



**THE DATASHEET OF  
ADP5063ACPZ-1-R7**



## FEATURES

- Default charging termination voltage at 3.6 V
- Fully compatible with USB 3.0 and USB Battery Charging 1.2 Compliance Plan Specification
- Operating input voltage from 4 V to 6.7 V
- Tolerant input voltage from  $-0.5$  V to +20 V (USB VBUS)
- Fully programmable via I<sup>2</sup>C
- Flexible digital control inputs
- Up to 2.1 A current from an ac charger in LDO mode
- Built-in current sensing and limiting
- As low as 55 mΩ battery isolation FET between battery and charger output
- Thermal regulation prevents overheating
- Compliant with JEITA1 and JEITA2 Li-Ion battery charging temperature specifications
- SYS\_EN flag permits the system to be disabled until battery is at the minimum required level for guaranteed system start-up
- 4 mm × 4 mm LFCSP package

## APPLICATIONS

- Single cell lithium iron phosphate (LiFePO<sub>4</sub>) portable equipment
- Portable medical devices
- Portable instrumentation devices
- Portable consumer devices

## GENERAL DESCRIPTION

The ADP5063 charger is fully compliant with USB 3.0 and the USB Battery Charging 1.2 Compliance Plan Specification, and enables charging via the mini USB VBUS pin from a wall charger, car charger, or USB host port.

The ADP5063 operates from a 4 V to 6.7 V input voltage range but is tolerant of voltages up to 20 V, thereby alleviating concerns about USB bus spikes during disconnection or connection scenarios.

The ADP5063 features an internal field effect transistor (FET) between the linear charger output and the battery. This permits battery isolation and, therefore, system powering under a dead battery or no battery scenario, which allows immediate system function upon connection to a USB power supply.

Based on the type of USB source, which is detected by an external USB detection chip, the ADP5063 can be set to apply the correct current limit for optimal charging and USB compliance.

The ADP5063 has three factory-programmable digital input/output pins that provide maximum flexibility for different systems. These digital input/output pins permit a combination of features, such as input current limits, charging enable and disable, charging current limits, and a dedicated interrupt output pin.

## TYPICAL APPLICATION CIRCUIT

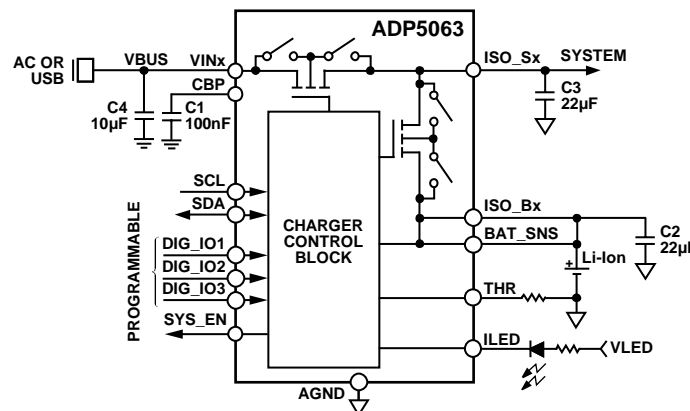


Figure 1.

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Rev. 0

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## REVISION HISTORY

7/13—Revision 0: Initial Version

## SPECIFICATIONS

$-40^{\circ}\text{C} < T_J < +125^{\circ}\text{C}$ ,  $V_{\text{VINx}} = 5.0\text{ V}$ ,  $R_{\text{HOT\_RISE}} < R_{\text{THR}} < R_{\text{COLD\_FALL}}$ ,  $V_{\text{BAT\_SNS}} = 3.6\text{ V}$ ,  $V_{\text{ISO\_Bx}} = V_{\text{BAT\_SNS}}$ ,  $C_{\text{VINx}} = 10\text{ }\mu\text{F}$ ,  $C_{\text{ISO\_Sx}} = 22\text{ }\mu\text{F}$ ,  $C_{\text{ISO\_Bx}} = 22\text{ }\mu\text{F}$ ,  $C_{\text{CBP}} = 100\text{ nF}$ , all registers are at default values, unless otherwise noted.

