-	-
	PF12	-	-	-	-	-
	PF13	-	-	-	-	-
	PF14	-	-	-	-	-
	PF15	-	-	-	-	-

Table 17. Alternate function AF0 to AF7⁽¹⁾ (continued)

Port	AF0	AF1	AF2	AF3	AF4	AF5
	SYS_AF	TIM1/TIM2/ TIM5/TIM8/ LPTIM1	TIM1/TIM2/ TIM3/TIM4/ TIM5	TIM8	I2C1/I2C2/I2C3	SPI1/SPI2
Port G	PG0	-	-	-	-	-
	PG1	-	-	-	-	-
	PG2	-	-	-	-	SPI1_SCK
	PG3	-	-	-	-	SPI1_MISO
	PG4	-	-	-	-	SPI1_MOSI
	PG5	-	-	-	-	SPI1_NSS
	PG6	-	-	-	I2C3_SMBA	-
	PG7	-	-	-	I2C3_SCL	-
	PG8	-	-	-	I2C3_SDA	-
	PG9	-	-	-	-	-
	PG10	-	LPTIM1_IN1	-	-	-
	PG11	-	LPTIM1_IN2	-	-	-
	PG12	-	LPTIM1_ETR	-	-	-
	PG13	-	-	-	I2C1_SDA	-
	PG14	-	-	-	I2C1_SCL	-
	PG15	-	LPTIM1_OUT	-	I2C1_SMBA	-
Port H	PH0	-	-	-	-	-
	PH1	-	-	-	-	-

1. Please refer to [Table 18](#) for AF8 to AF15.

Table 18. Alternate function AF8 to AF15⁽¹⁾

Port	AF8	AF9	AF10	AF11	AF12	AF15
	UART4, UART5, LPUART1	CAN1, TSC	OTG_FS, QUADSPI	LCD	SDMMC1, COMP1, COMP2, FMC, SWPMI1	SAI1, SAI2
Port A	PA0	-	-	-	-	SAI1_E
	PA1	-	-	LCD_SEG0	-	-
	PA2	-	-	LCD_SEG1	-	SAI2_E
	PA3	-	-	LCD_SEG2	-	-
	PA4	-	-	-	-	SAI1_F
	PA5	-	-	-	-	-
	PA6	-	QUADSPI_BK1_IO3	LCD_SEG3	TIM1_BKIN_ COMP2	TIM8_E COM
	PA7	-	QUADSPI_BK1_IO2	LCD_SEG4	-	-
	PA8	-	OTG_FS_SOF	LCD_COM0	-	-
	PA9	-	-	LCD_COM1	-	-
	PA10	-	OTG_FS_ID	LCD_COM2	-	-
	PA11	CAN1_RX	OTG_FS_DM	-	TIM1_BKIN2_ COMP1	-
	PA12	CAN1_TX	OTG_FS_DP	-	-	-
	PA13	-	OTG_FS_NOE	-	-	-
	PA14	-	-	-	-	-
PA15	UART4_RTS _DE	TSC_G3_IO1	-	LCD_SEG17	-	SAI2_F

Table 18. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF1
	UART4, UART5, LPUART1	CAN1, TSC	OTG_FS, QUADSPI	LCD	SDMMC1, COMP1, COMP2, FMC, SWPMI1	SAI1, S
PB0	-	-	QUADSPI_BK1_IO1	LCD_SEG5	COMP1_OUT	-
PB1	-	-	QUADSPI_BK1_IO0	LCD_SEG6	-	-
PB2	-	-	-	-	-	-
PB3	-	-	-	LCD_SEG7	-	SAI1_S
PB4	UART5_RTS _DE	TSC_G2_IO1	-	LCD_SEG8	-	SAI1_M B
PB5	UART5_CTS	TSC_G2_IO2	-	LCD_SEG9	COMP2_OUT	SAI1_S
PB6	-	TSC_G2_IO3	-	-	TIM8_BKIN2_ COMP2	SAI1_F
PB7	UART4_CTS	TSC_G2_IO4	-	LCD_SEG21	FMC_NL	TIM8_B COM
PB8	-	CAN1_RX	-	LCD_SEG16	SDMMC1_D4	SAI1_M A
PB9	-	CAN1_TX	-	LCD_COM3	SDMMC1_D5	SAI1_F
PB10	LPUART1_ RX	-	QUADSPI_CLK	LCD_SEG10	COMP1_OUT	SAI1_S
PB11	LPUART1_TX	-	QUADSPI_NCS	LCD_SEG11	COMP2_OUT	-
PB12	LPUART1_ RTS_DE	TSC_G1_IO1	-	LCD_SEG12	SWPMI1_IO	SAI2_F
PB13	LPUART1_ CTS	TSC_G1_IO2	-	LCD_SEG13	SWPMI1_TX	SAI2_S
PB14	-	TSC_G1_IO3	-	LCD_SEG14	SWPMI1_RX	SAI2_M A
PB15	-	TSC_G1_IO4	-	LCD_SEG15	SWPMI1_SUSPEND	SAI2_S

Port B





Table 18. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF1
	UART4, UART5, LPUART1	CAN1, TSC	OTG_FS, QUADSPI	LCD	SDMMC1, COMP1, COMP2, FMC, SWPMI1	SAI1, SAI2
Port C	PC0	-	-	LCD_SEG18	-	-
	PC1	-	-	LCD_SEG19	-	-
	PC2	-	-	LCD_SEG20	-	-
	PC3	-	-	LCD_VLCD	-	SAI1_S
	PC4	-	-	LCD_SEG22	-	-
	PC5	-	-	LCD_SEG23	-	-
	PC6	TSC_G4_IO1	-	LCD_SEG24	SDMMC1_D6	SAI2_M_A
	PC7	TSC_G4_IO2	-	LCD_SEG25	SDMMC1_D7	SAI2_M_B
	PC8	TSC_G4_IO3	-	LCD_SEG26	SDMMC1_D0	-
	PC9	TSC_G4_IO4	OTG_FS_NOE	LCD_SEG27	SDMMC1_D1	SAI2_E
	PC10	UART4_TX	-	LCD_COM4/ LCD_SEG28/ LCD_SEG40	SDMMC1_D2	SAI2_S
	PC11	UART4_RX	-	LCD_COM5/ LCD_SEG29/ LCD_SEG41	SDMMC1_D3	SAI2_M_B
	PC12	UART5_TX	-	LCD_COM6/ LCD_SEG30/ LCD_SEG42	SDMMC1_CK	SAI2_S
	PC13	-	-	-	-	-
	PC14	-	-	-	-	-
	PC15	-	-	-	-	-

Table 18. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF15
	UART4, UART5, LPUART1	CAN1, TSC	OTG_FS, QUADSPI	LCD	SDMMC1, COMP1, COMP2, FMC, SWPMI1	SAI1, SAI2
Port D	PD0	CAN1_RX	-	-	FMC_D2	-
	PD1	CAN1_TX	-	-	FMC_D3	-
	PD2	UART5_RX TSC_SYNC	-	LCD_COM7/ LCD_SEG31/ LCD_SEG43	SDMMC1_CMD	-
	PD3	-	-	-	FMC_CLK	-
	PD4	-	-	-	FMC_NOE	-
	PD5	-	-	-	FMC_NWE	-
	PD6	-	-	-	FMC_NWAIT	SAI1_S
	PD7	-	-	-	FMC_NE1	-
	PD8	-	-	LCD_SEG28	FMC_D13	-
	PD9	-	-	LCD_SEG29	FMC_D14	SAI2_M SAI2_A
	PD10	TSC_G6_IO1	-	LCD_SEG30	FMC_D15	SAI2_S
	PD11	TSC_G6_IO2	-	LCD_SEG31	FMC_A16	SAI2_S
	PD12	TSC_G6_IO3	-	LCD_SEG32	FMC_A17	SAI2_F
	PD13	TSC_G6_IO4	-	LCD_SEG33	FMC_A18	-
	PD14	-	-	LCD_SEG34	FMC_D0	-
	PD15	-	-	LCD_SEG35	FMC_D1	-

Table 18. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF1
	UART4, UART5, LPUART1	CAN1, TSC	OTG_FS, QUADSPI	LCD	SDMMC1, COMP1, COMP2, FMC, SWPMI1	SAI1, SAI2, SAI3
Port E	PE0	-	-	LCD_SEG36	FMC_NBL0	-
	PE1	-	-	LCD_SEG37	FMC_NBL1	-
	PE2	TSC_G7_IO1	-	LCD_SEG38	FMC_A23	SAI1_M_A
	PE3	TSC_G7_IO2	-	LCD_SEG39	FMC_A19	SAI1_S
	PE4	TSC_G7_IO3	-	-	FMC_A20	SAI1_F
	PE5	TSC_G7_IO4	-	-	FMC_A21	SAI1_S
	PE6	-	-	-	FMC_A22	SAI1_S
	PE7	-	-	-	FMC_D4	SAI1_S
	PE8	-	-	-	FMC_D5	SAI1_S
	PE9	-	-	-	FMC_D6	SAI1_F
	PE10	TSC_G5_IO1	QUADSPI_CLK	-	FMC_D7	SAI1_M_B
	PE11	TSC_G5_IO2	QUADSPI_NCS	-	FMC_D8	-
	PE12	TSC_G5_IO3	QUADSPI_BK1_IO0	-	FMC_D9	-
	PE13	TSC_G5_IO4	QUADSPI_BK1_IO1	-	FMC_D10	-
	PE14	-	QUADSPI_BK1_IO2	-	FMC_D11	-
	PE15	-	QUADSPI_BK1_IO3	-	FMC_D12	-

Table 18. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF1
	UART4, UART5, LPUART1	CAN1, TSC	OTG_FS, QUADSPI	LCD	SDDMMC1, COMP1, COMP2, FMC, SWPMI1	SAI1, S
Port F	PF0	-	-	-	FMC_A0	-
	PF1	-	-	-	FMC_A1	-
	PF2	-	-	-	FMC_A2	-
	PF3	-	-	-	FMC_A3	-
	PF4	-	-	-	FMC_A4	-
	PF5	-	-	-	FMC_A5	-
	PF6	-	-	-	-	SAI1_S
	PF7	-	-	-	-	SAI1_M B
	PF8	-	-	-	-	SAI1_S
	PF9	-	-	-	-	SAI1_F
	PF10	-	-	-	-	-
	PF11	-	-	-	-	-
	PF12	-	-	-	FMC_A6	-
	PF13	-	-	-	FMC_A7	-
	PF14	-	TSC_G8_IO1	-	FMC_A8	-
	PF15	-	TSC_G8_IO2	-	FMC_A9	-



Table 18. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF1
	UART4, UART5, LPUART1	CAN1, TSC	OTG_FS, QUADSPI	LCD	SDMMC1, COMP1, COMP2, FMC, SWPMI1	SAI1, S
Port G	PG0	TSC_G8_IO3	-	-	FMC_A10	-
	PG1	TSC_G8_IO4	-	-	FMC_A11	-
	PG2	-	-	-	FMC_A12	SAI2_S
	PG3	-	-	-	FMC_A13	SAI2_F
	PG4	-	-	-	FMC_A14	SAI2_M_B
	PG5	LPUART1_CTS	-	-	FMC_A15	SAI2_S
	PG6	LPUART1_RTS_DE	-	-	-	-
	PG7	LPUART1_TX	-	-	FMC_INT	-
	PG8	LPUART1_RX	-	-	-	-
	PG9	-	-	-	FMC_NCE/ FMC_NE2	SAI2_S
	PG10	-	-	-	FMC_NE3	SAI2_F
	PG11	-	-	-	-	SAI2_M_A
	PG12	-	-	-	FMC_NE4	SAI2_S
	PG13	-	-	-	FMC_A24	-
	PG14	-	-	-	FMC_A25	-
	PG15	-	-	-	-	-

Table 18. Alternate function AF8 to AF15⁽¹⁾ (continued)

Port	AF8	AF9	AF10	AF11	AF12	AF13
	UART4, UART5, LPUART1	CAN1, TSC	OTG_FS, QUADSPI	LCD	SDMMC1, COMP1, COMP2, FMC, SWPMI1	SAI1, SAI2
Port H	PH0	-	-	-	-	-
	PH1	-	-	-	-	-

1. Please refer to [Table 17](#) for AF0 to AF7.



5 Memory mapping

Figure 17. STM32L476xx memory map

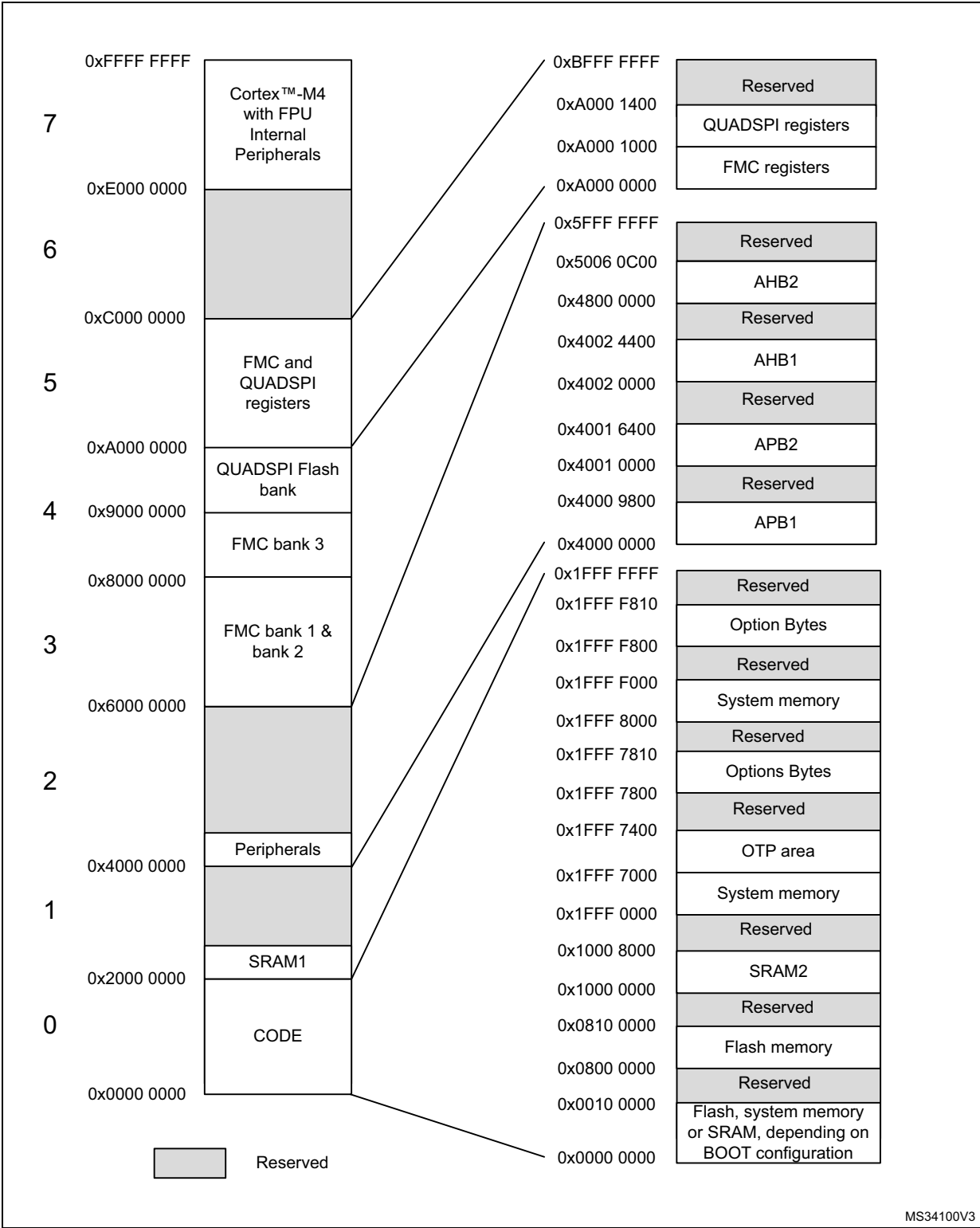


Table 19. STM32L476xx memory map and peripheral register boundary addresses⁽¹⁾

Bus	Boundary address	Size (bytes)	Peripheral
AHB3	0xA000 1000 - 0xA000 13FF	1 KB	QUADSPI
	0xA000 0000 - 0xA000 0FFF	4 KB	FMC
AHB2	0x5006 0800 - 0x5006 0BFF	1 KB	RNG
	0x5004 0400 - 0x5006 07FF	129 KB	Reserved
	0x5004 0000 - 0x5004 03FF	1 KB	ADC
	0x5000 0000 - 0x5003 FFFF	16 KB	OTG_FS
	0x4800 2000 - 0x4FFF FFFF	~127 MB	Reserved
	0x4800 1C00 - 0x4800 1FFF	1 KB	GPIOH
	0x4800 1800 - 0x4800 1BFF	1 KB	GPIOG
	0x4800 1400 - 0x4800 17FF	1 KB	GPIOF
	0x4800 1000 - 0x4800 13FF	1 KB	GPIOE
	0x4800 0C00 - 0x4800 0FFF	1 KB	GPIOD
	0x4800 0800 - 0x4800 0BFF	1 KB	GPIOC
	0x4800 0400 - 0x4800 07FF	1 KB	GPIOB
	0x4800 0000 - 0x4800 03FF	1 KB	GPIOA
-	0x4002 4400 - 0x47FF FFFF	~127 MB	Reserved
AHB1	0x4002 4000 - 0x4002 43FF	1 KB	TSC
	0x4002 3400 - 0x4002 3FFF	1 KB	Reserved
	0x4002 3000 - 0x4002 33FF	1 KB	CRC
	0x4002 2400 - 0x4002 2FFF	3 KB	Reserved
	0x4002 2000 - 0x4002 23FF	1 KB	FLASH registers
	0x4002 1400 - 0x4002 1FFF	3 KB	Reserved
	0x4002 1000 - 0x4002 13FF	1 KB	RCC
	0x4002 0800 - 0x4002 0FFF	2 KB	Reserved
	0x4002 0400 - 0x4002 07FF	1 KB	DMA2
	0x4002 0000 - 0x4002 03FF	1 KB	DMA1

Table 19. STM32L476xx memory map and peripheral register boundary addresses⁽¹⁾
(continued)

Bus	Boundary address	Size (bytes)	Peripheral
APB2	0x4001 6400 - 0x4001 FFFF	39 KB	Reserved
	0x4001 6000 - 0x4000 63FF	1 KB	DFSDM1
	0x4001 5C00 - 0x4000 5FFF	1 KB	Reserved
	0x4001 5800 - 0x4000 5BFF	1 KB	SAI2
	0x4001 5400 - 0x4000 57FF	1 KB	SAI1
	0x4001 4C00 - 0x4000 53FF	2 KB	Reserved
	0x4001 4800 - 0x4001 4BFF	1 KB	TIM17
	0x4001 4400 - 0x4001 47FF	1 KB	TIM16
	0x4001 4000 - 0x4001 43FF	1 KB	TIM15
APB2	0x4001 3C00 - 0x4001 3FFF	1 KB	Reserved
	0x4001 3800 - 0x4001 3BFF	1 KB	USART1
	0x4001 3400 - 0x4001 37FF	1 KB	TIM8
	0x4001 3000 - 0x4001 33FF	1 KB	SPI1
	0x4001 2C00 - 0x4001 2FFF	1 KB	TIM1
	0x4001 2800 - 0x4001 2BFF	1 KB	SDMMC1
	0x4001 2000 - 0x4001 27FF	2 KB	Reserved
	0x4001 1C00 - 0x4001 1FFF	1 KB	FIREWALL
	0x4001 0800 - 0x4001 1BFF	5 KB	Reserved
	0x4001 0400 - 0x4001 07FF	1 KB	EXTI
	0x4001 0200 - 0x4001 03FF	1 KB	COMP
	0x4001 0030 - 0x4001 01FF		VREFBUF
	0x4001 0000 - 0x4001 002F		SYSCFG

Table 19. STM32L476xx memory map and peripheral register boundary addresses⁽¹⁾
(continued)

Bus	Boundary address	Size (bytes)	Peripheral
APB1	0x4000 9800 - 0x4000 FFFF	26 KB	Reserved
	0x4000 9400 - 0x4000 97FF	1 KB	LPTIM2
	0x4000 8C00 - 0x4000 93FF	2 KB	Reserved
	0x4000 8800 - 0x4000 8BFF	1 KB	SWPMI1
	0x4000 8400 - 0x4000 87FF	1 KB	Reserved
	0x4000 8000 - 0x4000 83FF	1 KB	LPUART1
	0x4000 7C00 - 0x4000 7FFF	1 KB	LPTIM1
	0x4000 7800 - 0x4000 7BFF	1 KB	OPAMP
	0x4000 7400 - 0x4000 77FF	1 KB	DAC1
	0x4000 7000 - 0x4000 73FF	1 KB	PWR
	0x4000 6800 - 0x4000 6FFF	1 KB	Reserved
	0x4000 6400 - 0x4000 67FF	1 KB	CAN1
	0x4000 6000 - 0x4000 63FF	1 KB	Reserved
	0x4000 5C00 - 0x4000 5FFF	1 KB	I2C3
	0x4000 5800 - 0x4000 5BFF	1 KB	I2C2
	0x4000 5400 - 0x4000 57FF	1 KB	I2C1
	0x4000 5000 - 0x4000 53FF	1 KB	UART5
	0x4000 4C00 - 0x4000 4FFF	1 KB	UART4
	0x4000 4800 - 0x4000 4BFF	1 KB	USART3
	0x4000 4400 - 0x4000 47FF	1 KB	USART2

Table 19. STM32L476xx memory map and peripheral register boundary addresses⁽¹⁾
(continued)

Bus	Boundary address	Size (bytes)	Peripheral
APB1	0x4000 4000 - 0x4000 43FF	1 KB	Reserved
	0x4000 3C00 - 0x4000 3FFF	1 KB	SPI3
	0x4000 3800 - 0x4000 3BFF	1 KB	SPI2
	0x4000 3400 - 0x4000 37FF	1 KB	Reserved
	0x4000 3000 - 0x4000 33FF	1 KB	IWDG
	0x4000 2C00 - 0x4000 2FFF	1 KB	WWDG
	0x4000 2800 - 0x4000 2BFF	1 KB	RTC
	0x4000 2400 - 0x4000 27FF	1 KB	LCD
	0x4000 1800 - 0x4000 23FF	3 KB	Reserved
	0x4000 1400 - 0x4000 17FF	1 KB	TIM7
	0x4000 1000 - 0x4000 13FF	1 KB	TIM6
	0x4000 0C00 - 0x4000 0FFF	1 KB	TIM5
	0x4000 0800 - 0x4000 0BFF	1 KB	TIM4
	0x4000 0400 - 0x4000 07FF	1 KB	TIM3
	0x4000 0000 - 0x4000 03FF	1 KB	TIM2

1. The gray color is used for reserved boundary addresses.

6 Electrical characteristics

6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to V_{SS} .

6.1.1 Minimum and maximum values

Unless otherwise specified, the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at $T_A = 25\text{ °C}$ and $T_A = T_{Amax}$ (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation (mean $\pm 3\sigma$).

6.1.2 Typical values

Unless otherwise specified, typical data are based on $T_A = 25\text{ °C}$, $V_{DD} = V_{DDA} = 3\text{ V}$. They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated (mean $\pm 2\sigma$).

6.1.3 Typical curves

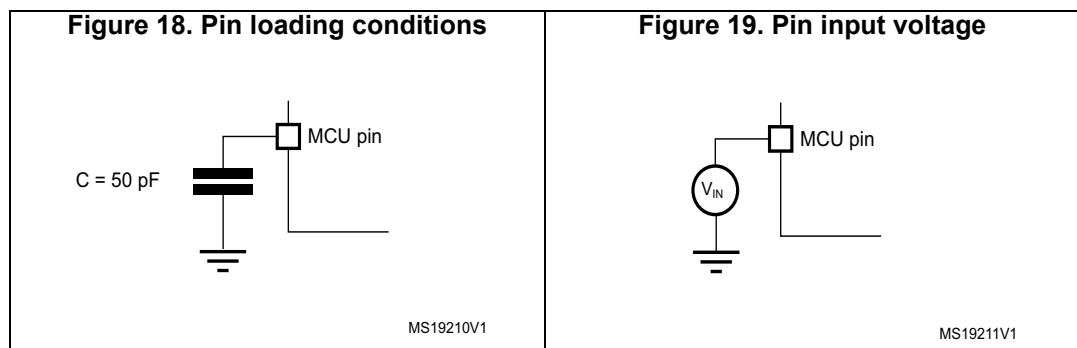
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in [Figure 18](#).

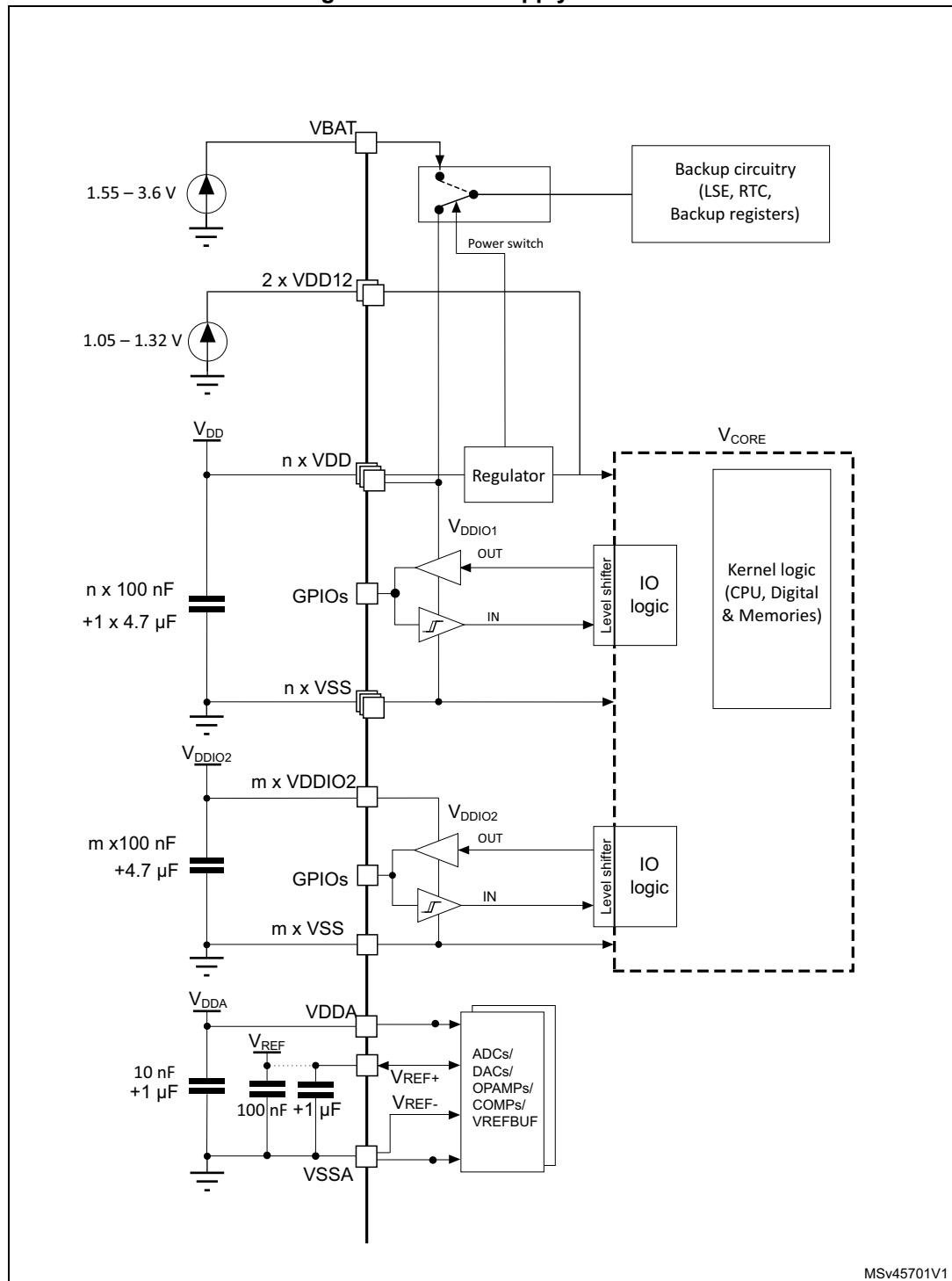
6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in [Figure 19](#).



6.1.6 Power supply scheme

Figure 20. Power supply scheme

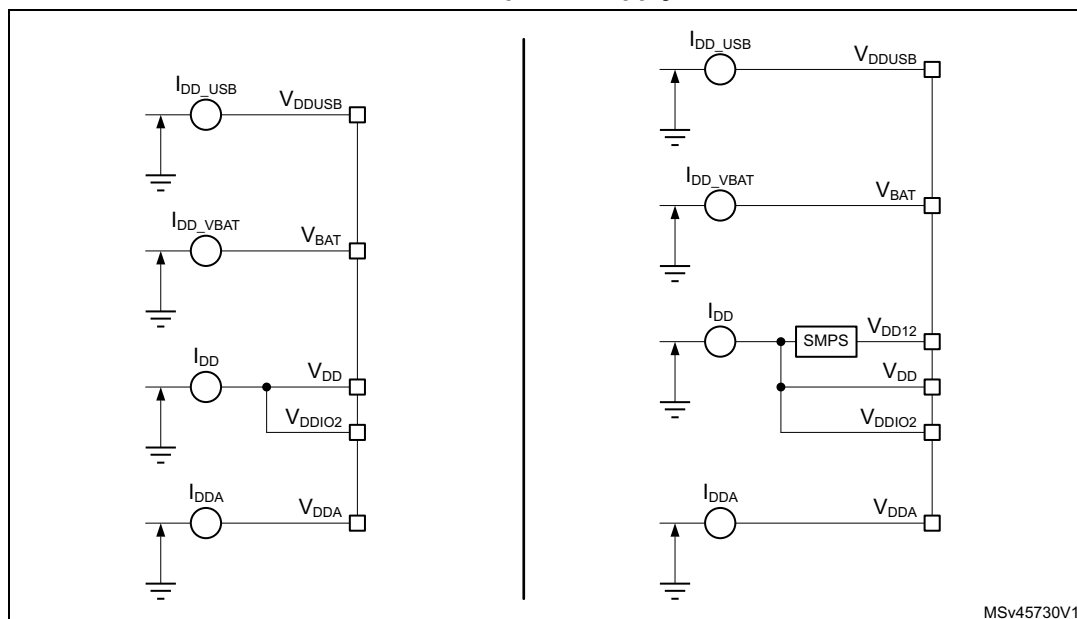


Caution: Each power supply pair (V_{DD}/V_{SS} , V_{DDA}/V_{SSA} etc.) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or

below, the appropriate pins on the underside of the PCB to ensure the good functionality of the device.

6.1.7 Current consumption measurement

Figure 21. Current consumption measurement scheme with and without external SMPS power supply



6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in [Table 20: Voltage characteristics](#), [Table 21: Current characteristics](#) and [Table 22: Thermal characteristics](#) may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability. Device mission profile (application conditions) is compliant with JEDEC JESD47 qualification standard, extended mission profiles are available on demand.

Table 20. Voltage characteristics⁽¹⁾

Symbol	Ratings		Min	Max	Unit
$V_{DDX} - V_{SS}$	External main supply voltage (including V_{DD} , V_{DDA} , V_{DDIO2} , V_{DDUSB} , V_{LCD} , V_{BAT})		-0.3	4.0	V
$V_{DD12} - V_{SS}$	External SMPS supply voltage	Range 1	-0.3	1.4	V
		Range 2	-0.3		
$V_{IN}^{(2)}$	Input voltage on FT_XXX pins		$V_{SS}-0.3$	$\min(V_{DD}, V_{DDA}, V_{DDIO2}, V_{DDUSB}, V_{LCD}) + 4.0^{(3)(4)}$	V
	Input voltage on TT_XX pins		$V_{SS}-0.3$	4.0	
	Input voltage on BOOT0 pin		V_{SS}	9.0	
	Input voltage on any other pins		$V_{SS}-0.3$	4.0	

Table 20. Voltage characteristics⁽¹⁾ (continued)

Symbol	Ratings	Min	Max	Unit
$ \Delta V_{DDx} $	Variations between different V_{DDx} power pins of the same domain	-	50	mV
$ V_{SSx}-V_{SS} $	Variations between all the different ground pins ⁽⁵⁾	-	50	mV

1. All main power (V_{DD} , V_{DDA} , V_{DDIO2} , V_{DDUSB} , V_{LCD} , V_{BAT}) and ground (V_{SS} , V_{SSA}) pins must always be connected to the external power supply, in the permitted range.
2. V_{IN} maximum must always be respected. Refer to [Table 21: Current characteristics](#) for the maximum allowed injected current values.
3. This formula has to be applied only on the power supplies related to the IO structure described in the pin definition table.
4. To sustain a voltage higher than 4 V the internal pull-up/pull-down resistors must be disabled.
5. Include VREF- pin.

Table 21. Current characteristics

Symbol	Ratings	Max	Unit
$\Sigma I_{V_{DD}}$	Total current into sum of all V_{DD} power lines (source) ⁽¹⁾⁽²⁾	150	mA
$\Sigma I_{V_{SS}}$	Total current out of sum of all V_{SS} ground lines (sink) ⁽¹⁾	150	
$I_{V_{DD}(PIN)}$	Maximum current into each V_{DD} power pin (source) ⁽¹⁾⁽²⁾	100	
$I_{V_{SS}(PIN)}$	Maximum current out of each V_{SS} ground pin (sink) ⁽¹⁾	100	
$I_{IO(PIN)}$	Output current sunk by any I/O and control pin except FT_f	20	
	Output current sunk by any FT_f pin	20	
	Output current sourced by any I/O and control pin	20	
$\Sigma I_{IO(PIN)}$	Total output current sunk by sum of all I/Os and control pins ⁽³⁾	100	
	Total output current sourced by sum of all I/Os and control pins ⁽³⁾	100	
$I_{INJ(PIN)}^{(4)}$	Injected current on FT_xxx, TT_xx, RST and B pins, except PA4, PA5	-5/+0 ⁽⁵⁾	
	Injected current on PA4, PA5	-5/0	
$\Sigma I_{INJ(PIN)} $	Total injected current (sum of all I/Os and control pins) ⁽⁶⁾	25	

1. All main power (V_{DD} , V_{DDA} , V_{DDIO2} , V_{DDUSB} , V_{LCD} , V_{BAT}) and ground (V_{SS} , V_{SSA}) pins must always be connected to the external power supplies, in the permitted range.
2. Valid also for V_{DD12} on SMPS packages.
3. This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count QFP packages.
4. Positive injection (when $V_{IN} > V_{DDIOx}$) is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
5. A negative injection is induced by $V_{IN} < V_{SS}$. $I_{INJ(PIN)}$ must never be exceeded. Refer also to [Table 20: Voltage characteristics](#) for the minimum allowed input voltage values.
6. When several inputs are submitted to a current injection, the maximum $\Sigma |I_{INJ(PIN)}|$ is the absolute sum of the negative injected currents (instantaneous values).

Table 22. Thermal characteristics

Symbol	Ratings	Value	Unit
T_{STG}	Storage temperature range	–65 to +150	°C
T_J	Maximum junction temperature	150	°C

6.3 Operating conditions

6.3.1 General operating conditions

Table 23. General operating conditions

Symbol	Parameter	Conditions	Min	Max	Unit
f_{HCLK}	Internal AHB clock frequency	-	0	80	MHz
f_{PCLK1}	Internal APB1 clock frequency	-	0	80	
f_{PCLK2}	Internal APB2 clock frequency	-	0	80	
V_{DD}	Standard operating voltage	-	1.71 (1)	3.6	V
V_{DD12}	Standard operating voltage	Full frequency range	1.08	1.32	V
		Up to 26 MHz	1.05		
V_{DDIO2}	PG[15:2] I/Os supply voltage	At least one I/O in PG[15:2] used	1.08	3.6	V
		PG[15:2] not used	0	3.6	
V_{DDA}	Analog supply voltage	ADC or COMP used	1.62	3.6	V
		DAC or OPAMP used	1.8		
		VREFBUF used	2.4		
		ADC, DAC, OPAMP, COMP, VREFBUF not used	0		
V_{BAT}	Backup operating voltage	-	1.55	3.6	V
V_{DDUSB}	USB supply voltage	USB used	3.0	3.6	V
		USB not used	0	3.6	
V_{IN}	I/O input voltage	TT_xx I/O	-0.3	$V_{DDIOx}+0.3$	V
		BOOT0	0	9	
		All I/O except BOOT0 and TT_xx	-0.3	$\text{Min}(\text{Min}(V_{DD}, V_{DDA}, V_{DDIO2}, V_{DDUSB}, V_{LCD})+3.6 \text{ V}, 5.5 \text{ V})^{(2)(3)}$	
P_D	Power dissipation at $T_A = 85^\circ\text{C}$ for suffix 6 or $T_A = 105^\circ\text{C}$ for suffix 7 ⁽⁴⁾	LQFP144	-	625	mW
		LQFP100	-	476	
		LQFP64	-	444	
		UFBGA144	-	377	
		UFBGA132	-	363	
		WLCSP81	-	487	
		WLCSP72	-	434	

Table 23. General operating conditions (continued)

Symbol	Parameter	Conditions	Min	Max	Unit
P_D	Power dissipation at $T_A = 125\text{ °C}$ for suffix 3 ⁽⁴⁾	LQFP144	-	156	mW
		LQFP100	-	119	
		LQFP64	-	111	
		UFBGA144	-	94	
		UFBGA132	-	90	
		WLCSP81	-	121	
		WLCSP72	-	108	
T_A	Ambient temperature for the suffix 6 version	Maximum power dissipation	-40	85	°C
		Low-power dissipation ⁽⁵⁾	-40	105	
	Ambient temperature for the suffix 7 version	Maximum power dissipation	-40	105	
		Low-power dissipation ⁽⁵⁾	-40	125	
	Ambient temperature for the suffix 3 version	Maximum power dissipation	-40	125	
		Low-power dissipation ⁽⁵⁾	-40	130	
T_J	Junction temperature range	Suffix 6 version	-40	105	°C
		Suffix 7 version	-40	125	
		Suffix 3 version	-40	130	

- When RESET is released functionality is guaranteed down to V_{BOR0} Min.
- This formula has to be applied only on the power supplies related to the IO structure described by the pin definition table. Maximum I/O input voltage is the smallest value between $\text{Min}(V_{DD}, V_{DDA}, V_{DDIO2}, V_{DDUSB}, V_{LCD}) + 3.6\text{ V}$ and 5.5 V .
- For operation with voltage higher than $\text{Min}(V_{DD}, V_{DDA}, V_{DDIO2}, V_{DDUSB}, V_{LCD}) + 0.3\text{ V}$, the internal Pull-up and Pull-Down resistors must be disabled.
- If T_A is lower, higher P_D values are allowed as long as T_J does not exceed T_{Jmax} (see [Section 7.8: Thermal characteristics](#)).
- In low-power dissipation state, T_A can be extended to this range as long as T_J does not exceed T_{Jmax} (see [Section 7.8: Thermal characteristics](#)).

6.3.2 Operating conditions at power-up / power-down

The parameters given in [Table 24](#) are derived from tests performed under the ambient temperature condition summarized in [Table 23](#).

Table 24. Operating conditions at power-up / power-down⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
t_{VDD}	V_{DD} rise time rate	-	0	∞	$\mu\text{s/V}$
	V_{DD} fall time rate		10	∞	
t_{VDDA}	V_{DDA} rise time rate	-	0	∞	$\mu\text{s/V}$
	V_{DDA} fall time rate		10	∞	
t_{VDDUSB}	V_{DDUSB} rise time rate	-	0	∞	$\mu\text{s/V}$
	V_{DDUSB} fall time rate		10	∞	

Table 24. Operating conditions at power-up / power-down⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Max	Unit
t_{VDDIO2}	V_{DDIO2} rise time rate	-	0	∞	$\mu\text{s/V}$
	V_{DDIO2} fall time rate		10	∞	

1. At power-up, the V_{DD12} voltage should not be forced externally.

The requirements for power-up/down sequence specified in [Section 3.9.1: Power supply schemes](#) must be respected.

6.3.3 Embedded reset and power control block characteristics

The parameters given in [Table 25](#) are derived from tests performed under the ambient temperature conditions summarized in [Table 23: General operating conditions](#).

Table 25. Embedded reset and power control block characteristics

Symbol	Parameter	Conditions ⁽¹⁾	Min	Typ	Max	Unit
$t_{RSTTEMPO}^{(2)}$	Reset temporization after BOR0 is detected	V_{DD} rising	-	250	400	μs
$V_{BOR0}^{(2)}$	Brown-out reset threshold 0	Rising edge	1.62	1.66	1.7	V
		Falling edge	1.6	1.64	1.69	
V_{BOR1}	Brown-out reset threshold 1	Rising edge	2.06	2.1	2.14	V
		Falling edge	1.96	2	2.04	
V_{BOR2}	Brown-out reset threshold 2	Rising edge	2.26	2.31	2.35	V
		Falling edge	2.16	2.20	2.24	
V_{BOR3}	Brown-out reset threshold 3	Rising edge	2.56	2.61	2.66	V
		Falling edge	2.47	2.52	2.57	
V_{BOR4}	Brown-out reset threshold 4	Rising edge	2.85	2.90	2.95	V
		Falling edge	2.76	2.81	2.86	
V_{PVD0}	Programmable voltage detector threshold 0	Rising edge	2.1	2.15	2.19	V
		Falling edge	2	2.05	2.1	
V_{PVD1}	PVD threshold 1	Rising edge	2.26	2.31	2.36	V
		Falling edge	2.15	2.20	2.25	
V_{PVD2}	PVD threshold 2	Rising edge	2.41	2.46	2.51	V
		Falling edge	2.31	2.36	2.41	
V_{PVD3}	PVD threshold 3	Rising edge	2.56	2.61	2.66	V
		Falling edge	2.47	2.52	2.57	
V_{PVD4}	PVD threshold 4	Rising edge	2.69	2.74	2.79	V
		Falling edge	2.59	2.64	2.69	
V_{PVD5}	PVD threshold 5	Rising edge	2.85	2.91	2.96	V
		Falling edge	2.75	2.81	2.86	

Table 25. Embedded reset and power control block characteristics (continued)

Symbol	Parameter	Conditions ⁽¹⁾	Min	Typ	Max	Unit
V_{PVD6}	PVD threshold 6	Rising edge	2.92	2.98	3.04	V
		Falling edge	2.84	2.90	2.96	
V_{hyst_BORH0}	Hysteresis voltage of BORH0	Hysteresis in continuous mode	-	20	-	mV
		Hysteresis in other mode	-	30	-	
$V_{hyst_BOR_PVD}$	Hysteresis voltage of BORH (except BORH0) and PVD	-	-	100	-	mV
I_{DD} (BOR_PVD) ⁽²⁾	BOR ⁽³⁾ (except BOR0) and PVD consumption from V_{DD}	-	-	1.1	1.6	μ A
V_{PVM1}	V_{DDUSB} peripheral voltage monitoring	-	1.18	1.22	1.26	V
V_{PVM2}	V_{DDIO2} peripheral voltage monitoring	-	0.92	0.96	1	V
V_{PVM3}	V_{DDA} peripheral voltage monitoring	Rising edge	1.61	1.65	1.69	V
		Falling edge	1.6	1.64	1.68	
V_{PVM4}	V_{DDA} peripheral voltage monitoring	Rising edge	1.78	1.82	1.86	V
		Falling edge	1.77	1.81	1.85	
V_{hyst_PVM3}	PVM3 hysteresis	-	-	10	-	mV
V_{hyst_PVM4}	PVM4 hysteresis	-	-	10	-	mV
I_{DD} (PVM1/PVM2) (2)	PVM1 and PVM2 consumption from V_{DD}	-	-	0.2	-	μ A
I_{DD} (PVM3/PVM4) (2)	PVM3 and PVM4 consumption from V_{DD}	-	-	2	-	μ A

1. Continuous mode means Run/Sleep modes, or temperature sensor enable in Low-power run/Low-power sleep modes.
2. Guaranteed by design.
3. BOR0 is enabled in all modes (except shutdown) and its consumption is therefore included in the supply current characteristics tables.

6.3.4 Embedded voltage reference

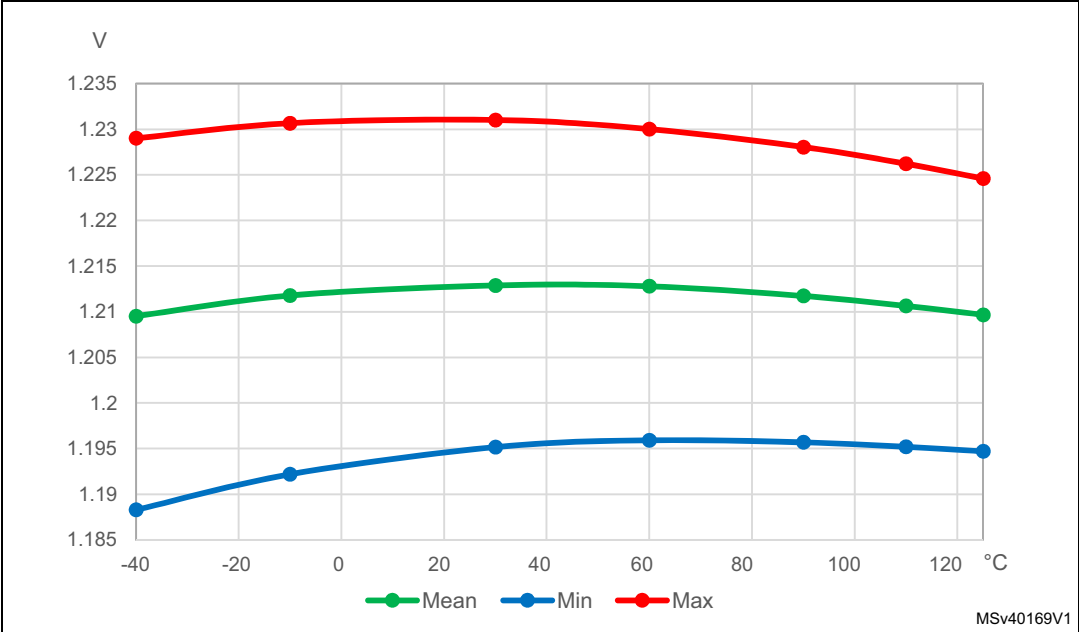
The parameters given in [Table 26](#) are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 23: General operating conditions](#).

Table 26. Embedded internal voltage reference

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{REFINT}	Internal reference voltage	$-40\text{ }^{\circ}\text{C} < T_A < +130\text{ }^{\circ}\text{C}$	1.182	1.212	1.232	V
$t_{S_vrefint}^{(1)}$	ADC sampling time when reading the internal reference voltage	-	4 ⁽²⁾	-	-	μs
$t_{start_vrefint}$	Start time of reference voltage buffer when ADC is enable	-	-	8	12 ⁽²⁾	μs
$I_{DD}(V_{REFINTBUF})$	V_{REFINT} buffer consumption from V_{DD} when converted by ADC	-	-	12.5	20 ⁽²⁾	μA
ΔV_{REFINT}	Internal reference voltage spread over the temperature range	$V_{DD} = 3\text{ V}$	-	5	7.5 ⁽²⁾	mV
T_{Coeff}	Average temperature coefficient	$-40\text{ }^{\circ}\text{C} < T_A < +130\text{ }^{\circ}\text{C}$	-	30	50 ⁽²⁾	ppm/ $^{\circ}\text{C}$
A_{Coeff}	Long term stability	1000 hours, $T = 25\text{ }^{\circ}\text{C}$	-	300	1000 ⁽²⁾	ppm
$V_{DDCoeff}$	Average voltage coefficient	$3.0\text{ V} < V_{DD} < 3.6\text{ V}$	-	250	1200 ⁽²⁾	ppm/V
V_{REFINT_DIV1}	1/4 reference voltage	-	24	25	26	% V_{REFINT}
V_{REFINT_DIV2}	1/2 reference voltage		49	50	51	
V_{REFINT_DIV3}	3/4 reference voltage		74	75	76	

1. The shortest sampling time can be determined in the application by multiple iterations.
2. Guaranteed by design.

Figure 22. V_{REFINT} versus temperature



6.3.5 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in [Figure 21: Current consumption measurement scheme with and without external SMPS power supply](#).

The I_{DD_ALL} parameters given in [Table 27](#) to [Table 49](#) represent the total MCU consumption including the current supplying V_{DD} , V_{DD12} , V_{DDIO2} , V_{DDA} , V_{LCD} , V_{DDUSB} and V_{BAT} .

Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in analog input mode
- All peripherals are disabled except when explicitly mentioned
- The Flash memory access time is adjusted with the minimum wait states number, depending on the f_{HCLK} frequency (refer to the table “Number of wait states according to CPU clock (HCLK) frequency” available in the RM0351 reference manual).
- When the peripherals are enabled $f_{PCLK} = f_{HCLK}$

The parameters given in [Table 27](#) to [Table 50](#) are derived from tests performed under ambient temperature and supply voltage conditions summarized in [Table 23: General operating conditions](#).

Table 27. Current consumption in Run and Low-power run modes, code with data running from Flash, ART enable (Cache ON Prefetch OFF)

Symbol	Parameter	Conditions		TYP						
		-	Voltage scaling	f _{HCLK}	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD,ALL} (Run)	Supply current in Run mode	f _{HCLK} = f _{HSE} up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	2.88	2.93	3.05	3.23	3.58	3.20
				16 MHz	1.83	1.87	1.98	2.16	2.49	2.01
				8 MHz	0.98	1.02	1.12	1.29	1.62	1.10
				4 MHz	0.55	0.59	0.69	0.85	1.18	0.61
				2 MHz	0.34	0.37	0.47	0.64	0.96	0.37
				1 MHz	0.23	0.26	0.36	0.53	0.85	0.27
				100 kHz	0.14	0.17	0.27	0.43	0.75	0.17
		Range 1	80 MHz	10.2	10.3	10.5	10.7	11.1	11.22	
			72 MHz	9.24	9.31	9.47	9.69	10.1	10.16	
			64 MHz	8.25	8.32	8.46	8.68	9.09	9.08	
I _{DD,ALL} (LPRun)	Supply current in Low-power run mode	f _{HCLK} = f _{MSI} all peripherals disable	48 MHz	6.28	6.35	6.5	6.72	7.11	6.91	
			32 MHz	4.24	4.30	4.44	4.65	5.04	4.66	
			24 MHz	3.21	3.27	3.4	3.61	3.98	3.53	
			16 MHz	2.19	2.24	2.36	2.56	2.94	2.41	
			2 MHz	272	303	413	592	958	330	
			1 MHz	154	184	293	473	835	195	
			400 kHz	78	108	217	396	758	110	
			100 kHz	42	73	182	360	723	75	

1. Guaranteed by characterization results, unless otherwise specified.

**Table 28. Current consumption in Run modes, code with data processing running
ART enable (Cache ON Prefetch OFF) and power supplied by external SMPS**
($V_{DD12} = 1.10\text{ V}$)

Symbol	Parameter	Conditions ⁽¹⁾	
		f_{HCLK}	25 °C
$I_{DD_ALL}(Run)$	Supply current in Run mode	- $f_{HCLK} = f_{HSE}$ up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	80 MHz
			3
			72 MHz
			3
			64 MHz
			2
			48 MHz
			2
			32 MHz
			1
			24 MHz
			1
			16 MHz
			0
			8 MHz
			0
			4 MHz
			0
			2 MHz
			0
			1 MHz
			0
			100 kHz
			0

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters: SMPS input = $V_{DD12} = 1.10\text{ V}$



Table 29. Current consumption in Run and Low-power run modes, code with data running from Flash, ART disable

Symbol	Parameter	Conditions		TYP						
		-	Voltage scaling	f _{HCLK}	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD,ALL} (Run)	Supply current in Run mode	f _{HCLK} = f _{HSE} up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	3.15	3.19	3.31	3.50	3.85	3.47
				16 MHz	2.24	2.28	2.39	2.57	2.90	2.46
				8 MHz	1.26	1.29	1.40	1.57	1.89	1.40
				4 MHz	0.71	0.75	0.85	1.02	1.34	0.79
				2 MHz	0.42	0.45	0.55	0.72	1.04	0.46
				1 MHz	0.27	0.30	0.40	0.57	0.89	0.30
			Range 1	100 kHz	0.14	0.17	0.27	0.43	0.75	0.17
				80 MHz	10.0	10.1	10.3	10.6	11.0	11.00
				72 MHz	9.06	9.13	9.28	9.51	9.92	9.97
				64 MHz	8.96	9.04	9.22	9.48	9.92	9.86
				48 MHz	7.64	7.72	7.91	8.17	8.62	8.40
				32 MHz	5.49	5.57	5.74	5.98	6.40	6.04
I _{DD,ALL} (LPRun)	Supply current in Low-power run	f _{HCLK} = f _{MSI} all peripherals disable	24 MHz	4.16	4.22	4.36	4.57	4.96	4.60	
			16 MHz	2.93	2.99	3.13	3.35	3.75	3.22	
			2 MHz	358	392	503	683	1050	435	
			1 MHz	197	230	340	519	880	245	
			400 kHz	97	126	235	414	778	130	
			100 kHz	47	77	186	365	726	85	

1. Guaranteed by characterization results, unless otherwise specified.



Table 30. Current consumption in Run modes, code with data processing running
ART disable and power supplied by external SMPS ($V_{DD12} = 1.10\text{ V}$)

Symbol	Parameter	Conditions ⁽¹⁾	
		f_{HCLK}	25 °C
$I_{DD_ALL}(Run)$	Supply current in Run mode	80 MHz	3
		72 MHz	3
		64 MHz	3
		48 MHz	2
		32 MHz	2
		24 MHz	1
		16 MHz	1
		8 MHz	0
		4 MHz	0
		2 MHz	0
		1 MHz	0
		100 kHz	0

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters: SMPS input = $V_{DD12} = 1.10\text{ V}$

Table 31. Current consumption in Run and Low-power run modes, code with data running from SRAM1

Symbol	Parameter	Conditions		TYP						
		-	Voltage scaling	f _{HCLK}	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD,ALL} (Run)	Supply current in Run mode	f _{HCLK} = f _{HSE} up to 48MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	2.88	2.94	3.05	3.23	3.58	3.18
				16 MHz	1.83	1.87	1.98	2.15	2.50	2.01
				8 MHz	0.97	1.00	1.11	1.27	1.62	1.07
				4 MHz	0.54	0.57	0.67	0.84	1.18	0.59
				2 MHz	0.33	0.36	0.46	0.62	0.96	0.37
				1 MHz	0.22	0.25	0.35	0.51	0.85	0.25
		Range 1	100 kHz	0.12	0.15	0.25	0.41	0.75	0.15	
			80 MHz	10.2	10.3	10.5	10.7	11.1	11.22	
			72 MHz	9.25	9.31	9.46	9.68	10.1	10.18	
			64 MHz	8.25	8.31	8.46	8.67	9.08	9.08	
			48 MHz	6.26	6.33	6.48	6.69	7.11	6.89	
			32 MHz	4.22	4.28	4.42	4.63	5.03	4.64	
			24 MHz	3.20	3.25	3.38	3.59	3.99	3.52	
I _{DD,ALL} (LPRun)	Supply current in low-power run mode	f _{HCLK} = f _{MSI} all peripherals disable FLASH in power-down	16 MHz	2.18	2.22	2.35	2.55	2.94	2.40	
			2 MHz	242	275	384	562	924	300	
			1 MHz	130	162	269	445	809	180	
			400 kHz	61	90	197	374	734	95	
			100 kHz	26	56	163	339	702	55	

1. Guaranteed by characterization results, unless otherwise specified.

Table 32. Current consumption in Run, code with data processing running
SRAM1 and power supplied by external SMPS ($V_{DD12} = 1.10\text{ V}$)

Symbol	Parameter	Conditions ⁽¹⁾	
		f_{HCLK}	25 °C
$I_{DD_ALL}(Run)$	Supply current in Run mode	80 MHz	3.67
		72 MHz	3.33
		64 MHz	2.97
		48 MHz	2.25
		32 MHz	1.52
		24 MHz	1.15
		16 MHz	0.78
		8 MHz	0.42
		4 MHz	0.23
		2 MHz	0.14
		1 MHz	0.09
		100 kHz	0.05

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters: SMPS input = $V_{DD12} = 1.10\text{ V}$



Table 33. Typical current consumption in Run and Low-power run modes, with different codes running from Flash, ART enable (Cache ON Prefetch OFF)

Symbol	Parameter	Conditions			TYP	Unit	TYP	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I_{DD_ALL} (Run)	Supply current in Run mode	$f_{HCLK} = f_{HSE}$ up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2 $f_{HCLK} = 26$ MHz	Reduced code ⁽¹⁾	2.9	mA	111	$\mu A/MHz$
				Coremark	3.1		118	
				Dhrystone 2.1	3.1		119	
				Fibonacci	2.9		112	
				While(1)	2.8		108	
			Range 1 $f_{HCLK} = 80$ MHz	Reduced code ⁽¹⁾	10.2	mA	127	$\mu A/MHz$
				Coremark	10.9		136	
				Dhrystone 2.1	11.0		137	
				Fibonacci	10.5		131	
				While(1)	9.9		124	
I_{DD_ALL} (LPRun)	Supply current in Low-power run	$f_{HCLK} = f_{MSI} = 2$ MHz all peripherals disable		Reduced code ⁽¹⁾	272	μA	136	$\mu A/MHz$
				Coremark	291		145	
				Dhrystone 2.1	302		151	
				Fibonacci	269		135	
				While(1)	269		135	

1. Reduced code used for characterization results provided in [Table 27](#), [Table 29](#), [Table 31](#).

Table 34. Typical current consumption in Run, with different codes running from Flash, ART enable (Cache ON Prefetch OFF) and power supplied by external SMPS ($V_{DD12} = 1.10$ V)

Symbol	Parameter	Conditions ⁽¹⁾			TYP	Unit	TYP	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I_{DD_ALL} (Run)	Supply current in Run mode	$f_{HCLK} = f_{HSE}$ up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	$f_{HCLK} = 26$ MHz	Reduced code ⁽²⁾	1.25	mA	48	$\mu A/MHz$
				Coremark	1.34		51	
				Dhrystone 2.1	1.34		51	
				Fibonacci	1.25		48	
				While(1)	1.21		46	
			$f_{HCLK} = 80$ MHz	Reduced code ⁽²⁾	3.67		46	
				Coremark	3.92		49	
				Dhrystone 2.1	3.95		49	
				Fibonacci	3.77		47	
				While(1)	3.56		44	

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters:
SMPS input = 3.3 V, SMPS efficiency = 85%, $V_{DD12} = 1.10$ V
2. Reduced code used for characterization results provided in [Table 27](#), [Table 29](#), [Table 31](#).

Table 35. Typical current consumption in Run, with different codes running from Flash, ART enable (Cache ON Prefetch OFF) and power supplied by external SMPS ($V_{DD12} = 1.05$ V)

Symbol	Parameter	Conditions ⁽¹⁾			TYP	Unit	TYP	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I_{DD_ALL} (Run)	Supply current in Run mode	$f_{HCLK} = f_{HSE}$ up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	$f_{HCLK} = 26$ MHz	Reduced code ⁽²⁾	1.14	mA	44	$\mu A/MHz$
				Coremark	1.22		47	
				Dhrystone 2.1	1.22		47	
				Fibonacci	1.14		44	
				While(1)	1.10		42	

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters:
SMPS input = 3.3 V, SMPS efficiency = 85%, $V_{DD12} = 1.05$ V
2. Reduced code used for characterization results provided in [Table 27](#), [Table 29](#), [Table 31](#).

Table 36. Typical current consumption in Run and Low-power run modes, with different codes running from Flash, ART disable

Symbol	Parameter	Conditions			TYP	Unit	TYP	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I_{DD_ALL} (Run)	Supply current in Run mode	$f_{HCLK} = f_{HSE}$ up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2 $f_{HCLK} = 26$ MHz	Reduced code ⁽¹⁾	3.1	mA	119	$\mu A/MHz$
				Coremark	2.9		111	
				Dhrystone 2.1	2.8		111	
				Fibonacci	2.7		104	
				While(1)	2.6		100	
			Range 1 $f_{HCLK} = 80$ MHz	Reduced code ⁽¹⁾	10.0	mA	125	$\mu A/MHz$
				Coremark	9.4		117	
				Dhrystone 2.1	9.1		114	
				Fibonacci	9.0		112	
				While(1)	9.3		116	
I_{DD_ALL} (LPRun)	Supply current in Low-power run	$f_{HCLK} = f_{MSI} = 2$ MHz all peripherals disable		Reduced code ⁽¹⁾	358	μA	179	$\mu A/MHz$
				Coremark	392		196	
				Dhrystone 2.1	390		195	
				Fibonacci	385		192	
				While(1)	385		192	

1. Reduced code used for characterization results provided in [Table 27](#), [Table 29](#), [Table 31](#).

Table 37. Typical current consumption in Run modes, with different codes running from Flash, ART disable and power supplied by external SMPS ($V_{DD12} = 1.10$ V)

Symbol	Parameter	Conditions ⁽¹⁾			TYP	Unit	TYP	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I_{DD_ALL} (Run)	Supply current in Run mode	$f_{HCLK} = f_{HSE}$ up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	$f_{HCLK} = 26$ MHz	Reduced code ⁽²⁾	1.34	mA	51	$\mu A/MHz$
				Coremark	1.25		48	
				Dhrystone 2.1	1.21		46	
				Fibonacci	1.16		45	
				While(1)	1.12		43	
			$f_{HCLK} = 80$ MHz	Reduced code ⁽²⁾	3.59		45	
				Coremark	3.38		42	
				Dhrystone 2.1	3.27		41	
				Fibonacci	3.24		40	
				While(1)	3.34		42	

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters: SMPS input = 3.3 V, SMPS efficiency = 85%, $V_{DD12} = 1.10$ V
2. Reduced code used for characterization results provided in [Table 27](#), [Table 29](#), [Table 31](#).

Table 38. Typical current consumption in Run modes, with different codes running from Flash, ART disable and power supplied by external SMPS ($V_{DD12} = 1.05$ V)

Symbol	Parameter	Conditions ⁽¹⁾			TYP	Unit	TYP	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I_{DD_ALL} (Run)	Supply current in Run mode	$f_{HCLK} = f_{HSE}$ up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals	$f_{HCLK} = 26$ MHz	Reduced code ⁽²⁾	1.22	mA	47	$\mu A/MHz$
				Coremark	1.14		44	
				Dhrystone 2.1	1.10		42	
				Fibonacci	1.06		41	
				While(1)	1.02		39	

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters: SMPS input = 3.3 V, SMPS efficiency = 85%, $V_{DD12} = 1.05$ V
2. Reduced code used for characterization results provided in [Table 27](#), [Table 29](#), [Table 31](#).

Table 39. Typical current consumption in Run and Low-power run modes, with different codes running from SRAM1

Symbol	Parameter	Conditions			TYP	Unit	TYP	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I_{DD_ALL} (Run)	Supply current in Run mode	$f_{HCLK} = f_{HSE}$ up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2 $f_{HCLK} = 26$ MHz	Reduced code ⁽¹⁾	2.9	mA	111	$\mu A/MHz$
				Coremark	2.9		111	
				Dhrystone 2.1	2.9		111	
				Fibonacci	2.6		100	
				While(1)	2.6		100	
			Range 1 $f_{HCLK} = 80$ MHz	Reduced code ⁽¹⁾	10.2	mA	127	$\mu A/MHz$
				Coremark	10.4		130	
				Dhrystone 2.1	10.3		129	
				Fibonacci	9.6		120	
				While(1)	9.3		116	
I_{DD_ALL} (LPRun)	Supply current in Low-power run	$f_{HCLK} = f_{MSI} = 2$ MHz all peripherals disable		Reduced code ⁽¹⁾	242	μA	121	$\mu A/MHz$
				Coremark	242		121	
				Dhrystone 2.1	242		121	
				Fibonacci	225		112	
				While(1)	242		121	

1. Reduced code used for characterization results provided in [Table 27](#), [Table 29](#), [Table 31](#).

Table 40. Typical current consumption in Run mode, with different codes running from SRAM1 and power supplied by external SMPS ($V_{DD12} = 1.10$ V)

Symbol	Parameter	Conditions ⁽¹⁾			TYP	Unit	TYP	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I_{DD_ALL} (Run)	Supply current in Run mode	$f_{HCLK} = f_{HSE}$ up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	$f_{HCLK} = 26$ MHz	Reduced code ⁽²⁾	1.25	mA	48	$\mu A/MHz$
				Coremark	1.25		48	
				Dhrystone 2.1	1.25		48	
				Fibonacci	1.12		43	
				While(1)	1.12		43	
			$f_{HCLK} = 80$ MHz	Reduced code ⁽²⁾	3.67		46	
				Coremark	3.74		47	
				Dhrystone 2.1	3.70		46	
				Fibonacci	3.45		43	
				While(1)	3.34		42	

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters: SMPS input = 3.3 V, SMPS efficiency = 85%, $V_{DD12} = 1.10$ V

2. Reduced code used for characterization results provided in [Table 27](#), [Table 29](#), [Table 31](#).

Table 41. Typical current consumption in Run mode, with different codes running from SRAM1 and power supplied by external SMPS ($V_{DD12} = 1.05\text{ V}$)

Symbol	Parameter	Conditions ⁽¹⁾			TYP	Unit	TYP	Unit
		-	Voltage scaling	Code	25 °C		25 °C	
I_{DD_ALL} (Run)	Supply current in Run mode	$f_{HCLK} = f_{HSE}$ up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	$f_{HCLK} = 26\text{ MHz}$	Reduced code ⁽²⁾	1.14	mA	44	$\mu\text{A/MHz}$
				Coremark	1.14		44	
				Dhrystone 2.1	1.14		44	
				Fibonacci	1.02		39	
				While(1)	1.02		39	

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters: SMPS input = 3.3 V, SMPS efficiency = 85%, $V_{DD12} = 1.05\text{ V}$

2. Reduced code used for characterization results provided in [Table 27](#), [Table 29](#), [Table 31](#).



Table 42. Current consumption in Sleep and Low-power sleep modes, Flash

Symbol	Parameter	Conditions		TYP						
		-	Voltage scaling	f _{HCLK}	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD,ALL} (Sleep)	Supply current in sleep mode,	f _{HCLK} = f _{HSE} up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable	Range 2	26 MHz	0.92	0.96	1.07	1.25	1.59	1.012
				16 MHz	0.61	0.65	0.75	0.92	1.27	0.69
				8 MHz	0.36	0.40	0.50	0.66	1.01	0.42
				4 MHz	0.24	0.27	0.37	0.53	0.87	0.28
				2 MHz	0.18	0.20	0.30	0.47	0.81	0.215
				1 MHz	0.15	0.17	0.27	0.43	0.77	0.18
				100 kHz	0.12	0.14	0.24	0.41	0.74	0.15
			Range 1	80 MHz	2.96	3.00	3.13	3.33	3.73	3.26
				72 MHz	2.69	2.73	2.85	3.05	3.45	2.96
				64 MHz	2.41	2.45	2.58	2.77	3.17	2.65
I _{DD,ALL} (LPSleep)	Supply current in low-power sleep mode	f _{HCLK} = f _{MSI} all peripherals disable	48 MHz	1.88	1.93	2.07	2.27	2.67	2.10	
			32 MHz	1.30	1.35	1.48	1.68	2.08	1.43	
			24 MHz	1.01	1.05	1.17	1.37	1.76	1.11	
			16 MHz	0.71	0.75	0.87	1.07	1.45	0.80	
			2 MHz	96	126	233	412	775	130	
			1 MHz	65	94	202	381	742	95	
			400 kHz	43	73	181	359	718	75	
			100 kHz	33	63	171	348	708	65	

1. Guaranteed by characterization results, unless otherwise specified.

Table 43. Current consumption in Sleep, Flash ON and power supplied by external source (V_{DD12} = 1.10 V)

Symbol	Parameter	Conditions ⁽¹⁾		
		-	f _{HCLK}	25 °C
I _{DD_ALL} (Sleep)	Supply current in sleep mode, f _{HCLK} = f _{HSE} up to 48 MHz included, bypass mode PLL ON above 48 MHz all peripherals disable		80 MHz	1.06
			72 MHz	0.97
			64 MHz	0.87
			48 MHz	0.68
			32 MHz	0.47
			24 MHz	0.36
			16 MHz	0.26
			8 MHz	0.16
			4 MHz	0.10
			2 MHz	0.08
			1 MHz	0.06
			100 kHz	0.05

1. All values are obtained by calculation based on measurements done without SMPS and using following parameters: SMPS input = 1.10 V

Table 44. Current consumption in Low-power sleep modes, Flash in power

Symbol	Parameter	Conditions		TYP				
		-	Voltage scaling	f _{HCLK}	25 °C	55 °C	85 °C	105 °C
I _{DD_ALL} (LPSleep)	Supply current in low-power sleep mode f _{HCLK} = f _{MSI} all peripherals disable			2 MHz	81	110	217	395
				1 MHz	50	78	185	362
				400 kHz	28	57	163	340
				100 kHz	18	47	155	332

1. Guaranteed by characterization results, unless otherwise specified.

Table 45. Current consumption in Stop 2 mode

Symbol	Parameter	Conditions	TYP						
			V _{DD}	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD_ALL} (Stop 2)	Supply current in Stop 2 mode, RTC disabled	-	1.8 V	1.14	3.77	14.7	34.7	77	2.7
			2.4 V	1.15	3.86	15	35.5	79.1	2.7
			3 V	1.18	3.97	15.4	36.4	81.3	2.8
			3.6 V	1.26	4.11	16	38	85.1	3.0
I _{DD_ALL} (Stop 2 with RTC)	Supply current in Stop 2 mode, RTC enabled	LCD disabled	1.8 V	1.43	3.98	15	35	77.3	3.2
			2.4 V	1.49	4.07	15.3	35.8	79.4	3.2
		LCD enabled ⁽³⁾ clocked by LSI	3 V	1.54	4.24	15.7	36.7	81.6	3.3
			3.6 V	1.75	4.47	16.1	38.3	85.4	3.5
		RTC clocked by LSI, LCD disabled	1.8 V	1.42	4.04	15	34.9	77.2	3.1
			2.4 V	1.5	4.22	15.4	35.7	79.2	3.2
		RTC clocked by LSI, LCD enabled ⁽³⁾	3 V	1.64	4.37	15.8	36.7	81.4	3.4
			3.6 V	1.79	4.65	16.6	38.4	85.4	3.6
	Supply current in Stop 2 mode, RTC enabled	RTC clocked by LSE bypassed at 32768Hz, LCD disabled	1.8 V	1.53	4.07	15.1	35.1	77.4	3.3
			2.4 V	1.62	4.32	15.5	35.9	79.5	3.4
		RTC clocked by LSE quartz ⁽⁴⁾ in low drive mode, LCD disabled	3 V	1.69	4.43	15.9	36.8	81.7	3.5
			3.6 V	1.86	4.65	16.7	38.5	85.5	3.7
		RTC clocked by LSE quartz ⁽⁴⁾ in low drive mode, LCD disabled	1.8 V	1.5	4.13	15.2	35.3	77.6	3.2
			2.4 V	1.63	4.33	15.6	36	79.6	3.4
		RTC clocked by LSE quartz ⁽⁴⁾ in low drive mode, LCD disabled	3 V	1.79	4.55	16.1	37	81.8	3.6
			3.6 V	2.04	4.9	16.8	38.7	85.6	3.9
	Supply current in Stop 2 mode, RTC enabled	-	1.8 V	1.43	3.99	14.7	35	-	3.2
			2.4 V	1.54	4.11	15	35.8	-	3.3
			3 V	1.67	4.29	15.5	36.7	-	3.4
			3.6 V	1.87	4.57	16.2	38.3	-	3.7

Table 45. Current consumption in Stop 2 mode (continued)

Symbol	Parameter	Conditions		TYP					
		-	V _{DD}	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD_ALL} (wake up from Stop 2)	Supply current during wake up from Stop 2 mode	Wake up clock is MSI = 48 MHz, voltage Range 1. See ⁽⁵⁾ .	3 V	1.9	-	-	-	-	-
		Wake up clock is MSI = 4 MHz, voltage Range 2. See ⁽⁵⁾ .	3 V	2.24	-	-	-	-	-
		Wake up clock is HSI16 = 16 MHz, voltage Range 1. See ⁽⁵⁾ .	3 V	2.1	-	-	-	-	-

1. Guaranteed by characterization results, unless otherwise specified.

2. Guaranteed by test in production.

3. LCD enabled with external voltage source. Consumption from VLCD excluded. Refer to LCD controller characteristics for I_{VLCD}.

4. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

5. Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in [Table 52: Low-power mode](#)

Table 46. Current consumption in Stop 1 mode

Symbol	Parameter	Conditions		V _{DD}	TYP						
		-	-		25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	
I _{DD_ALL} (Stop 1)	Supply current in Stop 1 mode, RTC disabled	-	LCD disabled	1.8 V	6.59	24.7	92.7	208	437	16	
				2.4 V	6.65	24.8	92.9	209	439	17	
				3 V	6.65	24.9	93.3	210	442	17	
				3.6 V	6.70	25.1	93.8	212	447	17	
		-	LCD enabled ⁽²⁾ clocked by LSI	1.8 V	7.00	25.2	97.2	219	461	18	
				2.4 V	7.14	25.4	97.5	220	463	18	
I _{DD_ALL} (Stop 1 with RTC)	Supply current in stop 1 mode, RTC enabled	RTC clocked by LSI	LCD disabled	3 V	7.24	25.7	97.7	221	465	18	
				3.6 V	7.36	26.1	98.7	223	471	18	
				1.8 V	6.88	25.0	93.1	209	439	17	
			LCD enabled ⁽²⁾	2.4 V	7.02	25.2	93.7	210	441	18	
				3 V	7.12	25.4	94.2	212	444	18	
				3.6 V	7.25	25.7	95.2	214	449	18	
		RTC clocked by LSE bypassed at 32768 Hz	LCD disabled	1.8 V	7.01	26.1	99.0	223	467	18	
				2.4 V	7.14	26.3	99.6	225	470	18	
				3 V	7.31	26.6	100.0	226	474	18	
				3.6 V	7.41	26.9	102.0	229	480	19	
			RTC clocked by LSE quartz ⁽³⁾ in low drive mode	LCD disabled	1.8 V	6.91	25.2	93.4	210	440	17
					2.4 V	7.04	25.3	94.2	211	443	18
					3 V	7.19	25.7	95.0	212	446	18
					3.6 V	7.97	26.0	96.1	215	451	20
		1.8 V	6.85	25.0	93.0	208.3	-	17			
		2.4 V	6.94	25.1	93.2	209.3	-	17			
		3 V	7.10	25.2	93.6	210.3	-	18			
		3.6 V	7.34	25.4	94.1	212.3	-	18			

Table 46. Current consumption in Stop 1 mode (continued)

Symbol	Parameter	Conditions		V _{DD}	TYP					
		-	-		25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD, ALL} (wake-up from Stop1)	Supply current during wake-up from Stop 1	Wakeup clock MSI = 48 MHz, voltage Range 1. See (4).		3 V	1.47	-	-	-	-	-
		Wakeup clock MSI = 4 MHz, voltage Range 2. See (4).		3 V	1.7	-	-	-	-	-
		Wakeup clock HSI16 = 16 MHz, voltage Range 1. See (4).		3 V	1.62	-	-	-	-	-

1. Guaranteed by characterization results, unless otherwise specified.

2. LCD enabled with external voltage source. Consumption from VLCD excluded. Refer to LCD controller characteristics for I_{VLCD}.

3. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

4. Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in [Table 52: Low-power mode](#)



Table 47. Current consumption in Stop 0 mode

Symbol	Parameter	Conditions	TYP				
			25 °C	55 °C	85 °C	105 °C	125 °C
I _{DD,ALL} (Stop 0)	Supply current in Stop 0 mode, RTC disabled	V _{DD}					
		1.8 V	108	132	217	356	631
		2.4 V	110	134	219	358	634
		3 V	111	135	220	360	637
		3.6 V	113	137	222	363	642
							153
							158
							161
							166

- 1. Guaranteed by characterization results, unless otherwise specified.
- 2. Guaranteed by test in production.

Table 48. Current consumption in Standby mode

Symbol	Parameter	Conditions	TYP							
			V _{DD}	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C	
I _{DD_ALL} (Standby)	Supply current in Standby mode (backup registers retained), RTC disabled	-	1.8 V	114	355	1540	4146	10735	176	
		no independent watchdog	2.4 V	138	407	1795	4828	12451	223	
			3 V	150	486	2074	5589	14291	263	
			3.6 V	198	618	2608	6928	17499	383	
			with independent watchdog	1.8 V	317	-	-	-	-	-
		2.4 V		391	-	-	-	-	-	
		RTC clocked by LSI, no independent watchdog	3 V	438	-	-	-	-	-	
			3.6 V	566	-	-	-	-	-	
		Supply current in Standby mode (backup registers retained), RTC enabled	RTC clocked by LSI, with independent watchdog	1.8 V	377	621	1873	4564	11318	491
				2.4 V	464	756	2210	5348	13166	614
	3 V			572	913	2599	6219	15197	770	
	3.6 V			722	1144	3253	7724	18696	1012	
	I _{DD_ALL} (Standby with RTC)	Supply current in Standby mode (backup registers retained), RTC enabled	RTC clocked by LSE bypassed at 32768Hz	1.8 V	456	-	-	-	-	-
				2.4 V	557	-	-	-	-	-
3 V				663	-	-	-	-	-	
3.6 V				885	-	-	-	-	-	
RTC clocked by LSE quartz ⁽³⁾ in low drive mode		RTC clocked by LSE	1.8 V	289	527	1747	4402	11009	-	
			2.4 V	396	671	2108	5202	12869	-	
			3 V	528	853	2531	6095	14915	-	
			3.6 V	710	1111	3115	7470	18221	-	
			1.8 V	416	640	1862	4479	11908	-	
			2.4 V	514	796	2193	5236	13689	-	
RTC clocked by LSE quartz ⁽³⁾ in low drive mode	RTC clocked by LSE	3 V	652	961	2589	6103	15598	-		
		3.6 V	821	1226	3235	7551	17947	-		



Table 48. Current consumption in Standby mode (continued)

Symbol	Parameter	Conditions		TYP					
		-	V _{DD}	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD_ALL} (SRAM2) ⁽⁴⁾	Supply current to be added in Standby mode when SRAM2 is retained	-	1.8 V	235	641	2293	5192	11213	588
			2.4 V	237	645	2303	5213	11246	593
			3 V	236	647	2306	5221	11333	593
			3.6 V	235	646	2308	5200	11327	595
I _{DD_ALL} (wake-up from Standby)	Supply current during wake-up from Standby mode	Wake-up clock is MSI = 4 MHz. See ⁽⁵⁾ .	3 V	1.7	-	-	-	-	

1. Guaranteed by characterization results, unless otherwise specified.
2. Guaranteed by test in production.
3. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.
4. The supply current in Standby with SRAM2 mode is: I_{DD_ALL} (Standby) + I_{DD_ALL} (SRAM2). The supply current in Standby with RTC mode is: I_{DD_ALL} (Standby) + I_{DD_ALL} (RTC).
5. Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in [Table 52: Low-power mode](#).

Table 49. Current consumption in Shutdown mode

Symbol	Parameter	Conditions		TYP					
		-	V _{DD}	25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD_ALL} (Shutdown)	Supply current in Shutdown mode (backup registers retained) RTC disabled	-	1.8 V	29.8	194	1110	3250	9093	75
			2.4 V	44.3	237	1310	3798	10473	111
			3 V	64.1	293	1554	4461	12082	160
			3.6 V	112	420	2041	5689	15186	280

Table 49. Current consumption in Shutdown mode (continued)

Symbol	Parameter	Conditions	V _{DD}	TYP					
				25 °C	55 °C	85 °C	105 °C	125 °C	25 °C
I _{DD_ALL} (Shutdown with RTC)	Supply current in Shutdown mode (backup registers retained) RTC enabled	-	1.8 V	210	378	1299	3437	9357	-
			2.4 V	303	499	1577	4056	10825	-
		RTC clocked by LSE bypassed at 32768 Hz	3 V	422	655	1925	4820	12569	-
			3.6 V	584	888	2511	6158	15706	-
		RTC clocked by LSE quartz ⁽²⁾ in low drive mode	1.8 V	329	499	1408	3460	-	-
			2.4 V	431	634	1688	4064	-	-
I _{DD_ALL} (wake-up from Shutdown)	Supply current during wake-up from Shutdown mode	Wake-up clock is MSI = 4 MHz. See ⁽³⁾ .	3 V	554	791	2025	4795	-	-
			3.6 V	729	1040	2619	6129	-	-
			3 V	0.6	-	-	-	-	-

1. Guaranteed by characterization results, unless otherwise specified.

2. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

3. Wakeup with code execution from Flash. Average value given for a typical wakeup time as specified in [Table 52: Low-power mode](#)



Table 50. Current consumption in VBAT mode

Symbol	Parameter	Conditions	TYP					25 °C	125 °C	25 °C
			V _{BAT}	25 °C	55 °C	85 °C	105 °C			
I _{DD_VBAT}	Backup domain supply current	RTC disabled	1.8 V	4	29	196	587	1663	10.8	
			2.4 V	5.27	36	226	673	1884	13.2	
			3 V	6	42	264	775	2147	15.5	
			3.6 V	10	58	323	919	2488	25.8	
		RTC enabled and clocked by LSE bypassed at 32768 Hz	1.8 V	183	201	367	729	-	-	
			2.4 V	268	295	486	901	-	-	
			3 V	376	412	602	1075	-	-	
			3.6 V	508	558	752	1299	-	-	
		RTC enabled and clocked by LSE quartz ⁽²⁾	1.8 V	302	344	521	915	1978	-	
			2.4 V	388	436	639	1091	2289	-	
			3 V	494	549	784	1301	2656	-	
			3.6 V	630	692	971	1571	3115	-	

1. Guaranteed by characterization results, unless otherwise specified.

2. Based on characterization done with a 32.768 kHz crystal (MC306-G-06Q-32.768, manufacturer JFVNY) with two 6.8 pF loading capacitors.

I/O system current consumption

The current consumption of the I/O system has two components: static and dynamic.

I/O static current consumption

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in [Table 70: I/O static characteristics](#).

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

Caution: Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

I/O dynamic current consumption

In addition to the internal peripheral current consumption measured previously (see [Table 51: Peripheral current consumption](#)), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the I/O supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DDIOx} \times f_{SW} \times C$$

where

I_{SW} is the current sunk by a switching I/O to charge/discharge the capacitive load

V_{DDIOx} is the I/O supply voltage

f_{SW} is the I/O switching frequency

C is the total capacitance seen by the I/O pin: $C = C_{INT} + C_{EXT} + C_S$

C_S is the PCB board capacitance including the pad pin.

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

On-chip peripheral current consumption

The current consumption of the on-chip peripherals is given in [Table 51](#). The MCU is placed under the following conditions:

- All I/O pins are in Analog mode
- The given value is calculated by measuring the difference of the current consumptions:
 - when the peripheral is clocked on
 - when the peripheral is clocked off
- Ambient operating temperature and supply voltage conditions summarized in [Table 20: Voltage characteristics](#)
- The power consumption of the digital part of the on-chip peripherals is given in [Table 51](#). The power consumption of the analog part of the peripherals (where applicable) is indicated in each related section of the datasheet.

Table 51. Peripheral current consumption

Peripheral		Range 1	Range 2	Low-power run and sleep	Unit
AHB	Bus Matrix ⁽¹⁾	4.5	3.7	4.1	μA/MHz
	ADC independent clock domain	0.4	0.1	0.2	
	ADC AHB clock domain	5.5	4.7	5.5	
	CRC	0.4	0.2	0.3	
	DMA1	1.4	1.3	1.4	
	DMA2	1.5	1.3	1.4	
	FLASH	6.2	5.2	5.8	
	FMC	8.9	7.5	8.4	
	GPIOA ⁽²⁾	4.8	3.8	4.4	
	GPIOB ⁽²⁾	4.8	4.0	4.6	
	GPIOC ⁽²⁾	4.5	3.8	4.3	
	GPIOD ⁽²⁾	4.6	3.9	4.4	
	GPIOE ⁽²⁾	5.2	4.5	4.9	
	GPIOF ⁽²⁾	5.9	4.9	5.7	
	GPIOG ⁽²⁾	4.3	3.8	4.2	
	GPIOH ⁽²⁾	0.7	0.6	0.8	
	OTG_FS independent clock domain	23.2	N/A	N/A	
	OTG_FS AHB clock domain	16.4	N/A	N/A	
	QUADSPI	7.8	6.7	7.3	
	RNG independent clock domain	2.2	N/A	N/A	
	RNG AHB clock domain	0.6	N/A	N/A	
	SRAM1	0.9	0.8	0.9	

Table 51. Peripheral current consumption (continued)

Peripheral		Range 1	Range 2	Low-power run and sleep	Unit
AHB	SRAM2	1.6	1.4	1.6	μA/MHz
	TSC	1.8	1.4	1.6	
	All AHB Peripherals	118.5	77.3	87.6	
APB1	AHB to APB1 bridge ⁽³⁾	0.9	0.7	0.9	μA/MHz
	CAN1	4.6	4.0	4.4	
	DAC1	2.4	1.9	2.2	
	I2C1 independent clock domain	3.7	3.1	3.2	
	I2C1 APB clock domain	1.3	1.1	1.5	
	I2C2 independent clock domain	3.7	3.0	3.2	
	I2C2 APB clock domain	1.4	1.1	1.5	
	I2C3 independent clock domain	2.9	2.3	2.5	
	I2C3 APB clock domain	0.9	0.9	1.1	
	LCD	1.0	0.8	0.9	
	LPUART1 independent clock domain	2.1	1.6	2.0	
	LPUART1 APB clock domain	0.6	0.6	0.6	
	LPTIM1 independent clock domain	3.3	2.6	2.9	
	LPTIM1 APB clock domain	0.9	0.8	1.0	
	LPTIM2 independent clock domain	3.1	2.7	2.9	
	LPTIM2 APB clock domain	0.8	0.6	0.7	
	OPAMP	0.4	0.4	0.3	
	PWR	0.5	0.5	0.4	
	SPI2	1.8	1.6	1.6	
	SPI3	2.1	1.7	1.8	
	SWPMI1 independent clock domain	2.3	1.8	2.2	
	SWPMI1 APB clock domain	1.1	1.1	1.0	
	TIM2	6.8	5.7	6.3	
	TIM3	5.4	4.6	5.0	
	TIM4	5.2	4.4	4.9	
	TIM5	6.5	5.5	6.1	
	TIM6	1.1	1.0	1.0	
	TIM7	1.1	0.9	1.0	

Table 51. Peripheral current consumption (continued)

Peripheral		Range 1	Range 2	Low-power run and sleep	Unit
APB1	USART2 independent clock domain	4.1	3.6	3.8	μA/MHz
	USART2 APB clock domain	1.4	1.1	1.5	
	USART3 independent clock domain	4.7	4.1	4.2	
	USART3 APB clock domain	1.5	1.3	1.7	
	UART4 independent clock domain	3.9	3.2	3.5	
	UART4 APB clock domain	1.5	1.3	1.6	
	UART5 independent clock domain	3.9	3.2	3.5	
	UART5 APB clock domain	1.3	1.2	1.4	
	WWDG	0.5	0.5	0.5	
	All APB1 on	84.2	70.7	80.2	
APB2	AHB to APB2 bridge ⁽⁴⁾	1.0	0.9	0.9	
	DFSDM1	5.6	4.6	5.3	
	FW	0.7	0.5	0.7	
	SAI1 independent clock domain	2.6	2.1	2.3	
	SAI1 APB clock domain	2.1	1.8	2.0	
	SAI2 independent clock domain	3.3	2.7	3.0	
	SAI2 APB clock domain	2.4	2.1	2.2	
	SDMMC1 independent clock domain	4.7	3.9	4.2	
	SDMMC1 APB clock domain	2.5	1.9	2.1	
	SPI1	2.0	1.6	1.9	
	SYSCFG/VREFBUF/COMP	0.6	0.4	0.5	
	TIM1	8.3	6.9	7.9	
	TIM8	8.6	7.1	8.1	
	TIM15	4.1	3.4	3.9	
	TIM16	3.0	2.5	2.9	
	TIM17	3.0	2.4	2.9	
	USART1 independent clock domain	4.9	4.0	4.4	
	USART1 APB clock domain	1.5	1.3	1.7	
	All APB2 on	56.8	43.3	48.2	
ALL		256.8	189.6	215.5	

1. The BusMatrix is automatically active when at least one master is ON (CPU, DMA).
2. The GPIOx (x= A...H) dynamic current consumption is approximately divided by a factor two versus this table values when the GPIO port is locked thanks to LCKK and LCKy bits in the GPIOx_LCKR register. In order to save the full GPIOx current consumption, the GPIOx clock should be disabled in the RCC when all port I/Os are used in alternate function or analog mode (clock is only required to read or write into GPIO registers, and is not used in AF or analog modes).
3. The AHB to APB1 Bridge is automatically active when at least one peripheral is ON on the APB1.
4. The AHB to APB2 Bridge is automatically active when at least one peripheral is ON on the APB2.

The consumption for the peripherals when using SMPS can be found using STM32CubeMX PCC tool.

6.3.6 Wakeup time from low-power modes and voltage scaling transition times

The wakeup times given in [Table 52](#) are the latency between the event and the execution of the first user instruction.

The device goes in low-power mode after the WFE (Wait For Event) instruction.

Table 52. Low-power mode wakeup timings⁽¹⁾

Symbol	Parameter	Conditions		Typ	Max	Unit
$t_{WUSLEEP}$	Wakeup time from Sleep mode to Run mode	-		6	6	Nb of CPU cycles
$t_{WULPSLEEP}$	Wakeup time from Low-power sleep mode to Low-power run mode	Wakeup in Flash with Flash in power-down during low-power sleep mode (SLEEP_PD=1 in FLASH_ACR) and with clock MSI = 2 MHz		6	9.3	
$t_{WUSTOP0}$	Wake up time from Stop 0 mode to Run mode in Flash	Range 1	Wakeup clock MSI = 48 MHz	5.6	10.9	μs
			Wakeup clock HSI16 = 16 MHz	4.7	10.4	
		Range 2	Wakeup clock MSI = 24 MHz	5.7	11.1	
			Wakeup clock HSI16 = 16 MHz	4.5	10.5	
			Wakeup clock MSI = 4 MHz	6.6	14.2	
			Wakeup clock MSI = 4 MHz	6.6	14.2	
	Wake up time from Stop 0 mode to Run mode in SRAM1	Range 1	Wakeup clock MSI = 48 MHz	0.7	2.05	
			Wakeup clock HSI16 = 16 MHz	1.7	2.8	
		Range 2	Wakeup clock MSI = 24 MHz	0.8	2.72	
			Wakeup clock HSI16 = 16 MHz	1.7	2.8	
			Wakeup clock MSI = 4 MHz	2.4	11.32	
			Wakeup clock MSI = 4 MHz	2.4	11.32	

Table 52. Low-power mode wakeup timings⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Typ	Max	Unit
t _{WUSTOP1}	Wake up time from Stop 1 mode to Run mode in Flash	Range 1	Wakeup clock MSI = 48 MHz	6.2	10.2	μs
			Wakeup clock HSI16 = 16 MHz	6.3	8.99	
		Range 2	Wakeup clock MSI = 24 MHz	6.3	10.46	
			Wakeup clock HSI16 = 16 MHz	6.3	8.87	
			Wakeup clock MSI = 4 MHz	8.0	13.23	
	Wake up time from Stop 1 mode to Run mode in SRAM1	Range 1	Wakeup clock MSI = 48 MHz	4.5	5.78	
			Wakeup clock HSI16 = 16 MHz	5.5	7.1	
		Range 2	Wakeup clock MSI = 24 MHz	5.0	6.5	
			Wakeup clock HSI16 = 16 MHz	5.5	7.1	
			Wakeup clock MSI = 4 MHz	8.2	13.5	
	Wake up time from Stop 1 mode to Low-power run mode in Flash	Regulator in low-power mode (LPR=1 in PWR_CR1)	Wakeup clock MSI = 2 MHz	12.7	20	
	Wake up time from Stop 1 mode to Low-power run mode in SRAM1			10.7	21.5	
t _{WUSTOP2}	Wake up time from Stop 2 mode to Run mode in Flash	Range 1	Wakeup clock MSI = 48 MHz	8.0	9.4	μs
			Wakeup clock HSI16 = 16 MHz	7.3	9.3	
		Range 2	Wakeup clock MSI = 24 MHz	8.2	9.9	
			Wakeup clock HSI16 = 16 MHz	7.3	9.3	
			Wakeup clock MSI = 4 MHz	10.6	15.8	
	Wake up time from Stop 2 mode to Run mode in SRAM1	Range 1	Wakeup clock MSI = 48 MHz	5.1	6.7	
			Wakeup clock HSI16 = 16 MHz	5.7	8	
		Range 2	Wakeup clock MSI = 24 MHz	5.5	6.65	
			Wakeup clock HSI16 = 16 MHz	5.7	7.53	
			Wakeup clock MSI = 4 MHz	8.2	16.6	
t _{WUSTBY}	Wakeup time from Standby mode to Run mode	Range 1	Wakeup clock MSI = 8 MHz	14.3	20.8	μs
			Wakeup clock MSI = 4 MHz	20.1	35.5	
t _{WUSTBY} SRAM2	Wakeup time from Standby with SRAM2 to Run mode	Range 1	Wakeup clock MSI = 8 MHz	14.3	24.3	μs
			Wakeup clock MSI = 4 MHz	20.1	38.5	
t _{WUSHDN}	Wakeup time from Shutdown mode to Run mode	Range 1	Wakeup clock MSI = 4 MHz	256	330.6	μs

1. Guaranteed by characterization results.

Table 53. Regulator modes transition times⁽¹⁾

Symbol	Parameter	Conditions	Typ	Max	Unit
$t_{WULPRUN}$	Wakeup time from Low-power run mode to Run mode ⁽²⁾	Code run with MSI 2 MHz	5	7	μs
t_{VOST}	Regulator transition time from Range 2 to Range 1 or Range 1 to Range 2 ⁽³⁾	Code run with MSI 24 MHz	20	40	

1. Guaranteed by characterization results.

2. Time until REGLPF flag is cleared in PWR_SR2.

3. Time until VOSF flag is cleared in PWR_SR2.

Table 54. Wakeup time using USART/LPUART⁽¹⁾

Symbol	Parameter	Conditions	Typ	Max	Unit
$t_{WUUSART}$ $t_{WULPUART}$	Wakeup time needed to calculate the maximum USART/LPUART baudrate allowing to wakeup up from stop mode when USART/LPUART clock source is HSI16	Stop 0 mode	-	1.7	μs
		Stop 1 mode and Stop 2 mode	-	8.5	

1. Guaranteed by design.

6.3.7 External clock source characteristics

High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard GPIO.