Table 1.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
<b>GENERAL PARAMETERS</b>						
Undervoltage Lockout	$V_{\text{UVLO}}$	2.25	2.35	2.5	V	Falling threshold, higher of $V_{\text{VINx}}$ or $V_{\text{BAT\_SNS}}$ <sup>1</sup>
Hysteresis		50	100	150	mV	Hysteresis, higher of $V_{\text{VINx}}$ or $V_{\text{BAT\_SNS}}$ rising <sup>1</sup>
Total Input Current	$I_{\text{LIM}}$	74	92	100	mA	Nominal USB initialized current level <sup>2</sup>
				150	mA	USB super speed
				300	mA	USB enumerated current level (specification for China)
				425	mA	USB enumerated current level
VINx Current Consumption	$I_{\text{QVIN}}$		2	500	mA	Dedicated charger input
				900	mA	Dedicated wall charger
Battery Current Consumption	$I_{\text{QBATT}}$		20	1500	mA	Charging or LDO mode
				5	$\mu\text{A}$	DIS_LDO = high, $T_J = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$
			0.5	0.9	mA	LDO mode, $V_{\text{ISO\_Sx}} > V_{\text{BAT\_SNS}}$ Standby, includes ISO_Sx pin leakage, $V_{\text{VINx}} = 0\text{ V}$ , $T_J = -40^{\circ}\text{C}$ to $+85^{\circ}\text{C}$ Standby, battery monitor active
<b>CHARGER</b>						
Fast Charge Current Constant Current (CC) Mode	$I_{\text{CHG}}$		750		mA	
Fast Charge Current Accuracy						
Trickle Charge Current <sup>2</sup>	$I_{\text{TRK\_DEAD}}$	16	20	25	mA	
Weak Charge Current <sup>2, 3</sup>	$I_{\text{CHG\_WEAK}}$		$I_{\text{TRK\_DEAD}} + I_{\text{CHG}}$		mA	
Trickle to Weak Charge Threshold						
Dead Battery	$V_{\text{TRK\_DEAD}}$	1.9	2.0	2.1	V	$V_{\text{TRK\_DEAD}} < V_{\text{BAT\_SNS}} < V_{\text{WEAK}}$ <sup>2, 4</sup>
Hysteresis	$\Delta V_{\text{TRK\_DEAD}}$		100		mV	On BAT_SNS <sup>2</sup>
Weak Battery Threshold						
Weak to Fast Charge Threshold	$V_{\text{WEAK}}$	2.89	3.0	3.11	V	On BAT_SNS <sup>2, 4</sup>
	$\Delta V_{\text{WEAK}}$		100		mV	
Battery Termination Voltage	$V_{\text{TRM}}$		3.600		V	
Termination Voltage Accuracy		-0.6		+0.6	%	On BAT_SNS, $T_J = 25^{\circ}\text{C}$ , $I_{\text{END}} = 52.5\text{ mA}$ <sup>2</sup>
		-1.55		+1.45	%	$T_J = 0^{\circ}\text{C}$ to $115^{\circ}\text{C}$ <sup>2</sup>
		-1.7		+1.7	%	$T_J = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$
Battery Overvoltage Threshold	$V_{\text{BATOv}}$		$V_{\text{IN}} - 0.075$		V	Relative to VINx voltage, BAT_SNS rising
Charge Complete Current	$I_{\text{END}}$	15	52.5	98	mA	$V_{\text{BAT\_SNS}} = V_{\text{TRM}}$
Charging Complete Current		17		83	mA	$I_{\text{END}} = 52.5\text{ mA}$ , $T_J = 0^{\circ}\text{C}$ to $115^{\circ}\text{C}$ <sup>2</sup>
Threshold Accuracy		59		123	mA	$I_{\text{END}} = 92.5\text{ mA}$ , $T_J = 0^{\circ}\text{C}$ to $115^{\circ}\text{C}$
Recharge Voltage Differential	$V_{\text{RCH}}$	160	260	390	mV	Relative to $V_{\text{TRM}}$ , BAT_SNS falling <sup>2</sup>
Battery Node Short Threshold Voltage <sup>2</sup>	$V_{\text{BAT\_SHR}}$	2.2	2.4	2.5	V	
Battery Short Detection Current	$I_{\text{TRK\_SHORT}}$		20		mA	$I_{\text{TRK\_SHORT}} = I_{\text{TRK\_DEAD}}$ <sup>2</sup>
Charging Start Voltage Limit	$V_{\text{CHG\_VLIM}}$	3.1	3.2	3.3	V	Voltage limit is not active by default
Charging Soft Start Current	$I_{\text{CHG\_START}}$	185	260	365	mA	$V_{\text{BAT\_SNS}} > V_{\text{TRK\_DEAD}}$
Charging Soft Start Time	$t_{\text{CHG\_START}}$		3		ms	

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
<b>BATTERY ISOLATION FET</b>						
Pin to Pin Resistance Between ISO_Sx and ISO_Bx	R <sub>DSON_ISO</sub>		55	89	mΩ	On battery supplement mode, VINx = 0 V, V <sub>ISO_Bx</sub> = 3.6 V, I <sub>ISO_Bx</sub> = 500 mA
Regulated System Voltage: V <sub>BAT</sub> Low	V <sub>ISO_SFC</sub>	3.6	3.8	4.0	V	VTRM[5:0] programming ≥ 4.00 V
		3.2	3.4	3.5	V	VTRM[5:0] programming < 4.00 V
Battery Supplementary Threshold	V <sub>THISO</sub>	0	5	12	mV	V <sub>ISO_Sx</sub> < V <sub>ISO_Bx</sub> , system voltage rising
<b>LDO AND HIGH VOLTAGE BLOCKING</b>						
Regulated System Voltage	V <sub>ISO_STRK</sub>	4.214	4.3	4.386	V	VSYSTEM[2:0] = 000 (binary) = 4.3 V, I <sub>ISO_Sx</sub> = 100 mA, LDO mode <sup>2</sup>
Load Regulation			-0.56		%/A	I <sub>ISO_Sx</sub> = 0 mA to 1500 mA
High Voltage Blocking FET (LDO FET) On Resistance	R <sub>DS(ON)HV</sub>		330	485	mΩ	I <sub>VINx</sub> = 500 mA
Maximum Output Current			2.1		A	V <sub>ISO_Sx</sub> = 4.3 V, LDO mode
VINx Input Voltage, Good Threshold Rising	V <sub>VIN_OK_RISE</sub>	3.75	3.9	4.0	V	
VINx Falling	V <sub>VIN_OK_FALL</sub>		3.6	3.7	V	
VINx Input Overvoltage Threshold	V <sub>VIN_OV</sub>	6.7	6.9	7.2	V	
Hysteresis	ΔV <sub>VIN_OV</sub>		0.1		V	
VINx Transition Timing	t <sub>VIN_RISE</sub>	10			μs	Minimum rise time for VINx from 5 V to 20 V
	t <sub>VIN_FALL</sub>	10			μs	Minimum fall time for VINx from 4 V to 0 V
<b>THERMAL CONTROL</b>						
Isothermal Charging Temperature	T <sub>LIM</sub>		115		°C	
Thermal Early Warning Temperature	T <sub>SDL</sub>		130		°C	
Thermal Shutdown Temperature	T <sub>SD</sub>		140		°C	T <sub>J</sub> rising
			110		°C	T <sub>J</sub> falling
<b>THERMISTOR CONTROL</b>						
Thermistor Current						
10,000 NTC (Negative Temperature Coefficient) Resistor	I <sub>NTC_10k</sub>			400	μA	
100,000 NTC Resistor	I <sub>NTC_100k</sub>			40	μA	
Thermistor Capacitance	C <sub>NTC</sub>		100		pF	
Cold Temperature Threshold	T <sub>NTC_COLD</sub>		0		°C	No battery charging occurs
Resistance Thresholds						
Cool to Cold Resistance	R <sub>COLD_FALL</sub>	20,500	25,600	30,720	Ω	
Cold to Cool Resistance	R <sub>COLD_RISE</sub>		24,400		Ω	
Hot Temperature Threshold	T <sub>NTC_HOT</sub>		60		°C	No battery charging occurs
Resistance Thresholds						
Hot to Typical Resistance	R <sub>HOT_FALL</sub>		3700		Ω	
Typical to Hot Resistance	R <sub>HOT_RISE</sub>	2750	3350	3950	Ω	
<b>JEITA1 LI-ION BATTERY CHARGING SPECIFICATION DEFAULTS<sup>5</sup></b>						
JEITA Cold Temperature	T <sub>JEITA_COLD</sub>		0		°C	No battery charging occurs
Resistance Thresholds						
Cool to Cold Resistance	R <sub>COLD_FALL</sub>	20,500	25,600	30,720	Ω	
Cold to Cool Resistance	R <sub>COLD_RISE</sub>		24,400		Ω	
JEITA Cool Temperature	T <sub>JEITA_COOL</sub>		10		°C	Battery charging occurs at 50% of programmed level
Resistance Thresholds						
Typical to Cool Resistance	R <sub>TYP_FALL</sub>	13,200	16,500	19,800	Ω	
Cool to Typical Resistance	R <sub>TYP_RISE</sub>		15,900		Ω	
JEITA Warm Temperature	T <sub>JEITA_WARM</sub>		45		°C	Battery termination voltage (V <sub>TRM</sub> ) is reduced by 100 mV
Resistance Thresholds						
Warm to Typical Resistance	R <sub>WARM_FALL</sub>		5800		Ω	
Typical to Warm Resistance	R <sub>WARM_RISE</sub>	4260	5200	6140	Ω	

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
JEITA Hot Temperature	T <sub>JEITA_HOT</sub>		60		°C	No battery charging occurs
Resistance Thresholds						
Hot to Warm Resistance	R <sub>HOT_FALL</sub>		3700		Ω	
Warm to Hot Resistance	R <sub>HOT_RISE</sub>	2750	3350	3950	Ω	
<b>JEITA2 Li-ION BATTERY CHARGING SPECIFICATION DEFAULTS</b>						
JEITA Cold Temperature	T <sub>JEITA_COLD</sub>		0		°C	No battery charging occurs
Resistance Thresholds						
Cool to Cold Resistance	R <sub>COLD_FALL</sub>	20,500	25,600	30,720	Ω	
Cold to Cool Resistance	R <sub>COLD_RISE</sub>		24,400		Ω	
JEITA Cool Temperature	T <sub>JEITA_COOL</sub>		10		°C	Battery termination voltage (V <sub>TRM</sub> ) is reduced by 100 mV
Resistance Thresholds						
Typical to Cool Resistance	R <sub>TYP_FALL</sub>	13,200	16,500	19,800	Ω	
Cool to Typical Resistance	R <sub>TYP_RISE</sub>		15,900		Ω	
JEITA Warm Temperature	T <sub>JEITA_WARM</sub>		45		°C	Battery termination voltage (V <sub>TRM</sub> ) is reduced by 100 mV
Resistance Thresholds						
Warm to Typical Resistance	R <sub>WARM_FALL</sub>		5800		Ω	
Typical to Warm Resistance	R <sub>WARM_RISE</sub>	4260	5200	6140	Ω	
JEITA Hot Temperature	T <sub>JEITA_HOT</sub>		60		°C	No battery charging occurs
Resistance Thresholds						
Hot to Warm Resistance	R <sub>HOT_FALL</sub>		3700		Ω	
Warm to Hot Resistance	R <sub>HOT_RISE</sub>	2750	3350	3950	Ω	
<b>BATTERY DETECTION</b>						
Sink Current	I <sub>SINK</sub>	13	20	34	mA	
Source Current	I <sub>SOURCE</sub>	7	10	13	mA	
Battery Threshold						
Low	V <sub>BATL</sub>	1.8	1.9	2.0	V	
High	V <sub>BATH</sub>		3.4		V	
Battery Detection Timer	t <sub>BATOK</sub>		333		ms	
<b>TIMERS</b>						
Clock Oscillator Frequency	f <sub>CLK</sub>	2.7	3	3.3	MHz	
Start Charging Delay	t <sub>START</sub>		1		sec	
Trickle Charge	t <sub>TRK</sub>		60		min	
Fast Charge	t <sub>CHG</sub>		600		min	
Charge Complete	t <sub>END</sub>		7.5		min	
Deglitch	t <sub>DG</sub>		31		ms	
Watchdog <sup>2</sup>	t <sub>WD</sub>		32		sec	
Safety	t <sub>SAFE</sub>	36	40	44	min	
Battery Short <sup>2</sup>	t <sub>BAT_SHR</sub>		30		sec	
<b>ILED OUTPUT PINS</b>						
Voltage Drop over ILED	V <sub>ILED</sub>		200		mV	I <sub>ILED</sub> = 20 mA
Maximum Operating Voltage over ILED	V <sub>MAXILED</sub>			5.5	V	
<b>SYS_EN OUTPUT Pin</b>						
SYS_EN FET On Resistance	R <sub>ON_SYS_EN</sub>		10		Ω	I <sub>SYS_EN</sub> = 20 mA

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
LOGIC INPUT PINS						
Maximum Voltage on Digital Inputs	V <sub>DIN_MAX</sub>			5.5	V	Applies to SCL, SDA, DIG_IO1, DIG_IO2, DIG_IO3
Maximum Logic Low Input Voltage	V <sub>IL</sub>			0.5	V	Applies to SCL, SDA, DIG_IO1, DIG_IO2, DIG_IO3
Minimum Logic High Input Voltage	V <sub>IH</sub>	1.2			V	Applies to SCL, SDA, DIG_IO1, DIG_IO2, DIG_IO3
Pull-Down Resistance		215	350	610	kΩ	Applies to DIG_IO1, DIG_IO2, DIG_IO3

<sup>1</sup> Undervoltage lockout generated normally from ISO\_Sx or ISO\_Bx; in certain transition cases, it can be generated from VINx.