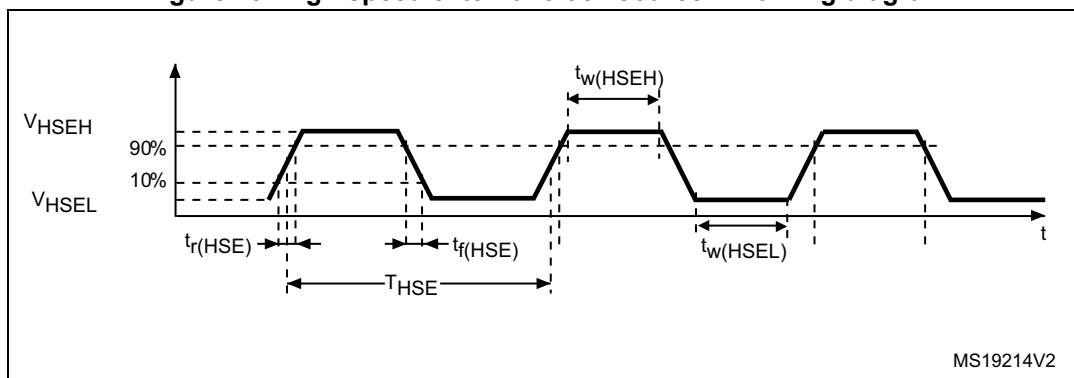
The external clock signal has to respect the I/O characteristics in [Section 6.3.14](#). However, the recommended clock input waveform is shown in [Figure 23: High-speed external clock source AC timing diagram](#).

Table 55. High-speed external user clock characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{HSE_ext}	User external clock source frequency	Voltage scaling Range 1	-	8	48	MHz
		Voltage scaling Range 2	-	8	26	
V_{HSEH}	OSC_IN input pin high level voltage	-	$0.7 V_{DDIOx}$	-	V_{DDIOx}	V
V_{HSEL}	OSC_IN input pin low level voltage	-	V_{SS}	-	$0.3 V_{DDIOx}$	
$t_{w(HSEH)}$ $t_{w(HSEL)}$	OSC_IN high or low time	Voltage scaling Range 1	7	-	-	ns
		Voltage scaling Range 2	18	-	-	

1. Guaranteed by design.

Figure 23. High-speed external clock source AC timing diagram



Low-speed external user clock generated from an external source

In bypass mode the LSE oscillator is switched off and the input pin is a standard GPIO.

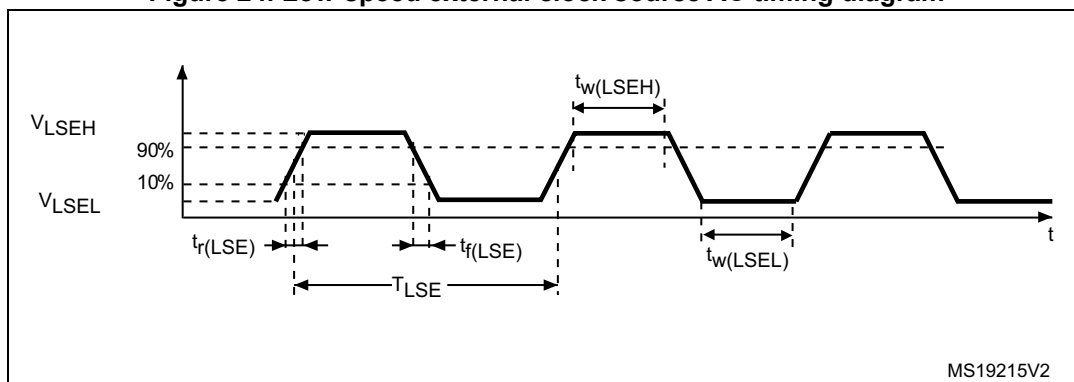
The external clock signal has to respect the I/O characteristics in [Section 6.3.14](#). However, the recommended clock input waveform is shown in [Figure 24](#).

Table 56. Low-speed external user clock characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{LSE_ext}	User external clock source frequency	-	-	32.768	1000	kHz
V_{LSEH}	OSC32_IN input pin high level voltage	-	$0.7 V_{DDIOx}$	-	V_{DDIOx}	V
V_{LSEL}	OSC32_IN input pin low level voltage	-	V_{SS}	-	$0.3 V_{DDIOx}$	
$t_{w(LSEH)}$ $t_{w(LSEL)}$	OSC32_IN high or low time	-	250	-	-	ns

1. Guaranteed by design.

Figure 24. Low-speed external clock source AC timing diagram



High-speed external clock generated from a crystal/ceramic resonator

The high-speed external (HSE) clock can be supplied with a 4 to 48 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in [Table 57](#). In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 57. HSE oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions ⁽²⁾	Min	Typ	Max	Unit
f_{OSC_IN}	Oscillator frequency	-	4	8	48	MHz
R_F	Feedback resistor	-	-	200	-	k Ω
$I_{DD(HSE)}$	HSE current consumption	During startup ⁽³⁾	-	-	5.5	mA
		$V_{DD} = 3\text{ V}$, $R_m = 30\ \Omega$, $CL = 10\text{ pF}@8\text{ MHz}$	-	0.44	-	
		$V_{DD} = 3\text{ V}$, $R_m = 45\ \Omega$, $CL = 10\text{ pF}@8\text{ MHz}$	-	0.45	-	
		$V_{DD} = 3\text{ V}$, $R_m = 30\ \Omega$, $CL = 5\text{ pF}@48\text{ MHz}$	-	0.68	-	
		$V_{DD} = 3\text{ V}$, $R_m = 30\ \Omega$, $CL = 10\text{ pF}@48\text{ MHz}$	-	0.94	-	
		$V_{DD} = 3\text{ V}$, $R_m = 30\ \Omega$, $CL = 20\text{ pF}@48\text{ MHz}$	-	1.77	-	
G_m	Maximum critical crystal transconductance	Startup	-	-	1.5	mA/V
$t_{SU(HSE)}^{(4)}$	Startup time	V_{DD} is stabilized	-	2	-	ms

1. Guaranteed by design.

2. Resonator characteristics given by the crystal/ceramic resonator manufacturer.

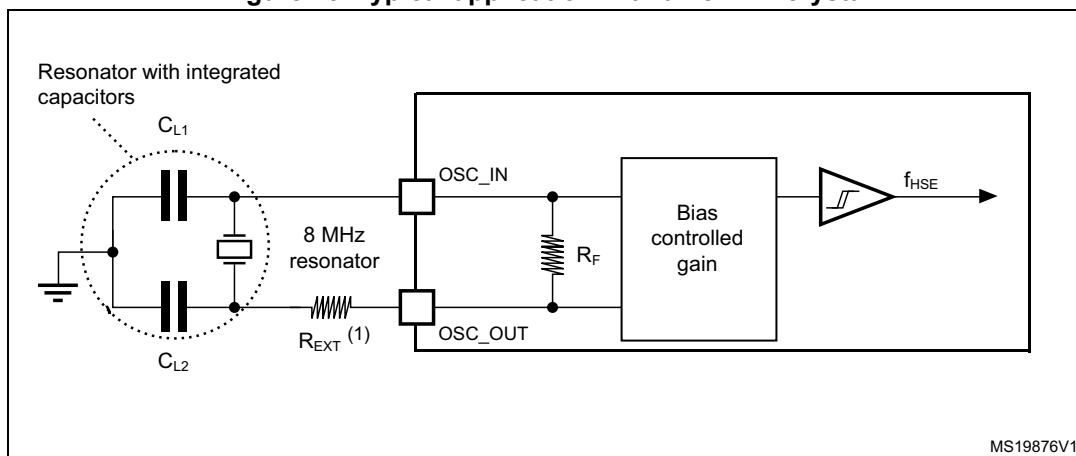
3. This consumption level occurs during the first 2/3 of the $t_{SU(HSE)}$ startup time

4. $t_{SU(HSE)}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer

For C_{L1} and C_{L2} , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 20 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see [Figure 25](#)). C_{L1} and C_{L2} are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of C_{L1} and C_{L2} . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing C_{L1} and C_{L2} .

Note: For information on selecting the crystal, refer to the application note AN2867 “Oscillator design guide for ST microcontrollers” available from the ST website www.st.com.

Figure 25. Typical application with an 8 MHz crystal



1. R_{EXT} value depends on the crystal characteristics.

Low-speed external clock generated from a crystal resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal resonator oscillator. All the information given in this paragraph are based on design simulation results obtained with typical external components specified in [Table 58](#). In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

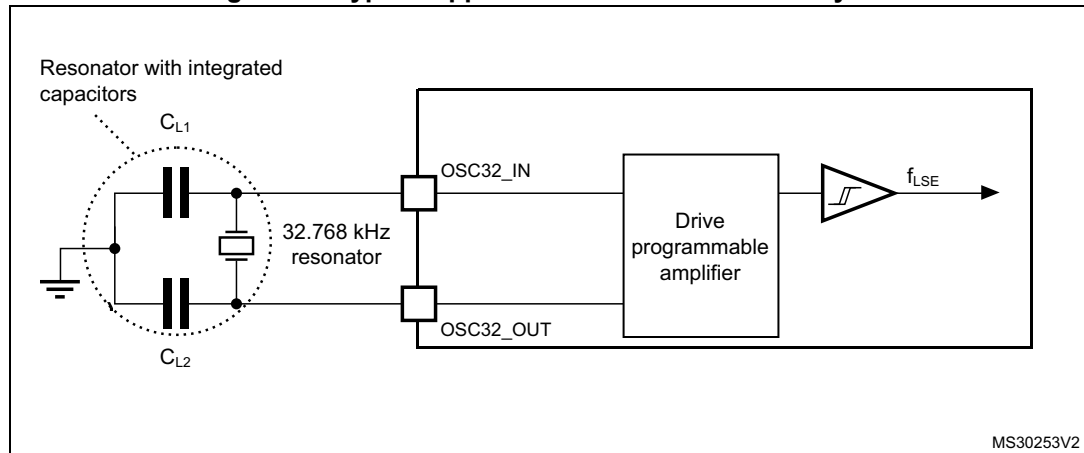
Table 58. LSE oscillator characteristics ($f_{LSE} = 32.768$ kHz)⁽¹⁾

Symbol	Parameter	Conditions ⁽²⁾	Min	Typ	Max	Unit
$I_{DD(LSE)}$	LSE current consumption	LSEDRV[1:0] = 00 Low drive capability	-	250	-	nA
		LSEDRV[1:0] = 01 Medium low drive capability	-	315	-	
		LSEDRV[1:0] = 10 Medium high drive capability	-	500	-	
		LSEDRV[1:0] = 11 High drive capability	-	630	-	
$G_{m_{critmax}}$	Maximum critical crystal gm	LSEDRV[1:0] = 00 Low drive capability	-	-	0.5	$\mu A/V$
		LSEDRV[1:0] = 01 Medium low drive capability	-	-	0.75	
		LSEDRV[1:0] = 10 Medium high drive capability	-	-	1.7	
		LSEDRV[1:0] = 11 High drive capability	-	-	2.7	
$t_{SU(LSE)}^{(3)}$	Startup time	V_{DD} is stabilized	-	2	-	s

1. Guaranteed by design.
2. Refer to the note and caution paragraphs below the table, and to the application note AN2867 "Oscillator design guide for ST microcontrollers".
3. $t_{SU(LSE)}$ is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is measured for a standard crystal and it can vary significantly with the crystal manufacturer

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website www.st.com.

Figure 26. Typical application with a 32.768 kHz crystal



Note: An external resistor is not required between $OSC32_IN$ and $OSC32_OUT$ and it is forbidden to add one.

6.3.8 Internal clock source characteristics

The parameters given in [Table 59](#) are derived from tests performed under ambient temperature and supply voltage conditions summarized in [Table 23: General operating conditions](#). The provided curves are characterization results, not tested in production.

High-speed internal (HSI16) RC oscillator

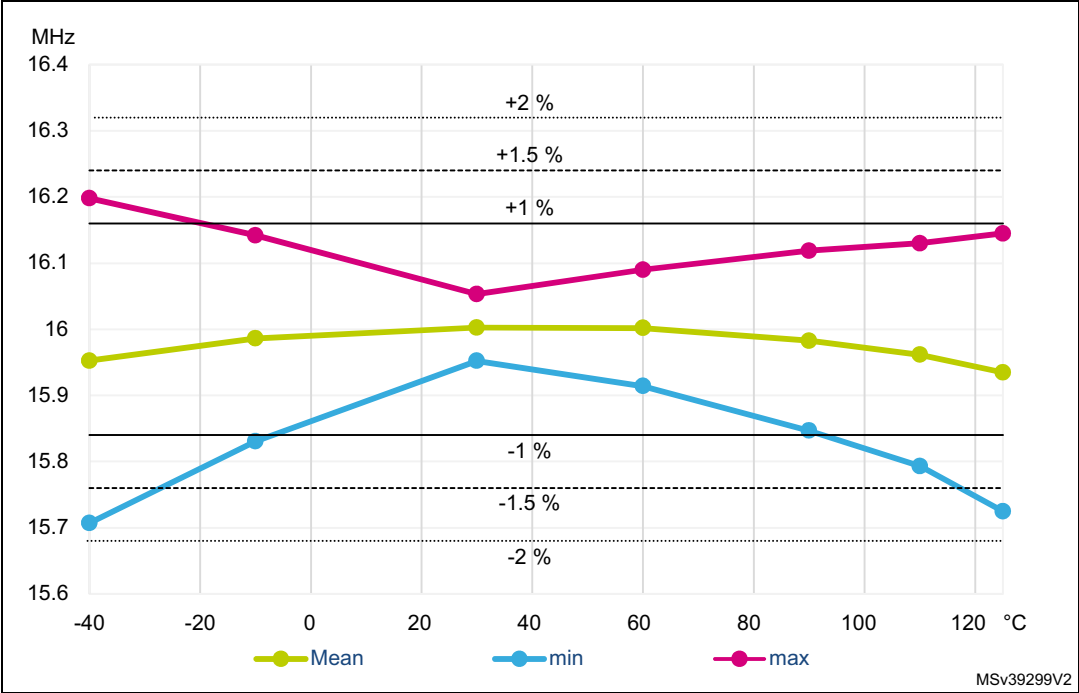
Table 59. HSI16 oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{HSI16}	HSI16 Frequency	$V_{\text{DD}}=3.0\text{ V}$, $T_{\text{A}}=30\text{ }^{\circ}\text{C}$	15.88	-	16.08	MHz
TRIM	HSI16 user trimming step	Trimming code is not a multiple of 64	0.2	0.3	0.4	%
		Trimming code is a multiple of 64	-4	-6	-8	
$\text{DuCy}(\text{HSI16})^{(2)}$	Duty Cycle	-	45	-	55	%
$\Delta_{\text{Temp}}(\text{HSI16})$	HSI16 oscillator frequency drift over temperature	$T_{\text{A}}=0\text{ to }85\text{ }^{\circ}\text{C}$	-1	-	1	%
		$T_{\text{A}}=-40\text{ to }125\text{ }^{\circ}\text{C}$	-2	-	1.5	%
$\Delta_{\text{VDD}}(\text{HSI16})$	HSI16 oscillator frequency drift over V_{DD}	$V_{\text{DD}}=1.62\text{ V to }3.6\text{ V}$	-0.1	-	0.05	%
$t_{\text{su}}(\text{HSI16})^{(2)}$	HSI16 oscillator start-up time	-	-	0.8	1.2	μs
$t_{\text{stab}}(\text{HSI16})^{(2)}$	HSI16 oscillator stabilization time	-	-	3	5	μs
$I_{\text{DD}}(\text{HSI16})^{(2)}$	HSI16 oscillator power consumption	-	-	155	190	μA

1. Guaranteed by characterization results.

2. Guaranteed by design.

Figure 27. HSI16 frequency versus temperature



Multi-speed internal (MSI) RC oscillator

Table 60. MSI oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
f_{MSI}	MSI frequency after factory calibration, done at $V_{\text{DD}}=3\text{ V}$ and $T_{\text{A}}=30\text{ }^{\circ}\text{C}$	MSI mode	Range 0	99	100	101	kHz
			Range 1	198	200	202	
			Range 2	396	400	404	
			Range 3	792	800	808	
			Range 4	0.99	1	1.01	MHz
			Range 5	1.98	2	2.02	
			Range 6	3.96	4	4.04	
			Range 7	7.92	8	8.08	
			Range 8	15.8	16	16.16	
			Range 9	23.8	24	24.4	
			Range 10	31.7	32	32.32	
			Range 11	47.5	48	48.48	
		PLL mode XTAL= 32.768 kHz	Range 0	-	98.304	-	kHz
			Range 1	-	196.608	-	
			Range 2	-	393.216	-	
			Range 3	-	786.432	-	
			Range 4	-	1.016	-	MHz
			Range 5	-	1.999	-	
			Range 6	-	3.998	-	
			Range 7	-	7.995	-	
			Range 8	-	15.991	-	
			Range 9	-	23.986	-	
			Range 10	-	32.014	-	
			Range 11	-	48.005	-	
$\Delta_{\text{TEMP}}(\text{MSI})^{(2)}$	MSI oscillator frequency drift over temperature	MSI mode	$T_{\text{A}} = -0\text{ to }85\text{ }^{\circ}\text{C}$	-3.5	-	3	%
			$T_{\text{A}} = -40\text{ to }125\text{ }^{\circ}\text{C}$	-8	-	6	

Table 60. MSI oscillator characteristics⁽¹⁾ (continued)

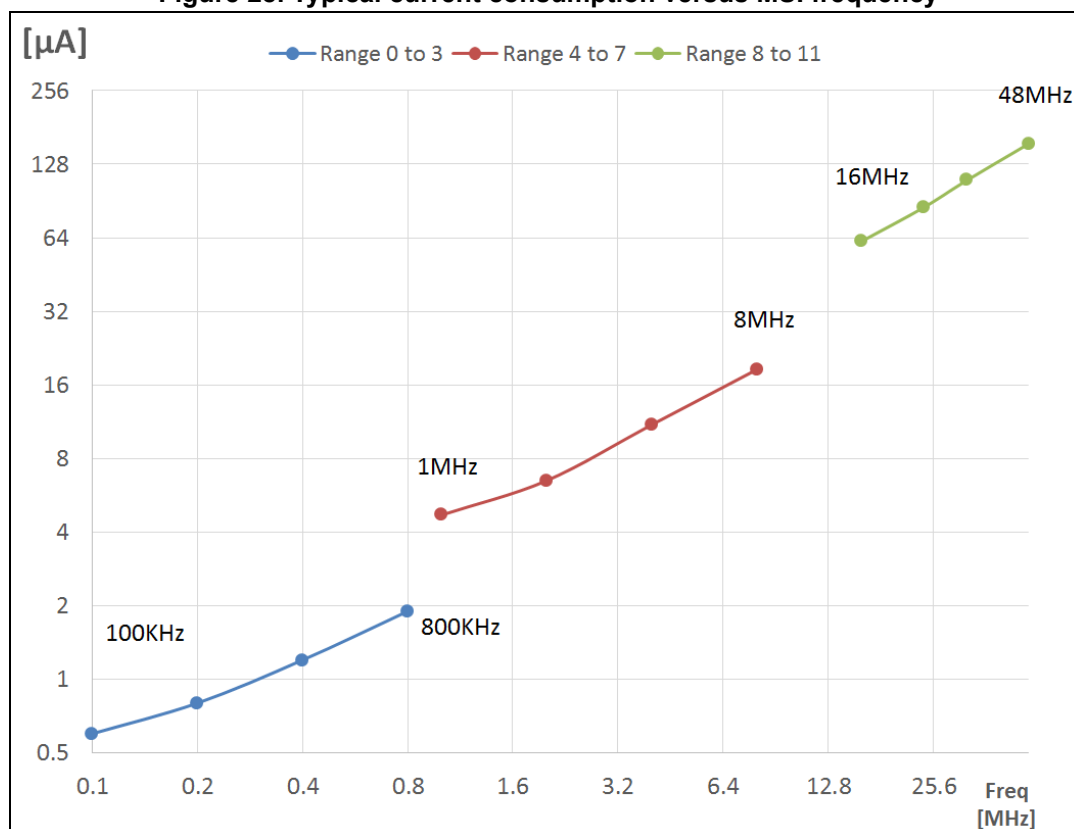
Symbol	Parameter	Conditions			Min	Typ	Max	Unit	
$\Delta V_{DD}(MSI)^{(2)}$	MSI oscillator frequency drift over V_{DD} (reference is 3 V)	MSI mode	Range 0 to 3	$V_{DD}=1.62$ V to 3.6 V	-1.2	-	0.5	%	
				$V_{DD}=2.4$ V to 3.6 V	-0.5	-			
			Range 4 to 7	$V_{DD}=1.62$ V to 3.6 V	-2.5	-	0.7		
				$V_{DD}=2.4$ V to 3.6 V	-0.8	-			
			Range 8 to 11	$V_{DD}=1.62$ V to 3.6 V	-5	-	1		
				$V_{DD}=2.4$ V to 3.6 V	-1.6	-			
$\Delta F_{SAMPLING}(MSI)^{(2)(6)}$	Frequency variation in sampling mode ⁽³⁾	MSI mode	$T_A= -40$ to 85 °C		-	1	2	%	
			$T_A= -40$ to 125 °C		-	2	4		
P_USB Jitter(MSI) ⁽⁶⁾	Period jitter for USB clock ⁽⁴⁾	PLL mode Range 11	for next transition	-	-	-	3.458	ns	
			for paired transition	-	-	-	3.916		
MT_USB Jitter(MSI) ⁽⁶⁾	Medium term jitter for USB clock ⁽⁵⁾	PLL mode Range 11	for next transition	-	-	-	2	ns	
			for paired transition	-	-	-	1		
CC jitter(MSI) ⁽⁶⁾	RMS cycle-to-cycle jitter	PLL mode Range 11			-	-	60	-	ps
P jitter(MSI) ⁽⁶⁾	RMS Period jitter	PLL mode Range 11			-	-	50	-	ps
$t_{SU}(MSI)^{(6)}$	MSI oscillator start-up time	Range 0		-	-	10	20	us	
		Range 1		-	-	5	10		
		Range 2		-	-	4	8		
		Range 3		-	-	3	7		
		Range 4 to 7		-	-	3	6		
		Range 8 to 11		-	-	2.5	6		
$t_{STAB}(MSI)^{(6)}$	MSI oscillator stabilization time	PLL mode Range 11	10 % of final frequency	-	-	0.25	0.5	ms	
			5 % of final frequency	-	-	0.5	1.25		
			1 % of final frequency	-	-	-	2.5		

Table 60. MSI oscillator characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$I_{DD}(MSI)^{(6)}$	MSI oscillator power consumption	MSI and PLL mode	Range 0	-	-	0.6	1
			Range 1	-	-	0.8	1.2
			Range 2	-	-	1.2	1.7
			Range 3	-	-	1.9	2.5
			Range 4	-	-	4.7	6
			Range 5	-	-	6.5	9
			Range 6	-	-	11	15
			Range 7	-	-	18.5	25
			Range 8	-	-	62	80
			Range 9	-	-	85	110
			Range 10	-	-	110	130
			Range 11	-	-	155	190

1. Guaranteed by characterization results.
2. This is a deviation for an individual part once the initial frequency has been measured.
3. Sampling mode means Low-power run/Low-power sleep modes with Temperature sensor disable.
4. Average period of MSI @48 MHz is compared to a real 48 MHz clock over 28 cycles. It includes frequency tolerance + jitter of MSI @48 MHz clock.
5. Only accumulated jitter of MSI @48 MHz is extracted over 28 cycles.
For next transition: min. and max. jitter of 2 consecutive frame of 28 cycles of the MSI @48 MHz, for 1000 captures over 28 cycles.
For paired transitions: min. and max. jitter of 2 consecutive frame of 56 cycles of the MSI @48 MHz, for 1000 captures over 56 cycles.
6. Guaranteed by design.

Figure 28. Typical current consumption versus MSI frequency



Low-speed internal (LSI) RC oscillator

Table 61. LSI oscillator characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{LSI}	LSI Frequency	$V_{\text{DD}} = 3.0 \text{ V}$, $T_{\text{A}} = 30 \text{ }^{\circ}\text{C}$	31.04	-	32.96	kHz
		$V_{\text{DD}} = 1.62 \text{ to } 3.6 \text{ V}$, $T_{\text{A}} = -40 \text{ to } 125 \text{ }^{\circ}\text{C}$	29.5	-	34	
$t_{\text{SU}}(\text{LSI})^{(2)}$	LSI oscillator start-up time	-	-	80	130	µs
$t_{\text{STAB}}(\text{LSI})^{(2)}$	LSI oscillator stabilization time	5% of final frequency	-	125	180	µs
$I_{\text{DD}}(\text{LSI})^{(2)}$	LSI oscillator power consumption	-	-	110	180	nA

1. Guaranteed by characterization results.

2. Guaranteed by design.

6.3.9 PLL characteristics

The parameters given in [Table 62](#) are derived from tests performed under temperature and V_{DD} supply voltage conditions summarized in [Table 23: General operating conditions](#).

Table 62. PLL, PLLSAI1, PLLSAI2 characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{PLL_IN}	PLL input clock ⁽²⁾	-	4	-	16	MHz
	PLL input clock duty cycle	-	45	-	55	%
$f_{PLL_P_OUT}$	PLL multiplier output clock P	Voltage scaling Range 1	2.0645	-	80	MHz
		Voltage scaling Range 2	2.0645	-	26	
$f_{PLL_Q_OUT}$	PLL multiplier output clock Q	Voltage scaling Range 1	8	-	80	MHz
		Voltage scaling Range 2	8	-	26	
$f_{PLL_R_OUT}$	PLL multiplier output clock R	Voltage scaling Range 1	8	-	80	MHz
		Voltage scaling Range 2	8	-	26	
f_{VCO_OUT}	PLL VCO output	Voltage scaling Range 1	64	-	344	MHz
		Voltage scaling Range 2	64	-	128	
t_{LOCK}	PLL lock time	-	-	15	40	μ s
Jitter	RMS cycle-to-cycle jitter	System clock 80 MHz	-	40	-	$\pm$ ps
	RMS period jitter		-	30	-	
$I_{DD}(PLL)$	PLL power consumption on V_{DD} ⁽¹⁾	VCO freq = 64 MHz	-	150	200	μ A
		VCO freq = 96 MHz	-	200	260	
		VCO freq = 192 MHz	-	300	380	
		VCO freq = 344 MHz	-	520	650	

1. Guaranteed by design.

2. Take care of using the appropriate division factor M to obtain the specified PLL input clock values. The M factor is shared between the 3 PLLs.

6.3.10 Flash memory characteristics

Table 63. Flash memory characteristics⁽¹⁾

Symbol	Parameter	Conditions	Typ	Max	Unit
t_{prog}	64-bit programming time	-	81.69	90.76	μ s
t_{prog_row}	one row (32 double word) programming time	normal programming	2.61	2.90	ms
		fast programming	1.91	2.12	
t_{prog_page}	one page (2 Kbyte) programming time	normal programming	20.91	23.24	
		fast programming	15.29	16.98	
t_{ERASE}	Page (2 KB) erase time	-	22.02	24.47	
t_{prog_bank}	one bank (512 Kbyte) programming time	normal programming	5.35	5.95	s
		fast programming	3.91	4.35	
t_{ME}	Mass erase time (one or two banks)	-	22.13	24.59	ms

Table 63. Flash memory characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Typ	Max	Unit
I_{DD}	Average consumption from V_{DD}	Write mode	3.4	-	mA
		Erase mode	3.4	-	
	Maximum current (peak)	Write mode	7 (for 2 μ s)	-	
		Erase mode	7 (for 41 μ s)	-	

1. Guaranteed by design.

Table 64. Flash memory endurance and data retention

Symbol	Parameter	Conditions	Min ⁽¹⁾	Unit
N_{END}	Endurance	$T_A = -40$ to $+105$ °C	10	kcycles
t_{RET}	Data retention	1 kcycle ⁽²⁾ at $T_A = 85$ °C	30	Years
		1 kcycle ⁽²⁾ at $T_A = 105$ °C	15	
		1 kcycle ⁽²⁾ at $T_A = 125$ °C	7	
		10 kcycles ⁽²⁾ at $T_A = 55$ °C	30	
		10 kcycles ⁽²⁾ at $T_A = 85$ °C	15	
		10 kcycles ⁽²⁾ at $T_A = 105$ °C	10	

1. Guaranteed by characterization results.

2. Cycling performed over the whole temperature range.

6.3.11 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports), the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- **Electrostatic discharge (ESD)** (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- **FTB**: A Burst of Fast Transient voltage (positive and negative) is applied to V_{DD} and V_{SS} through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in [Table 65](#). They are based on the EMS levels and classes defined in application note AN1709.

Table 65. EMS characteristics

Symbol	Parameter	Conditions	Level/Class
V_{FESD}	Voltage limits to be applied on any I/O pin to induce a functional disturbance	$V_{DD} = 3.3\text{ V}$, $T_A = +25\text{ }^{\circ}\text{C}$, $f_{HCLK} = 80\text{ MHz}$, conforming to IEC 61000-4-2	3B
V_{EFTB}	Fast transient voltage burst limits to be applied through 100 pF on V_{DD} and V_{SS} pins to induce a functional disturbance	$V_{DD} = 3.3\text{ V}$, $T_A = +25\text{ }^{\circ}\text{C}$, $f_{HCLK} = 80\text{ MHz}$, conforming to IEC 61000-4-4	4A

Designing hardened software to avoid noise problems

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

Electromagnetic Interference (EMI)

The electromagnetic field emitted by the device are monitored while a simple application is executed (toggling 2 LEDs through the I/O ports). This emission test is compliant with IEC 61967-2 standard which specifies the test board and the pin loading.

Table 66. EMI characteristics

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. [f _{HSE} /f _{HCLK}]		Unit
				f _{MSI} = 24 MHz	8 MHz / 80 MHz	
S _{EMI}	Peak level	V _{DD} = 3.6 V, T _A = 25 °C, LQFP144 package compliant with IEC 61967-2	0.1 MHz to 30 MHz	-9	2	dBμV
			30 MHz to 130 MHz	-8	3	
			130 MHz to 1 GHz	-10	14	
			EMI Level	1.5	3.5	-

6.3.12 Electrical sensitivity characteristics

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the ANSI/JEDEC standard.

Table 67. ESD absolute maximum ratings

Symbol	Ratings	Conditions	Class	Maximum value ⁽¹⁾	Unit
V _{ESD(HBM)}	Electrostatic discharge voltage (human body model)	T _A = +25 °C, conforming to ANSI/ESDA/JEDEC JS-001	2	2000	V
V _{ESD(CDM)}	Electrostatic discharge voltage (charge device model)	T _A = +25 °C, conforming to ANSI/ESD STM5.3.1	C3	250	

1. Guaranteed by characterization results.

Static latch-up

Two complementary static tests are required on six parts to assess the latch-up performance:

- A supply overvoltage is applied to each power supply pin.
- A current injection is applied to each input, output and configurable I/O pin.

These tests are compliant with EIA/JESD 78A IC latch-up standard.

Table 68. Electrical sensitivities

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	$T_A = +105\text{ }^{\circ}\text{C}$ conforming to JESD78A	II level A ⁽¹⁾

1. Negative injection is limited to -30 mA for PF0, PF1, PG6, PG7, PG8, PG12, PG13, PG14.

6.3.13 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below V_{SS} or above V_{DDIOx} (for standard, 3.3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (higher than 5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of the -5 μA /+0 μA range) or other functional failure (for example reset occurrence or oscillator frequency deviation).

The characterization results are given in [Table 69](#).

Negative induced leakage current is caused by negative injection and positive induced leakage current is caused by positive injection.

Table 69. I/O current injection susceptibility

Symbol	Description	Functional susceptibility		Unit
		Negative injection	Positive injection	
I_{INJ}	Injected current on BOOT0 pin	-0	0	mA
	Injected current on pins except PA4, PA5, BOOT0	-5	N/A ⁽¹⁾	
	Injected current on PA4, PA5 pins	-5	0	

1. Injection is not possible.

6.3.14 I/O port characteristics

General input/output characteristics

Unless otherwise specified, the parameters given in [Table 70](#) are derived from tests performed under the conditions summarized in [Table 23: General operating conditions](#). All I/Os are designed as CMOS- and TTL-compliant (except BOOT0).

Table 70. I/O static characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IL}^{(1)}$	I/O input low level voltage except BOOT0	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	-	-	$0.3 \times V_{DDIOx}^{(2)}$	V
	I/O input low level voltage except BOOT0	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	-	-	$0.39 \times V_{DDIOx} - 0.06^{(3)}$	
	I/O input low level voltage except BOOT0	$1.08\text{ V} < V_{DDIOx} < 1.62\text{ V}$	-	-	$0.43 \times V_{DDIOx} - 0.1^{(3)}$	
	BOOT0 I/O input low level voltage	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	-	-	$0.17 \times V_{DDIOx}^{(3)}$	
$V_{IH}^{(1)}$	I/O input high level voltage except BOOT0	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	$0.7 \times V_{DDIOx}^{(2)}$	-	-	V
	I/O input high level voltage except BOOT0	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	$0.49 \times V_{DDIOx} + 0.26^{(3)}$	-	-	
	I/O input high level voltage except BOOT0	$1.08\text{ V} < V_{DDIOx} < 1.62\text{ V}$	$0.61 \times V_{DDIOx} + 0.05^{(3)}$	-	-	
	BOOT0 I/O input high level voltage	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	$0.77 \times V_{DDIOx}^{(3)}$	-	-	
$V_{hys}^{(3)}$	TT_xx, FT_xxx and NRST I/O input hysteresis	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	-	200	-	mV
	FT_sx	$1.08\text{ V} < V_{DDIOx} < 1.62\text{ V}$	-	150	-	
	BOOT0 I/O input hysteresis	$1.62\text{ V} < V_{DDIOx} < 3.6\text{ V}$	-	200	-	

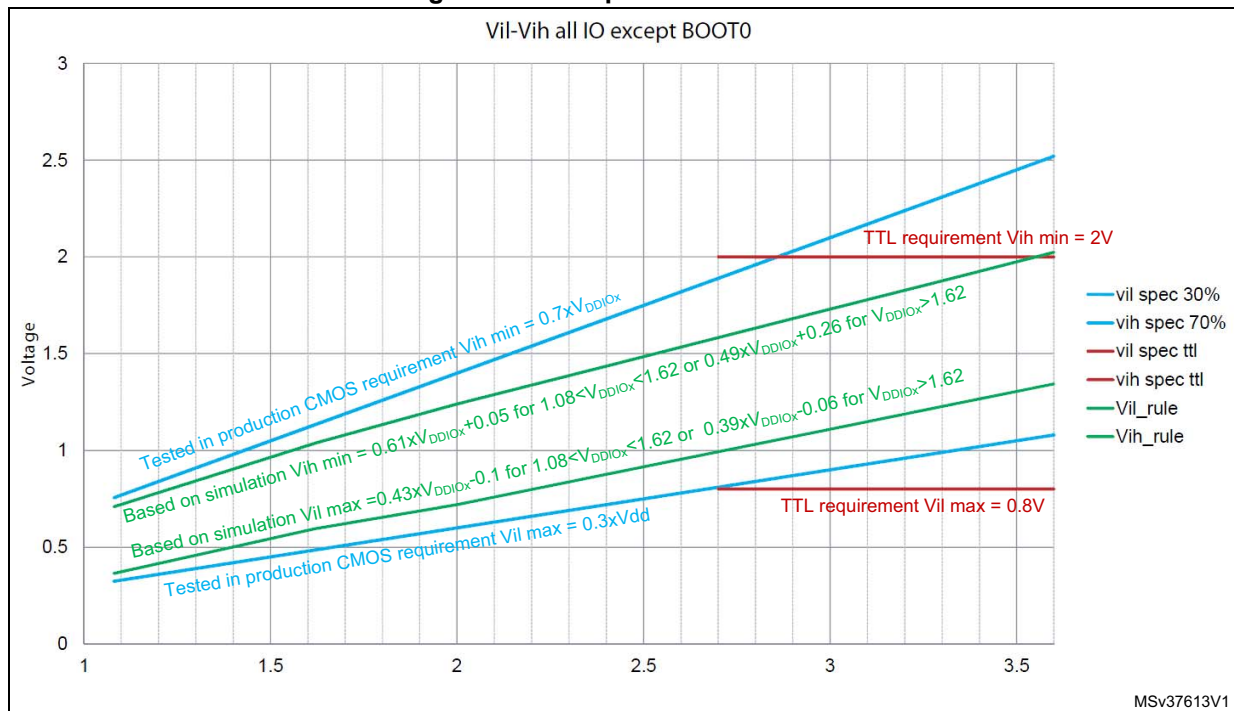
Table 70. I/O static characteristics (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{lkg}^{(4)}$	FT_xx input leakage current ⁽³⁾⁽⁵⁾	$V_{IN} \leq \text{Max}(V_{DDXXX})^{(6)(7)}$	-	-	±100	nA
		$\text{Max}(V_{DDXXX}) \leq V_{IN} \leq \text{Max}(V_{DDXXX})+1 \text{ V}^{(6)(7)}$	-	-	650	
		$\text{Max}(V_{DDXXX})+1 \text{ V} < V_{IN} \leq 5.5 \text{ V}^{(6)(7)}$	-	-	200	
	FT_lu, FT_u and PC3 I/Os	$V_{IN} \leq \text{Max}(V_{DDXXX})^{(6)(7)}$	-	-	±150	
		$\text{Max}(V_{DDXXX}) \leq V_{IN} \leq \text{Max}(V_{DDXXX})+1 \text{ V}^{(6)(7)}$	-	-	2500 ⁽³⁾	
		$\text{Max}(V_{DDXXX})+1 \text{ V} < V_{IN} \leq 5.5 \text{ V}^{(6)(7)}$	-	-	250	
	TT_xx input leakage current	$V_{IN} \leq \text{Max}(V_{DDXXX})^{(6)}$	-	-	±150	
		$\text{Max}(V_{DDXXX}) \leq V_{IN} < 3.6 \text{ V}^{(6)}$	-	-	2000 ⁽³⁾	
	OPAMPx_VINM (x=1,2) dedicated input leakage current (UFBGA132 and UFBGA144 only)	-	-	-	(8)	
R_{PU}	Weak pull-up equivalent resistor ⁽⁹⁾	$V_{IN} = V_{SS}$	25	40	55	kΩ
R_{PD}	Weak pull-down equivalent resistor ⁽⁹⁾	$V_{IN} = V_{DDIOx}$	25	40	55	kΩ
C_{IO}	I/O pin capacitance	-	-	5	-	pF

1. Refer to [Figure 29: I/O input characteristics](#).
2. Guaranteed by test in production.
3. Guaranteed by design.
4. This value represents the pad leakage of the IO itself. The total product pad leakage is provided by this formula:
 $I_{Total_leak_max} = 10 \mu A + [\text{number of IOs where } V_{IN} \text{ is applied on the pad}] \times I_{lkg}(\text{Max})$.
5. All FT_xx GPIOs except FT_lu, FT_u and PC3 I/Os.
6. $\text{Max}(V_{DDXXX})$ is the maximum value of all the I/O supplies.
7. To sustain a voltage higher than $\text{Min}(V_{DD}, V_{DDA}, V_{DDIO2}, V_{DDUSB}, V_{LCD}) + 0.3 \text{ V}$, the internal Pull-up and Pull-Down resistors must be disabled.
8. Refer to I_{bias} in [Table 86: OPAMP characteristics](#) for the values of the OPAMP dedicated input leakage current.
9. Pull-up and pull-down resistors are designed with a true resistance in series with a switchable PMOS/NMOS. This PMOS/NMOS contribution to the series resistance is minimal (~10% order).

All I/Os are CMOS- and TTL-compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements is shown in [Figure 29](#) for standard I/Os, and in [Figure 29](#) for 5 V tolerant I/Os.

Figure 29. I/O input characteristics



Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to ± 8 mA, and sink or source up to ± 20 mA (with a relaxed V_{OL}/V_{OH}).

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in [Section 6.2](#):

- The sum of the currents sourced by all the I/Os on V_{DDIOX} , plus the maximum consumption of the MCU sourced on V_{DD} , cannot exceed the absolute maximum rating ΣI_{VDD} (see [Table 20: Voltage characteristics](#)).
- The sum of the currents sunk by all the I/Os on V_{SS} , plus the maximum consumption of the MCU sunk on V_{SS} , cannot exceed the absolute maximum rating ΣI_{VSS} (see [Table 20: Voltage characteristics](#)).

Output voltage levels

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 23: General operating conditions](#). All I/Os are CMOS- and TTL-compliant (FT OR TT unless otherwise specified).

Table 71. Output voltage characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
V_{OL}	Output low level voltage for an I/O pin	CMOS port ⁽²⁾ $ I_{IO} = 8 \text{ mA}$ $V_{DDIOx} \geq 2.7 \text{ V}$	-	0.4	V
V_{OH}	Output high level voltage for an I/O pin		$V_{DDIOx} - 0.4$	-	
$V_{OL}^{(3)}$	Output low level voltage for an I/O pin	TTL port ⁽²⁾ $ I_{IO} = 8 \text{ mA}$ $V_{DDIOx} \geq 2.7 \text{ V}$	-	0.4	
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		2.4	-	
$V_{OL}^{(3)}$	Output low level voltage for an I/O pin	$ I_{IO} = 20 \text{ mA}$ $V_{DDIOx} \geq 2.7 \text{ V}$	-	1.3	
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$V_{DDIOx} - 1.3$	-	
$V_{OL}^{(3)}$	Output low level voltage for an I/O pin	$ I_{IO} = 4 \text{ mA}$ $V_{DDIOx} \geq 1.62 \text{ V}$	-	0.45	
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$V_{DDIOx} - 0.45$	-	
$V_{OL}^{(3)}$	Output low level voltage for an I/O pin	$ I_{IO} = 2 \text{ mA}$ $1.62 \text{ V} \geq V_{DDIOx} \geq 1.08 \text{ V}$	-	$0.35 \times V_{DDIOx}$	
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$0.65 \times V_{DDIOx}$	-	
$V_{OLFM+}^{(3)}$	Output low level voltage for an FT I/O pin in FM+ mode (FT I/O with "f" option)	$ I_{IO} = 20 \text{ mA}$ $V_{DDIOx} \geq 2.7 \text{ V}$	-	0.4	
		$ I_{IO} = 10 \text{ mA}$ $V_{DDIOx} \geq 1.62 \text{ V}$	-	0.4	
		$ I_{IO} = 2 \text{ mA}$ $1.62 \text{ V} \geq V_{DDIOx} \geq 1.08 \text{ V}$	-	0.4	

1. The I_{IO} current sourced or sunk by the device must always respect the absolute maximum rating specified in [Table 20: Voltage characteristics](#), and the sum of the currents sourced or sunk by all the I/Os (I/O ports and control pins) must always respect the absolute maximum ratings ΣI_{IO} .
2. TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.
3. Guaranteed by design.

Input/output AC characteristics

The definition and values of input/output AC characteristics are given in [Figure 30](#) and [Table 72](#), respectively.

Unless otherwise specified, the parameters given are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 23: General operating conditions](#).

Table 72. I/O AC characteristics⁽¹⁾⁽²⁾

Speed	Symbol	Parameter	Conditions	Min	Max	Unit
00	Fmax	Maximum frequency	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	5	MHz
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	1	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	0.1	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	10	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	1.5	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	0.1	
	Tr/Tf	Output rise and fall time	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	25	ns
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	52	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	140	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	17	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	37	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	110	
01	Fmax	Maximum frequency	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	25	MHz
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	10	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	1	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	50	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	15	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	1	
	Tr/Tf	Output rise and fall time	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	9	ns
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	16	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	40	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	4.5	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	9	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	21	

Table 72. I/O AC characteristics⁽¹⁾⁽²⁾ (continued)

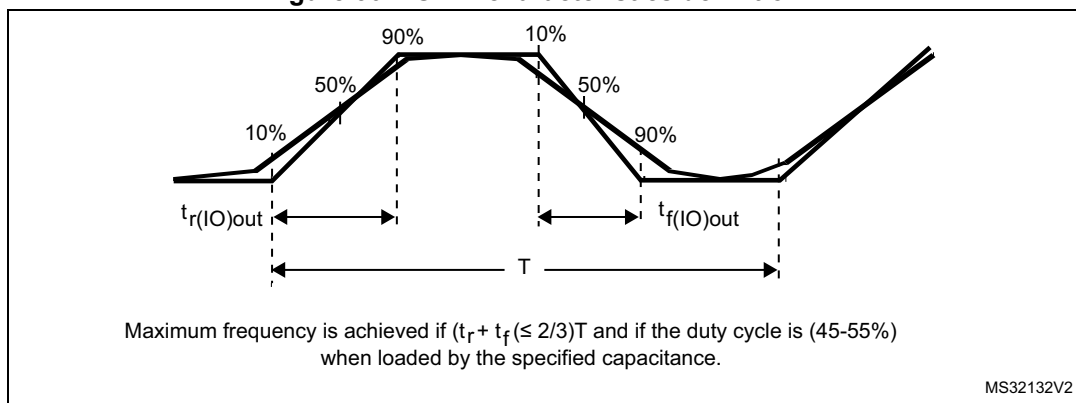
Speed	Symbol	Parameter	Conditions	Min	Max	Unit
10	Fmax	Maximum frequency	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	50	MHz
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	25	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	5	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	100 ⁽³⁾	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	37.5	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	5	
	Tr/Tf	Output rise and fall time	C=50 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	5.8	ns
			C=50 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	11	
			C=50 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	28	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	2.5	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	5	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	12	
11	Fmax	Maximum frequency	C=30 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	120 ⁽³⁾	MHz
			C=30 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	50	
			C=30 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	10	
			C=10 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	180 ⁽³⁾	
			C=10 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	75	
			C=10 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	10	
	Tr/Tf	Output rise and fall time	C=30 pF, 2.7 V ≤ V _{DDIOx} ≤ 3.6 V	-	3.3	ns
			C=30 pF, 1.62 V ≤ V _{DDIOx} ≤ 2.7 V	-	6	
			C=30 pF, 1.08 V ≤ V _{DDIOx} ≤ 1.62 V	-	16	
Fm+	Fmax	Maximum frequency	C=50 pF, 1.6 V ≤ V _{DDIOx} ≤ 3.6 V	-	1	MHz
	Tf	Output fall time ⁽⁴⁾		-	5	ns

1. The I/O speed is configured using the OSPEEDRy[1:0] bits. The Fm+ mode is configured in the SYSCFG_CFGR1 register. Refer to the RM0351 reference manual for a description of GPIO Port configuration register.