<sup>2</sup> These values are programmable via I<sup>2</sup>C. Values are given with default register values.

<sup>3</sup> The output current during charging may be limited by the input current limit or by the isothermal charging mode.

<sup>4</sup> During weak charging mode, the charger provides at least 20 mA of charging current via the trickle charge branch to the battery unless trickle charging is disabled. Any residual current that is not required by the system is also used to charge the battery.

<sup>5</sup> Either JEITA1 (default) or JEITA2 can be selected in I<sup>2</sup>C, or both JEITA functions can be enabled or disabled in I<sup>2</sup>C.

## RECOMMENDED INPUT AND OUTPUT CAPACITANCES

Table 2.

Parameter	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
CAPACITANCES						
VINx	C <sub>VINx</sub>	4		10	μF	Effective capacitance
CBP	C <sub>CBP</sub>	60	100	140	nF	
ISO_Sx	C <sub>ISO_Sx</sub>	10	22	100	μF	
ISO_Bx	C <sub>ISO_Bx</sub>	10	22		μF	

## I<sup>2</sup>C-COMPATIBLE INTERFACE TIMING SPECIFICATIONS

Table 3.

Parameter <sup>1</sup>	Symbol	Min	Typ	Max	Unit
I <sup>2</sup> C-COMPATIBLE INTERFACE <sup>2</sup>					
Capacitive Load for Each Bus Line	C <sub>S</sub>			400	pF
SCL Clock Frequency	f <sub>SCL</sub>			400	kHz
SCL High Time	t <sub>HIGH</sub>	0.6			μs
SCL Low Time	t <sub>LOW</sub>	1.3			μs
Data Setup Time	t <sub>SU, DAT</sub>	100			ns
Data Hold Time	t <sub>HD, DAT</sub>	0		0.9	μs
Setup Time for Repeated Start	t <sub>SU, STA</sub>	0.6			μs
Hold Time for Start/Repeated Start	t <sub>HD, STA</sub>	0.6			μs
Bus Free Time Between a Stop and a Start Condition	t <sub>BUF</sub>	1.3			μs
Setup Time for Stop Condition	t <sub>SU, STO</sub>	0.6			μs
Rise Time of SCL/SDA	t <sub>R</sub>	20		300	ns
Fall Time of SCL/SDA	t <sub>F</sub>	20		300	ns
Pulse Width of Suppressed Spike	t <sub>SP</sub>	0		50	ns

<sup>1</sup> Guaranteed by design.

<sup>2</sup> A master device must provide a hold time of at least 300 ns for the SDA signal to bridge the undefined region of the falling edge of SCL (see Figure 2).

Timing Diagram

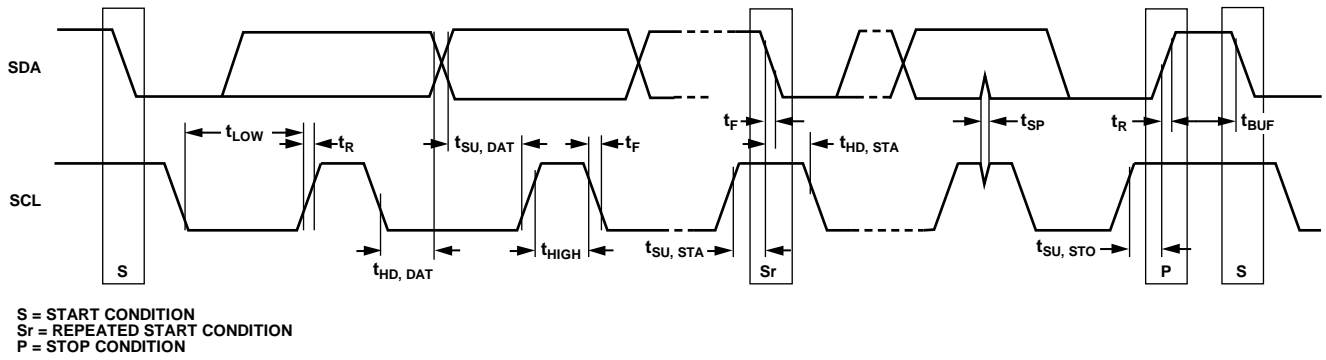


Figure 2. I<sup>2</sup>C Timing Diagram

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## ABSOLUTE MAXIMUM RATINGS

Table 4.

Parameter	Rating
VIN1, VIN2, VIN3 to AGND	-0.5 V to +20 V
All Other Pins to AGND	-0.3 V to +6 V
Continuous Drain Current, Battery Supplementary Mode, from ISO_Bx to ISO_Sx	2.1 A
Storage Temperature Range	-65°C to +150°C
Operating Junction Temperature Range	-40°C to +125°C
Soldering Conditions	JEDEC J-STD-020

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### THERMAL RESISTANCE

$\theta_{JA}$  is specified for the worst-case conditions, that is,  $\theta_{JA}$  is specified for a device soldered in a circuit board for surface-mount packages.

Table 5.

Package Type	$\theta_{JA}$	$\theta_{JC}$	Unit
20-Lead LFCSP	35.6	3.65	°C/W

### Maximum Power Dissipation

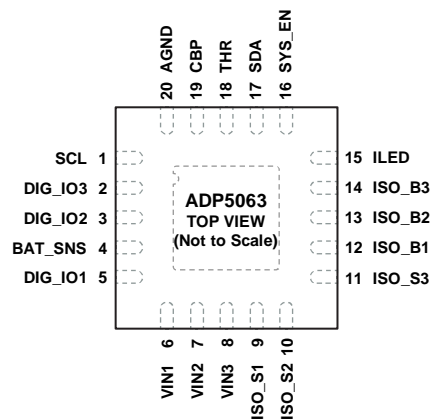
The maximum safe power dissipation in the ADP5063 package is limited by the associated rise in junction temperature ( $T_j$ ) on the die. At a die temperature of approximately 150°C (the glass transition temperature), the properties of the plastic change. Even temporarily exceeding this temperature limit may change the stresses that the package exerts on the die, thereby permanently shifting the parametric performance of the ADP5063. Exceeding a junction temperature of 175°C for an extended period can result in changes in the silicon devices, potentially causing failure.

### ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



## NOTES

1. CONNECTION OF THE EXPOSED PAD IS NOT REQUIRED. THE EXPOSED PAD CAN BE CONNECTED TO ANALOG GROUND TO IMPROVE HEAT DISSIPATION FROM THE PACKAGE TO BOARD.

11993-003

Figure 3. Pin Configuration

Table 6. Pin Function Descriptions

Pin No.	Name	Type <sup>1</sup>	Description
1	SCL	I	I <sup>2</sup> C-Compatible Interface Serial Clock.
2	DIG_IO3	GPIO	Charging Enable. When DIG_IO3 = low or high-Z, charging is disabled. When DIG_IO3 = high, charging is enabled. <sup>2,3</sup>
3	DIG_IO2	GPIO	Set Input Current Limit. When DIG_IO2 = low or high-Z, the input limit is defined by DIG_IO1 setting. When DIG_IO2 = high, the input limit is 1500 mA. <sup>2,3</sup>
4	BAT_SNS	I	Battery Voltage Sense Pin.
5	DIG_IO1	GPIO	Set Input Current Limit. This pin sets the input current limit directly. When DIG_IO1 = low or high-Z, the input limit is 100 mA. When DIG_IO1 = high, the input limit is 500 mA. <sup>2,3</sup>
6, 7, 8	VIN1, VIN2, VIN3	I/O	Power Connections to USB VBUS. These pins are high current inputs when in charging mode.
9, 10, 11	ISO_S1, ISO_S2, ISO_S3	I/O	Linear Charger Supply Side Input to Internal Isolation FET/Battery Current Regulation FET. High current input/output.
12, 13, 14	ISO_B1, ISO_B2, ISO_B3	I/O	Battery Supply Side Input to Internal Isolation FET/Battery Current Regulation FET.
15	ILED	O	Open-Drain Output to Indicator LED.
16	SYS_EN	O	System Enable. This pin is the battery OK flag/open-drain pull-down FET to enable the system when the battery reaches the $V_{WEAK}$ level.
17	SDA	I/O	I <sup>2</sup> C-Compatible Interface Serial Data.
18	THR	I	Battery Pack Thermistor Connection. If this pin is not used, connect a dummy 10 k $\Omega$ resistor from THR to AGND.
19	CBP	I/O	Bypass Capacitor Input.
20	AGND	G	Analog Ground.
N/A <sup>4</sup>	EP	N/A <sup>4</sup>	Exposed Pad. Connection of the exposed pad is not required. The exposed pad can be connected to analog ground to improve heat dissipation from the package to the board.

<sup>1</sup> I is input, O is output, I/O is input/output, G is ground, and GPIO is the factory programmable general-purpose input/output.

<sup>2</sup> See the Digital Input and Output Options section for details.

<sup>3</sup> The DIG\_IOx setting defines the initial state of the ADP5063. If the parameter or the mode that is related to each DIG\_IOx pin setting is changed (by programming an equivalent I<sup>2</sup>C register bit or bits), the I<sup>2</sup>C register setting takes precedence over the DIG\_IOx pin setting. VINx connection or disconnection resets control to the DIG\_IOx pin.

<sup>4</sup> N/A = not applicable.

## TYPICAL PERFORMANCE CHARACTERISTICS

$V_{VINx} = 5.0\text{ V}$ ,  $C_{VINx} = 10\ \mu\text{F}$ ,  $C_{ISO\_Sx} = 44\ \mu\text{F}$ ,  $C_{ISO\_Bx} = 22\ \mu\text{F}$ ,  $C_{CBP} = 100\ \text{nF}$ , all registers are at default values, unless otherwise noted.

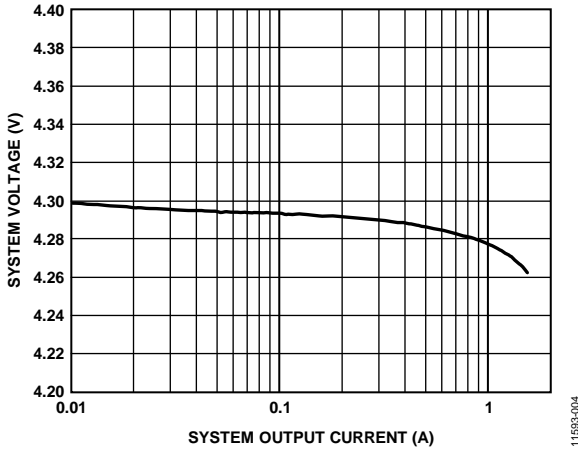


Figure 4. System Voltage vs. System Output Current, LDO Mode,  $V_{SYSTEM}[2:0] = 000$  (Binary) = 4.3 V