2. Guaranteed by design.

3. This value represents the I/O capability but the maximum system frequency is limited to 80 MHz.

4. The fall time is defined between 70% and 30% of the output waveform accordingly to I²C specification.

Figure 30. I/O AC characteristics definition⁽¹⁾

1. Refer to [Table 72: I/O AC characteristics](#).

6.3.15 NRST pin characteristics

The NRST pin input driver uses the CMOS technology. It is connected to a permanent pull-up resistor, R_{PU} .

Unless otherwise specified, the parameters given in the table below are derived from tests performed under the ambient temperature and supply voltage conditions summarized in [Table 23: General operating conditions](#).

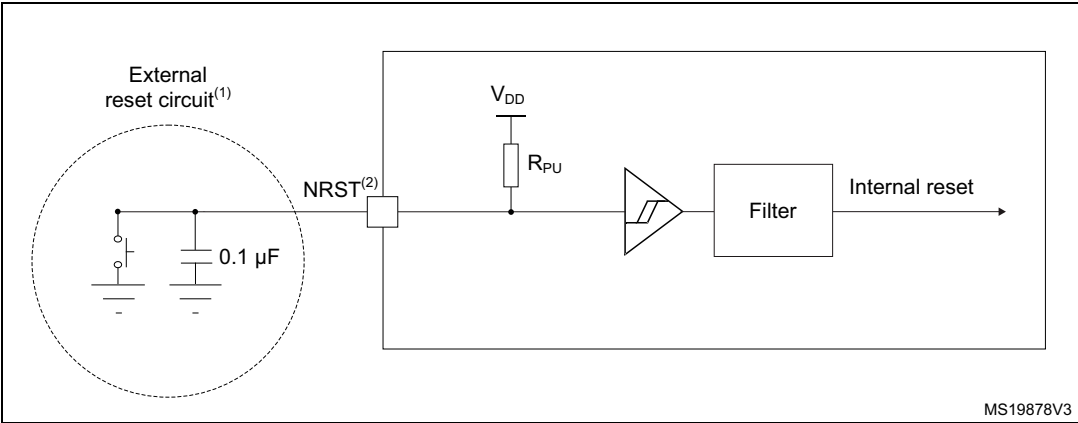
Table 73. NRST pin characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{IL(NRST)}$	NRST input low level voltage	-	-	-	$0.3 \times V_{DDIOx}$	V
$V_{IH(NRST)}$	NRST input high level voltage	-	$0.7 \times V_{DDIOx}$	-	-	
$V_{hys(NRST)}$	NRST Schmitt trigger voltage hysteresis	-	-	200	-	mV
R_{PU}	Weak pull-up equivalent resistor ⁽²⁾	$V_{IN} = V_{SS}$	25	40	55	k Ω
$V_{F(NRST)}$	NRST input filtered pulse	-	-	-	70	ns
$V_{NF(NRST)}$	NRST input not filtered pulse	$1.71 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	350	-	-	ns

1. Guaranteed by design.

2. The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimal (~10% order).

Figure 31. Recommended NRST pin protection



1. The reset network protects the device against parasitic resets.
2. The user must ensure that the level on the NRST pin can go below the $V_{IL(NRST)}$ max level specified in [Table 73: NRST pin characteristics](#). Otherwise the reset will not be taken into account by the device.
3. The external capacitor on NRST must be placed as close as possible to the device.

6.3.16 Extended interrupt and event controller input (EXTI) characteristics

The pulse on the interrupt input must have a minimal length in order to guarantee that it is detected by the event controller.

Table 74. EXTI input characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
PLEC	Pulse length to event controller	-	20	-	-	ns

1. Guaranteed by design.

6.3.17 Analog switches booster

Table 75. Analog switches booster characteristics⁽¹⁾

Symbol	Parameter	Min	Typ	Max	Unit
V_{DD}	Supply voltage	1.62	-	3.6	V
$t_{SU(BOOST)}$	Booster startup time	-	-	240	µs
$I_{DD(BOOST)}$	Booster consumption for $1.62\text{ V} \leq V_{DD} \leq 2.0\text{ V}$	-	-	250	µA
	Booster consumption for $2.0\text{ V} \leq V_{DD} \leq 2.7\text{ V}$	-	-	500	
	Booster consumption for $2.7\text{ V} \leq V_{DD} \leq 3.6\text{ V}$	-	-	900	

1. Guaranteed by design.

6.3.18 Analog-to-Digital converter characteristics

Unless otherwise specified, the parameters given in [Table 76](#) are preliminary values derived from tests performed under ambient temperature, f_{PCLK} frequency and V_{DDA} supply voltage conditions summarized in [Table 23: General operating conditions](#).

Note: *It is recommended to perform a calibration after each power-up.*

Table 76. ADC characteristics^{(1) (2)}

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{DDA}	Analog supply voltage	-	1.62	-	3.6	V
$V_{\text{REF+}}$	Positive reference voltage	$V_{\text{DDA}} \geq 2 \text{ V}$	2	-	V_{DDA}	V
		$V_{\text{DDA}} < 2 \text{ V}$	V_{DDA}			V
$V_{\text{REF-}}$	Negative reference voltage	-	V_{SSA}			V
f_{ADC}	ADC clock frequency	Range 1	0.14	-	80	MHz
		Range 2	0.14	-	26	
f_{s}	Sampling rate for FAST channels	Resolution = 12 bits	-	-	5.33	MSPS
		Resolution = 10 bits	-	-	6.15	
		Resolution = 8 bits	-	-	7.27	
		Resolution = 6 bits	-	-	8.88	
	Sampling rate for SLOW channels	Resolution = 12 bits	-	-	4.21	
		Resolution = 10 bits	-	-	4.71	
		Resolution = 8 bits	-	-	5.33	
		Resolution = 6 bits	-	-	6.15	
f_{TRIG}	External trigger frequency	$f_{\text{ADC}} = 80 \text{ MHz}$ Resolution = 12 bits	-	-	5.33	MHz
		Resolution = 12 bits	-	-	15	$1/f_{\text{ADC}}$
$V_{\text{AIN}}^{(3)}$	Conversion voltage range(2)	-	0	-	$V_{\text{REF+}}$	V
R_{AIN}	External input impedance	-	-	-	50	k Ω
C_{ADC}	Internal sample and hold capacitor	-	-	5	-	pF
t_{STAB}	Power-up time	-	1			conversion cycle
t_{CAL}	Calibration time	$f_{\text{ADC}} = 80 \text{ MHz}$	1.45			μs
		-	116			$1/f_{\text{ADC}}$
t_{LATR}	Trigger conversion latency Regular and injected channels without conversion abort	CKMODE = 00	1.5	2	2.5	$1/f_{\text{ADC}}$
		CKMODE = 01	-	-	2.0	
		CKMODE = 10	-	-	2.25	
		CKMODE = 11	-	-	2.125	

Table 76. ADC characteristics^{(1) (2)} (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
t_{LATINJ}	Trigger conversion latency Injected channels aborting a regular conversion	CKMODE = 00	2.5	3	3.5	$1/f_{ADC}$
		CKMODE = 01	-	-	3.0	
		CKMODE = 10	-	-	3.25	
		CKMODE = 11	-	-	3.125	
t_s	Sampling time	$f_{ADC} = 80 \text{ MHz}$	0.03125	-	8.00625	μs
		-	2.5	-	640.5	$1/f_{ADC}$
t_{ADCVRG_STUP}	ADC voltage regulator start-up time	-	-	-	20	μs
t_{CONV}	Total conversion time (including sampling time)	$f_{ADC} = 80 \text{ MHz}$ Resolution = 12 bits	0.1875	-	8.1625	μs
		Resolution = 12 bits	$t_s + 12.5 \text{ cycles for successive approximation} = 15 \text{ to } 653$			$1/f_{ADC}$
$I_{DDA}(ADC)$	ADC consumption from the V_{DDA} supply	$f_s = 5 \text{ Msps}$	-	730	830	μA
		$f_s = 1 \text{ Msps}$	-	160	220	
		$f_s = 10 \text{ ksps}$	-	16	50	
$I_{DDV_S}(ADC)$	ADC consumption from the V_{REF+} single ended mode	$f_s = 5 \text{ Msps}$	-	130	160	μA
		$f_s = 1 \text{ Msps}$	-	30	40	
		$f_s = 10 \text{ ksps}$	-	0.6	2	
$I_{DDV_D}(ADC)$	ADC consumption from the V_{REF+} differential mode	$f_s = 5 \text{ Msps}$	-	260	310	μA
		$f_s = 1 \text{ Msps}$	-	60	70	
		$f_s = 10 \text{ ksps}$	-	1.3	3	

1. Guaranteed by design
2. The I/O analog switch voltage booster is enable when $V_{DDA} < 2.4 \text{ V}$ (BOOSTEN = 1 in the SYSCFG_CFGR1 when $V_{DDA} < 2.4 \text{ V}$). It is disable when $V_{DDA} \geq 2.4 \text{ V}$.
3. V_{REF+} can be internally connected to V_{DDA} and V_{REF-} can be internally connected to V_{SSA} , depending on the package. Refer to [Section 4: Pinouts and pin description](#) for further details.

The maximum value of R_{AIN} can be found in [Table 77: Maximum ADC RAIN](#).

Table 77. Maximum ADC $R_{AIN}^{(1)(2)}$

Resolution	Sampling cycle @80 MHz	Sampling time [ns] @80 MHz	R_{AIN} max (Ω)	
			Fast channels ⁽³⁾	Slow channels ⁽⁴⁾
12 bits	2.5	31.25	100	N/A
	6.5	81.25	330	100
	12.5	156.25	680	470
	24.5	306.25	1500	1200
	47.5	593.75	2200	1800
	92.5	1156.25	4700	3900
	247.5	3093.75	12000	10000
	640.5	8006.75	39000	33000
10 bits	2.5	31.25	120	N/A
	6.5	81.25	390	180
	12.5	156.25	820	560
	24.5	306.25	1500	1200
	47.5	593.75	2200	1800
	92.5	1156.25	5600	4700
	247.5	3093.75	12000	10000
	640.5	8006.75	47000	39000
8 bits	2.5	31.25	180	N/A
	6.5	81.25	470	270
	12.5	156.25	1000	680
	24.5	306.25	1800	1500
	47.5	593.75	2700	2200
	92.5	1156.25	6800	5600
	247.5	3093.75	15000	12000
	640.5	8006.75	50000	50000
6 bits	2.5	31.25	220	N/A
	6.5	81.25	560	330
	12.5	156.25	1200	1000
	24.5	306.25	2700	2200
	47.5	593.75	3900	3300
	92.5	1156.25	8200	6800
	247.5	3093.75	18000	15000
	640.5	8006.75	50000	50000

1. Guaranteed by design.

2. The I/O analog switch voltage booster is enable when $V_{DDA} < 2.4\text{ V}$ (BOOSTEN = 1 in the SYSCFG_CFGR1 when $V_{DDA} < 2.4\text{V}$). It is disable when $V_{DDA} \geq 2.4\text{ V}$.
3. Fast channels are: PC0, PC1, PC2, PC3, PA0.
4. Slow channels are: all ADC inputs except the fast channels.

Table 78. ADC accuracy - limited test conditions 1⁽¹⁾(2)(3)

Sym- bol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit
ET	Total unadjusted error	Single ended	Fast channel (max speed)	-	4	5	LSB
			Slow channel (max speed)	-	4	5	
		Differential	Fast channel (max speed)	-	3.5	4.5	
			Slow channel (max speed)	-	3.5	4.5	
EO	Offset error	Single ended	Fast channel (max speed)	-	1	2.5	
			Slow channel (max speed)	-	1	2.5	
		Differential	Fast channel (max speed)	-	1.5	2.5	
			Slow channel (max speed)	-	1.5	2.5	
EG	Gain error	Single ended	Fast channel (max speed)	-	2.5	4.5	
			Slow channel (max speed)	-	2.5	4.5	
		Differential	Fast channel (max speed)	-	2.5	3.5	
			Slow channel (max speed)	-	2.5	3.5	
ED	Differential linearity error	Single ended	Fast channel (max speed)	-	1	1.5	
			Slow channel (max speed)	-	1	1.5	
		Differential	Fast channel (max speed)	-	1	1.2	
			Slow channel (max speed)	-	1	1.2	
EL	Integral linearity error	Single ended	Fast channel (max speed)	-	1.5	2.5	
			Slow channel (max speed)	-	1.5	2.5	
		Differential	Fast channel (max speed)	-	1	2	
			Slow channel (max speed)	-	1	2	
ENOB	Effective number of bits	Single ended	Fast channel (max speed)	10.4	10.5	-	bits
			Slow channel (max speed)	10.4	10.5	-	
		Differential	Fast channel (max speed)	10.8	10.9	-	
			Slow channel (max speed)	10.8	10.9	-	
SINAD	Signal-to-noise and distortion ratio	Single ended	Fast channel (max speed)	64.4	65	-	dB
			Slow channel (max speed)	64.4	65	-	
		Differential	Fast channel (max speed)	66.8	67.4	-	
			Slow channel (max speed)	66.8	67.4	-	
SNR	Signal-to-noise ratio	Single ended	Fast channel (max speed)	65	66	-	
			Slow channel (max speed)	65	66	-	
		Differential	Fast channel (max speed)	67	68	-	
			Slow channel (max speed)	67	68	-	

Table 78. ADC accuracy - limited test conditions 1⁽¹⁾(2)(3) (continued)

Sym- bol	Parameter	Conditions ⁽⁴⁾			Min	Typ	Max	Unit
THD	Total harmonic distortion	ADC clock frequency $\leq$ 80 MHz, Sampling rate $\leq$ 5.33 Msps, $V_{DDA} = V_{REF+} = 3$ V, $T_A = 25$ °C	Single ended	Fast channel (max speed)	-	-74	-73	dB
				Slow channel (max speed)	-	-74	-73	
			Differential	Fast channel (max speed)	-	-79	-76	
				Slow channel (max speed)	-	-79	-76	

1. Guaranteed by design.
2. ADC DC accuracy values are measured after internal calibration.
3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
4. The I/O analog switch voltage booster is enable when $V_{DDA} < 2.4$ V (BOOSTEN = 1 in the SYSCFG_CFGR1 when $V_{DDA} < 2.4$ V). It is disable when $V_{DDA} \geq 2.4$ V. No oversampling.

Table 79. ADC accuracy - limited test conditions 2⁽¹⁾(2)(3)

Sym- bol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit
ET	Total unadjusted error	Single ended	Fast channel (max speed)	-	4	6.5	LSB
			Slow channel (max speed)	-	4	6.5	
		Differential	Fast channel (max speed)	-	3.5	5.5	
			Slow channel (max speed)	-	3.5	5.5	
EO	Offset error	Single ended	Fast channel (max speed)	-	1	4.5	
			Slow channel (max speed)	-	1	5	
		Differential	Fast channel (max speed)	-	1.5	3	
			Slow channel (max speed)	-	1.5	3	
EG	Gain error	Single ended	Fast channel (max speed)	-	2.5	6	
			Slow channel (max speed)	-	2.5	6	
		Differential	Fast channel (max speed)	-	2.5	3.5	
			Slow channel (max speed)	-	2.5	3.5	
ED	Differential linearity error	Single ended	Fast channel (max speed)	-	1	1.5	
			Slow channel (max speed)	-	1	1.5	
		Differential	Fast channel (max speed)	-	1	1.2	
			Slow channel (max speed)	-	1	1.2	
EL	Integral linearity error	Single ended	Fast channel (max speed)	-	1.5	3.5	
			Slow channel (max speed)	-	1.5	3.5	
		Differential	Fast channel (max speed)	-	1	3	
			Slow channel (max speed)	-	1	2.5	
ENOB	Effective number of bits	Single ended	Fast channel (max speed)	10	10.5	-	bits
			Slow channel (max speed)	10	10.5	-	
		Differential	Fast channel (max speed)	10.7	10.9	-	
			Slow channel (max speed)	10.7	10.9	-	
SINAD	Signal-to-noise and distortion ratio	Single ended	Fast channel (max speed)	62	65	-	dB
			Slow channel (max speed)	62	65	-	
		Differential	Fast channel (max speed)	66	67.4	-	
			Slow channel (max speed)	66	67.4	-	
SNR	Signal-to-noise ratio	Single ended	Fast channel (max speed)	64	66	-	
			Slow channel (max speed)	64	66	-	
		Differential	Fast channel (max speed)	66.5	68	-	
			Slow channel (max speed)	66.5	68	-	

Table 79. ADC accuracy - limited test conditions 2⁽¹⁾(2)(3) (continued)

Sym- bol	Parameter	Conditions ⁽⁴⁾			Min	Typ	Max	Unit
THD	Total harmonic distortion	ADC clock frequency $\leq$ 80 MHz, Sampling rate $\leq$ 5.33 Msps, $2\text{ V} \leq V_{DDA}$	Single ended	Fast channel (max speed)	-	-74	-65	dB
				Slow channel (max speed)	-	-74	-67	
			Differential	Fast channel (max speed)	-	-79	-70	
				Slow channel (max speed)	-	-79	-71	

1. Guaranteed by design.
2. ADC DC accuracy values are measured after internal calibration.
3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
4. The I/O analog switch voltage booster is enable when $V_{DDA} < 2.4\text{ V}$ (BOOSTEN = 1 in the SYSCFG_CFGR1 when $V_{DDA} < 2.4\text{ V}$). It is disable when $V_{DDA} \geq 2.4\text{ V}$. No oversampling.

Table 80. ADC accuracy - limited test conditions 3⁽¹⁾⁽²⁾⁽³⁾

Sym- bol	Parameter	Conditions ⁽⁴⁾			Min	Typ	Max	Unit
ET	Total unadjusted error	ADC clock frequency ≤ 80 MHz, Sampling rate ≤ 5.33 Msps, 1.65 V ≤ V _{DDA} = V _{REF+} ≤ 3.6 V, Voltage scaling Range 1	Single ended	Fast channel (max speed)	-	5.5	7.5	LSB
				Slow channel (max speed)	-	4.5	6.5	
			Differential	Fast channel (max speed)	-	4.5	7.5	
				Slow channel (max speed)	-	4.5	5.5	
EO	Offset error		Single ended	Fast channel (max speed)	-	2	5	
				Slow channel (max speed)	-	2.5	5	
			Differential	Fast channel (max speed)	-	2	3.5	
				Slow channel (max speed)	-	2.5	3	
EG	Gain error		Single ended	Fast channel (max speed)	-	4.5	7	
				Slow channel (max speed)	-	3.5	6	
			Differential	Fast channel (max speed)	-	3.5	4	
				Slow channel (max speed)	-	3.5	5	
ED	Differential linearity error	Single ended	Fast channel (max speed)	-	1.2	1.5		
			Slow channel (max speed)	-	1.2	1.5		
		Differential	Fast channel (max speed)	-	1	1.2		
			Slow channel (max speed)	-	1	1.2		
EL	Integral linearity error	Single ended	Fast channel (max speed)	-	3	3.5		
			Slow channel (max speed)	-	2.5	3.5		
		Differential	Fast channel (max speed)	-	2	2.5		
			Slow channel (max speed)	-	2	2.5		
ENOB	Effective number of bits	Single ended	Fast channel (max speed)	10	10.4	-	bits	
			Slow channel (max speed)	10	10.4	-		
		Differential	Fast channel (max speed)	10.6	10.7	-		
			Slow channel (max speed)	10.6	10.7	-		
SINAD	Signal-to-noise and distortion ratio	Single ended	Fast channel (max speed)	62	64	-	dB	
			Slow channel (max speed)	62	64	-		
		Differential	Fast channel (max speed)	65	66	-		
			Slow channel (max speed)	65	66	-		
SNR	Signal-to-noise ratio	Single ended	Fast channel (max speed)	63	65	-		
			Slow channel (max speed)	63	65	-		
		Differential	Fast channel (max speed)	66	67	-		
			Slow channel (max speed)	66	67	-		

Table 80. ADC accuracy - limited test conditions 3⁽¹⁾(2)(3) (continued)

Sym- bol	Parameter	Conditions ⁽⁴⁾			Min	Typ	Max	Unit
THD	Total harmonic distortion	ADC clock frequency $\leq$ 80 MHz, Sampling rate $\leq$ 5.33 Msps, $1.65\text{ V} \leq V_{\text{DDA}} = V_{\text{REF+}} \leq 3.6\text{ V}$, Voltage scaling Range 1	Single ended	Fast channel (max speed)	-	-69	-67	dB
				Slow channel (max speed)	-	-71	-67	
			Differential	Fast channel (max speed)	-	-72	-71	
				Slow channel (max speed)	-	-72	-71	

1. Guaranteed by design.
2. ADC DC accuracy values are measured after internal calibration.
3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
4. The I/O analog switch voltage booster is enable when $V_{\text{DDA}} < 2.4\text{ V}$ (BOOSTEN = 1 in the SYSCFG_CFGR1 when $V_{\text{DDA}} < 2.4\text{ V}$). It is disable when $V_{\text{DDA}} \geq 2.4\text{ V}$. No oversampling.

Table 81. ADC accuracy - limited test conditions 4⁽¹⁾⁽²⁾⁽³⁾

Sym- bol	Parameter	Conditions ⁽⁴⁾		Min	Typ	Max	Unit
ET	Total unadjusted error	Single ended	Fast channel (max speed)	-	5	5.4	LSB
			Slow channel (max speed)	-	4	5	
		Differential	Fast channel (max speed)	-	4	5	
			Slow channel (max speed)	-	3.5	4.5	
EO	Offset error	Single ended	Fast channel (max speed)	-	2	4	
			Slow channel (max speed)	-	2	4	
		Differential	Fast channel (max speed)	-	2	3.5	
			Slow channel (max speed)	-	2	3.5	
EG	Gain error	Single ended	Fast channel (max speed)	-	4	4.5	
			Slow channel (max speed)	-	4	4.5	
		Differential	Fast channel (max speed)	-	3	4	
			Slow channel (max speed)	-	3	4	
ED	Differential linearity error	Single ended	Fast channel (max speed)	-	1	1.5	
			Slow channel (max speed)	-	1	1.5	
		Differential	Fast channel (max speed)	-	1	1.2	
			Slow channel (max speed)	-	1	1.2	
EL	Integral linearity error	Single ended	Fast channel (max speed)	-	2.5	3	
			Slow channel (max speed)	-	2.5	3	
		Differential	Fast channel (max speed)	-	2	2.5	
			Slow channel (max speed)	-	2	2.5	
ENOB	Effective number of bits	Single ended	Fast channel (max speed)	10.2	10.5	-	bits
			Slow channel (max speed)	10.2	10.5	-	
		Differential	Fast channel (max speed)	10.6	10.7	-	
			Slow channel (max speed)	10.6	10.7	-	
SINAD	Signal-to-noise and distortion ratio	Single ended	Fast channel (max speed)	63	65	-	dB
			Slow channel (max speed)	63	65	-	
		Differential	Fast channel (max speed)	65	66	-	
			Slow channel (max speed)	65	66	-	
SNR	Signal-to-noise ratio	Single ended	Fast channel (max speed)	64	65	-	
			Slow channel (max speed)	64	65	-	
		Differential	Fast channel (max speed)	66	67	-	
			Slow channel (max speed)	66	67	-	

Table 81. ADC accuracy - limited test conditions 4⁽¹⁾(2)(3) (continued)

Sym- bol	Parameter	Conditions ⁽⁴⁾			Min	Typ	Max	Unit
THD	Total harmonic distortion	ADC clock frequency $\leq$ 26 MHz, $1.65\text{ V} \leq V_{DDA} = V_{REF+} \leq 3.6\text{ V}$, Voltage scaling Range 2	Single ended	Fast channel (max speed)	-	-71	-69	dB
				Slow channel (max speed)	-	-71	-69	
			Differential	Fast channel (max speed)	-	-73	-72	
				Slow channel (max speed)	-	-73	-72	

1. Guaranteed by design.
2. ADC DC accuracy values are measured after internal calibration.
3. ADC accuracy vs. negative Injection Current: Injecting negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative current.
4. The I/O analog switch voltage booster is enable when $V_{DDA} < 2.4\text{ V}$ (BOOSTEN = 1 in the SYSCFG_CFGR1 when $V_{DDA} < 2.4\text{ V}$). It is disable when $V_{DDA} \geq 2.4\text{ V}$. No oversampling.

Figure 32. ADC accuracy characteristics

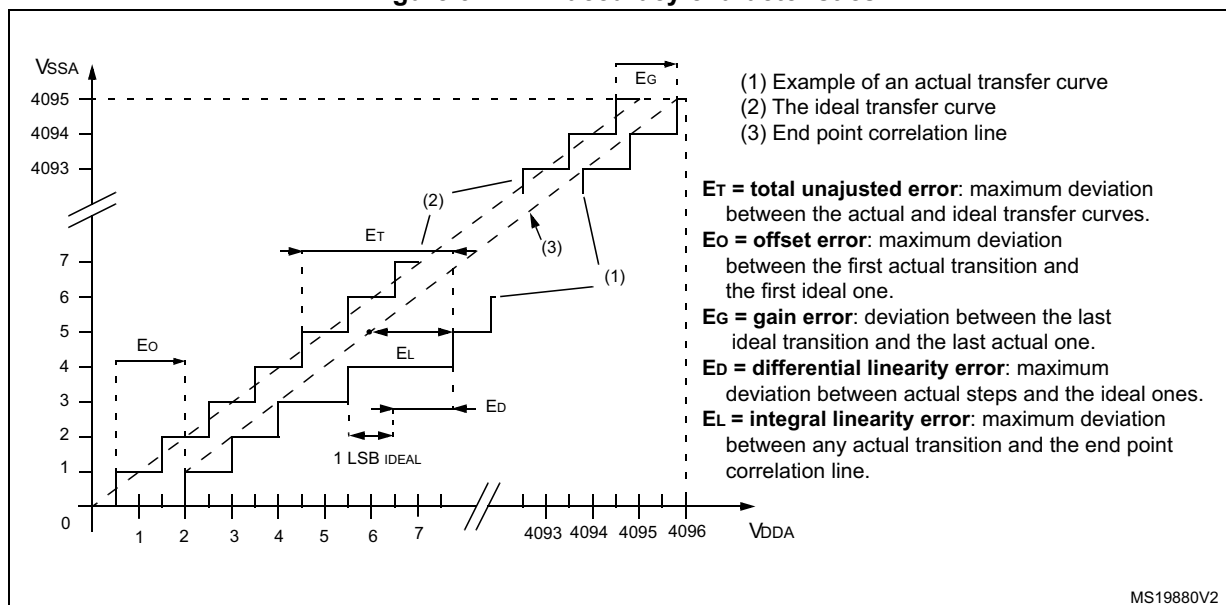
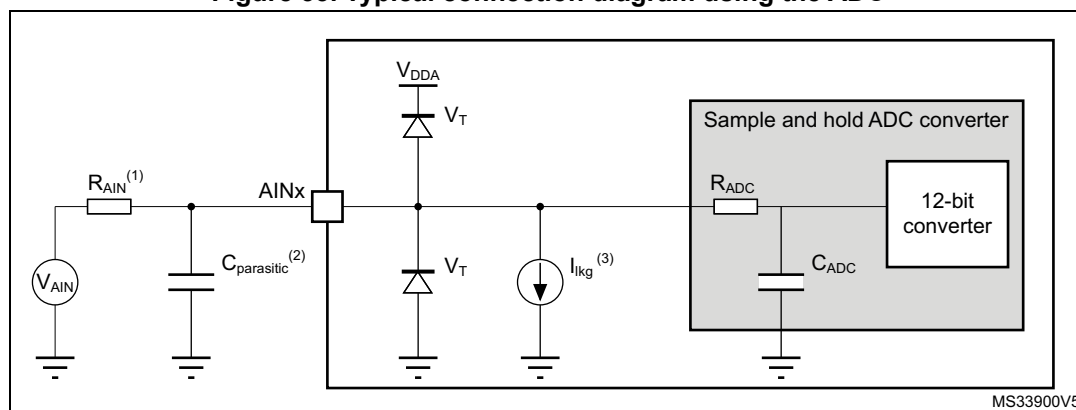


Figure 33. Typical connection diagram using the ADC



1. Refer to [Table 76: ADC characteristics](#) for the values of R_{AIN} and C_{ADC} .
2. $C_{parasitic}$ represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (refer to [Table 70: I/O static characteristics](#) for the value of the pad capacitance). A high $C_{parasitic}$ value will downgrade conversion accuracy. To remedy this, f_{ADC} should be reduced.
3. Refer to [Table 70: I/O static characteristics](#) for the values of I_{kg} .

General PCB design guidelines

Power supply decoupling should be performed as shown in [Figure 20: Power supply scheme](#). The 10 nF capacitor should be ceramic (good quality) and it should be placed as close as possible to the chip.

6.3.19 Digital-to-Analog converter characteristics

Table 82. DAC characteristics⁽¹⁾

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
V_{DDA}	Analog supply voltage for DAC ON	DAC output buffer OFF (no resistive load on DAC1_OUTx pin or internal connection)		1.71	-	3.6	V
		Other modes		1.80	-		
V_{REF+}	Positive reference voltage	DAC output buffer OFF (no resistive load on DAC1_OUTx pin or internal connection)		1.71	-	V_{DDA}	
		Other modes		1.80	-		
V_{REF-}	Negative reference voltage	-		V_{SSA}			
R_L	Resistive load	DAC output buffer ON	connected to V_{SSA}	5	-	-	k Ω
			connected to V_{DDA}	25	-	-	
R_O	Output Impedance	DAC output buffer OFF		9.6	11.7	13.8	k Ω
R_{BON}	Output impedance sample and hold mode, output buffer ON	$V_{DD} = 2.7\text{ V}$		-	-	2	k Ω
		$V_{DD} = 2.0\text{ V}$		-	-	3.5	
R_{BOFF}	Output impedance sample and hold mode, output buffer OFF	$V_{DD} = 2.7\text{ V}$		-	-	16.5	k Ω
		$V_{DD} = 2.0\text{ V}$		-	-	18.0	
C_L	Capacitive load	DAC output buffer ON		-	-	50	pF
C_{SH}		Sample and hold mode		-	0.1	1	μF
V_{DAC_OUT}	Voltage on DAC1_OUTx output	DAC output buffer ON		0.2	-	$V_{REF+} - 0.2$	V
		DAC output buffer OFF		0	-	V_{REF+}	
$t_{SETTLING}$	Settling time (full scale: for a 12-bit code transition between the lowest and the highest input codes when DAC1_OUTx reaches final value $\pm 0.5\text{LSB}$, $\pm 1\text{LSB}$, $\pm 2\text{LSB}$, $\pm 4\text{LSB}$, $\pm 8\text{LSB}$)	Normal mode DAC output buffer ON CL $\leq 50\text{ pF}$, RL $\geq 5\text{ k}\Omega$	$\pm 0.5\text{LSB}$	-	1.7	3	μs
			$\pm 1\text{LSB}$	-	1.6	2.9	
			$\pm 2\text{LSB}$	-	1.55	2.85	
			$\pm 4\text{LSB}$	-	1.48	2.8	
			$\pm 8\text{LSB}$	-	1.4	2.75	
		Normal mode DAC output buffer OFF, $\pm 1\text{LSB}$, CL = 10 pF		-	2	2.5	
$t_{WAKEUP}^{(2)}$	Wakeup time from off state (setting the ENx bit in the DAC Control register) until final value $\pm 1\text{LSB}$	Normal mode DAC output buffer ON CL $\leq 50\text{ pF}$, RL $\geq 5\text{ k}\Omega$		-	4.2	7.5	μs
		Normal mode DAC output buffer OFF, CL $\leq 10\text{ pF}$		-	2	5	
PSRR	V_{DDA} supply rejection ratio	Normal mode DAC output buffer ON CL $\leq 50\text{ pF}$, RL = 5 k Ω , DC		-	-80	-28	dB

Table 82. DAC characteristics⁽¹⁾ (continued)

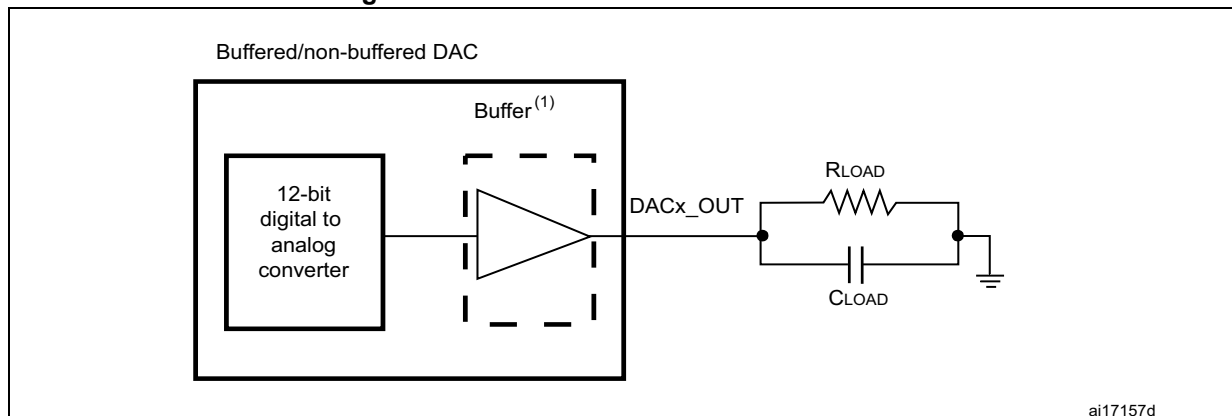
Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$T_{W_to_W}$	Minimal time between two consecutive writes into the DAC_DORx register to guarantee a correct DAC1_OUTx for a small variation of the input code (1 LSB) DAC_MCR:MODEx[2:0] = 000 or 001 DAC_MCR:MODEx[2:0] = 010 or 011	CL ≤ 50 pF, RL ≥ 5 kΩ CL ≤ 10 pF		1 1.4	-	-	μs
t_{SAMP}	Sampling time in sample and hold mode (code transition between the lowest input code and the highest input code when DAC1_OUTx reaches final value ±1LSB)	DAC1_OUTx pin connected	DAC output buffer ON, C _{SH} = 100 nF	-	0.7	3.5	ms
			DAC output buffer OFF, C _{SH} = 100 nF	-	10.5	18	
		DAC1_OUTx pin not connected (internal connection only)	DAC output buffer OFF	-	2	3.5	μs
I_{leak}	Output leakage current	Sample and hold mode, DAC1_OUTx pin connected		-	-	_(3)	nA
$C_{l_{int}}$	Internal sample and hold capacitor	-		5.2	7	8.8	pF
t_{TRIM}	Middle code offset trim time	DAC output buffer ON		50	-	-	μs
V_{offset}	Middle code offset for 1 trim code step	$V_{REF+} = 3.6\text{ V}$		-	1500	-	μV
		$V_{REF+} = 1.8\text{ V}$		-	750	-	
$I_{DDA(DAC)}$	DAC consumption from V _{DDA}	DAC output buffer ON	No load, middle code (0x800)	-	315	500	μA
			No load, worst code (0xF1C)	-	450	670	
		DAC output buffer OFF	No load, middle code (0x800)	-	-	0.2	
		Sample and hold mode, C _{SH} = 100 nF		-	$315 \times \frac{T_{on}}{T_{on} + T_{off}}$ (4)	$670 \times \frac{T_{on}}{T_{on} + T_{off}}$ (4)	

Table 82. DAC characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$I_{DDV}(DAC)$	DAC consumption from V_{REF+}	DAC output buffer ON	No load, middle code (0x800)	-	185	240	μA
			No load, worst code (0xF1C)	-	340	400	
		DAC output buffer OFF	No load, middle code (0x800)	-	155	205	
		Sample and hold mode, buffer ON, $C_{SH} = 100\text{ nF}$, worst case		-	$185 \times \text{Ton}/(\text{Ton} + \text{Toff})$ (4)	$400 \times \text{Ton}/(\text{Ton} + \text{Toff})$ (4)	
		Sample and hold mode, buffer OFF, $C_{SH} = 100\text{ nF}$, worst case		-	$155 \times \text{Ton}/(\text{Ton} + \text{Toff})$ (4)	$205 \times \text{Ton}/(\text{Ton} + \text{Toff})$ (4)	

1. Guaranteed by design.
2. In buffered mode, the output can overshoot above the final value for low input code (starting from min value).
3. Refer to [Table 70: I/O static characteristics](#).
4. Ton is the Refresh phase duration. Toff is the Hold phase duration. Refer to RM0351 reference manual for more details.

Figure 34. 12-bit buffered / non-buffered DAC



1. The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC_CR register.

Table 83. DAC accuracy⁽¹⁾

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
DNL	Differential non linearity ⁽²⁾	DAC output buffer ON		-	-	±2	LSB
		DAC output buffer OFF		-	-	±2	
-	monotonicity	10 bits		guaranteed			
INL	Integral non linearity ⁽³⁾	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±4	
		DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±4	
Offset	Offset error at code 0x800 ⁽³⁾	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	V _{REF+} = 3.6 V	-	-	±12	
			V _{REF+} = 1.8 V	-	-	±25	
		DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±8	
Offset1	Offset error at code 0x001 ⁽⁴⁾	DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±5	
OffsetCal	Offset Error at code 0x800 after calibration	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ	V _{REF+} = 3.6 V	-	-	±5	
			V _{REF+} = 1.8 V	-	-	±7	
Gain	Gain error ⁽⁵⁾	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±0.5	%
		DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±0.5	
TUE	Total unadjusted error	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±30	LSB
		DAC output buffer OFF CL ≤ 50 pF, no RL		-	-	±12	
TUECal	Total unadjusted error after calibration	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ		-	-	±23	LSB
SNR	Signal-to-noise ratio	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ 1 kHz, BW 500 kHz		-	71.2	-	dB
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz BW 500 kHz		-	71.6	-	
THD	Total harmonic distortion	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz		-	-78	-	dB
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz		-	-79	-	

Table 83. DAC accuracy⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
SINAD	Signal-to-noise and distortion ratio	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz	-	70.4	-	dB
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	71	-	
ENOB	Effective number of bits	DAC output buffer ON CL ≤ 50 pF, RL ≥ 5 kΩ, 1 kHz	-	11.4	-	bits
		DAC output buffer OFF CL ≤ 50 pF, no RL, 1 kHz	-	11.5	-	

1. Guaranteed by design.
2. Difference between two consecutive codes - 1 LSB.
3. Difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 4095.
4. Difference between the value measured at Code (0x001) and the ideal value.
5. Difference between ideal slope of the transfer function and measured slope computed from code 0x000 and 0xFFFF when buffer is OFF, and from code giving 0.2 V and (V_{REF+} - 0.2) V when buffer is ON.

6.3.20 Voltage reference buffer characteristics

Table 84. VREFBUF characteristics⁽¹⁾

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
V_{DDA}	Analog supply voltage	Normal mode	$V_{RS} = 0$	2.4	-	3.6	V
			$V_{RS} = 1$	2.8	-	3.6	
		Degraded mode ⁽²⁾	$V_{RS} = 0$	1.65	-	2.4	
			$V_{RS} = 1$	1.65	-	2.8	
V_{REFBUF_OUT}	Voltage reference output	Normal mode	$V_{RS} = 0$	2.046 ⁽³⁾	2.048	2.049 ⁽³⁾	
			$V_{RS} = 1$	2.498 ⁽³⁾	2.5	2.502 ⁽³⁾	
		Degraded mode ⁽²⁾	$V_{RS} = 0$	$V_{DDA} - 150 \text{ mV}$	-	V_{DDA}	
			$V_{RS} = 1$	$V_{DDA} - 150 \text{ mV}$	-	V_{DDA}	
TRIM	Trim step resolution	-	-	-	± 0.05	± 0.1	%
CL	Load capacitor	-	-	0.5	1	1.5	μF
esr	Equivalent Serial Resistor of Cloud	-	-	-	-	2	Ω
I_{load}	Static load current	-	-	-	-	4	mA
I_{line_reg}	Line regulation	$2.8 \text{ V} \leq V_{DDA} \leq 3.6 \text{ V}$	$I_{load} = 500 \mu\text{A}$	-	200	1000	ppm/V
			$I_{load} = 4 \text{ mA}$	-	100	500	
I_{load_reg}	Load regulation	$500 \mu\text{A} \leq I_{load} \leq 4 \text{ mA}$	Normal mode	-	50	500	ppm/mA
T_{Coeff}	Temperature coefficient	$-40 \text{ }^\circ\text{C} < T_J < +125 \text{ }^\circ\text{C}$		-	-	$T_{coeff_vrefint} + 50$	ppm/ $^\circ\text{C}$
		$0 \text{ }^\circ\text{C} < T_J < +50 \text{ }^\circ\text{C}$		-	-	$T_{coeff_vrefint} + 50$	
PSRR	Power supply rejection	DC		40	60	-	dB
		100 kHz		25	40	-	
t_{START}	Start-up time	$CL = 0.5 \mu\text{F}^{(4)}$		-	300	350	μs
		$CL = 1.1 \mu\text{F}^{(4)}$		-	500	650	
		$CL = 1.5 \mu\text{F}^{(4)}$		-	650	800	
I_{INRUSH}	Control of maximum DC current drive on VREFBUF_OUT during start-up phase ⁽⁵⁾	-	-	-	8	-	mA

Table 84. VREFBUF characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{DDA}(VREFBUF)$	VREFBUF consumption from V_{DDA}	$I_{load} = 0 \mu A$	-	16	25	μA
		$I_{load} = 500 \mu A$	-	18	30	
		$I_{load} = 4 mA$	-	35	50	

1. Guaranteed by design, unless otherwise specified.
2. In degraded mode, the voltage reference buffer can not maintain accurately the output voltage which will follow (V_{DDA} - drop voltage).
3. Guaranteed by test in production.
4. The capacitive load must include a 100 nF capacitor in order to cut-off the high frequency noise.
5. To correctly control the VREFBUF inrush current during start-up phase and scaling change, the V_{DDA} voltage should be in the range [2.4 V to 3.6 V] and [2.8 V to 3.6 V] respectively for $V_{RS} = 0$ and $V_{RS} = 1$.