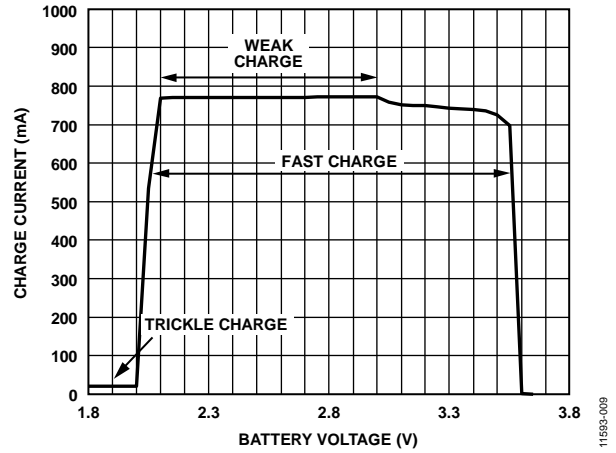


Figure 6. Battery Charge Current vs. Battery Voltage,  $ICHG[4:0] = 01001$  (Binary) = 500 mA,  $ILIM[3:0] = 1111$  (Binary) = 2100 mA

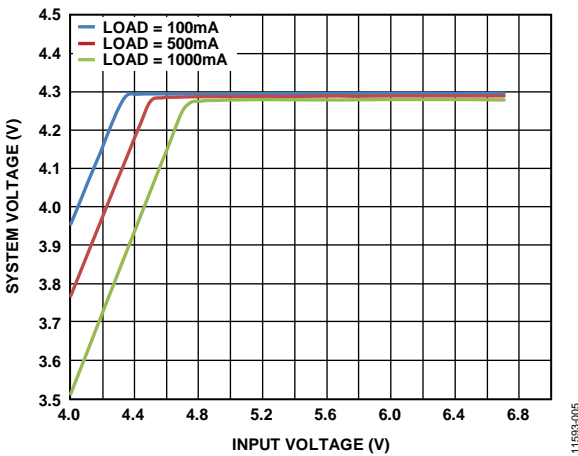


Figure 5. System Voltage vs. Input Voltage (in Dropout), LDO Mode,  $V_{SYSTEM}[2:0] = 000$  (Binary) = 4.3 V

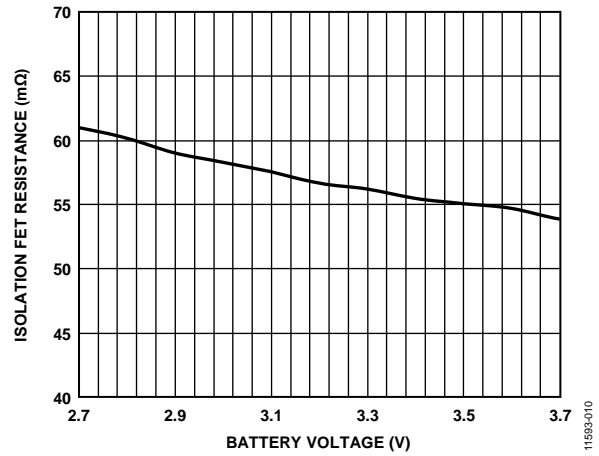


Figure 7. Ideal Diode  $R_{ON}$  vs. Battery Voltage,  $I_{ISO\_Sx} = 500\ \text{mA}$ ,  $V_{INx}$  Open

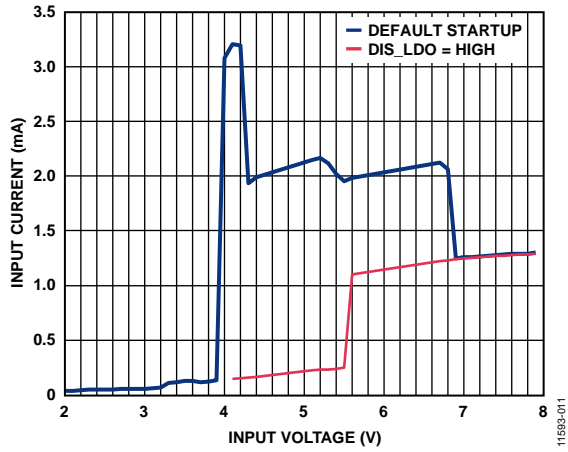


Figure 8. Input Current vs. Input Voltage,  $V_{ISO\_Bx} = 3.3\text{ V}$

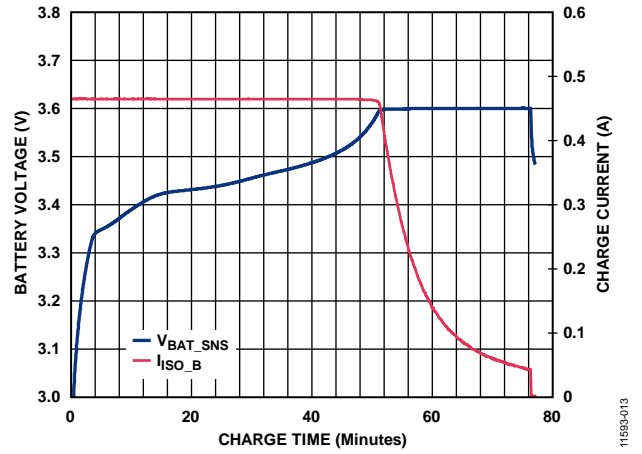


Figure 10. Charge Profile,  $ILIM[3:0] = 0110$  (Binary) = 500 mA, LiFePO<sub>4</sub>, Battery Capacity = 500 mAh

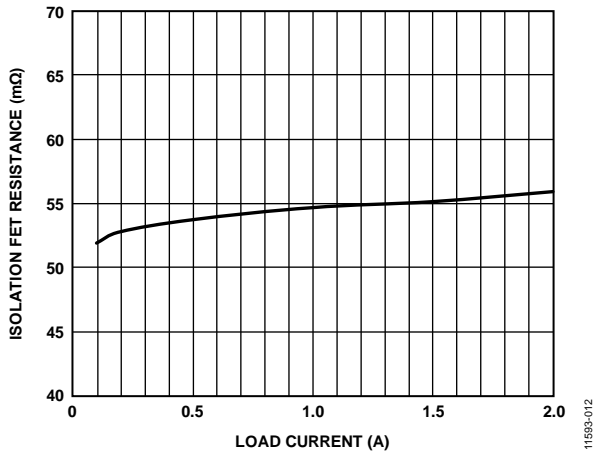


Figure 9. Ideal Diode  $R_{ON}$  vs. Load Current,  $V_{ISO\_Bx} = 3.6\text{ V}$

TEMPERATURE CHARACTERISTICS

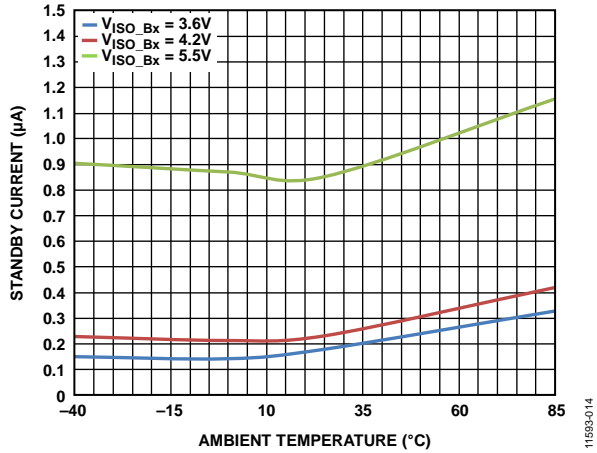


Figure 11. Battery Leakage (Standby) Current vs. Ambient Temperature, Standby Mode

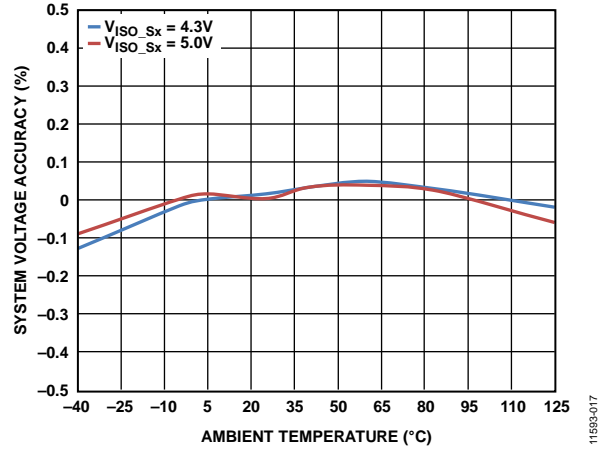


Figure 14. System Voltage Accuracy vs. Ambient Temperature, Trickle Charge Mode,  $V_{ISO\_Sx} = 4.3\text{ V}$  and  $V_{VINx} = 5.0\text{ V}$ , or  $V_{ISO\_Sx} = 5.0\text{ V}$  and  $V_{VINx} = 6.0\text{ V}$

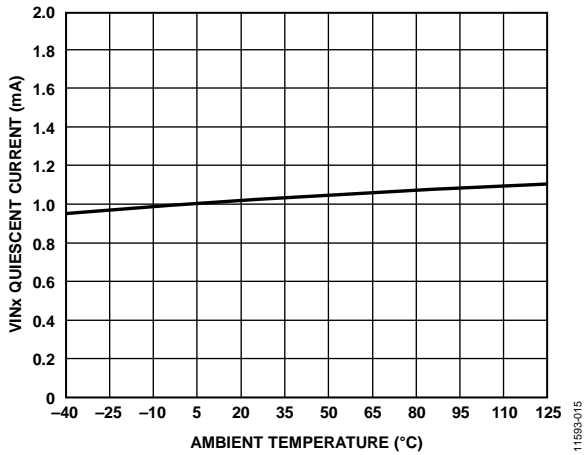


Figure 12.  $V_{INx}$  Quiescent Current vs. Ambient Temperature,  $DIS\_LDO = High$

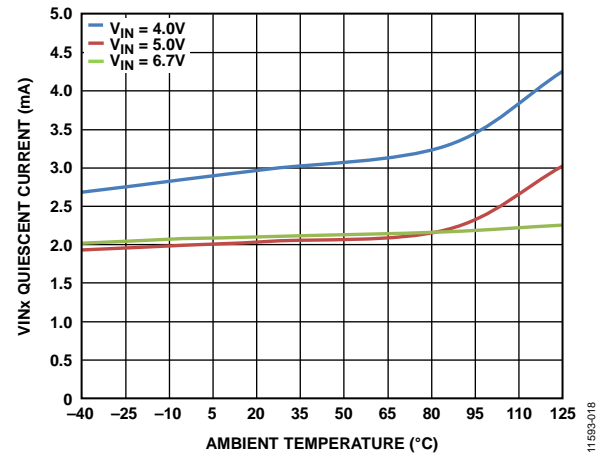


Figure 15.  $V_{INx}$  Quiescent Current vs. Ambient Temperature, LDO Mode

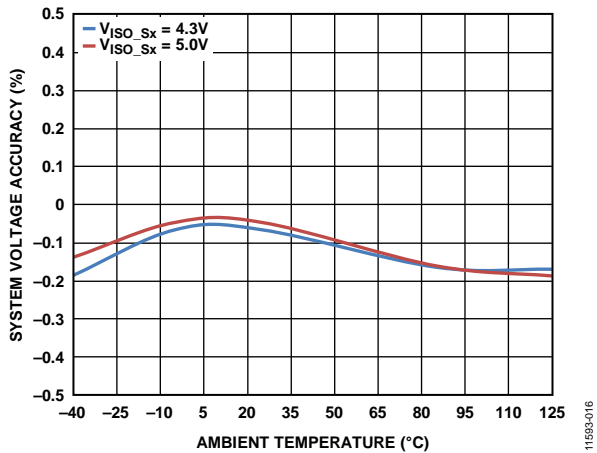


Figure 13. System Voltage Accuracy vs. Ambient Temperature, Load = 100 mA,  $V_{VINx} = 5.5\text{ V}$