6.3.21 Comparator characteristics

Table 85. COMP characteristics⁽¹⁾

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
V _{DDA}	Analog supply voltage	-		1.62	-	3.6	V
V _{IN}	Comparator input voltage range	-		0	-	V _{DDA}	
V _{BG} ⁽²⁾	Scaler input voltage	-		V _{REFINT}			
V _{SC}	Scaler offset voltage	-		-	±5	±10	mV
I _{DDA} (SCALER)	Scaler static consumption from V _{DDA}	BRG_EN=0 (bridge disable)		-	200	300	nA
		BRG_EN=1 (bridge enable)		-	0.8	1	μA
t _{START_SCALER}	Scaler startup time	-		-	100	200	μs
t _{START}	Comparator startup time to reach propagation delay specification	High-speed mode	V _{DDA} ≥ 2.7 V	-	-	5	μs
			V _{DDA} < 2.7 V	-	-	7	
		Medium mode	V _{DDA} ≥ 2.7 V	-	-	15	
			V _{DDA} < 2.7 V	-	-	25	
		Ultra-low-power mode		-	-	80	
t _D ⁽³⁾	Propagation delay for 200 mV step with 100 mV overdrive	High-speed mode	V _{DDA} ≥ 2.7 V	-	55	80	ns
			V _{DDA} < 2.7 V	-	65	100	
		Medium mode	V _{DDA} ≥ 2.7 V	-	0.55	0.9	μs
			V _{DDA} < 2.7 V	-	0.65	1	
		Ultra-low-power mode		-	5	12	
V _{offset}	Comparator offset error	Full common mode range	-	-	±5	±20	mV
V _{hys}	Comparator hysteresis	No hysteresis		-	0	-	mV
		Low hysteresis		4	8	16	
		Medium hysteresis		8	15	30	
		High hysteresis		15	27	52	

Table 85. COMP characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
$I_{DDA}(COMP)$	Comparator consumption from V_{DDA}	Ultra-low-power mode	Static	-	400	600	nA
			With 50 kHz ± 100 mV overdrive square signal	-	1200	-	
		Medium mode	Static	-	5	7	μA
			With 50 kHz ± 100 mV overdrive square signal	-	6	-	
		High-speed mode	Static	-	70	100	
			With 50 kHz ± 100 mV overdrive square signal	-	75	-	
I_{bias}	Comparator input bias current	-		-	-	_(4)	nA

1. Guaranteed by design, unless otherwise specified.

2. Refer to [Table 26: Embedded internal voltage reference](#).

3. Guaranteed by characterization results.

4. Mostly I/O leakage when used in analog mode. Refer to I_{lk} parameter in [Table 70: I/O static characteristics](#).

6.3.22 Operational amplifiers characteristics

Table 86. OPAMP characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{DDA}	Analog supply voltage ⁽²⁾	-	1.8	-	3.6	V
CMIR	Common mode input range	-	0	-	V_{DDA}	V
$V_{I\text{OFFSET}}$	Input offset voltage	25 °C, No Load on output.	-	-	± 1.5	mV
		All voltage/Temp.	-	-	± 3	
$\Delta V_{I\text{OFFSET}}$	Input offset voltage drift	Normal mode	-	± 5	-	$\mu V/^{\circ}C$
		Low-power mode	-	± 10	-	
TRIMOFFSETP TRIMLPOFFSETP	Offset trim step at low common input voltage ($0.1 \times V_{DDA}$)	-	-	0.8	1.1	mV
TRIMOFFSETN TRIMLPOFFSETN	Offset trim step at high common input voltage ($0.9 \times V_{DDA}$)	-	-	1	1.35	

Table 86. OPAMP characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
I _{LOAD}	Drive current	Normal mode	V _{DDA} ≥ 2 V	-	-	500	μA
		Low-power mode		-	-	100	
I _{LOAD_PGA}	Drive current in PGA mode	Normal mode	V _{DDA} ≥ 2 V	-	-	450	
		Low-power mode		-	-	50	
R _{LOAD}	Resistive load (connected to VSSA or to VDDA)	Normal mode	V _{DDA} < 2 V	4	-	-	kΩ
		Low-power mode		20	-	-	
R _{LOAD_PGA}	Resistive load in PGA mode (connected to VSSA or to VDDA)	Normal mode	V _{DDA} < 2 V	4.5	-	-	
		Low-power mode		40	-	-	
C _{LOAD}	Capacitive load	-		-	-	50	pF
CMRR	Common mode rejection ratio	Normal mode		-	-85	-	dB
		Low-power mode		-	-90	-	
PSRR	Power supply rejection ratio	Normal mode	C _{LOAD} ≤ 50 pf, R _{LOAD} ≥ 4 kΩ DC	70	85	-	dB
		Low-power mode	C _{LOAD} ≤ 50 pf, R _{LOAD} ≥ 20 kΩ DC	72	90	-	
GBW	Gain Bandwidth Product	Normal mode	V _{DDA} ≥ 2.4 V (OPA_RANGE = 1)	550	1600	2200	kHz
		Low-power mode		100	420	600	
		Normal mode	V _{DDA} < 2.4 V (OPA_RANGE = 0)	250	700	950	
		Low-power mode		40	180	280	
SR ⁽³⁾	Slew rate (from 10 and 90% of output voltage)	Normal mode	V _{DDA} ≥ 2.4 V	-	700	-	V/ms
		Low-power mode		-	180	-	
		Normal mode	V _{DDA} < 2.4 V	-	300	-	
		Low-power mode		-	80	-	
AO	Open loop gain	Normal mode		55	110	-	dB
		Low-power mode		45	110	-	
V _{OHSAT} ⁽³⁾	High saturation voltage	Normal mode	I _{load} = max or R _{load} = min Input at V _{DDA} .	V _{DDA} - 100	-	-	mV
		Low-power mode		V _{DDA} - 50	-	-	
V _{OLSAT} ⁽³⁾	Low saturation voltage	Normal mode	I _{load} = max or R _{load} = min Input at 0.	-	-	100	
		Low-power mode		-	-	50	
φ _m	Phase margin	Normal mode		-	74	-	°
		Low-power mode		-	66	-	

Table 86. OPAMP characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
GM	Gain margin	Normal mode		-	13	-	dB
		Low-power mode		-	20	-	
t_{WAKEUP}	Wake up time from OFF state.	Normal mode	$C_{\text{LOAD}} \leq 50 \text{ pf}$, $R_{\text{LOAD}} \geq 4 \text{ k}\Omega$ follower configuration	-	5	10	μs
		Low-power mode	$C_{\text{LOAD}} \leq 50 \text{ pf}$, $R_{\text{LOAD}} \geq 20 \text{ k}\Omega$ follower configuration	-	10	30	
I_{bias}	OPAMP input bias current	Dedicated input (UFBGA132 only)	$T_J \leq 75 \text{ }^\circ\text{C}$	-	-	1	nA
			$T_J \leq 85 \text{ }^\circ\text{C}$	-	-	3	
			$T_J \leq 105 \text{ }^\circ\text{C}$	-	-	8	
			$T_J \leq 125 \text{ }^\circ\text{C}$	-	-	15	
		General purpose input (all packages except UFBGA132)		-	-	_(4)	
PGA gain ⁽³⁾	Non inverting gain value	-		-	2	-	-
				-	4	-	
				-	8	-	
				-	16	-	
R_{network}	R2/R1 internal resistance values in PGA mode ⁽⁵⁾	PGA Gain = 2		-	80/80	-	k Ω /k Ω
		PGA Gain = 4		-	120/40	-	
		PGA Gain = 8		-	140/20	-	
		PGA Gain = 16		-	150/10	-	
Delta R	Resistance variation (R1 or R2)	-		-15	-	15	%
PGA gain error	PGA gain error	-		-1	-	1	%
PGA BW	PGA bandwidth for different non inverting gain	Gain = 2	-	-	GBW/2	-	MHz
		Gain = 4	-	-	GBW/4	-	
		Gain = 8	-	-	GBW/8	-	
		Gain = 16	-	-	GBW/16	-	

Table 86. OPAMP characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions		Min	Typ	Max	Unit
e _n	Voltage noise density	Normal mode	at 1 kHz, Output loaded with 4 kΩ	-	500	-	nV/√Hz
		Low-power mode	at 1 kHz, Output loaded with 20 kΩ	-	600	-	
		Normal mode	at 10 kHz, Output loaded with 4 kΩ	-	180	-	
		Low-power mode	at 10 kHz, Output loaded with 20 kΩ	-	290	-	
I _{DDA} (OPAMP) ⁽³⁾	OPAMP consumption from V _{DDA}	Normal mode	no Load, quiescent mode	-	120	260	μA
		Low-power mode		-	45	100	

1. Guaranteed by design, unless otherwise specified.
2. The temperature range is limited to 0 °C-125 °C when V_{DDA} is below 2 V
3. Guaranteed by characterization results.
4. Mostly I/O leakage, when used in analog mode. Refer to I_{lkg} parameter in [Table 70: I/O static characteristics](#).
5. R2 is the internal resistance between OPAMP output and OPAMP inverting input. R1 is the internal resistance between OPAMP inverting input and ground. The PGA gain = 1+R2/R1

6.3.23 Temperature sensor characteristics

Table 87. TS characteristics

Symbol	Parameter	Min	Typ	Max	Unit
$T_L^{(1)}$	V_{TS} linearity with temperature	-	± 1	± 2	$^{\circ}\text{C}$
Avg_Slope ⁽²⁾	Average slope	2.3	2.5	2.7	mV/ $^{\circ}\text{C}$
V_{30}	Voltage at 30 $^{\circ}\text{C}$ (± 5 $^{\circ}\text{C}$) ⁽³⁾	0.742	0.76	0.785	V
$t_{\text{START}}^{(1)}$ (TS_BUF) ⁽¹⁾	Sensor Buffer Start-up time in continuous mode ⁽⁴⁾	-	8	15	μs
$t_{\text{START}}^{(1)}$	Start-up time when entering in continuous mode ⁽⁴⁾	-	70	120	μs
$t_{\text{S_temp}}^{(1)}$	ADC sampling time when reading the temperature	5	-	-	μs
$I_{\text{DD}}(\text{TS})^{(1)}$	Temperature sensor consumption from V_{DD} , when selected by ADC	-	4.7	7	μA

1. Guaranteed by design.
2. Guaranteed by characterization results.
3. Measured at $V_{\text{DDA}} = 3.0 \text{ V} \pm 10 \text{ mV}$. The V_{30} ADC conversion result is stored in the TS_CAL1 byte. Refer to [Table 8: Temperature sensor calibration values](#).
4. Continuous mode means Run/Sleep modes, or temperature sensor enable in Low-power run/Low-power sleep modes.

6.3.24 V_{BAT} monitoring characteristics

Table 88. V_{BAT} monitoring characteristics

Symbol	Parameter	Min	Typ	Max	Unit
R	Resistor bridge for V_{BAT}	-	39	-	k Ω
Q	Ratio on V_{BAT} measurement	-	3	-	-
$E_r^{(1)}$	Error on Q	-10	-	10	%
$t_{\text{S_vbat}}^{(1)}$	ADC sampling time when reading the VBAT	12	-	-	μs

1. Guaranteed by design.

Table 89. V_{BAT} charging characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
R_{BC}	Battery charging resistor	VBRS = 0	-	5	-	k Ω
		VBRS = 1	-	1.5	-	

6.3.25 LCD controller characteristics

The devices embed a built-in step-up converter to provide a constant LCD reference voltage independently from the V_{DD} voltage. An external capacitor C_{ext} must be connected to the VLCD pin to decouple this converter.

Table 90. LCD controller characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V_{LCD}	LCD external voltage		-	-	3.6	V
V_{LCD0}	LCD internal reference voltage 0		-	2.62	-	
V_{LCD1}	LCD internal reference voltage 1		-	2.76	-	
V_{LCD2}	LCD internal reference voltage 2		-	2.89	-	
V_{LCD3}	LCD internal reference voltage 3		-	3.04	-	
V_{LCD4}	LCD internal reference voltage 4		-	3.19	-	
V_{LCD5}	LCD internal reference voltage 5		-	3.32	-	
V_{LCD6}	LCD internal reference voltage 6		-	3.46	-	
V_{LCD7}	LCD internal reference voltage 7		-	3.62	-	
C_{ext}	V_{LCD} external capacitance	Buffer OFF (BUFEN=0 is LCD_CR register)	0.2	-	2	μF
		Buffer ON (BUFEN=1 is LCD_CR register)	1	-	2	
$I_{LCD}^{(2)}$	Supply current from V_{DD} at $V_{DD} = 2.2 V$	Buffer OFF (BUFEN=0 is LCD_CR register)	-	3	-	μA
	Supply current from V_{DD} at $V_{DD} = 3.0 V$	Buffer OFF (BUFEN=0 is LCD_CR register)	-	1.5	-	
I_{VLCD}	Supply current from V_{LCD} ($V_{LCD} = 3 V$)	Buffer OFF (BUFFEN = 0, PON = 0)	-	0.5	-	μA
		Buffer ON (BUFFEN = 1, 1/2 Bias)	-	0.6	-	
		Buffer ON (BUFFEN = 1, 1/3 Bias)	-	0.8	-	
		Buffer ON (BUFFEN = 1, 1/4 Bias)	-	1	-	
R_{HN}	Total High Resistor value for Low drive resistive network		-	5.5	-	M Ω
R_{LN}	Total Low Resistor value for High drive resistive network		-	240	-	k Ω
V_{44}	Segment/Common highest level voltage		-	V_{LCD}	-	V
V_{34}	Segment/Common 3/4 level voltage		-	$3/4 V_{LCD}$	-	
V_{23}	Segment/Common 2/3 level voltage		-	$2/3 V_{LCD}$	-	
V_{12}	Segment/Common 1/2 level voltage		-	$1/2 V_{LCD}$	-	
V_{13}	Segment/Common 1/3 level voltage		-	$1/3 V_{LCD}$	-	
V_{14}	Segment/Common 1/4 level voltage		-	$1/4 V_{LCD}$	-	
V_0	Segment/Common lowest level voltage		-	0	-	

1. Guaranteed by design.
2. LCD enabled with 3 V internal step-up active, 1/8 duty, 1/4 bias, division ratio= 64, all pixels active, no LCD connected.

6.3.26 DFSDM characteristics

Unless otherwise specified, the parameters given in [Table 91](#) for DFSDM are derived from tests performed under the ambient temperature, f_{APB2} frequency and V_{DD} supply voltage conditions summarized in [Table 23: General operating conditions](#).

- Output speed is set to $OSPEEDRy[1:0] = 10$
- Capacitive load $C = 30$ pF
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$

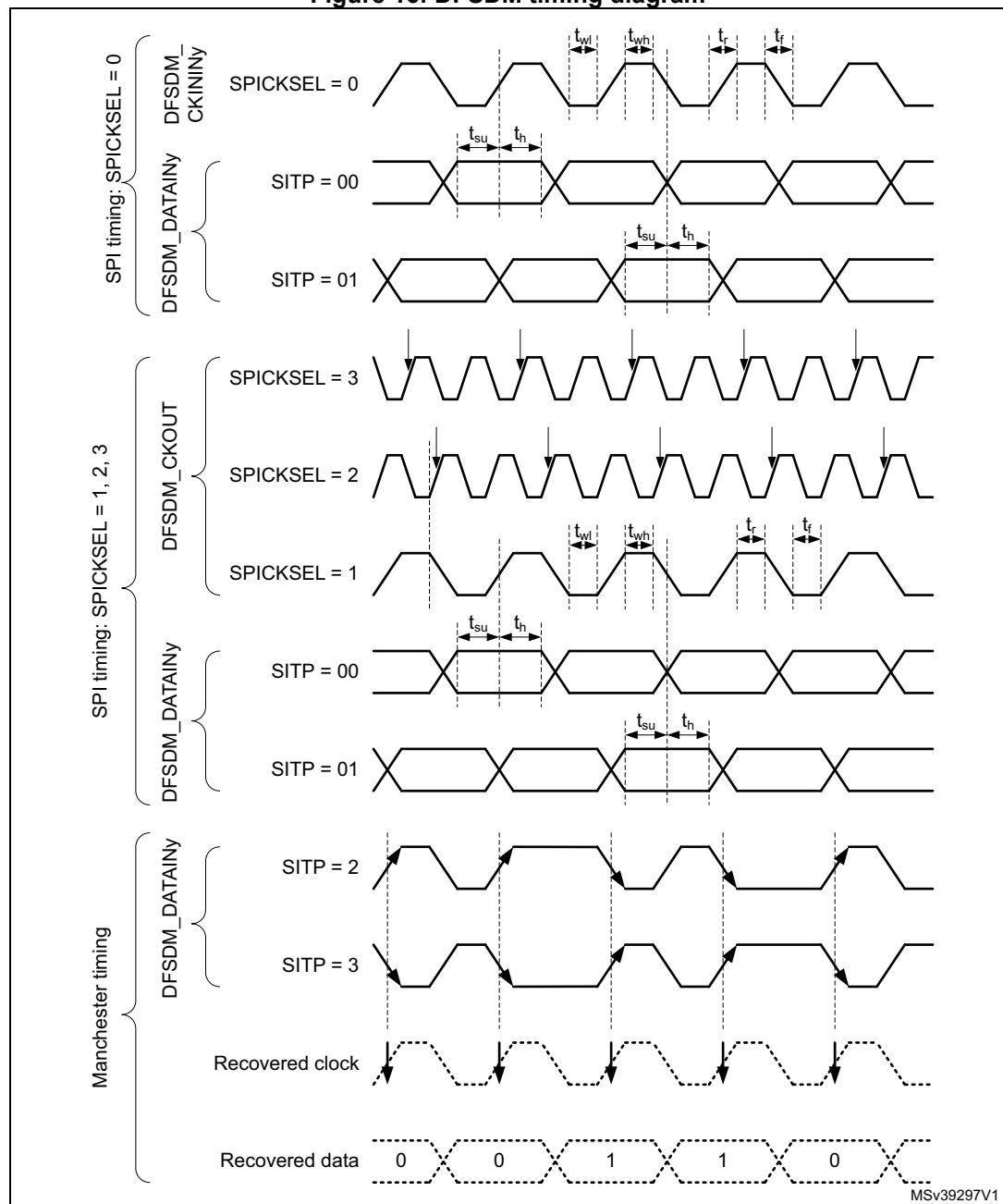
Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics (DFSDM1_CKINy, DFSDM1_DATINy, DFSDM1_CKOUT for DFSDM).

Table 91. DFSDM characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{DFSDMCLK}$	DFSDM clock	-	-	-	f_{SYSCLK}	MHz
f_{CKIN} ($1/T_{CKIN}$)	Input clock frequency	SPI mode (SITP[1:0] = 01)	-	-	20 ($f_{DFSDMCLK}/4$)	
f_{CKOUT}	Output clock frequency	-	-	-	20	MHz
$DuCy_{CKOUT}$	Output clock frequency duty cycle	-	45	50	55	%
$t_{wh(CKIN)}$ $t_{wl(CKIN)}$	Input clock high and low time	SPI mode (SITP[1:0] = 01), External clock mode (SPICKSEL[1:0] = 0)	$T_{CKIN}/2-0.5$	$T_{CKIN}/2$	-	ns
t_{su}	Data input setup time	SPI mode (SITP[1:0]=01), External clock mode (SPICKSEL[1:0] = 0)	0	-	-	
t_h	Data input hold time	SPI mode (SITP[1:0]=01), External clock mode (SPICKSEL[1:0] = 0)	2	-	-	
$T_{Manchester}$	Manchester data period (recovered clock period)	Manchester mode (SITP[1:0] = 10 or 11), Internal clock mode (SPICKSEL[1:0] ≠ 0)	$(CKOUT DIV+1) \times T_{DFSDMCLK}$	-	$(2 \times CKOUTDIV) \times T_{DFSDMCLK}$	

1. Guaranteed by characterization results.

Figure 16: DFSDM timing diagram



6.3.27 Timer characteristics

The parameters given in the following tables are guaranteed by design.

Refer to [Section 6.3.14: I/O port characteristics](#) for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 92. TIMx⁽¹⁾ characteristics

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{res(TIM)}$	Timer resolution time	-	1	-	$t_{TIMxCLK}$
		$f_{TIMxCLK} = 80\text{ MHz}$	12.5	-	ns
f_{EXT}	Timer external clock frequency on CH1 to CH4	-	0	$f_{TIMxCLK}/2$	MHz
		$f_{TIMxCLK} = 80\text{ MHz}$	0	40	MHz
Res_{TIM}	Timer resolution	TIMx (except TIM2 and TIM5)	-	16	bit
		TIM2 and TIM5	-	32	
$t_{COUNTER}$	16-bit counter clock period	-	1	65536	$t_{TIMxCLK}$
		$f_{TIMxCLK} = 80\text{ MHz}$	0.0125	819.2	μs
t_{MAX_COUNT}	Maximum possible count with 32-bit counter	-	-	65536×65536	$t_{TIMxCLK}$
		$f_{TIMxCLK} = 80\text{ MHz}$	-	53.68	s

1. TIMx is used as a general term in which x stands for 1,2,3,4,5,6,7,8,15,16 or 17.

Table 93. IWDG min/max timeout period at 32 kHz (LSI)⁽¹⁾

Prescaler divider	PR[2:0] bits	Min timeout RL[11:0]= 0x000	Max timeout RL[11:0]= 0xFFFF	Unit
/4	0	0.125	512	ms
/8	1	0.250	1024	
/16	2	0.500	2048	
/32	3	1.0	4096	
/64	4	2.0	8192	
/128	5	4.0	16384	
/256	6 or 7	8.0	32768	

1. The exact timings still depend on the phasing of the APB interface clock versus the LSI clock so that there is always a full RC period of uncertainty.

Table 94. WWDG min/max timeout value at 80 MHz (PCLK)

Prescaler	WDGTB	Min timeout value	Max timeout value	Unit
1	0	0.0512	3.2768	ms
2	1	0.1024	6.5536	
4	2	0.2048	13.1072	
8	3	0.4096	26.2144	

6.3.28 Communication interfaces characteristics

I²C interface characteristics

The I2C interface meets the timings requirements of the I²C-bus specification and user manual rev. 03 for:

- Standard-mode (Sm): with a bit rate up to 100 kbit/s
- Fast-mode (Fm): with a bit rate up to 400 kbit/s
- Fast-mode Plus (Fm+): with a bit rate up to 1 Mbit/s.

The I2C timings requirements are guaranteed by design when the I2C peripheral is properly configured (refer to RM0351 reference manual).

The SDA and SCL I/O requirements are met with the following restrictions: the SDA and SCL I/O pins are not “true” open-drain. When configured as open-drain, the PMOS connected between the I/O pin and V_{DDIOx} is disabled, but is still present. Only FT_f I/O pins support Fm+ low level output current maximum requirement. Refer to [Section 6.3.14: I/O port characteristics](#) for the I2C I/Os characteristics.

All I2C SDA and SCL I/Os embed an analog filter. Refer to the table below for the analog filter characteristics:

Table 95. I2C analog filter characteristics⁽¹⁾

Symbol	Parameter	Min	Max	Unit
t _{AF}	Maximum pulse width of spikes that are suppressed by the analog filter	50 ⁽²⁾	260 ⁽³⁾	ns

1. Guaranteed by design.
2. Spikes with widths below t_{AF(min)} are filtered.
3. Spikes with widths above t_{AF(max)} are not filtered

SPI characteristics

Unless otherwise specified, the parameters given in [Table 96](#) for SPI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and supply voltage conditions summarized in [Table 23: General operating conditions](#).

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI).

Table 96. SPI characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{SCK} $1/t_{c(SCK)}$	SPI clock frequency	Master mode receiver/full duplex $2.7 < V_{DD} < 3.6$ V Voltage Range 1	-	-	24	MHz
		Master mode receiver/full duplex $1.71 < V_{DD} < 3.6$ V Voltage Range 1			13	
		Master mode transmitter $1.71 < V_{DD} < 3.6$ V Voltage Range 1			40	
		Slave mode receiver $1.71 < V_{DD} < 3.6$ V Voltage Range 1			40	
		Slave mode transmitter/full duplex $2.7 < V_{DD} < 3.6$ V Voltage Range 1			26 ⁽²⁾	
		Slave mode transmitter/full duplex $1.71 < V_{DD} < 3.6$ V Voltage Range 1			16 ⁽²⁾	
		Voltage Range 2			13	
		$1.08 < V_{DDIO2} < 1.32$ V ⁽³⁾			8	
$t_{su(NSS)}$	NSS setup time	Slave mode, SPI prescaler = 2	$4 \times T_{PCLK}$	-	-	ns
$t_{h(NSS)}$	NSS hold time	Slave mode, SPI prescaler = 2	$2 \times T_{PCLK}$	-	-	ns
$t_{w(SCKH)}$ $t_{w(SCKL)}$	SCK high and low time	Master mode	$T_{PCLK}-2$	T_{PCLK}	$T_{PCLK}+2$	ns
$t_{su(MI)}$	Data input setup time	Master mode	3.5	-	-	ns
$t_{su(SI)}$		Slave mode	3	-	-	
$t_{h(MI)}$	Data input hold time	Master mode	6.5	-	-	ns
$t_{h(SI)}$		Slave mode	3	-	-	
$t_{a(SO)}$	Data output access time	Slave mode	9	-	36	ns
$t_{dis(SO)}$	Data output disable time	Slave mode	9	-	16	ns

Table 96. SPI characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{v(SO)}$	Data output valid time	Slave mode $2.7 < V_{DD} < 3.6$ V Voltage Range 1	-	12.5	19	ns
		Slave mode $1.71 < V_{DD} < 3.6$ V Voltage Range 1	-	12.5	30	
		Slave mode $1.71 < V_{DD} < 3.6$ V Voltage Range 2	-	12.5	33	
-		Slave mode $1.08 < V_{DDIO2} < 1.32$ V ⁽³⁾	-	25	62.5	
$t_{v(MO)}$		Master mode	-	2.5	12.5	
$t_{h(SO)}$	Data output hold time	Slave mode	9	-	-	ns
-		Slave mode $1.08 < V_{DDIO2} < 1.32$ V ⁽³⁾	24	-	-	
$t_{h(MO)}$		Master mode	0	-	-	

- Guaranteed by characterization results.
- Maximum frequency in Slave transmitter mode is determined by the sum of $t_{v(SO)}$ and $t_{su(MI)}$ which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having $t_{su(MI)} = 0$ while Duty(SCK) = 50 %.
- SPI mapped on Port G.

Figure 35. SPI timing diagram - slave mode and CPHA = 0

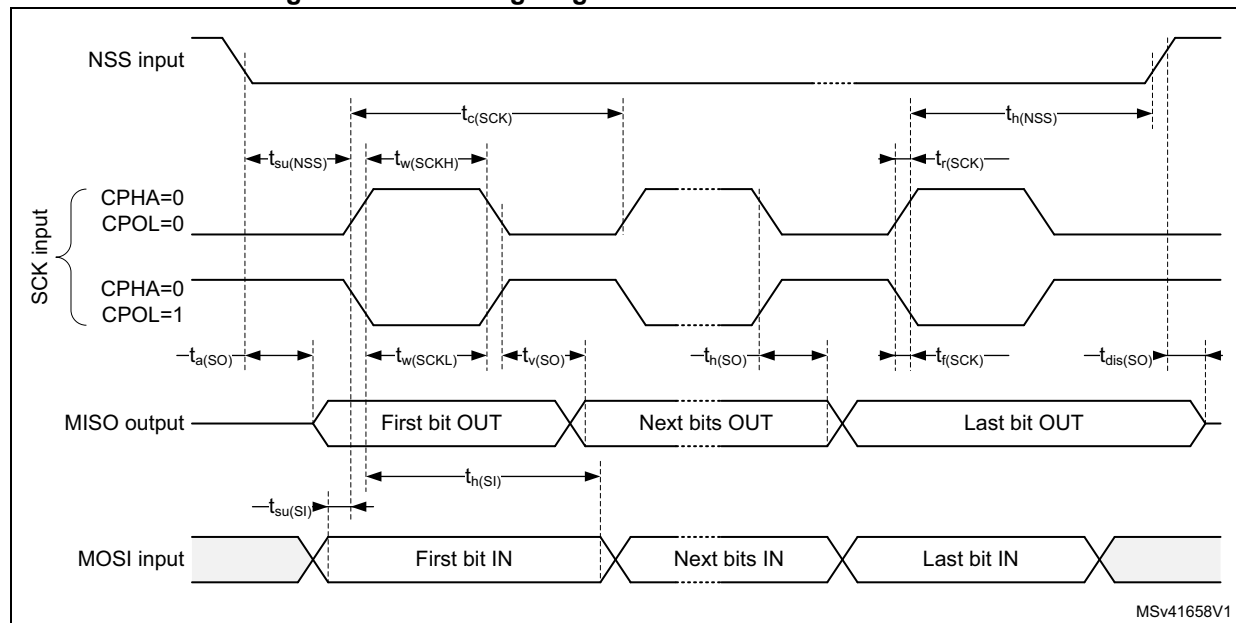
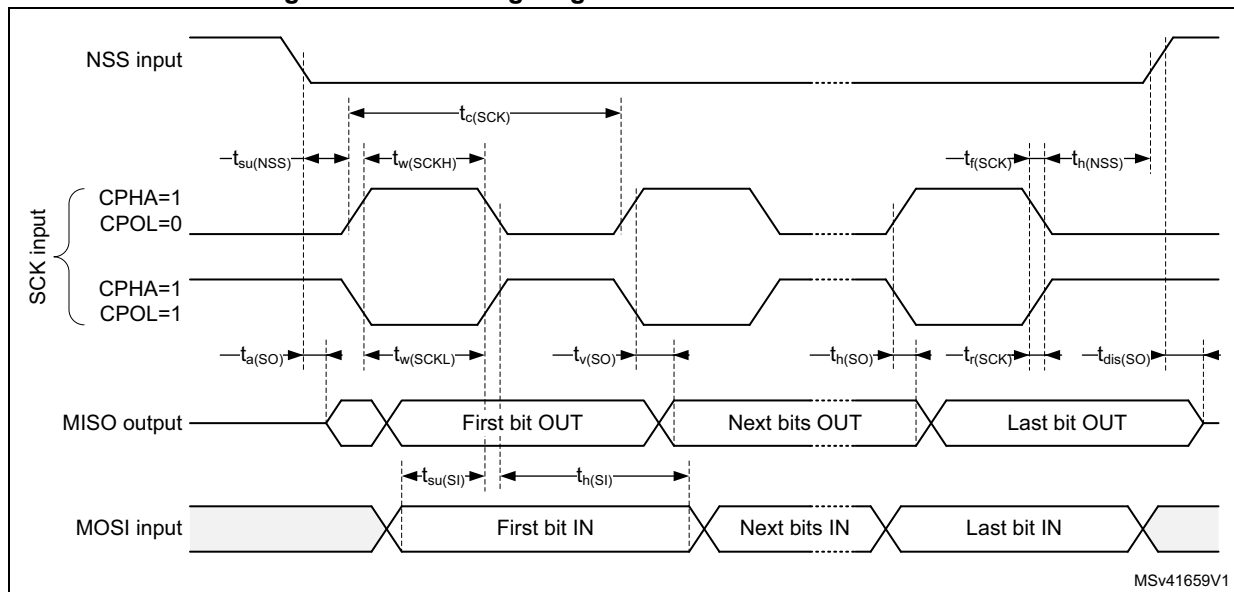
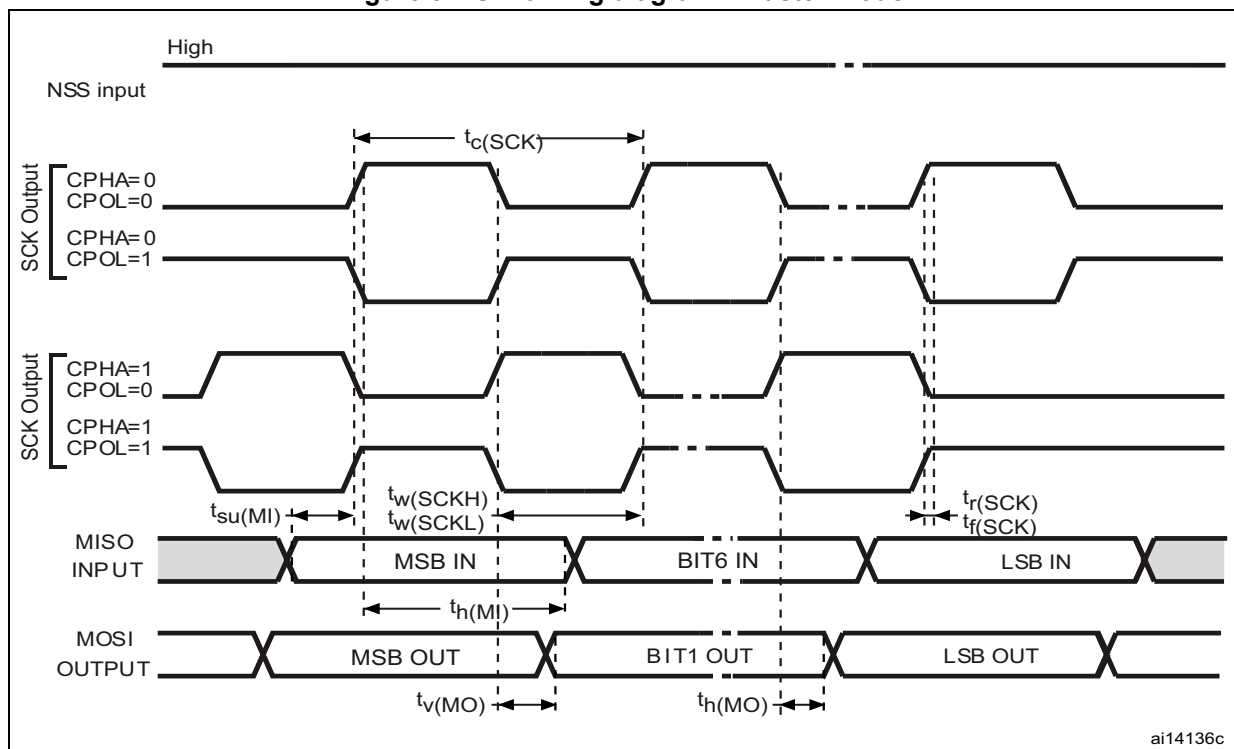


Figure 36. SPI timing diagram - slave mode and CPHA = 1



1. Measurement points are done at CMOS levels: 0.3 V_{DD} and 0.7 V_{DD} .

Figure 37. SPI timing diagram - master mode



1. Measurement points are done at CMOS levels: 0.3 V_{DD} and 0.7 V_{DD} .

Quad SPI characteristics

Unless otherwise specified, the parameters given in [Table 97](#) and [Table 98](#) for Quad SPI are derived from tests performed under the ambient temperature, f_{AHB} frequency and V_{DD} supply voltage conditions summarized in [Table 23: General operating conditions](#), with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 11
- Capacitive load C = 15 or 20 pF
- Measurement points are done at CMOS levels: $0.5 \times V_{\text{DD}}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics.

Table 97. Quad SPI characteristics in SDR mode⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
F_{CK} $1/t_{\text{CK}}$	Quad SPI clock frequency	$1.71 < V_{\text{DD}} < 3.6 \text{ V}$, $C_{\text{LOAD}} = 20 \text{ pF}$ Voltage Range 1	-	-	40	MHz
		$1.71 < V_{\text{DD}} < 3.6 \text{ V}$, $C_{\text{LOAD}} = 15 \text{ pF}$ Voltage Range 1	-	-	48	
		$2.7 < V_{\text{DD}} < 3.6 \text{ V}$, $C_{\text{LOAD}} = 15 \text{ pF}$ Voltage Range 1	-	-	60	
		$1.71 < V_{\text{DD}} < 3.6 \text{ V}$, $C_{\text{LOAD}} = 20 \text{ pF}$ Voltage Range 2	-	-	26	
$t_{\text{w(CKH)}}$	Quad SPI clock high and low time	$f_{\text{AHBCLK}} = 48 \text{ MHz}$, presc=0	$t_{\text{CK}}/2-2$	-	$t_{\text{CK}}/2$	ns
$t_{\text{w(CKL)}}$			$t_{\text{CK}}/2$	-	$t_{\text{CK}}/2+2$	
$t_{\text{s(IN)}}$	Data input setup time	Voltage Range 1	4	-	-	
		Voltage Range 2	3.5	-	-	
$t_{\text{h(IN)}}$	Data input hold time	Voltage Range 1	5.5	-	-	
		Voltage Range 2	6.5	-	-	
$t_{\text{v(OUT)}}$	Data output valid time	Voltage Range 1	-	2.5	5	
		Voltage Range 2	-	3	5	
$t_{\text{h(OUT)}}$	Data output hold time	Voltage Range 1	1.5	-	-	
		Voltage Range 2	2	-	-	

1. Guaranteed by characterization results.

Table 98. QUADSPI characteristics in DDR mode⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
F_{CK} $1/t_{(CK)}$	Quad SPI clock frequency	$1.71 < V_{DD} < 3.6 \text{ V}$, $C_{LOAD} = 20 \text{ pF}$ Voltage Range 1	-	-	40	MHz
		$2 < V_{DD} < 3.6 \text{ V}$, $C_{LOAD} = 20 \text{ pF}$ Voltage Range 1	-	-	48	
		$1.71 < V_{DD} < 3.6 \text{ V}$, $C_{LOAD} = 15 \text{ pF}$ Voltage Range 1	-	-	48	
		$1.71 < V_{DD} < 3.6 \text{ V}$, $C_{LOAD} = 20 \text{ pF}$ Voltage Range 2	-	-	26	
$t_{w(CKH)}$	Quad SPI clock high and low time	$f_{AHBCLK} = 48 \text{ MHz}$, presc=0	$t_{(CK)}/2-2$	-	$t_{(CK)}/2$	ns
$t_{w(CKL)}$			$t_{(CK)}/2$	-	$t_{(CK)}/2+2$	
$t_{sf(IN)}; t_{sr(IN)}$	Data input setup time	Voltage Range 1 and 2	3.5	-	-	
$t_{hf(IN)}; t_{hr(IN)}$	Data input hold time		6.5	-	-	
$t_{vf(OUT)}; t_{vr(OUT)}$	Data output valid time	Voltage Range 1	-	11	12	
		Voltage Range 2	-	15	19	
$t_{hf(OUT)}; t_{hr(OUT)}$	Data output hold time	Voltage Range 1	6	-	-	
		Voltage Range 2	8	-	-	

1. Guaranteed by characterization results.

Figure 38. Quad SPI timing diagram - SDR mode

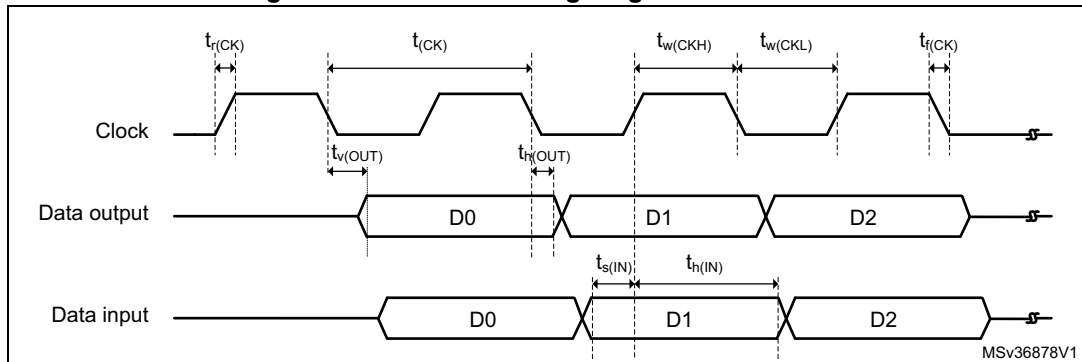
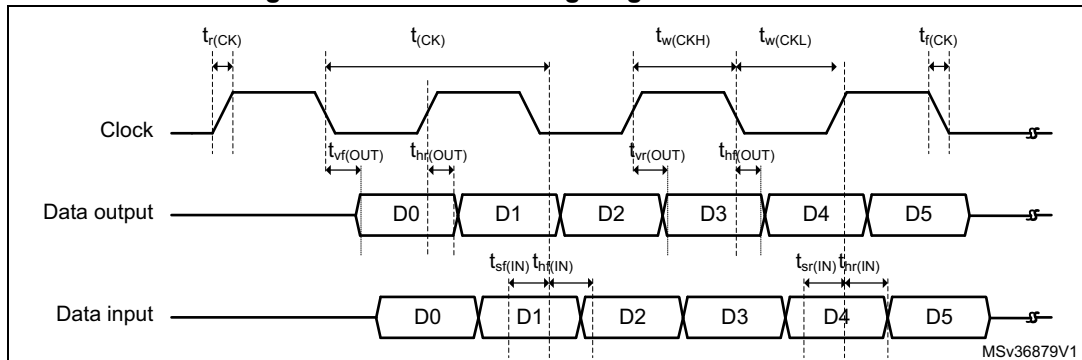


Figure 39. Quad SPI timing diagram - DDR mode



SAI characteristics

Unless otherwise specified, the parameters given in [Table 99](#) for SAI are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in [Table 23: General operating conditions](#), with the following configuration:

- Output speed is set to $OSPEEDRy[1:0] = 10$
- Capacitive load $C = 30$ pF
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics (CK,SD,FS).

Table 99. SAI characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Max	Unit
f_{MCLK}	SAI Main clock output	-	-	50	MHz
f_{CK}	SAI clock frequency ⁽²⁾	Master transmitter $2.7 \leq V_{DD} \leq 3.6$ Voltage Range 1	-	18.5	MHz
		Master transmitter $1.71 \leq V_{DD} \leq 3.6$ Voltage Range 1	-	12.5	
		Master receiver Voltage Range 1	-	25	
		Slave transmitter $2.7 \leq V_{DD} \leq 3.6$ Voltage Range 1	-	22.5	
		Slave transmitter $1.71 \leq V_{DD} \leq 3.6$ Voltage Range 1	-	14.5	
		Slave receiver Voltage Range 1	-	25	
		Voltage Range 2	-	12.5	
$t_{V(FS)}$	FS valid time	Master mode $2.7 \leq V_{DD} \leq 3.6$	-	22	ns
		Master mode $1.71 \leq V_{DD} \leq 3.6$	-	40	
$t_{h(FS)}$	FS hold time	Master mode	10	-	ns
$t_{su(FS)}$	FS setup time	Slave mode	1	-	ns
$t_{h(FS)}$	FS hold time	Slave mode	2	-	ns
$t_{su(SD_A_MR)}$	Data input setup time	Master receiver	2.5	-	ns
$t_{su(SD_B_SR)}$		Slave receiver	3	-	
$t_{h(SD_A_MR)}$	Data input hold time	Master receiver	8	-	ns
$t_{h(SD_B_SR)}$		Slave receiver	4	-	

Table 99. SAI characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{V(SD_B_ST)}$	Data output valid time	Slave transmitter (after enable edge) $2.7 \leq V_{DD} \leq 3.6$	-	22	ns
		Slave transmitter (after enable edge) $1.71 \leq V_{DD} \leq 3.6$	-	34	
$t_{h(SD_B_ST)}$	Data output hold time	Slave transmitter (after enable edge)	10	-	ns
$t_{V(SD_A_MT)}$	Data output valid time	Master transmitter (after enable edge) $2.7 \leq V_{DD} \leq 3.6$	-	27	ns
		Master transmitter (after enable edge) $1.71 \leq V_{DD} \leq 3.6$	-	40	
$t_{h(SD_A_MT)}$	Data output hold time	Master transmitter (after enable edge)	10	-	ns

1. Guaranteed by characterization results.
2. APB clock frequency must be at least twice SAI clock frequency.

Figure 40. SAI master timing waveforms

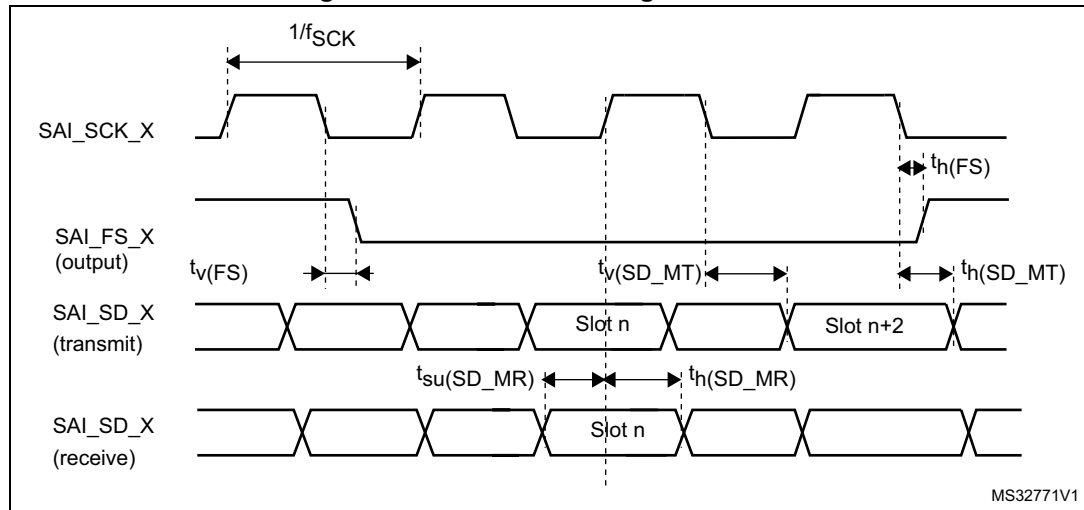
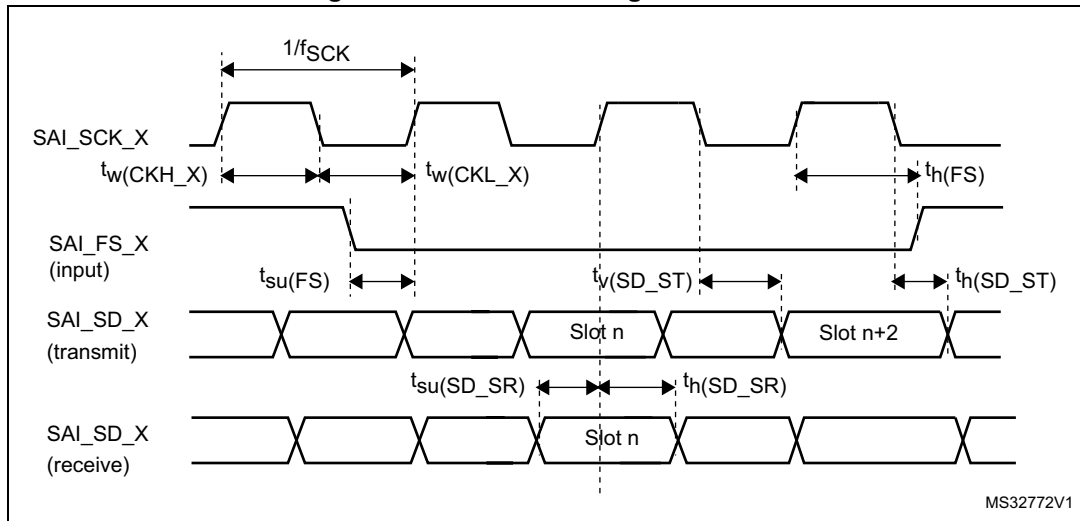


Figure 41. SAI slave timing waveforms



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SDMMC characteristics

Unless otherwise specified, the parameters given in [Table 100](#) for SDIO are derived from tests performed under the ambient temperature, f_{PCLKx} frequency and V_{DD} supply voltage conditions summarized in [Table 23: General operating conditions](#), with the following configuration:

- Output speed is set to $OSPEEDRy[1:0] = 11$
- Capacitive load $C = 30$ pF
- Measurement points are done at CMOS levels: $0.5 \times V_{DD}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output characteristics.

Table 100. SD / MMC dynamic characteristics, $V_{DD}=2.7$ V to 3.6 V⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{PP}	Clock frequency in data transfer mode	-	0	-	50	MHz
-	SDIO_CK/fPCLK2 frequency ratio	-	-	-	4/3	-
$t_{W(CKL)}$	Clock low time	$f_{PP} = 50$ MHz	8	10	-	ns
$t_{W(CKH)}$	Clock high time	$f_{PP} = 50$ MHz	8	10	-	ns
CMD, D inputs (referenced to CK) in MMC and SD HS mode						
t_{ISU}	Input setup time HS	$f_{PP} = 50$ MHz	2	-	-	ns
t_{IH}	Input hold time HS	$f_{PP} = 50$ MHz	4.5	-	-	ns
CMD, D outputs (referenced to CK) in MMC and SD HS mode						
t_{OV}	Output valid time HS	$f_{PP} = 50$ MHz	-	12	14	ns
t_{OH}	Output hold time HS	$f_{PP} = 50$ MHz	9	-	-	ns
CMD, D inputs (referenced to CK) in SD default mode						
t_{ISUD}	Input setup time SD	$f_{PP} = 50$ MHz	2	-	-	ns
t_{IHD}	Input hold time SD	$f_{PP} = 50$ MHz	4.5	-	-	ns

Table 100. SD / MMC dynamic characteristics, $V_{DD}=2.7\text{ V}$ to $3.6\text{ V}^{(1)}$ (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
CMD, D outputs (referenced to CK) in SD default mode						
t_{OVD}	Output valid default time SD	$f_{PP} = 50\text{ MHz}$	-	4.5	5	ns
t_{OHD}	Output hold default time SD	$f_{PP} = 50\text{ MHz}$	0	-	-	ns

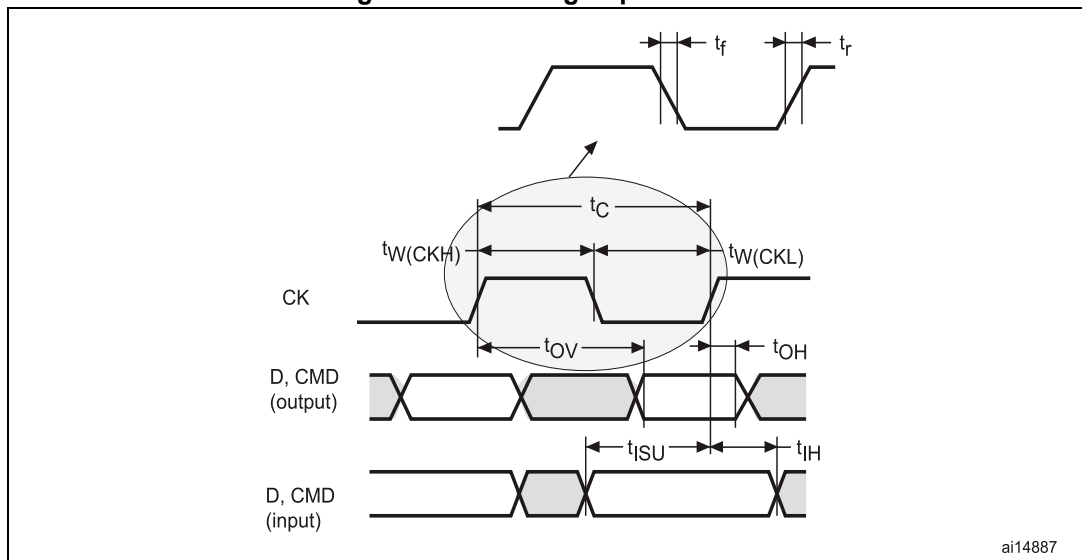
1. Guaranteed by characterization results.

Table 101. eMMC dynamic characteristics, $V_{DD} = 1.71\text{ V}$ to $1.9\text{ V}^{(1)(2)}$

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f_{PP}	Clock frequency in data transfer mode	-	0	-	50	MHz
-	SDIO_CK/ f_{PCLK2} frequency ratio	-	-	-	4/3	-
$t_{W(CKL)}$	Clock low time	$f_{PP} = 50\text{ MHz}$	8	10	-	ns
$t_{W(CKH)}$	Clock high time	$f_{PP} = 50\text{ MHz}$	8	10	-	ns
CMD, D inputs (referenced to CK) in eMMC mode						
t_{ISU}	Input setup time HS	$f_{PP} = 50\text{ MHz}$	0	-	-	ns
t_{IH}	Input hold time HS	$f_{PP} = 50\text{ MHz}$	5	-	-	ns
CMD, D outputs (referenced to CK) in eMMC mode						
t_{OV}	Output valid time HS	$f_{PP} = 50\text{ MHz}$	-	13.5	15.5	ns
t_{OH}	Output hold time HS	$f_{PP} = 50\text{ MHz}$	9	-	-	ns

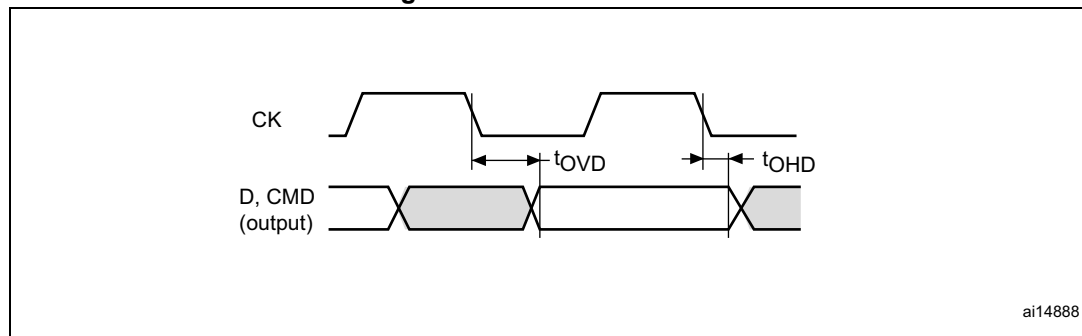
1. Guaranteed by characterization results.

2. $C_{LOAD} = 20\text{ pF}$.

Figure 42. SDIO high-speed mode

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Figure 43. SD default mode



USB OTG full speed (FS) characteristics

The STM32L476xx USB interface is fully compliant with the USB specification version 2.0 and is USB-IF certified (for Full-speed device operation).

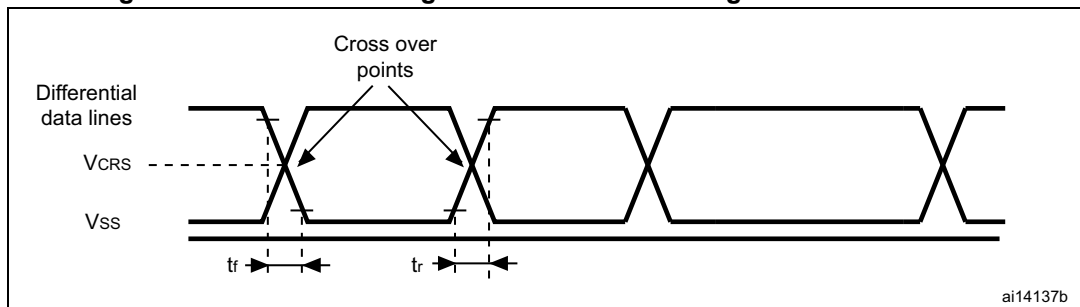
Table 102. USB OTG DC electrical characteristics

Symbol	Parameter	Conditions	Min ⁽¹⁾	Typ	Max ⁽¹⁾	Unit
V_{DDUSB}	USB OTG full speed transceiver operating voltage	-	3.0 ⁽²⁾	-	3.6	V
$V_{DI}^{(3)}$	Differential input sensitivity	Over VCM range	0.2	-	-	V
$V_{CM}^{(3)}$	Differential input common mode range	Includes V_{DI} range	0.8	-	2.5	
$V_{SE}^{(3)}$	Single ended receiver input threshold	-	0.8	-	2.0	
V_{OL}	Static output level low	R_L of 1.5 k Ω to 3.6 V ⁽⁴⁾	-	-	0.3	V
V_{OH}	Static output level high	R_L of 15 k Ω to $V_{SS}^{(4)}$	2.8	-	3.6	
$R_{PD}^{(3)}$	Pull down resistor on PA11, PA12 (USB_FS_DP/DM)	$V_{IN} = V_{DD}$	14.25	-	24.8	k Ω
$R_{PU}^{(3)}$	Pull Up Resistor on PA12 (USB_FS_DP)	$V_{IN} = V_{SS}$, during idle	0.9	1.25	1.575	k Ω
	Pull Up Resistor on PA12 (USB_FS_DP)	$V_{IN} = V_{SS}$ during reception	1.425	2.25	3.09	k Ω
	Pull Up Resistor on PA10 (OTG_FS_ID)	-	-	-	14.5	k Ω

1. All the voltages are measured from the local ground potential.
2. The USB OTG full speed transceiver functionality is ensured down to 2.7 V but not the full USB full speed electrical characteristics which are degraded in the 2.7-to-3.0 V V_{DD} voltage range.
3. Guaranteed by design.
4. R_L is the load connected on the USB OTG full speed drivers.

Note: When VBUS sensing feature is enabled, PA9 should be left at its default state (floating input), not as alternate function. A typical 200 μ A current consumption of the sensing block (current to voltage conversion to determine the different sessions) can be observed on PA9 when the feature is enabled.

Figure 44. USB OTG timings – definition of data signal rise and fall time

Table 103. USB OTG electrical characteristics⁽¹⁾

Driver characteristics					
Symbol	Parameter	Conditions	Min	Max	Unit
t_{rLS}	Rise time in LS ⁽²⁾	$C_L = 200$ to 600 pF	75	300	ns
t_{fLS}	Fall time in LS ⁽²⁾	$C_L = 200$ to 600 pF	75	300	ns
t_{rfmLS}	Rise/ fall time matching in LS	t_r/t_f	80	125	%
t_{rFS}	Rise time in FS ⁽²⁾	$C_L = 50$ pF	4	20	ns
t_{fFS}	Fall time in FS ⁽²⁾	$C_L = 50$ pF	4	20	ns
t_{rfmFS}	Rise/ fall time matching in FS	t_r/t_f	90	111	%
V_{CRS}	Output signal crossover voltage (LS/FS)	-	1.3	2.0	V
Z_{DRV}	Output driver impedance ⁽³⁾	Driving high or low	28	44	Ω

1. Guaranteed by design.
2. Measured from 10% to 90% of the data signal. For more detailed informations, please refer to USB Specification - Chapter 7 (version 2.0).
3. No external termination series resistors are required on DP (D+) and DM (D-) pins since the matching impedance is included in the embedded driver.

Table 104. USB BCD DC electrical characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
$I_{DD(USBBCD)}$	Primary detection mode consumption	-	-	-	300	μA
	Secondary detection mode consumption	-	-	-	300	μA
RDAT_LKG	Data line leakage resistance	-	300	-	-	k Ω
VDAT_LKG	Data line leakage voltage	-	0.0	-	3.6	V
RDCP_DAT	Dedicated charging port resistance across D+/D-	-	-	-	200	Ω
VLGC_HI	Logic high	-	2.0	-	3.6	V
VLGC_LOW	Logic low	-	-	-	0.8	V
VLGC	Logic threshold	-	0.8	-	2.0	V

Table 104. USB BCD DC electrical characteristics⁽¹⁾ (continued)

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
V _{DAT_REF}	Data detect voltage	-	0.25	-	0.4	V
V _{DP_SRC}	D+ source voltage	-	0.5	-	0.7	V
V _{DM_SRC}	D- source voltage	-	0.5	-	0.7	V
I _{DP_SINK}	D+ sink current	-	25	-	175	μA
I _{DM_SINK}	D- sink current	-	25	-	175	μA

1. Guaranteed by design.

CAN (controller area network) interface

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output alternate function characteristics (CAN_TX and CAN_RX).

6.3.29 FSMC characteristics

Unless otherwise specified, the parameters given in [Table 105](#) to [Table 118](#) for the FMC interface are derived from tests performed under the ambient temperature, f_{HCLK} frequency and V_{DD} supply voltage conditions summarized in [Table 23](#), with the following configuration:

- Output speed is set to $OSPEEDRy[1:0] = 11$
- Capacitive load $C = 30$ pF
- Measurement points are done at CMOS levels: $0.5V_{DD}$

Refer to [Section 6.3.14: I/O port characteristics](#) for more details on the input/output characteristics.

Asynchronous waveforms and timings

[Figure 45](#) through [Figure 48](#) represent asynchronous waveforms and [Table 105](#) through [Table 112](#) provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- $AddressSetupTime = 0x1$
- $AddressHoldTime = 0x1$
- $DataSetupTime = 0x1$ (except for asynchronous NWAIT mode, $DataSetupTime = 0x5$)
- $BusTurnAroundDuration = 0x0$

In all timing tables, the THCLK is the HCLK clock period.

Figure 45. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms

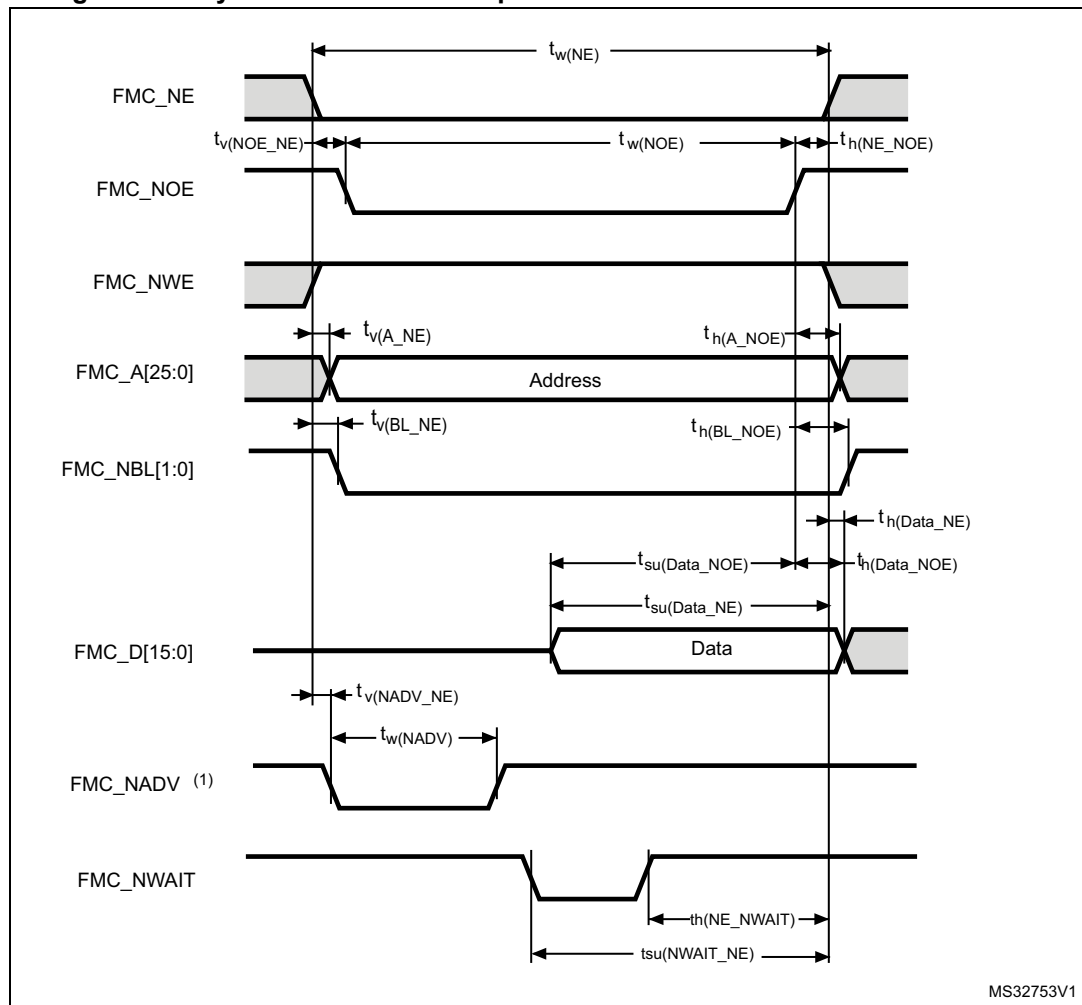


Table 105. Asynchronous non-multiplexed SRAM/PSRAM/NOR read timings⁽¹⁾⁽²⁾

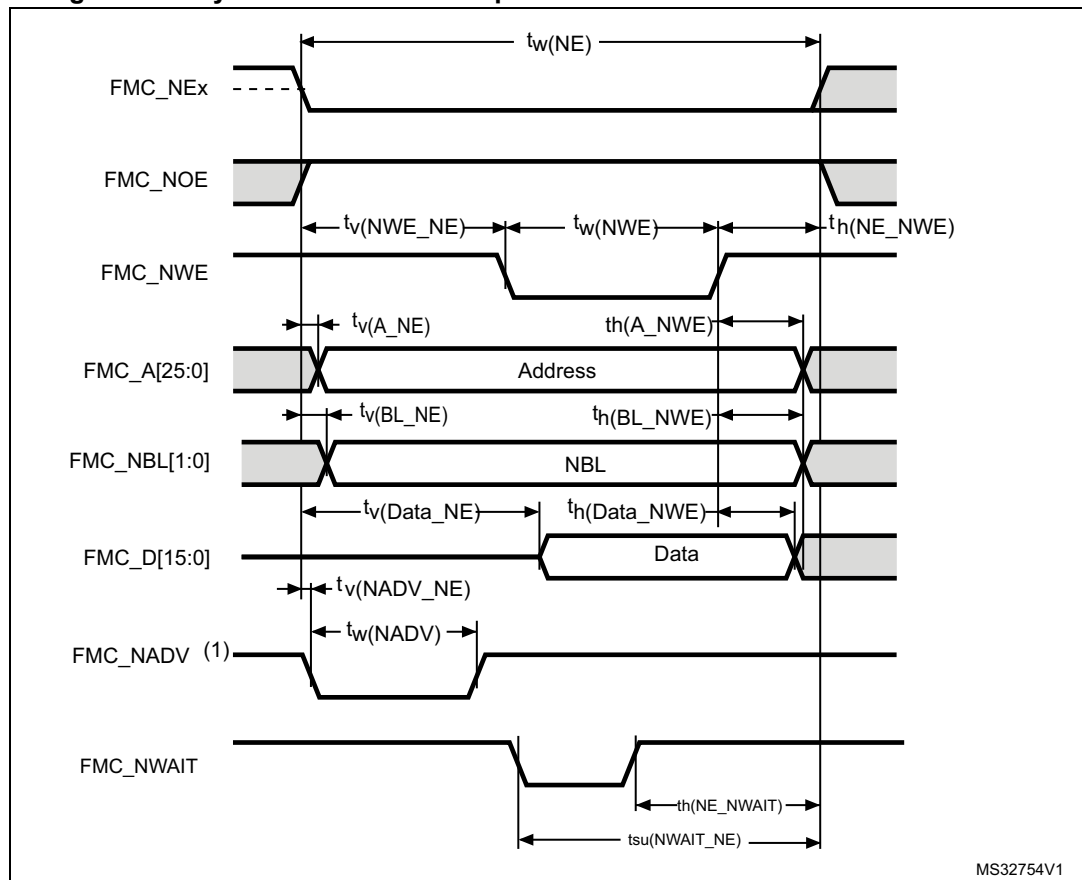
Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$2T_{HCLK}-0.5$	$2T_{HCLK}+0.5$	ns
$t_{v(NOE_NE)}$	FMC_NEx low to FMC_NOE low	0	1	
$t_{w(NOE)}$	FMC_NOE low time	$2T_{HCLK}-0.5$	$2T_{HCLK}+1$	
$t_{h(NE_NOE)}$	FMC_NOE high to FMC_NE high hold time	0	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	3.5	
$t_{h(A_NOE)}$	Address hold time after FMC_NOE high	0	-	
$t_{v(BL_NE)}$	FMC_NEx low to FMC_BL valid	-	2	
$t_{h(BL_NOE)}$	FMC_BL hold time after FMC_NOE high	0	-	
$t_{su(Data_NE)}$	Data to FMC_NEx high setup time	$T_{HCLK}-1$	-	
$t_{su(Data_NOE)}$	Data to FMC_NOEx high setup time	$T_{HCLK}-0.5$	-	
$t_{h(Data_NOE)}$	Data hold time after FMC_NOE high	0	-	
$t_{h(Data_NE)}$	Data hold time after FMC_NEx high	0	-	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	-	1	
$t_{w(NADV)}$	FMC_NADV low time	-	$T_{HCLK}+0.5$	

1. CL = 30 pF.
2. Guaranteed by characterization results.

Table 106. Asynchronous non-multiplexed SRAM/PSRAM/NOR read-NWAIT timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$7T_{HCLK}-0.5$	$7T_{HCLK}+0.5$	ns
$t_{w(NOE)}$	FMC_NWE low time	$5T_{HCLK}-0.5$	$5T_{HCLK}+0.5$	
$t_{w(NWAIT)}$	FMC_NWAIT low time	$T_{HCLK}-0.5$	-	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$5T_{HCLK}+2$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}$	-	

1. CL = 30 pF.
2. Guaranteed by characterization results.

Figure 46. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms**Table 107. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings⁽¹⁾⁽²⁾**

Symbol	Parameter	Min	Max	Unit
$t_w(NE)$	FMC_NE low time	$3T_{HCLK}-1$	$3T_{HCLK}+2$	ns
$t_v(NWE_NE)$	FMC_NEx low to FMC_NWE low	$T_{HCLK}-0.5$	$T_{HCLK}+1.5$	
$t_w(NWE)$	FMC_NWE low time	$T_{HCLK}-1$	$T_{HCLK}+1$	
$t_h(NE_NWE)$	FMC_NWE high to FMC_NE high hold time	$T_{HCLK}-0.5$	-	
$t_v(A_NE)$	FMC_NEx low to FMC_A valid	-	0	
$t_h(A_NWE)$	Address hold time after FMC_NWE high	$T_{HCLK}-1$	-	
$t_v(BL_NE)$	FMC_NEx low to FMC_BL valid	-	1.5	
$t_h(BL_NWE)$	FMC_BL hold time after FMC_NWE high	$T_{HCLK}-0.5$	-	
$t_v(Data_NE)$	Data to FMC_NEx low to Data valid	-	$T_{HCLK}+4$	
$t_h(Data_NWE)$	Data hold time after FMC_NWE high	$T_{HCLK}+1$	-	
$t_v(NADV_NE)$	FMC_NEx low to FMC_NADV low	-	1	
$t_w(NADV)$	FMC_NADV low time	-	$T_{HCLK}+0.5$	

1. CL = 30 pF.

2. Guaranteed by characterization results.

Table 108. Asynchronous non-multiplexed SRAM/PSRAM/NOR write-NWAIT timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$8T_{HCLK}+0.5$	$8T_{HCLK}+0.5$	ns
$t_{w(NWE)}$	FMC_NWE low time	$6T_{HCLK}-0.5$	$6T_{HCLK}+0.5$	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$6T_{HCLK}+2$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}+2$	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

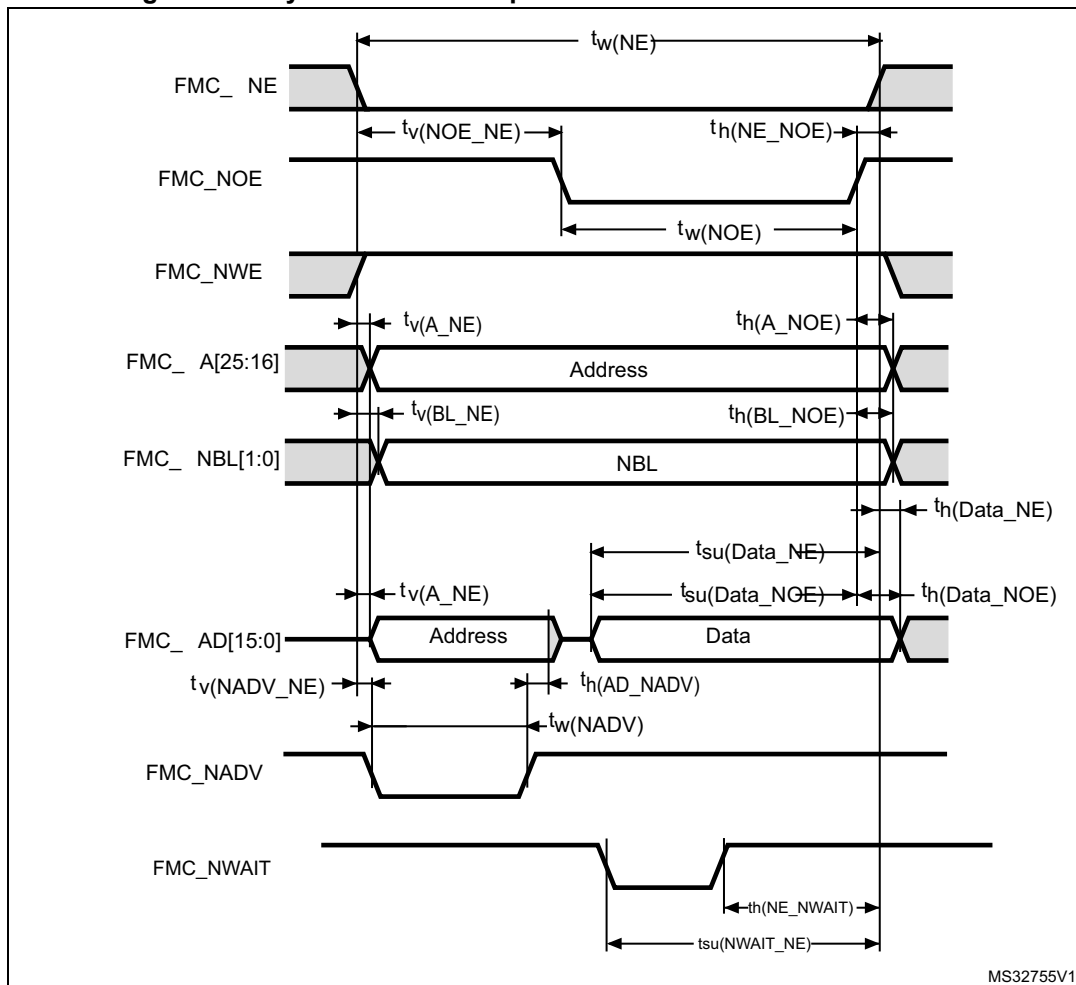
Figure 47. Asynchronous multiplexed PSRAM/NOR read waveforms

Table 109. Asynchronous multiplexed PSRAM/NOR read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$3T_{HCLK}-0.5$	$3T_{HCLK}+2$	ns
$t_{v(NOE_NE)}$	FMC_NEx low to FMC_NOE low	$2T_{HCLK}-0.5$	$2T_{HCLK}+0.5$	
$t_{w(NOE)}$	FMC_NOE low time	$T_{HCLK}+0.5$	$T_{HCLK}+1$	
$t_{h(NE_NOE)}$	FMC_NOE high to FMC_NE high hold time	0	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	3	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	0	1	
$t_{w(NADV)}$	FMC_NADV low time	$T_{HCLK}-0.5$	$T_{HCLK}+1$	
$t_{h(AD_NADV)}$	FMC_AD(address) valid hold time after FMC_NADV high	0	-	
$t_{h(A_NOE)}$	Address hold time after FMC_NOE high	$T_{HCLK}-0.5$	-	
$t_{h(BL_NOE)}$	FMC_BL time after FMC_NOE high	0	-	
$t_{v(BL_NE)}$	FMC_NEx low to FMC_BL valid	-	2	
$t_{su(Data_NE)}$	Data to FMC_NEx high setup time	$T_{HCLK}-2$	-	
$t_{su(Data_NOE)}$	Data to FMC_NOE high setup time	$T_{HCLK}-1$	-	
$t_{h(Data_NE)}$	Data hold time after FMC_NEx high	0	-	
$t_{h(Data_NOE)}$	Data hold time after FMC_NOE high	0	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

Table 110. Asynchronous multiplexed PSRAM/NOR read-NWAIT timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$8T_{HCLK}+2$	$8T_{HCLK}+4$	ns
$t_{w(NOE)}$	FMC_NWE low time	$5T_{HCLK}-1$	$5T_{HCLK}+1.5$	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$5T_{HCLK}+1.5$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}+1$	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

Figure 48. Asynchronous multiplexed PSRAM/NOR write waveforms

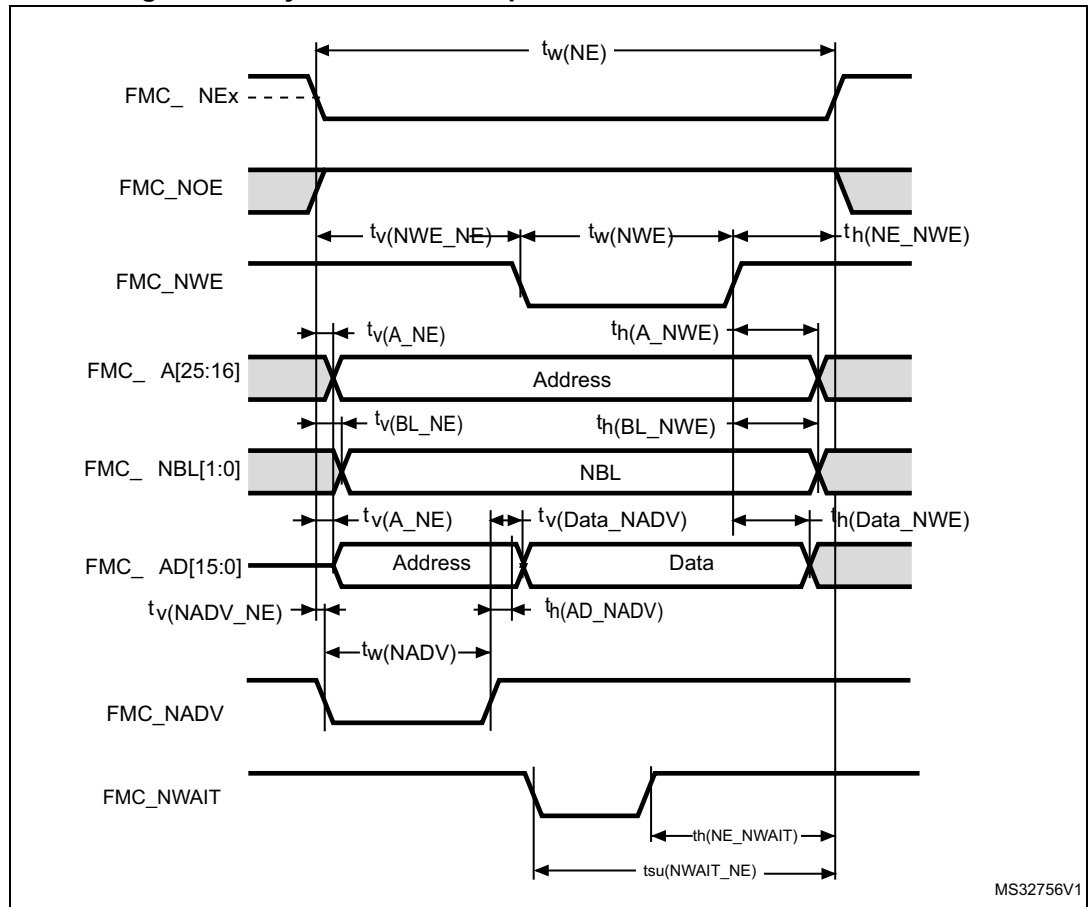


Table 111. Asynchronous multiplexed PSRAM/NOR write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$4T_{HCLK}-0.5$	$4T_{HCLK}+2$	ns
$t_{v(NWE_NE)}$	FMC_NEx low to FMC_NWE low	$T_{HCLK}-0.5$	$T_{HCLK}+1$	
$t_{w(NWE)}$	FMC_NWE low time	$2xT_{HCLK}-1.5$	$2xT_{HCLK}+1.5$	
$t_{h(NE_NWE)}$	FMC_NWE high to FMC_NE high hold time	$T_{HCLK}-0.5$	-	
$t_{v(A_NE)}$	FMC_NEx low to FMC_A valid	-	3	
$t_{v(NADV_NE)}$	FMC_NEx low to FMC_NADV low	0	1	
$t_{w(NADV)}$	FMC_NADV low time	$T_{HCLK}-0.5$	$T_{HCLK}+1$	
$t_{h(AD_NADV)}$	FMC_AD(address) valid hold time after FMC_NADV high	$T_{HCLK}-2$	-	
$t_{h(A_NWE)}$	Address hold time after FMC_NWE high	$T_{HCLK}-1$	-	
$t_{h(BL_NWE)}$	FMC_BL hold time after FMC_NWE high	$T_{HCLK}+0.5$	-	
$t_{v(BL_NE)}$	FMC_NEx low to FMC_BL valid	-	1.5	
$t_{v(Data_NADV)}$	FMC_NADV high to Data valid	-	$T_{HCLK} + 4$	
$t_{h(Data_NWE)}$	Data hold time after FMC_NWE high	$T_{HCLK} + 0.5$	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

Table 112. Asynchronous multiplexed PSRAM/NOR write-NWAIT timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$9T_{HCLK}-0.5$	$9T_{HCLK}+2$	ns
$t_{w(NWE)}$	FMC_NWE low time	$7T_{HCLK}-1.5$	$7T_{HCLK}+1.5$	
$t_{su(NWAIT_NE)}$	FMC_NWAIT valid before FMC_NEx high	$6T_{HCLK}+2$	-	
$t_{h(NE_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}-3$	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

Synchronous waveforms and timings

[Figure 49](#) through [Figure 52](#) represent synchronous waveforms and [Table 113](#) through [Table 116](#) provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- BurstAccessMode = FMC_BurstAccessMode_Enable
- MemoryType = FMC_MemoryType_CRAM
- WriteBurst = FMC_WriteBurst_Enable
- CLKDivision = 1
- DataLatency = 1 for NOR Flash; DataLatency = 0 for PSRAM

In all timing tables, the T_{HCLK} is the HCLK clock period.

Figure 49. Synchronous multiplexed NOR/PSRAM read timings

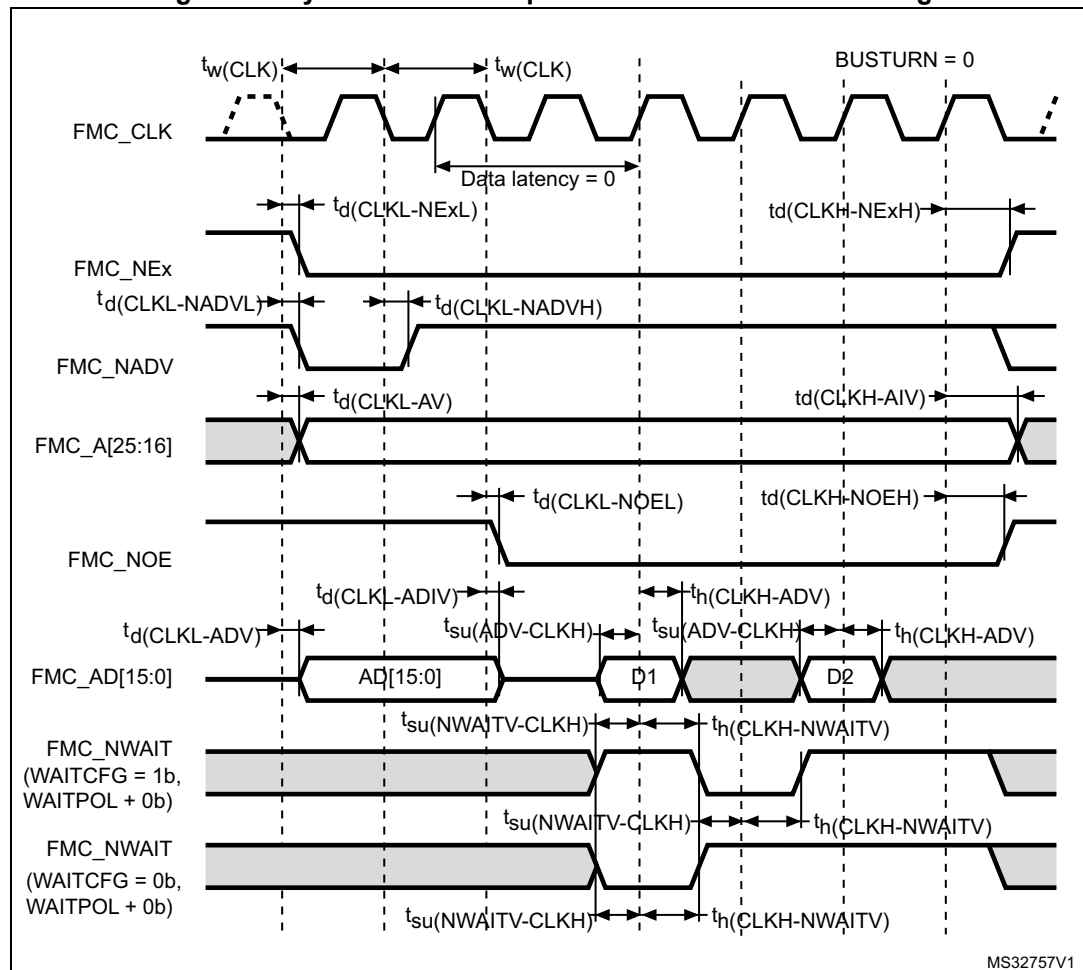


Table 113. Synchronous multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FMC_CLK period	$2T_{HCLK}-1$	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low ($x=0..2$)	-	2	
$t_{d(CLKH-NExH)}$	FMC_CLK high to FMC_NEx high ($x=0..2$)	$T_{HCLK}+0.5$	-	
$t_{d(CLKL-NADV_L)}$	FMC_CLK low to FMC_NADV low	-	2.5	
$t_{d(CLKL-NADV_H)}$	FMC_CLK low to FMC_NADV high	1	-	
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid ($x=16..25$)	-	3.5	
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid ($x=16..25$)	T_{HCLK}	-	
$t_{d(CLKL-NOEL)}$	FMC_CLK low to FMC_NOE low	-	1.5	
$t_{d(CLKH-NOEH)}$	FMC_CLK high to FMC_NOE high	$T_{HCLK}+1$	-	
$t_{d(CLKL-ADV)}$	FMC_CLK low to FMC_AD[15:0] valid	-	4	
$t_{d(CLKL-ADIV)}$	FMC_CLK low to FMC_AD[15:0] invalid	0	-	
$t_{su(ADV-CLKH)}$	FMC_A/D[15:0] valid data before FMC_CLK high	0	-	
$t_h(CLKH-ADV)$	FMC_A/D[15:0] valid data after FMC_CLK high	2.5	-	
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	0	-	
$t_h(CLKH-NWAIT)$	FMC_NWAIT valid after FMC_CLK high	4	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

Figure 50. Synchronous multiplexed PSRAM write timings

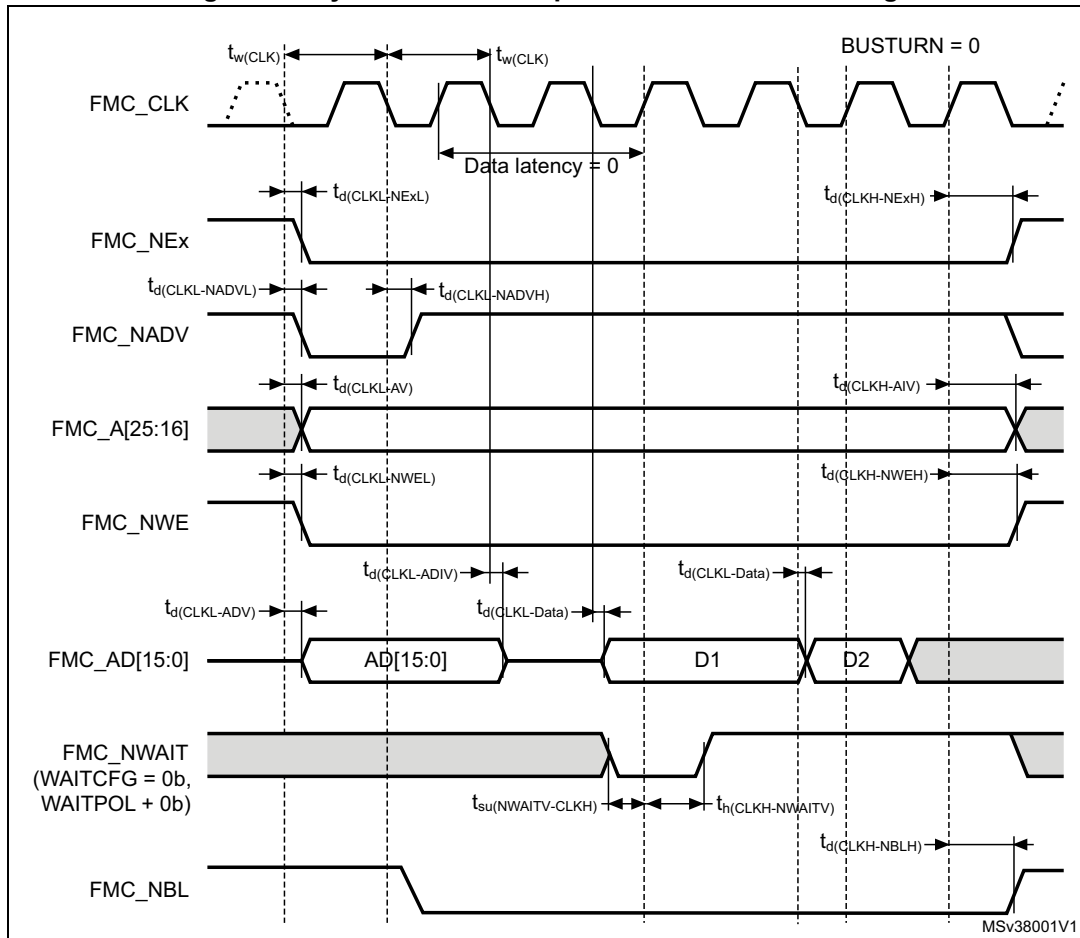


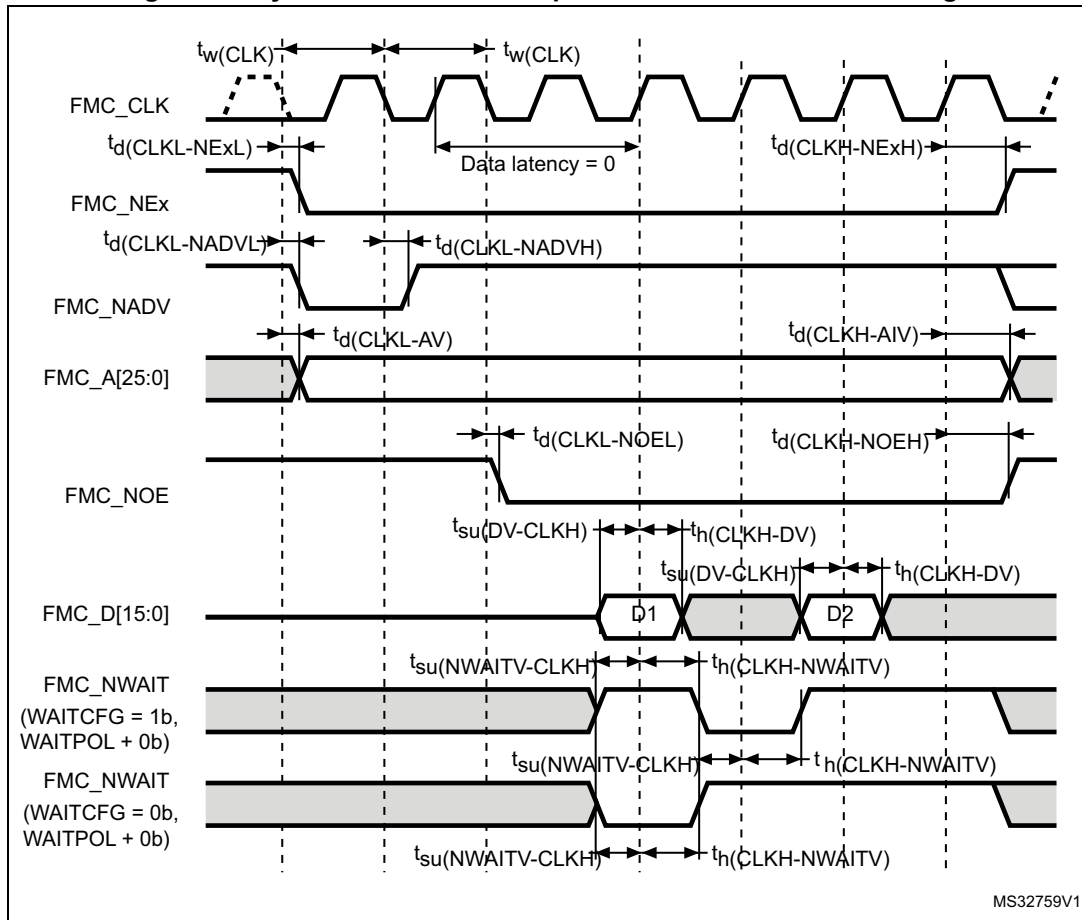
Table 114. Synchronous multiplexed PSRAM write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FMC_CLK period	$2T_{HCLK}-1$	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low ($x=0..2$)	-	2	
$t_{d(CLKH-NExH)}$	FMC_CLK high to FMC_NEx high ($x=0..2$)	$T_{HCLK}+0.5$	-	
$t_{d(CLKL-NADV_L)}$	FMC_CLK low to FMC_NADV low	-	2.5	
$t_{d(CLKL-NADV_H)}$	FMC_CLK low to FMC_NADV high	1	-	
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid ($x=16..25$)	-	3.5	
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid ($x=16..25$)	T_{HCLK}	-	
$t_{d(CLKL-NWEL)}$	FMC_CLK low to FMC_NWE low	-	2	
$t_{d(CLKH-NWEH)}$	FMC_CLK high to FMC_NWE high	$T_{HCLK}+1$	-	
$t_{d(CLKL-ADV)}$	FMC_CLK low to FMC_AD[15:0] valid	-	4	
$t_{d(CLKL-ADIV)}$	FMC_CLK low to FMC_AD[15:0] invalid	0	-	
$t_{d(CLKL-DATA)}$	FMC_A/D[15:0] valid data after FMC_CLK low	-	5.5	
$t_{d(CLKL-NBLL)}$	FMC_CLK low to FMC_NBL low	-	2.5	
$t_{d(CLKH-NBLH)}$	FMC_CLK high to FMC_NBL high	$T_{HCLK}+1$	-	
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	0	-	
$t_h(CLKH-NWAIT)$	FMC_NWAIT valid after FMC_CLK high	4	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

Figure 51. Synchronous non-multiplexed NOR/PSRAM read timings

Table 115. Synchronous non-multiplexed NOR/PSRAM read timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FMC_CLK period	$2T_{HCLK}$	-	ns
$t_{d(CLKxL-NExL)}$	FMC_CLK low to FMC_NEx low (x=0..2)	-	2.5	
$t_{d(CLKxH-NExH)}$	FMC_CLK high to FMC_NEx high (x= 0...2)	$T_{HCLK}-0.5$	-	
$t_{d(CLKxL-NADVxL)}$	FMC_CLK low to FMC_NADV low	-	2	
$t_{d(CLKxL-NADVxH)}$	FMC_CLK low to FMC_NADV high	0.5	-	
$t_{d(CLKxL-AV)}$	FMC_CLK low to FMC_Ax valid (x=16...25)	-	3.5	
$t_{d(CLKxH-AIV)}$	FMC_CLK high to FMC_Ax invalid (x=16...25)	T_{HCLK}	-	
$t_{d(CLKxL-NOEL)}$	FMC_CLK low to FMC_NOE low	-	2	
$t_{d(CLKxH-NOEH)}$	FMC_CLK high to FMC_NOE high	$T_{HCLK}-0.5$	-	
$t_{su(DV-CLKH)}$	FMC_D[15:0] valid data before FMC_CLK high	0	-	
$t_{h(CLKH-DV)}$	FMC_D[15:0] valid data after FMC_CLK high	5	-	
$t_{su(NWAITV-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	0	-	
$t_{h(CLKH-NWAITV)}$	FMC_NWAIT valid after FMC_CLK high	4	-	

1. CL = 30 pF.
2. Guaranteed by characterization results.

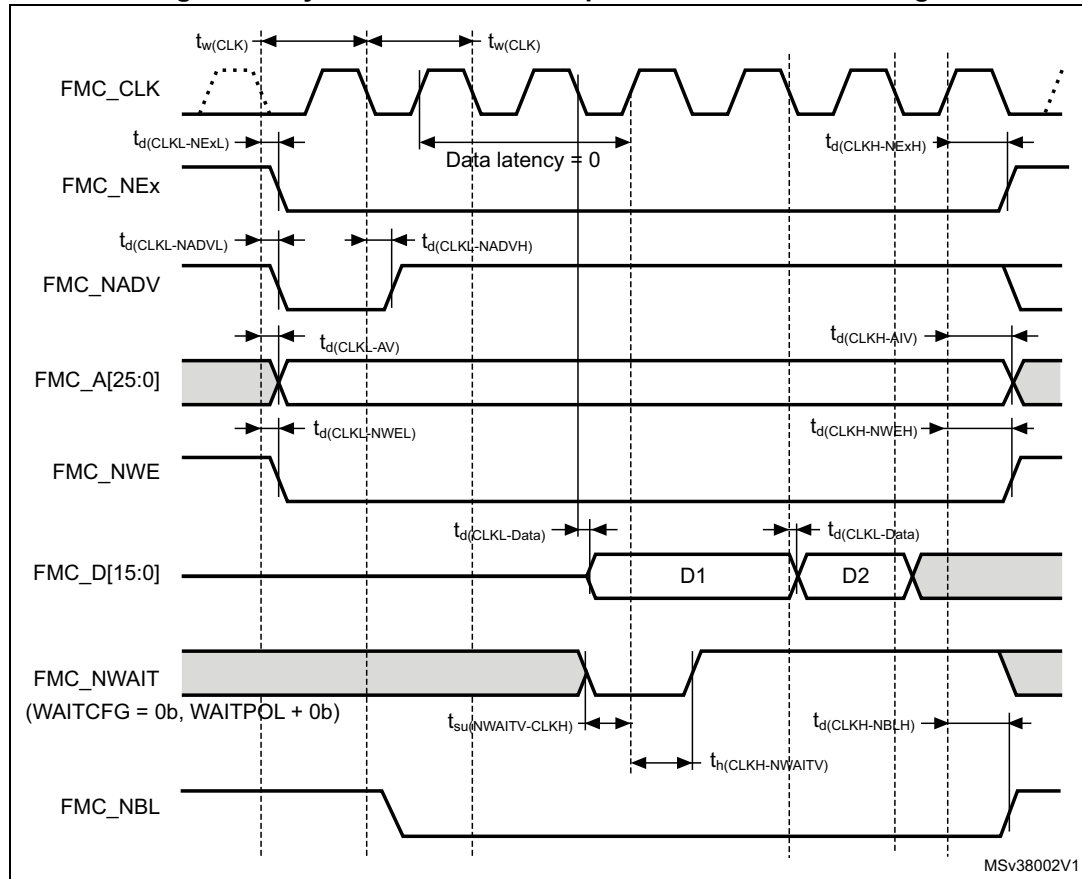
Figure 52. Synchronous non-multiplexed PSRAM write timings

Table 116. Synchronous non-multiplexed PSRAM write timings⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FMC_CLK period	$2T_{HCLK}-0.5$	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low ($x=0..2$)	-	2	
$t_{d(CLKH-NExH)}$	FMC_CLK high to FMC_NEx high ($x=0..2$)	$T_{HCLK}+0.5$	-	
$t_{d(CLKL-NADV_L)}$	FMC_CLK low to FMC_NADV low	-	2	
$t_{d(CLKL-NADV_H)}$	FMC_CLK low to FMC_NADV high	2.5	-	
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid ($x=16..25$)	-	5	
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid ($x=16..25$)	$T_{HCLK}-1$	-	
$t_{d(CLKL-NWE_L)}$	FMC_CLK low to FMC_NWE low	-	2	
$t_{d(CLKH-NWE_H)}$	FMC_CLK high to FMC_NWE high	$T_{HCLK}-1$	-	
$t_{d(CLKL-Data)}$	FMC_D[15:0] valid data after FMC_CLK low	-	4.5	
$t_{d(CLKL-NBL_L)}$	FMC_CLK low to FMC_NBL low	1.5	-	
$t_{d(CLKH-NBL_H)}$	FMC_CLK high to FMC_NBL high	$T_{HCLK}+1$	-	
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	0	-	
$t_h(CLKH-NWAIT)$	FMC_NWAIT valid after FMC_CLK high	4	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

NAND controller waveforms and timings

[Figure 53](#) through [Figure 56](#) represent synchronous waveforms, and [Table 117](#) and [Table 118](#) provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- COM.FMC_SetupTime = 0x02
- COM.FMC_WaitSetupTime = 0x03
- COM.FMC_HoldSetupTime = 0x02
- COM.FMC_HiZSetupTime = 0x03
- ATT.FMC_SetupTime = 0x01
- ATT.FMC_WaitSetupTime = 0x03
- ATT.FMC_HoldSetupTime = 0x02
- ATT.FMC_HiZSetupTime = 0x03
- Bank = FMC_Bank_NAND
- MemoryDataWidth = FMC_MemoryDataWidth_16b
- ECC = FMC_ECC_Enable
- ECCPageSize = FMC_ECCPageSize_512Bytes
- TCLRSetupTime = 0
- TARSetupTime = 0

In all timing tables, the T_{HCLK} is the HCLK clock period.

Figure 53. NAND controller waveforms for read access

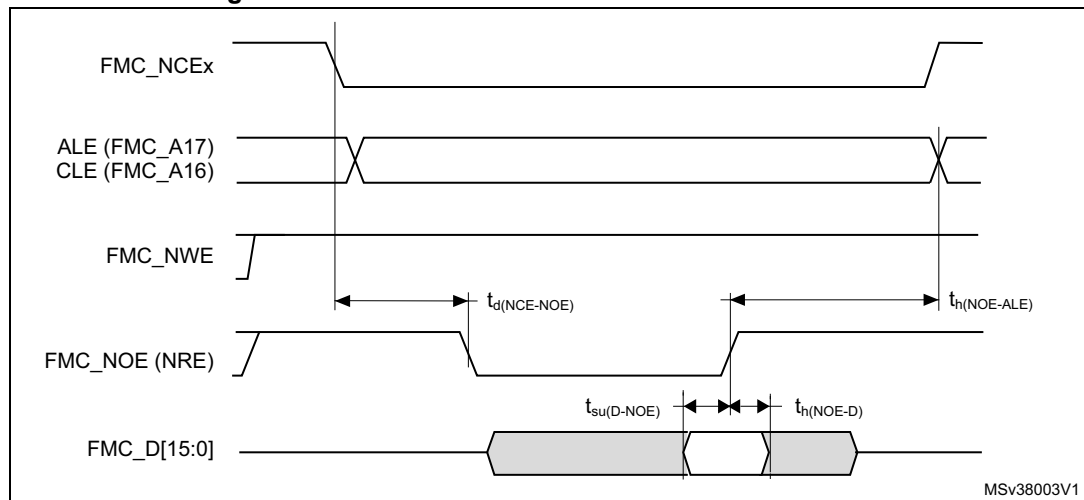


Figure 54. NAND controller waveforms for write access

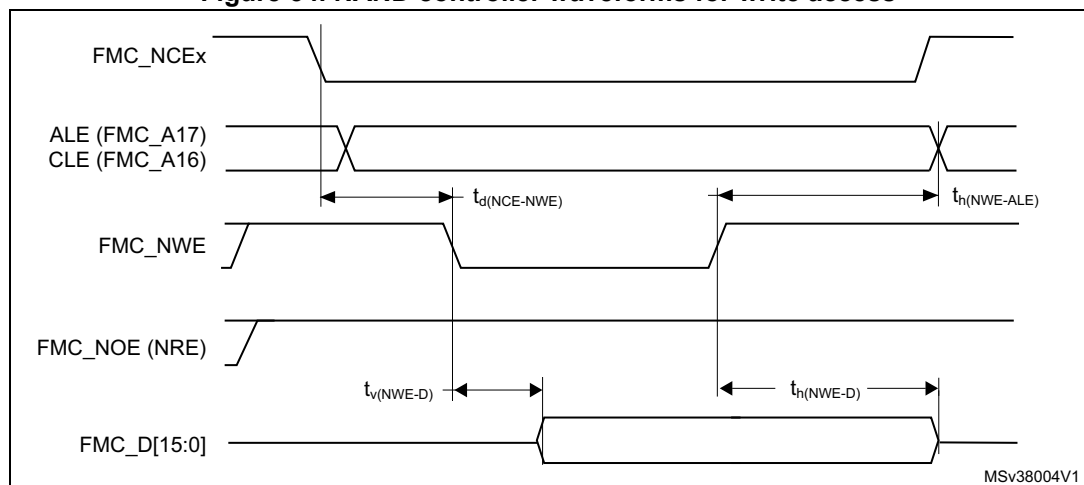


Figure 55. NAND controller waveforms for common memory read access

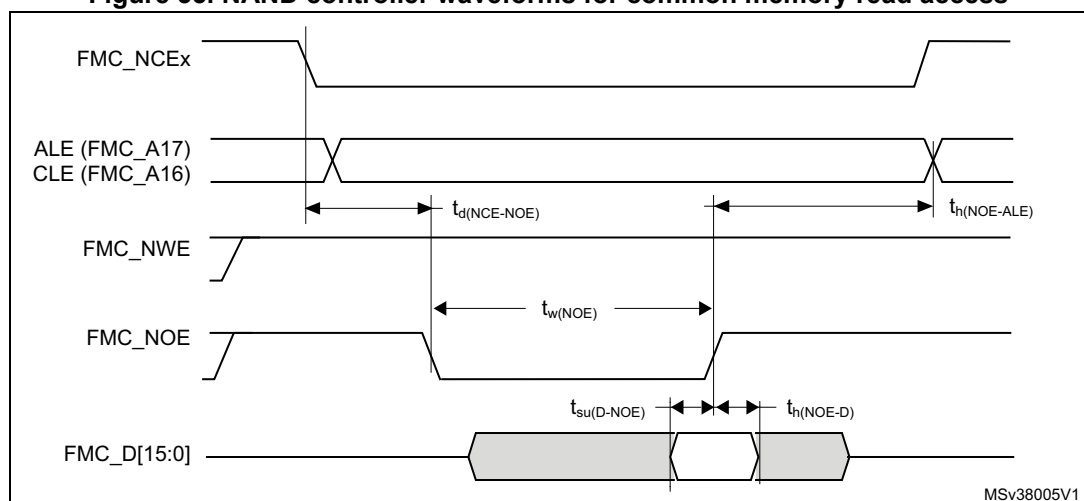
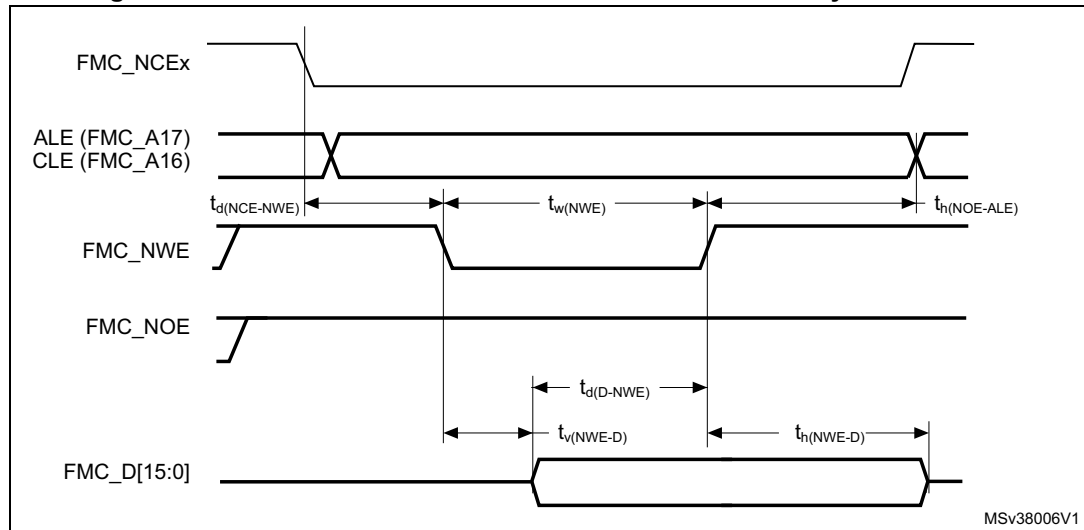


Figure 56. NAND controller waveforms for common memory write access**Table 117. Switching characteristics for NAND Flash read cycles⁽¹⁾⁽²⁾**

Symbol	Parameter	Min	Max	Unit
$T_{w(NOE)}$	FMC_NOE low width	$4T_{HCLK}-1$	$4T_{HCLK}+1$	ns
$T_{su(D-NOE)}$	FMC_D[15-0] valid data before FMC_NOE high	16	-	
$T_{h(NOE-D)}$	FMC_D[15-0] valid data after FMC_NOE high	6	-	
$T_{d(NCE-NOE)}$	FMC_NCE valid before FMC_NOE low	-	$3T_{HCLK}+1$	
$T_{h(NOE-ALE)}$	FMC_NOE high to FMC_ALE invalid	$2T_{HCLK}-2$	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

Table 118. Switching characteristics for NAND Flash write cycles⁽¹⁾⁽²⁾

Symbol	Parameter	Min	Max	Unit
$T_{w(NWE)}$	FMC_NWE low width	$4T_{HCLK}-1$	$4T_{HCLK}+1$	ns
$T_{v(NWE-D)}$	FMC_NWE low to FMC_D[15-0] valid	-	2.5	
$T_{h(NWE-D)}$	FMC_NWE high to FMC_D[15-0] invalid	$3T_{HCLK}-4$	-	
$T_{d(D-NWE)}$	FMC_D[15-0] valid before FMC_NWE high	$5T_{HCLK}-3$	-	
$T_{d(NCE-NWE)}$	FMC_NCE valid before FMC_NWE low	-	$3T_{HCLK}+1$	
$T_{h(NWE-ALE)}$	FMC_NWE high to FMC_ALE invalid	$2T_{HCLK}-2$	-	

1. CL = 30 pF.

2. Guaranteed by characterization results.

6.3.30 SWPMI characteristics

The Single Wire Protocol Master Interface (SWPMI) and the associated SWPMI_IO transceiver are compliant with the ETSI TS 102 613 technical specification.

Table 119. SWPMI electrical characteristics⁽¹⁾

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
t_{SWPSTART}	SWPMI regulator startup time	SWP Class B $2.7\text{ V} \leq V_{\text{DD}} \leq 3.3\text{ V}$	-	-	300	μs
t_{SWPBIT}	SWP bit duration	V_{CORE} voltage range 1	500	-	-	ns
		V_{CORE} voltage range 2	620	-	-	

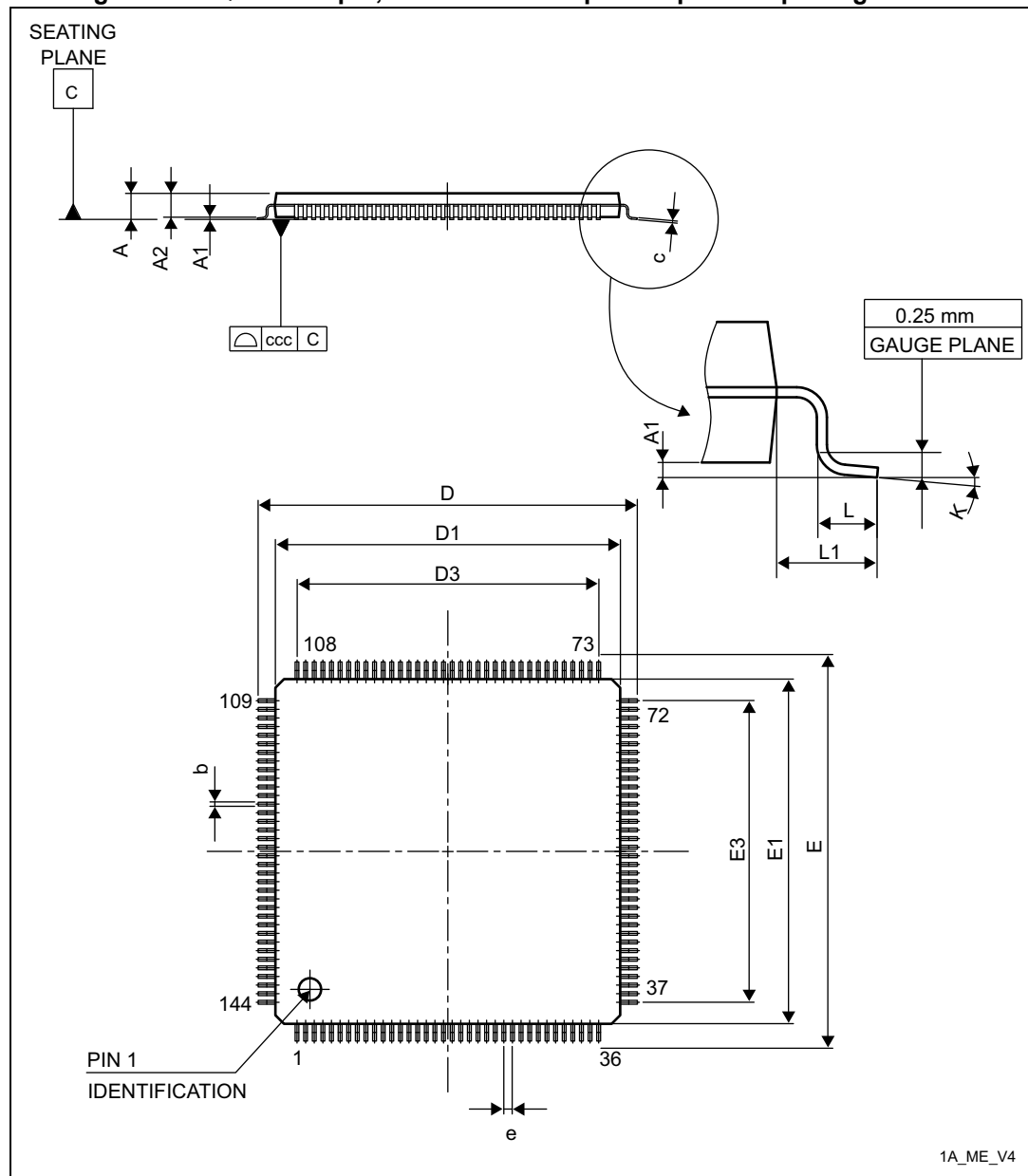
1. Guaranteed by design.

7 Package information

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK[®] packages, depending on their level of environmental compliance. ECOPACK[®] specifications, grade definitions and product status are available at: www.st.com. ECOPACK[®] is an ST trademark.

7.1 LQFP144 package information

Figure 57. LQFP - 144-pin, 20 x 20 mm low-profile quad flat package outline



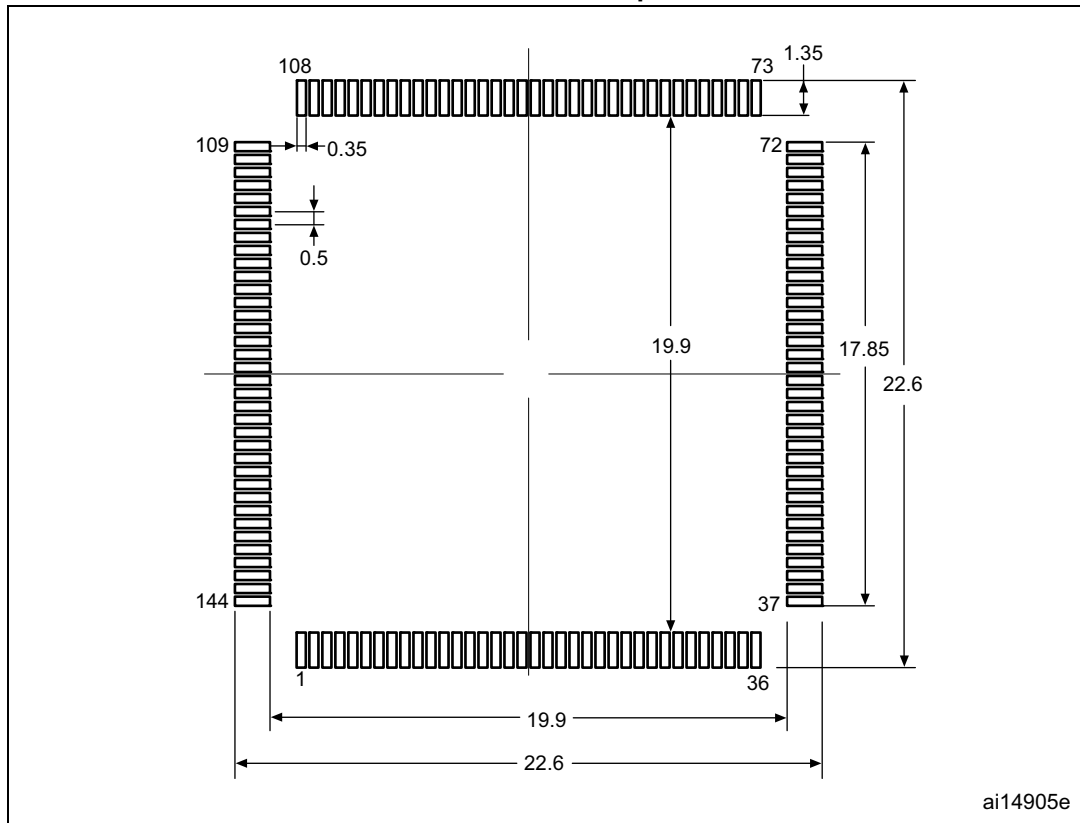
1. Drawing is not to scale.

**Table 120. LQFP - 144-pin, 20 x 20 mm low-profile quad flat package
mechanical data**

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	21.800	22.000	22.200	0.8583	0.8661	0.8740
D1	19.800	20.000	20.200	0.7795	0.7874	0.7953
D3	-	17.500	-	-	0.6890	-
E	21.800	22.000	22.200	0.8583	0.8661	0.8740
E1	19.800	20.000	20.200	0.7795	0.7874	0.7953
E3	-	17.500	-	-	0.6890	-
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 58. LQFP - 144-pin, 20 x 20 mm low-profile quad flat package recommended footprint



1. Dimensions are expressed in millimeters.

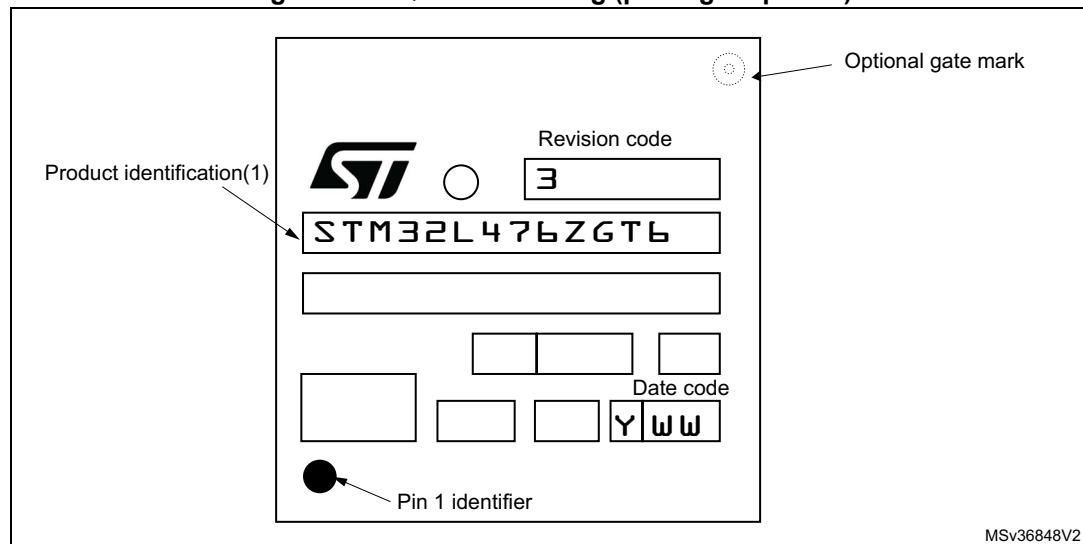
Device marking

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 59. LQFP144 marking (package top view)



1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.2 UFBGA144 package information

Figure 60. UFBGA - 144-pin, 10 x 10 mm, 0.80 mm pitch, ultra fine pitch ball grid array package outline

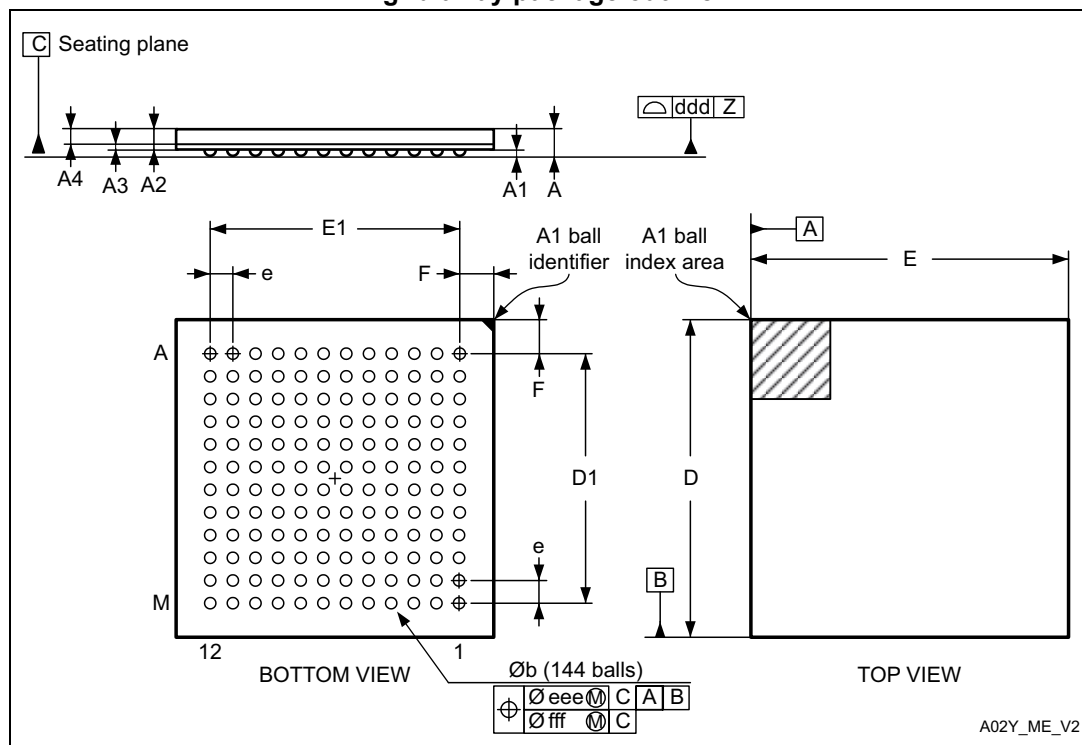


Table 121. UFBGA - 144-pin, 10 x 10 mm, 0.80 mm pitch, ultra fine pitch ball grid array package mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min.	Typ.	Max.	Min.	Typ.	Max.
A	0.460	0.530	0.600	0.0181	0.0209	0.0236
A1	0.050	0.080	0.110	0.0020	0.0031	0.0043
A2	0.400	0.450	0.500	0.0157	0.0177	0.0197
A3	-	0.130	-	-	0.0051	-
A4	-	0.320	-	-	0.0126	-
b	0.360	0.400	0.440	0.0091	0.0110	0.0130
D	9.950	10.000	10.050	0.2736	0.2756	0.2776
D1	8.750	8.800	8.850	0.2343	0.2362	0.2382
E	9.950	10.000	10.050	0.2736	0.2756	0.2776
E1	8.750	8.800	8.850	0.2343	0.2362	0.2382
e	0.750	0.800	0.850	-	0.0197	-
F	0.550	0.600	0.650	0.0177	0.0197	0.0217

Table 121. UFBGA - 144-pin, 10 x 10 mm, 0.80 mm pitch, ultra fine pitch ball grid array package mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min.	Typ.	Max.	Min.	Typ.	Max.
ddd	-	-	0.080	-	-	0.0039
eee	-	-	0.150	-	-	0.0059
fff	-	-	0.080	-	-	0.0020

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 61. UFBGA - 144-pin, 10 x 10 mm, 0.80 mm pitch, ultra fine pitch ball grid array package recommended footprint

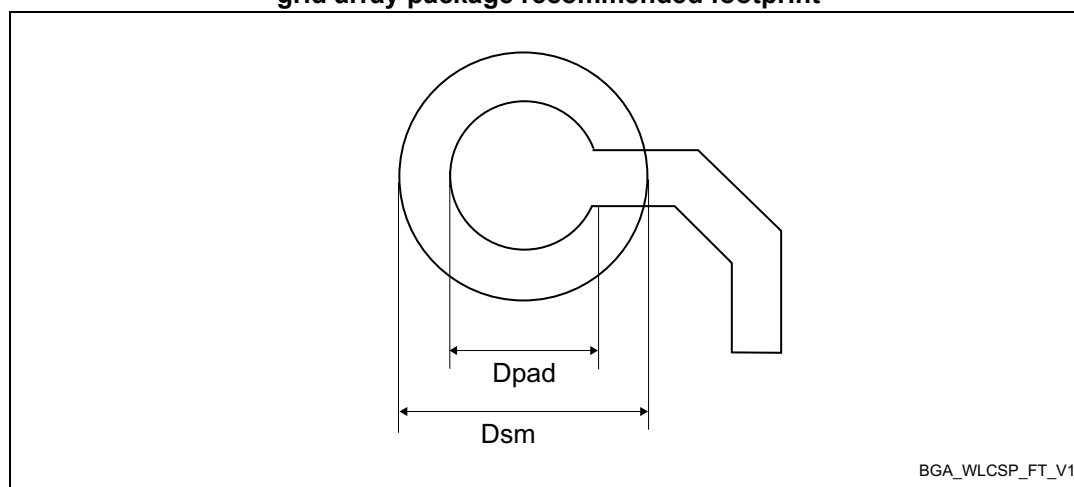


Table 122. UFBGA144 recommended PCB design rules (0.80 mm pitch BGA)

Dimension	Recommended values
Pitch	0.80 mm
Dpad	0.400 mm
Dsm	0.550 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.400 mm
Stencil thickness	Between 0.100 mm and 0.125 mm
Pad trace width	0.120 mm

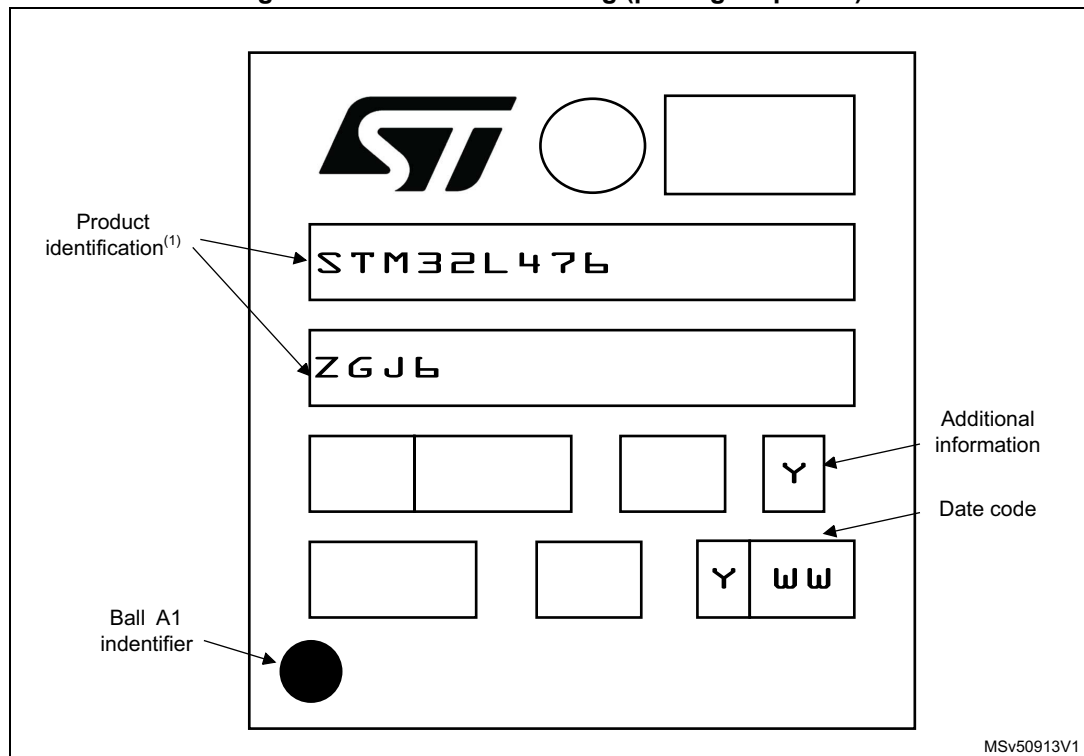
Device marking

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

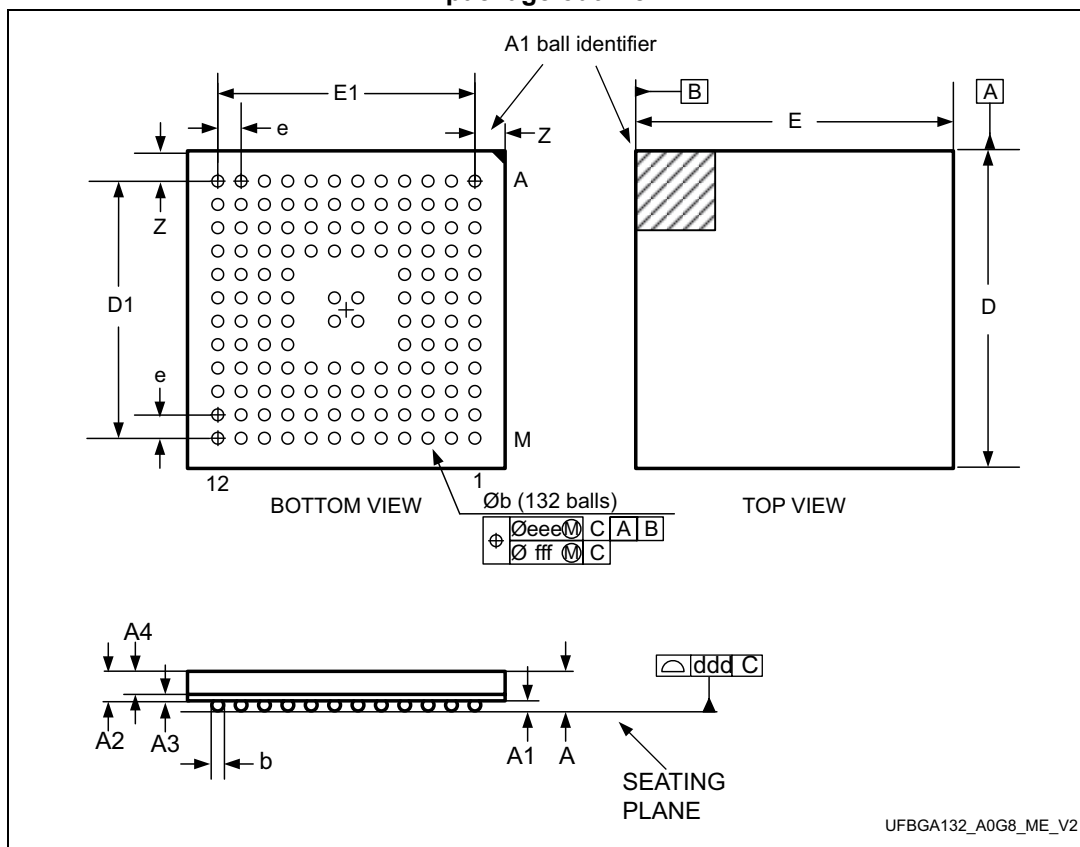
Figure 62. UFBGA144 marking (package top view)



1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.3 UFBGA132 package information

Figure 63. UFBGA - 132-ball, 7 x 7 mm ultra thin fine pitch ball grid array package outline



1. Drawing is not to scale.

Table 123. UFBGA - 132-ball, 7 x 7 mm ultra thin fine pitch ball grid array package mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	-	-	0.600	-	-	0.0236
A1	-	-	0.110	-	-	0.0043
A2	-	0.450	-	-	0.0177	-
A3	-	0.130	-	-	0.0051	-
A4	-	0.320	-	-	0.0126	-
b	0.240	0.290	0.340	0.0094	0.0114	0.0134
D	6.850	7.000	7.150	0.2697	0.2756	0.2815
D1	-	5.500	-	-	0.2165	-
E	6.850	7.000	7.150	0.2697	0.2756	0.2815
E1	-	5.500	-	-	0.2165	-

Table 123. UFBGA - 132-ball, 7 x 7 mm ultra thin fine pitch ball grid array package mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
e	-	0.500	-	-	0.0197	-
Z	-	0.750	-	-	0.0295	-
ddd	-	0.080	-	-	0.0031	-
eee	-	0.150	-	-	0.0059	-
fff	-	0.050	-	-	0.0020	-

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 64. UFBGA - 132-ball, 7 x 7 mm ultra thin fine pitch ball grid array package recommended footprint

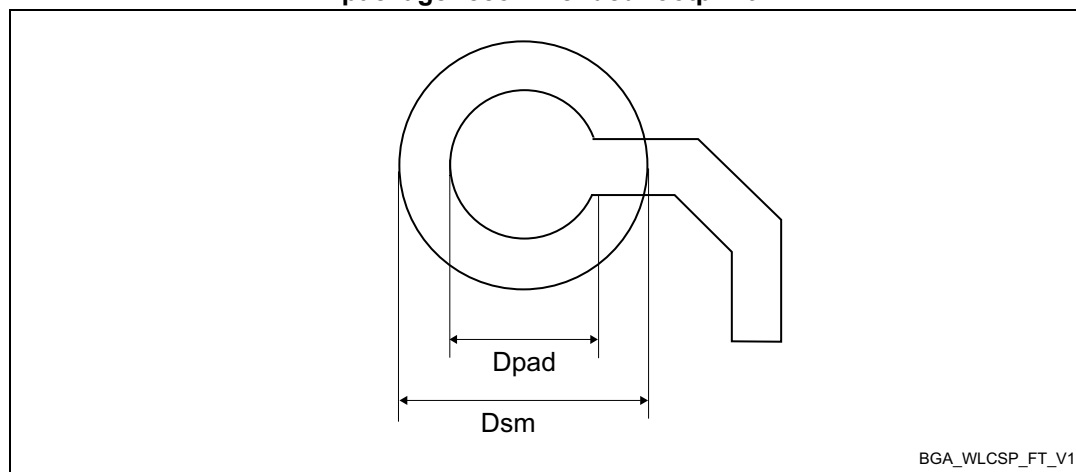


Table 124. UFBGA132 recommended PCB design rules (0.5 mm pitch BGA)

Dimension	Recommended values
Pitch	0.5 mm
Dpad	0.280 mm
Dsm	0.370 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.280 mm
Stencil thickness	Between 0.100 mm and 0.125 mm
Pad trace width	0.100 mm
Ball diameter	0.280 mm

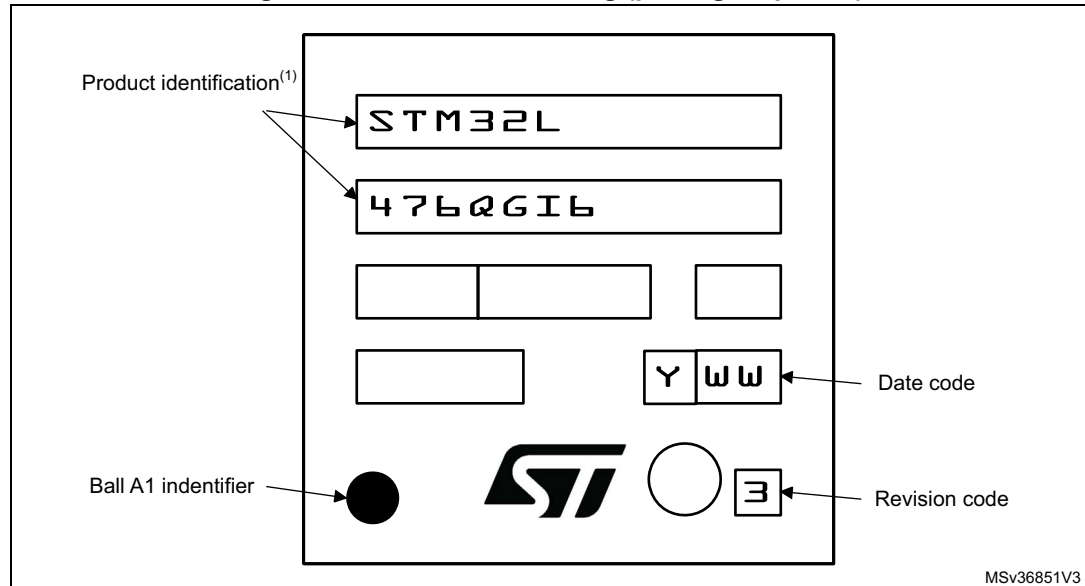
Device marking

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

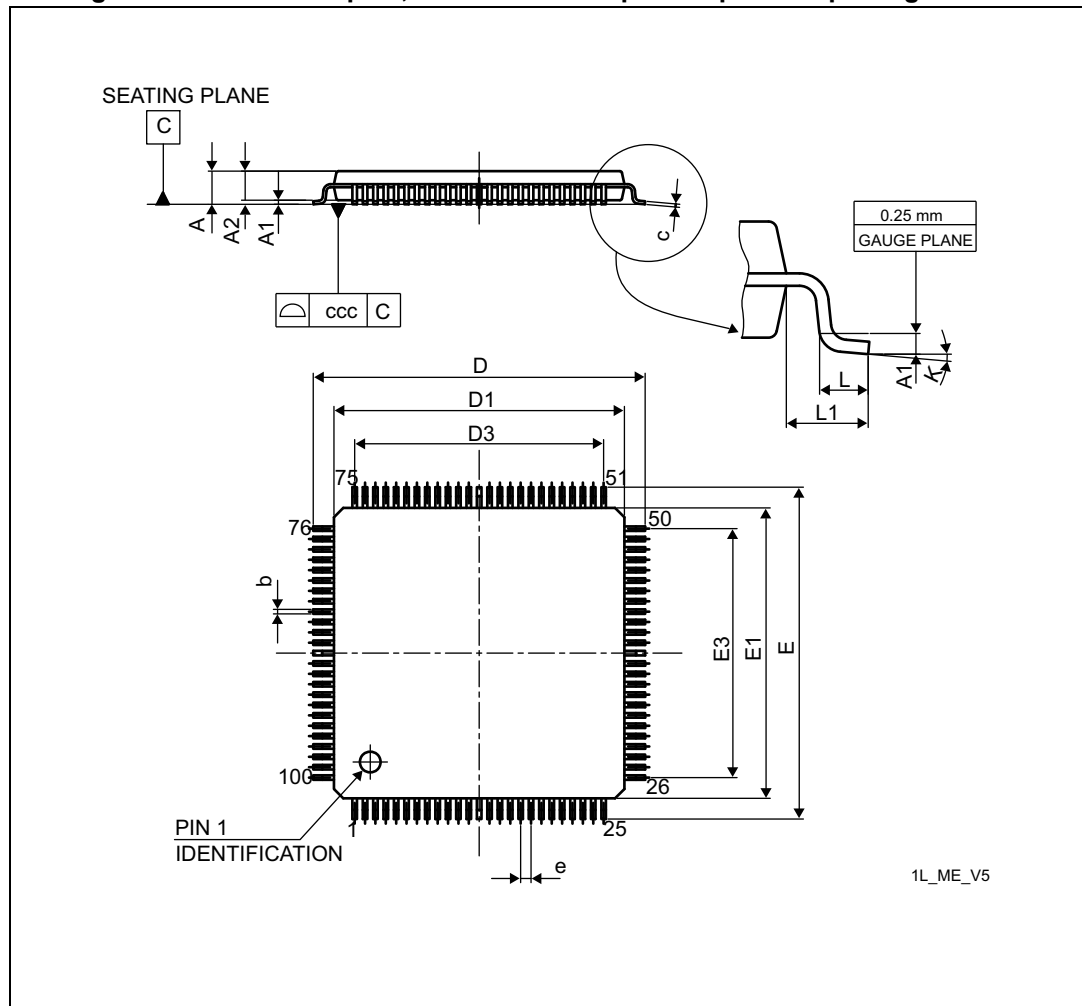
Figure 65. UFBGA132 marking (package top view)



1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.4 LQFP100 package information

Figure 66. LQFP - 100 pins, 14 x 14 mm low-profile quad flat package outline



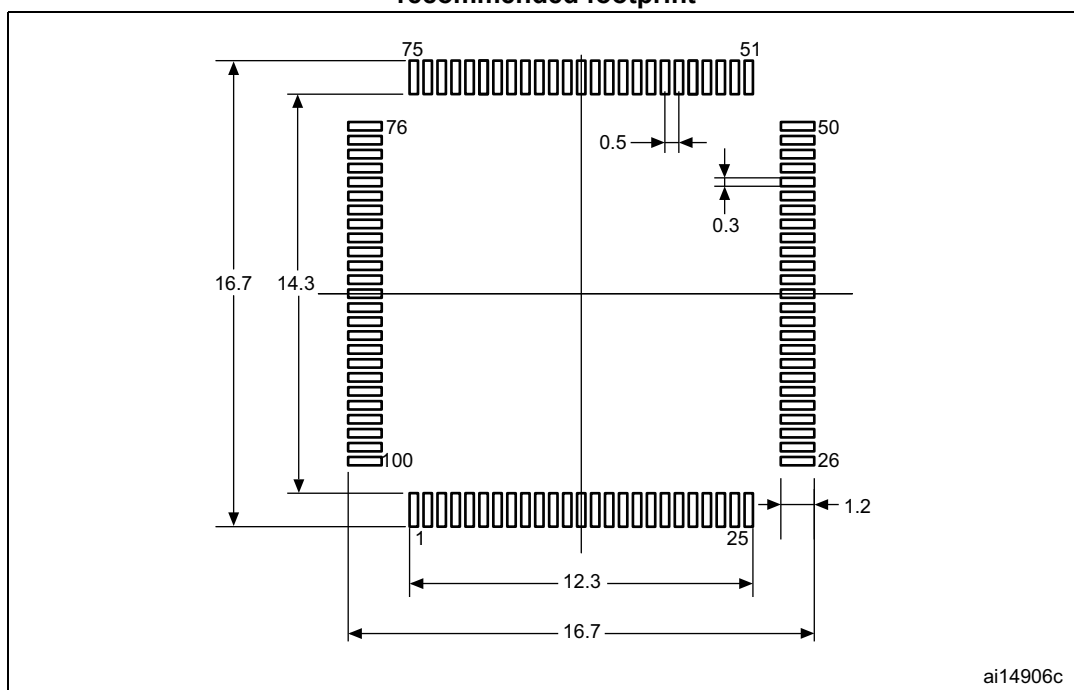
1. Drawing is not to scale.

Table 125. LQFP - 100 pins, 14 x 14 mm low-profile quad flat package mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	15.800	16.000	16.200	0.6220	0.6299	0.6378
D1	13.800	14.000	14.200	0.5433	0.5512	0.5591

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
D3	-	12.000	-	-	0.4724	-
E	15.800	16.000	16.200	0.6220	0.6299	0.6378
E1	13.800	14.000	14.200	0.5433	0.5512	0.5591
E3	-	12.000	-	-	0.4724	-
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0.0°	3.5°	7.0°	0.0°	3.5°	7.0°
ccc	-	-	0.080	-	-	0.0031

Figure 67. LQFP - 100 pins, 14 x 14 mm low-profile quad flat recommended footprint

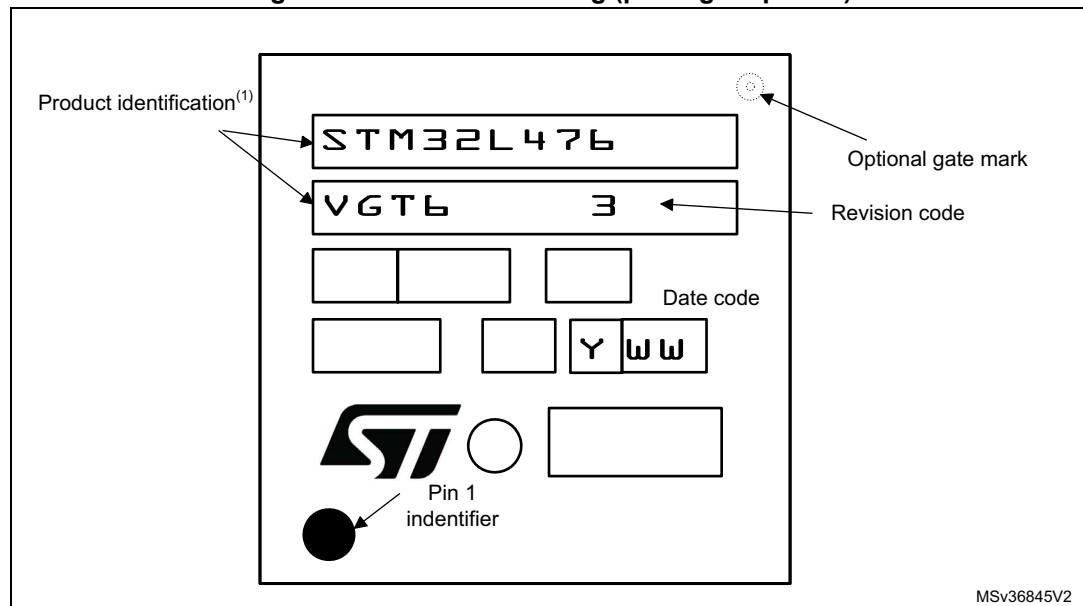


Device marking

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

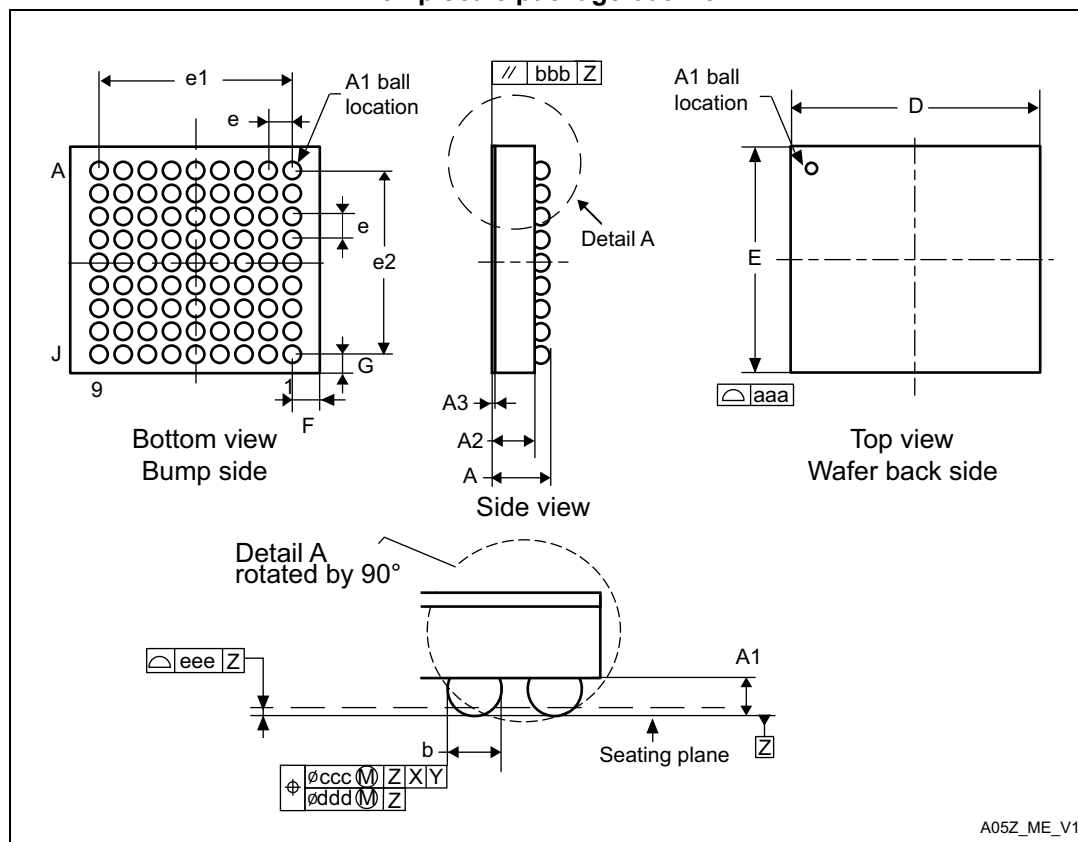
Figure 68. LQFP100 marking (package top view)



1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.5 WLCSP81 package information

Figure 69. WLCSP - 81-ball, 4.4084 x 3.7594 mm, 0.4 mm pitch wafer level chip scale package outline



1. Drawing is not to scale.

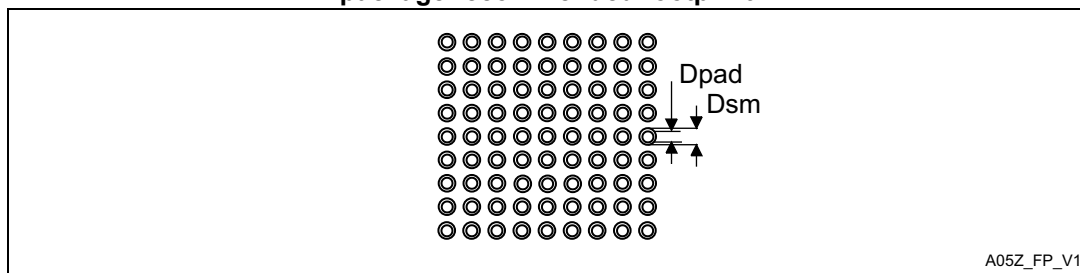
Table 126. WLCSP- 81-ball, 4.4084 x 3.7594 mm, 0.4 mm pitch wafer level chip scale package mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	0.525	0.555	0.585	0.0207	0.0219	0.0230
A1	-	0.175	-	-	0.0069	-
A2	-	0.380	-	-	0.0150	-
A3 ⁽²⁾	-	0.025	-	-	0.0010	-
b ⁽³⁾	0.220	0.250	0.280	0.0087	0.0098	0.0110
D	4.3734	4.4084	4.4434	0.1722	0.1736	0.1749
E	3.7244	3.7594	3.7944	0.1466	0.1480	0.1494
e	-	0.400	-	-	0.0157	-
e1	-	3.200	-	-	0.1260	-
e2	-	3.200	-	-	0.1260	-

Table 126. WLCSP- 81-ball, 4.4084 x 3.7594 mm, 0.4 mm pitch wafer level chip scale package mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
F	-	0.6042	-	-	0.0238	-
G	-	0.2797	-	-	0.0110	-
aaa	-	-	0.100	-	-	0.0039
bbb	-	-	0.100	-	-	0.0039
ccc	-	-	0.100	-	-	0.0039
ddd	-	-	0.050	-	-	0.0020
eee	-	-	0.050	-	-	0.0020

1. Values in inches are converted from mm and rounded to 4 decimal digits.
2. Back side coating
3. Dimension is measured at the maximum bump diameter parallel to primary datum Z.

Figure 70. WLCSP81- 81-ball, 4.4084 x 3.7594 mm, 0.4 mm pitch wafer level chip scale package recommended footprint**Table 127. WLCSP81 recommended PCB design rules (0.4 mm pitch)**

Dimension	Recommended values
Pitch	0.4 mm
Dpad	0.225 mm
Dsm	0.290 mm typ. (depends on the soldermask registration tolerance)
Stencil opening	0.250 mm
Stencil thickness	0.100 mm

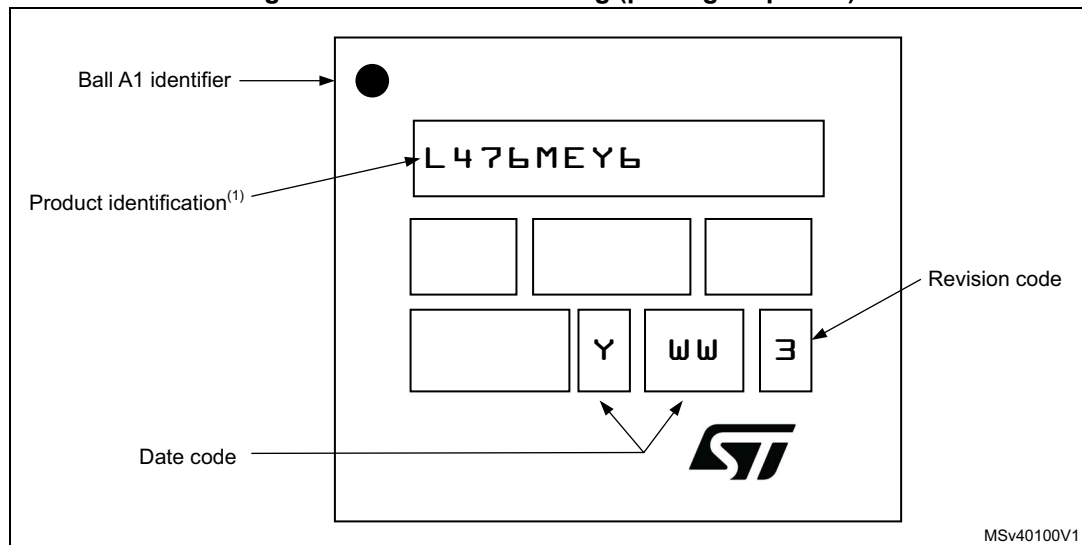
Device marking

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

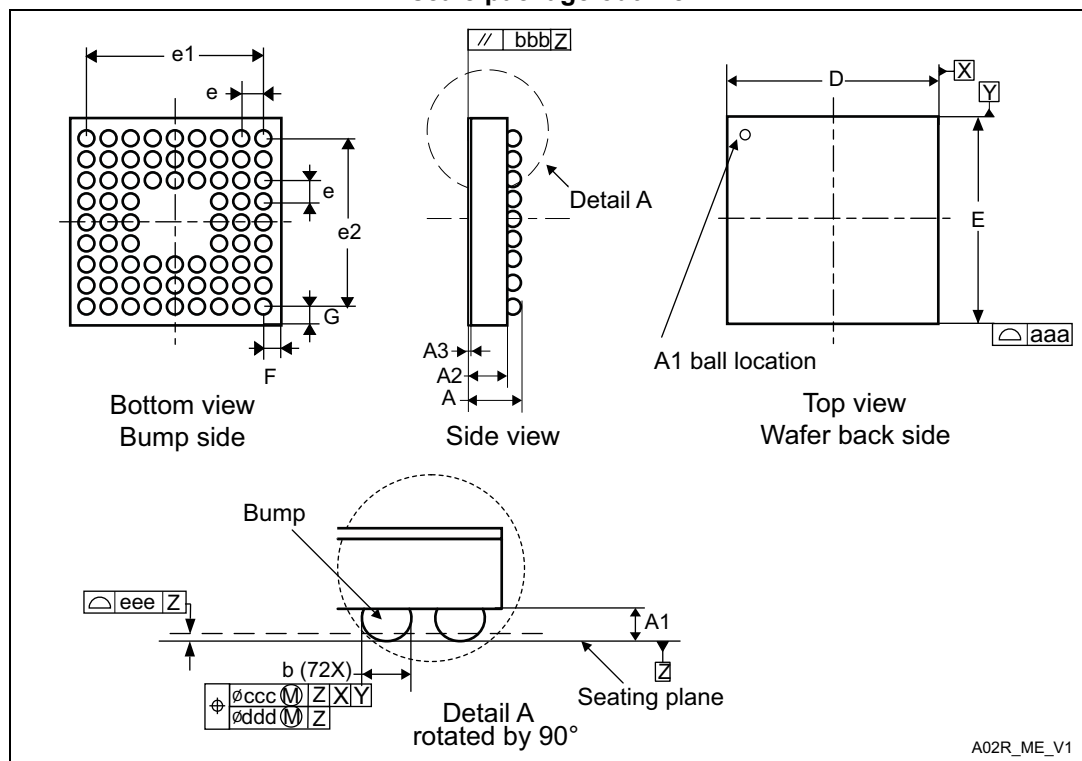
Figure 71. WLCSP81 marking (package top view)



1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.6 WLCSP72 package information

Figure 72. WLCSP - 72-ball, 4.4084 x 3.7594 mm, 0.4 mm pitch wafer level chip scale package outline



1. Drawing is not to scale.

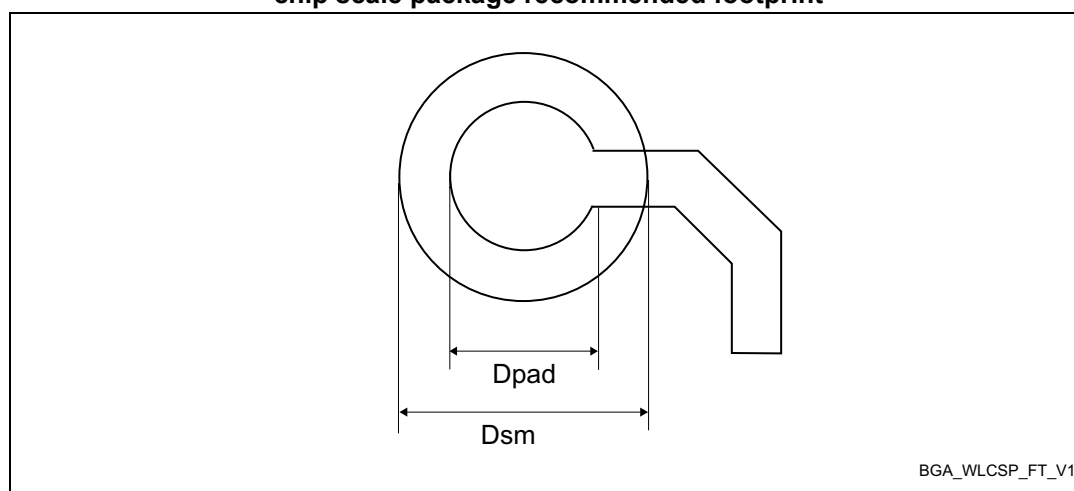
Table 128. WLCSP - 72-ball, 4.4084 x 3.7594 mm, 0.4 mm pitch wafer level chip scale package mechanical data

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	0.525	0.555	0.585	0.0207	0.0219	0.0230
A1	-	0.175	-	-	0.0069	-
A2	-	0.380	-	-	0.0150	-
A3 ⁽²⁾	-	0.025	-	-	0.0010	-
b ⁽³⁾	0.220	0.250	0.280	0.0087	0.0098	0.0110
D	4.3734	4.4084	4.4434	0.1722	0.1736	0.1749
E	3.7244	3.7594	3.7944	0.1466	0.1480	0.1494
e	-	0.400	-	-	0.0157	-
e1	-	3.200	-	-	0.1260	-
e2	-	3.200	-	-	0.1260	-
F	-	0.6042	-	-	0.0238	-
G	-	0.2797	-	-	0.0110	-

Table 128. WLCSP - 72-ball, 4.4084 x 3.7594 mm, 0.4 mm pitch wafer level chip scale package mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
aaa	-	0.100	-	-	0.0039	-
bbb	-	0.100	-	-	0.0039	-
ccc	-	0.100	-	-	0.0039	-
ddd	-	0.050	-	-	0.0020	-
eee	-	0.050	-	-	0.0020	-

1. Values in inches are converted from mm and rounded to 4 decimal digits.
2. Back side coating
3. Dimension is measured at the maximum bump diameter parallel to primary datum Z.

Figure 73. WLCSP72 - 72-ball, 4.4084 x 3.7594 mm, 0.4 mm pitch wafer level chip scale package recommended footprint**Table 129. WLCSP72 recommended PCB design rules (0.4 mm pitch BGA)**

Dimension	Recommended values
Pitch	0.4 mm
Dpad	0.225 mm
Dsm	0.290 mm typ. (depends on the solder mask registration tolerance)
Stencil opening	0.250 mm
Stencil thickness	0.100 mm

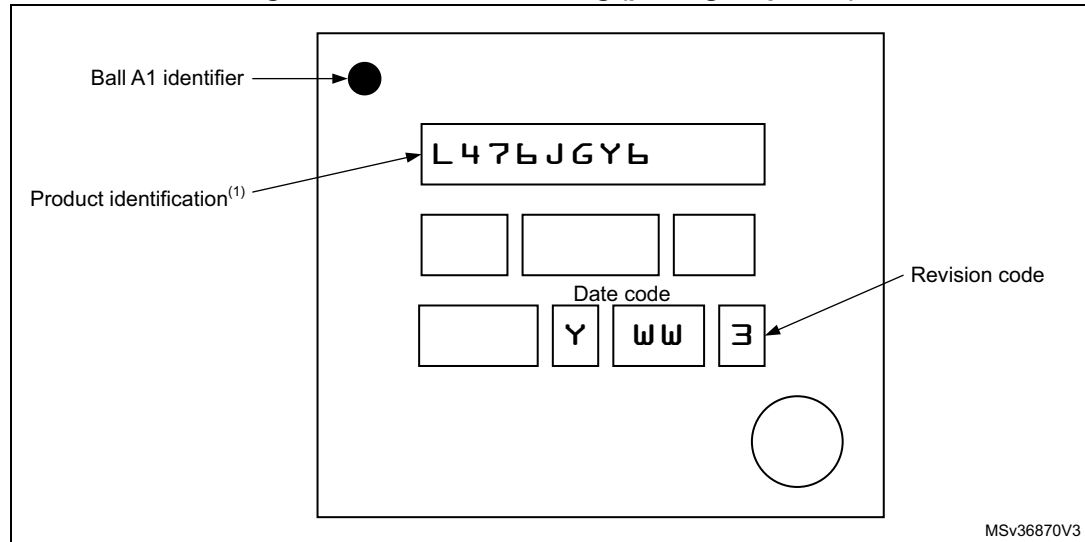
Device marking

The following figure gives an example of topside marking orientation versus ball A1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

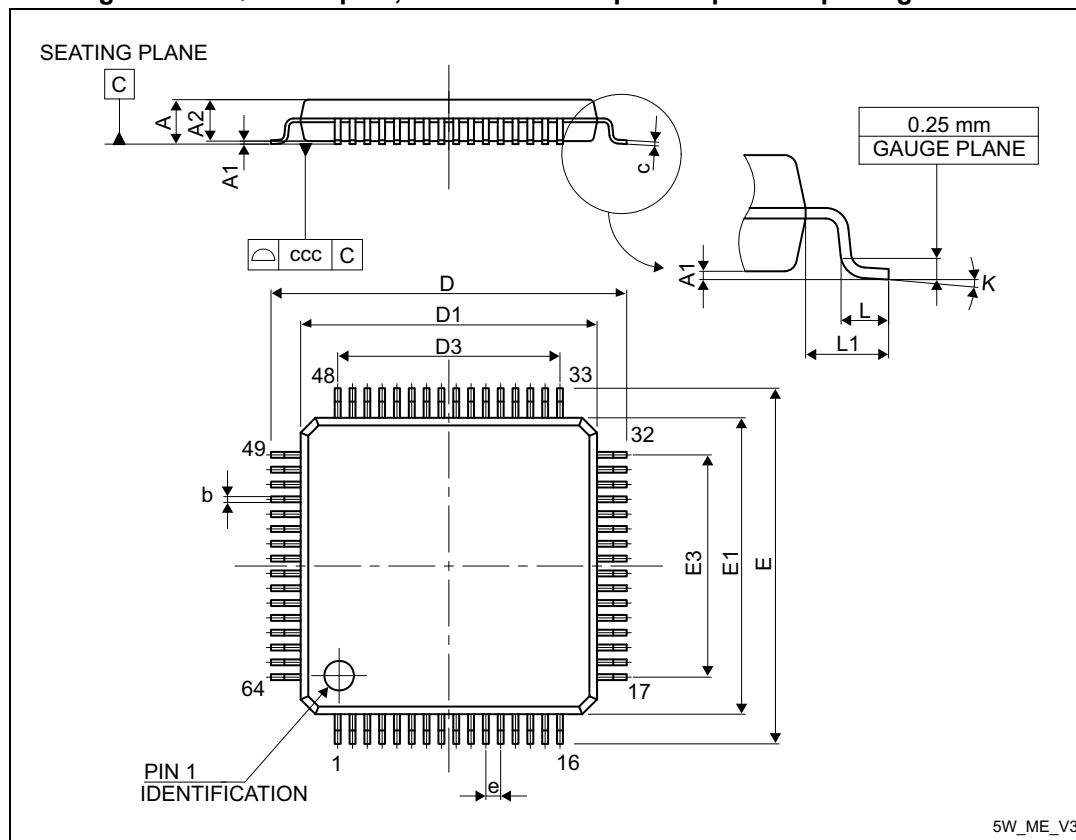
Figure 74. WLCSP72 marking (package top view)



1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.7 LQFP64 package information

Figure 75. LQFP - 64 pins, 10 x 10 mm low-profile quad flat package outline



1. Drawing is not to scale.

Table 130. LQFP - 64 pins, 10 x 10 mm low-profile quad flat package mechanical data

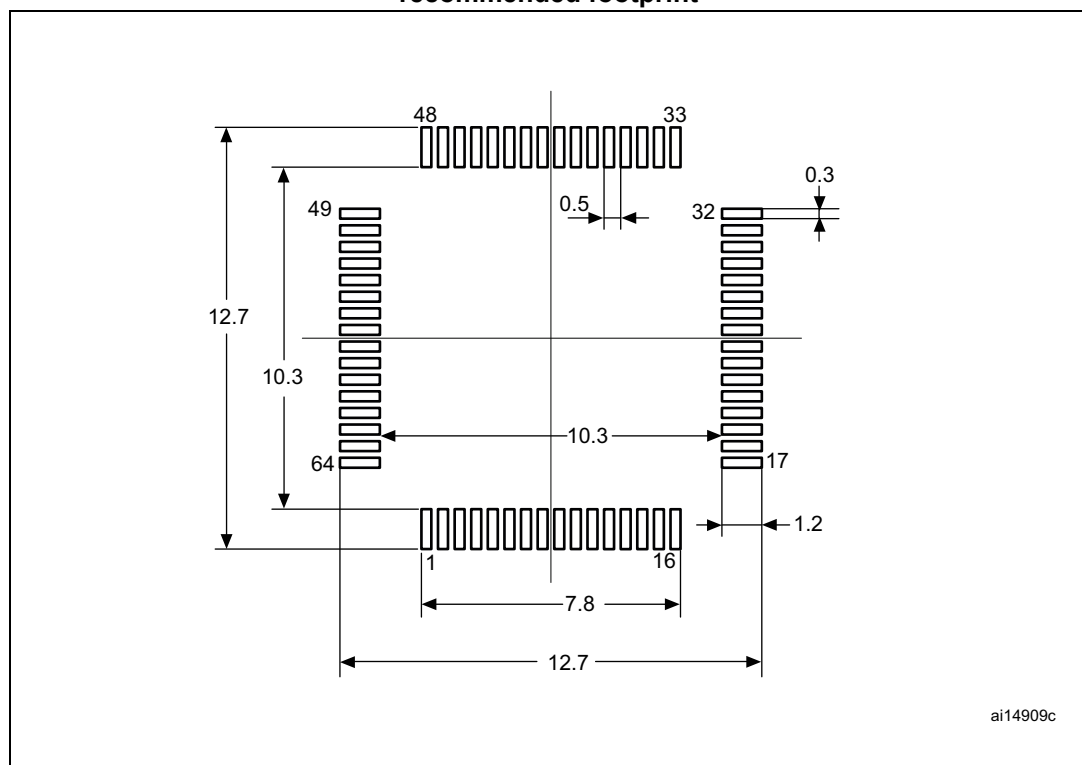
Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	-	12.000	-	-	0.4724	-
D1	-	10.000	-	-	0.3937	-
D3	-	7.500	-	-	0.2953	-
E	-	12.000	-	-	0.4724	-
E1	-	10.000	-	-	0.3937	-

Table 130. LQFP - 64 pins, 10 x 10 mm low-profile quad flat package mechanical data (continued)

Symbol	millimeters			inches ⁽¹⁾		
	Min	Typ	Max	Min	Typ	Max
E3	-	7.500	-	-	0.2953	-
e	-	0.500	-	-	0.0197	-
K	0°	3.5°	7°	0°	3.5°	7°
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
ccc	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 76. LQFP - 64 pins, 10 x 10 mm low-profile quad flat package recommended footprint



1. Dimensions are expressed in millimeters.

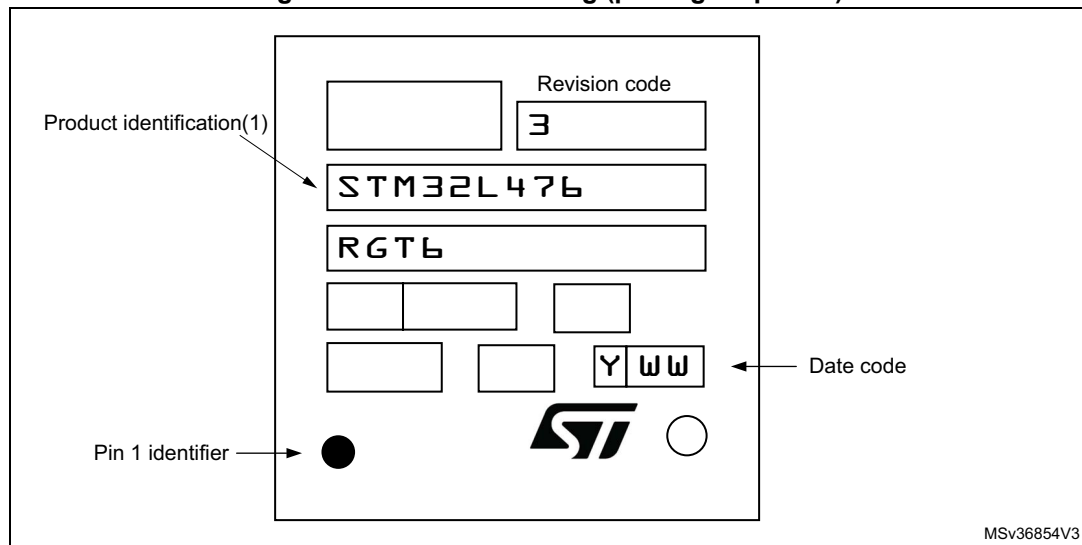
Device marking

The following figure gives an example of topside marking orientation versus pin 1 identifier location.

The printed markings may differ depending on the supply chain.

Other optional marking or inset/upset marks, which identify the parts throughout supply chain operations, are not indicated below.

Figure 77. LQFP64 marking (package top view)



1. Parts marked as ES or E or accompanied by an Engineering Sample notification letter are not yet qualified and therefore not approved for use in production. ST is not responsible for any consequences resulting from such use. In no event will ST be liable for the customer using any of these engineering samples in production. ST's Quality department must be contacted prior to any decision to use these engineering samples to run a qualification activity.

7.8 Thermal characteristics

The maximum chip junction temperature (T_{Jmax}) must never exceed the values given in [Table 22: General operating conditions](#).

The maximum chip-junction temperature, T_J max, in degrees Celsius, may be calculated using the following equation:

$$T_J \text{ max} = T_A \text{ max} + (P_D \text{ max} \times \Theta_{JA})$$

Where:

- T_A max is the maximum ambient temperature in °C,
- Θ_{JA} is the package junction-to-ambient thermal resistance, in °C/W,
- P_D max is the sum of P_{INT} max and $P_{I/O}$ max ($P_D \text{ max} = P_{INT} \text{ max} + P_{I/O} \text{ max}$),
- P_{INT} max is the product of all I_{DDXXX} and V_{DDXXX} , expressed in Watts. This is the maximum chip internal power.

$P_{I/O}$ max represents the maximum power dissipation on output pins where:

$$P_{I/O} \text{ max} = \sum (V_{OL} \times I_{OL}) + \sum ((V_{DDIOx} - V_{OH}) \times I_{OH}),$$

taking into account the actual V_{OL} / I_{OL} and V_{OH} / I_{OH} of the I/Os at low and high level in the application.

Table 131. Package thermal characteristics

Symbol	Parameter	Value	Unit
Θ_{JA}	Thermal resistance junction-ambient LQFP64 - 10 × 10 mm / 0.5 mm pitch	45	°C/W
	Thermal resistance junction-ambient LQFP100 - 14 × 14mm	42	
	Thermal resistance junction-ambient LQFP144 - 20 × 20 mm	32	
	Thermal resistance junction-ambient UFBGA144 - 10 × 10 mm	53	
	Thermal resistance junction-ambient UFBGA132 - 7 × 7 mm	55	
	Thermal resistance junction-ambient WLCSP72	46	
	Thermal resistance junction-ambient WLCSP81	41	

7.8.1 Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from www.jedec.org

7.8.2 Selecting the product temperature range

When ordering the microcontroller, the temperature range is specified in the ordering information scheme shown in [Section 8: Ordering information](#).

Each temperature range suffix corresponds to a specific guaranteed ambient temperature at maximum dissipation and, to a specific maximum junction temperature.

As applications do not commonly use the STM32L476xx at maximum dissipation, it is useful to calculate the exact power consumption and junction temperature to determine which temperature range will be best suited to the application.

The following examples show how to calculate the temperature range needed for a given application.

Example 1: High-performance application

Assuming the following application conditions:

Maximum ambient temperature $T_{Amax} = 82\text{ }^{\circ}\text{C}$ (measured according to JESD51-2),
 $I_{DDmax} = 50\text{ mA}$, $V_{DD} = 3.5\text{ V}$, maximum 20 I/Os used at the same time in output at low level with $I_{OL} = 8\text{ mA}$, $V_{OL} = 0.4\text{ V}$ and maximum 8 I/Os used at the same time in output at low level with $I_{OL} = 20\text{ mA}$, $V_{OL} = 1.3\text{ V}$

$$P_{INTmax} = 50\text{ mA} \times 3.5\text{ V} = 175\text{ mW}$$

$$P_{IOmax} = 20 \times 8\text{ mA} \times 0.4\text{ V} + 8 \times 20\text{ mA} \times 1.3\text{ V} = 272\text{ mW}$$

This gives: $P_{INTmax} = 175\text{ mW}$ and $P_{IOmax} = 272\text{ mW}$:

$$P_{Dmax} = 175 + 272 = 447\text{ mW}$$

Using the values obtained in [Table 131](#) T_{Jmax} is calculated as follows:

– For LQFP64, $45\text{ }^{\circ}\text{C/W}$

$$T_{Jmax} = 82\text{ }^{\circ}\text{C} + (45\text{ }^{\circ}\text{C/W} \times 447\text{ mW}) = 82\text{ }^{\circ}\text{C} + 20.115\text{ }^{\circ}\text{C} = 102.115\text{ }^{\circ}\text{C}$$

This is within the range of the suffix 6 version parts ($-40 < T_J < 105\text{ }^{\circ}\text{C}$) see [Section 8: Ordering information](#).

In this case, parts must be ordered at least with the temperature range suffix 6 (see Part numbering).

Note: With this given P_{Dmax} we can find the T_{Amax} allowed for a given device temperature range (order code suffix 6 or 7).

$$\text{Suffix 6: } T_{Amax} = T_{Jmax} - (45\text{ }^{\circ}\text{C/W} \times 447\text{ mW}) = 105 - 20.115 = 84.885\text{ }^{\circ}\text{C}$$

$$\text{Suffix 7: } T_{Amax} = T_{Jmax} - (45\text{ }^{\circ}\text{C/W} \times 447\text{ mW}) = 125 - 20.115 = 104.885\text{ }^{\circ}\text{C}$$

Example 2: High-temperature application

Using the same rules, it is possible to address applications that run at high ambient temperatures with a low dissipation, as long as junction temperature T_J remains within the specified range.

Assuming the following application conditions:

Maximum ambient temperature $T_{Amax} = 100\text{ }^{\circ}\text{C}$ (measured according to JESD51-2),
 $I_{DDmax} = 20\text{ mA}$, $V_{DD} = 3.5\text{ V}$, maximum 20 I/Os used at the same time in output at low level with $I_{OL} = 8\text{ mA}$, $V_{OL} = 0.4\text{ V}$

$$P_{INTmax} = 20\text{ mA} \times 3.5\text{ V} = 70\text{ mW}$$

$$P_{IOmax} = 20 \times 8\text{ mA} \times 0.4\text{ V} = 64\text{ mW}$$

This gives: $P_{INTmax} = 70\text{ mW}$ and $P_{IOmax} = 64\text{ mW}$:

$$P_{Dmax} = 70 + 64 = 134\text{ mW}$$

Thus: $P_{Dmax} = 134\text{ mW}$

Using the values obtained in [Table 131](#) T_{Jmax} is calculated as follows:

– For LQFP64, 45 °C/W

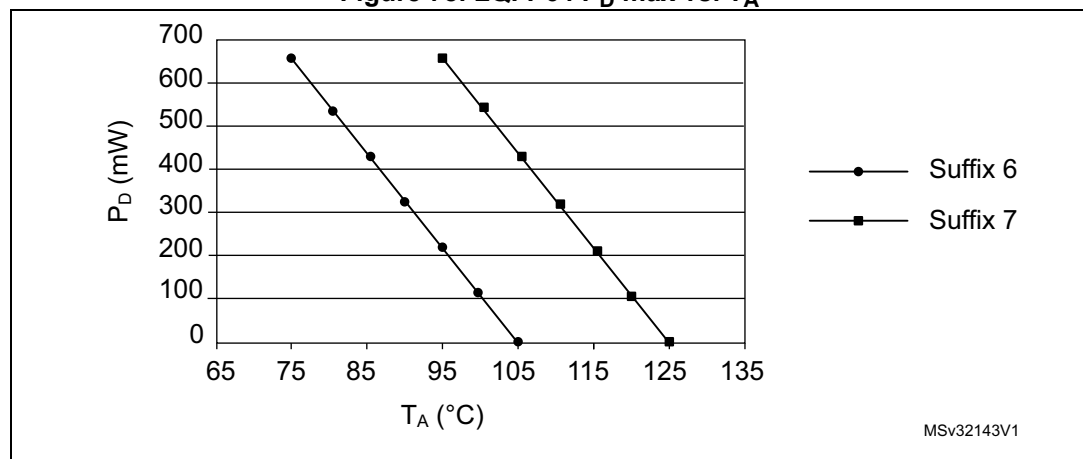
$$T_{Jmax} = 100\text{ °C} + (45\text{ °C/W} \times 134\text{ mW}) = 100\text{ °C} + 6.03\text{ °C} = 106.03\text{ °C}$$

This is above the range of the suffix 6 version parts ($-40 < T_J < 105\text{ °C}$).

In this case, parts must be ordered at least with the temperature range suffix 7 (see [Section 8: Ordering information](#)) unless we reduce the power dissipation in order to be able to use suffix 6 parts.

Refer to [Figure 78](#) to select the required temperature range (suffix 6 or 7) according to your ambient temperature or power requirements.

Figure 78. LQFP64 P_D max vs. T_A



8 Ordering information

Table 132. STM32L476xx ordering information scheme

Example:	STM32	L	476	R	G	T	6	P	M	TR
Device family										
STM32 = Arm® based 32-bit microcontroller										
Product type										
L = ultra-low-power										
Device subfamily										
476: STM32L476xx										
Pin count										
R = 64 pins J = 72 pins M = 81 pins V = 100 pins Q = 132 pins Z = 144 pins										
Flash memory size										
C = 256 KB of Flash memory E = 512 KB of Flash memory G = 1 MB of Flash memory										
Package										
T = LQFP ECOPACK®2 I = UFBGA (7 × 7 mm) ECOPACK®2 J = UFBGA (10 × 10 mm) ECOPACK®2 Y = CSP ECOPACK®2										
Temperature range										
6 = Industrial temperature range, -40 to 85 °C (105 °C junction) 7 = Industrial temperature range, -40 to 105 °C (125 °C junction) 3 = Industrial temperature range, -40 to 125 °C (130 °C junction)										
Option										
Blank = Standard production with integrated LDO P = Dedicated pinout supporting external SMPS										
OptionSpecific package										
M = Specific supply chain ⁽¹⁾ Blank = Standard										
Packing										
TR = tape and reel xxx = programmed parts										

1. This option is available only on STM32L476MGY6MTR part number under specific ordering conditions. Contact your nearest ST sales office for availability.

For a list of available options (speed, package, etc.) or for further information on any aspect of this device, please contact your nearest ST sales office.

9 Revision history

Table 133. Document revision history

Date	Revision	Changes
29-May-2015	1	Initial release.
15-Jun-2015	2	Updated Table 1: Device summary and Table 85: COMP characteristics .
18-Sep-2015	3	Changed alternate function pin name “SWDAT” into “SWDIO” in all the document. Updated Section 3.9.1: Power supply schemes. Updated Section 3.15.1: Temperature sensor. In all Section 6: Electrical characteristics, renamed table footnotes related to test and characterization. Added Note 2. Updated Table 51: Low-power mode wakeup timings. Updated Table 52: Regulator modes transition times. Updated Table 58: HSI16 oscillator characteristics. Added Table 29: HSI16 frequency versus temperature. Updated Table 59: MSI oscillator characteristics. Updated Table 61: LSI oscillator characteristics. Updated Table 69: I/O current injection susceptibility. Removed first Note in Table 70: I/O static characteristics. Removed second Note in Table 71: Output voltage characteristics. Updated Table 76: ADC characteristics. Updated Table 78: ADC accuracy - limited test conditions 1. Added Table 79: ADC accuracy - limited test conditions 2. Added Table 80: ADC accuracy - limited test conditions 3. Added Table 81: ADC accuracy - limited test conditions 4. Updated Table 83: DAC accuracy. Updated Table 84: VREFBUF characteristics. Added Section 6.3.26: DFSDM characteristics. Updated Section : Quad SPI characteristics. Updated Table 97: Quad SPI characteristics in SDR mode. Updated Table 98: QUADSPI characteristics in DDR mode. Updated Table 103: USB OTG electrical characteristics. Updated Section 7.3: UFBGA132 package information. Updated Section 7.6: WLCSP72 package information. Updated Table 77: LQFP64 marking (package top view).

Table 133. Document revision history (continued)

Date	Revision	Changes
03-Dec-2015	4	In all the document: – Stop 1 with main regulator becomes Stop 0 – Stop 1 with low-power regulator remains as Stop 1. In Section 4: Pinouts and pin description: – PC14/OSC32_IN becomes PC14-OSC32_IN (PC14) – PC15/OSC32_OUT becomes PC15-OSC32_OUT (PC15) – PH0/OSC_IN becomes PH0-OSC_IN (PH0) – PH1/OSC_OUT becomes PH1-OSC_OUT (PH1) – PA13 becomes PA13 (JTMS-SWDIO) – PA14 becomes PA14 (JTCK-SWCLK) – PA15 becomes PA15 (JTDI) – PB3 becomes PB3 (JTDO-TRACESWO) – PB4 becomes PB4 (NJTRST). Added Table 13: STM32L476xx USART/UART/LPUART features. Added Note 5. Updated Table 25: Embedded internal voltage reference. Updated Table 45: Current consumption in Stop 2 mode. Updated Table 46: Current consumption in Stop 1 mode. Updated Table 47: Current consumption in Stop 0 mode. Updated Table 48: Current consumption in Standby mode. Updated Table 49: Current consumption in Shutdown mode. Updated Table 51: Low-power mode wakeup timings. Added Figure 24: VREFINT versus temperature. Updated Figure 29: HSI16 frequency versus temperature. Updated Table 70: I/O static characteristics. Updated Table 82: DAC characteristics. Updated Figure 63: UFBGA - 132-ball, 7 x 7 mm ultra thin fine pitch ball grid array package outline. Updated Table 123: UFBGA - 132-ball, 7 x 7 mm ultra thin fine pitch ball grid array package mechanical data.
06-Jul-2017	5	In whole document: – DFSDM peripheral name updated to DFSDM1 – Introduced SMPS product variant Added: – Section 3.24.5: Infrared interface (IRTIM) – Section 6.3.16: Extended interrupt and event controller input (EXTI) characteristics – Section 6.3.30: SWPMI characteristics

Table 133. Document revision history (continued)

Date	Revision	Changes
06-Jul-2017	5 (continued)	– Figure 7: STM32L476Zx, external SMPS device, LQFP144 pinout⁽¹⁾ – Figure 12: STM32L476Mx WLCSP81 ballout⁽¹⁾ – Section 6.3.16: Extended interrupt and event controller input (EXTI) characteristics – Section 6.3.30: SWPMI characteristics – Table 28: Current consumption in Run modes, code with data processing running from Flash, ART enable (Cache ON Prefetch OFF) and power supplied by external SMPS (VDD12 = 1.10 V) – Table 30: Current consumption in Run modes, code with data processing running from Flash, ART disable and power supplied by external SMPS (VDD12 = 1.10 V) – Table 32: Current consumption in Run, code with data processing running from SRAM1 and power supplied by external SMPS (VDD12 = 1.10 V) – Table 34: Typical current consumption in Run, with different codes running from Flash, ART enable (Cache ON Prefetch OFF) and power supplied by external SMPS (VDD12 = 1.10 V) – Table 35: Typical current consumption in Run, with different codes running from Flash, ART enable (Cache ON Prefetch OFF) and power supplied by external SMPS (VDD12 = 1.05 V) – Table 37: Typical current consumption in Run modes, with different codes running from Flash, ART disable and power supplied by external SMPS (VDD12 = 1.10 V) – Table 38: Typical current consumption in Run modes, with different codes running from Flash, ART disable and power supplied by external SMPS (VDD12 = 1.05 V) – Table 40: Typical current consumption in Run mode, with different codes running from SRAM1 and power supplied by external SMPS (VDD12 = 1.10 V) – Table 41: Typical current consumption in Run mode, with different codes running from SRAM1 and power supplied by external SMPS (VDD12 = 1.05 V) – Table 43: Current consumption in Sleep, Flash ON and power supplied by external SMPS (VDD12 = 1.10 V) – Table 54: Wakeup time using USART/LPUART – Table 104: USB BCD DC electrical characteristics – Figure 5: Voltage reference buffer Sections updated: – Section : Features – Section 2: Description – Section 3.9.1: Power supply schemes – Section 3.9.3: Voltage regulator

Table 133. Document revision history (continued)

Date	Revision	Changes
06-Jul-2017	5 (continued)	– Section 3.14.2: Extended interrupt/event controller (EXTI) – Section 3.24.6: Independent watchdog (IWDG) – Section 3.27: Universal synchronous/asynchronous receiver transmitter (USART) – Section 3.28: Low-power universal asynchronous receiver transmitter (LPUART) – Section 3.34: Universal serial bus on-the-go full-speed (OTG_FS) – Section 6.2: Absolute maximum ratings – Section 6.3.5: Supply current characteristics – Section 6.3.18: Analog-to-Digital converter characteristics – Section 7: Package information – Section 8: Ordering information Tables updated: – Table 2: STM32L476xx family device features and peripheral counts – Table 4: STM32L476xx modes overview – Table 6: STM32L476xx peripherals interconnect matrix to add TIM16/TIM17 – Table 16: STM32L476xx pin definitions on pin PA3 updated I/O structure from TT to TT_Ia, on pin VSSA/VREF- updated type to supply pin, added SMPS packages – Table 17: Alternate function AF0 to AF7 – Table 18: Alternate function AF8 to AF15 – Table 20: Voltage characteristics – Table 21: Current characteristics – Table 23: General operating conditions – Table 24: Operating conditions at power-up / power-down – Table 26: Embedded internal voltage reference – Table 52: Low-power mode wakeup timings – Table 75: Analog switches booster characteristics: deleted VBOOST. – Table 82: DAC characteristics – Table 85: COMP characteristics to add Ibias parameter – Table 86: OPAMP characteristics – Table 102: USB OTG DC electrical characteristics – Table 103: USB OTG electrical characteristics – Table 132: STM32L476xx ordering information scheme

Table 133. Document revision history (continued)

Date	Revision	Changes
06-Jul-2017	5 (continued)	Figure updated: Figure 1: STM32L476xx block diagram Figure 63: UFBGA - 132-ball, 7 x 7 mm ultra thin fine pitch ball grid array package outline Footnotes updated for: – Table 16: STM32L476xx pin definitions – Table 20: Voltage characteristics – Table 70: I/O static characteristics – Table 84: VREFBUF characteristics – Table 85: COMP characteristics – Table 77: Maximum ADC RAIN – Figure 31: Recommended NRST pin protection – Figure 33: Typical connection diagram using the ADC
09-Mar-2018	6	Added SMPS option to UFBGA132 package. Aligned DAC instance (DAC1) and DAC output channels (DAC1_OUTx) terminology in all the document. Updated Table 1: STM32L476xx block diagram . Added Figure 10: STM32L476Qx, external SMPS device, UFBGA132 ballout . Updated Table 16: STM32L476xx pin definitions . Updated Table 20: Voltage characteristics . Updated Table 82: DAC characteristics .
24-May-2018	7	Added UFBGA144 package. Updated Section 3.9.1: Power supply schemes . Added Figure 3: Power-up/down sequence . Updated Figure 4: Clock tree . Updated Section 6.3.2: Operating conditions at power-up / power-down . Updated Table 132: STM32L476xx ordering information scheme .
07-Jun-2019	8	Updated Table 132: STM32L476xx ordering information scheme .

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