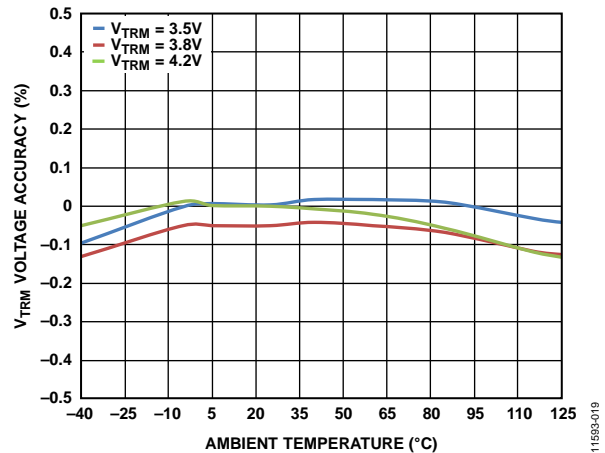


Figure 16. Termination ( $V_{TRM}$ ) Voltage Accuracy vs. Ambient Temperature

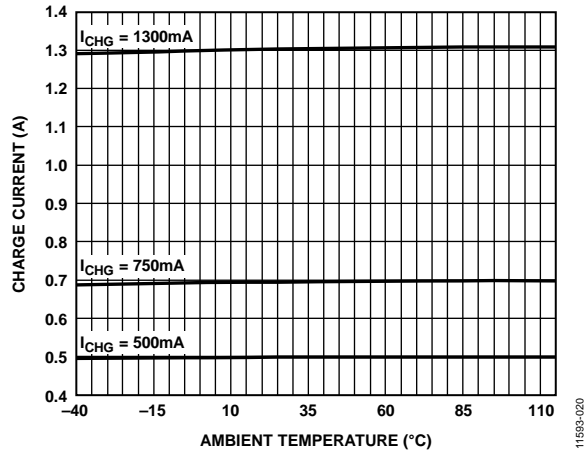


Figure 17. Fast Charge Current CC Mode vs. Ambient Temperature

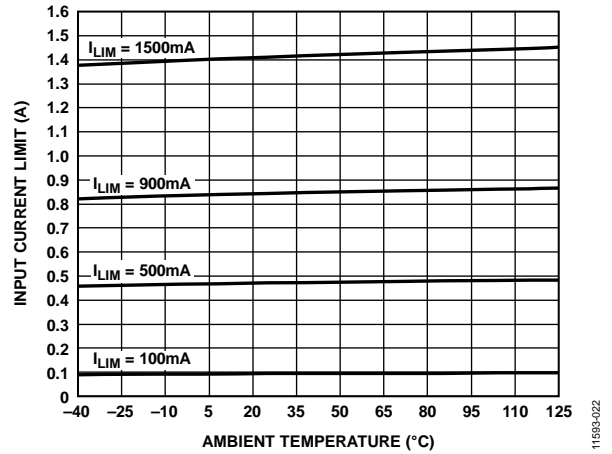


Figure 19. Input Current Limit vs. Ambient Temperature

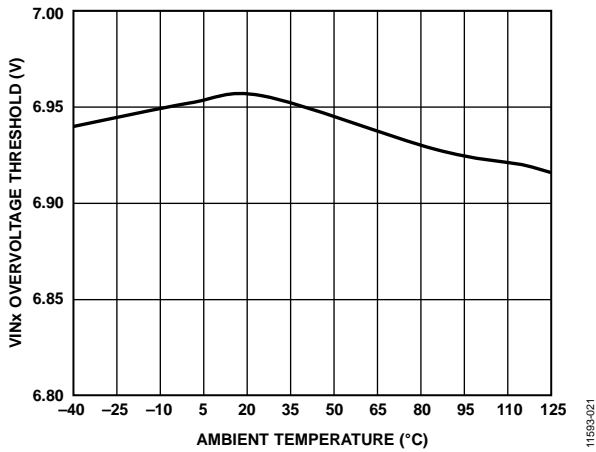


Figure 18. VINx Overvoltage Threshold vs. Ambient Temperature

TYPICAL WAVEFORMS

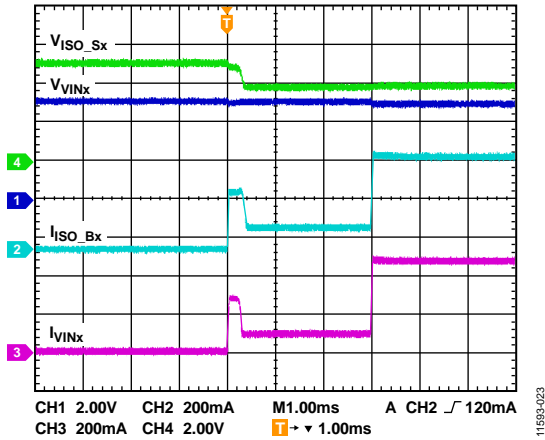


Figure 20. Charging Startup,  $V_{VINx} = 5.0\text{ V}$ ,  $ILIM[3:0] = 0110$  (Binary) = 500 mA,  $ICHG[4:0] = 01110$  (Binary) = 750 mA

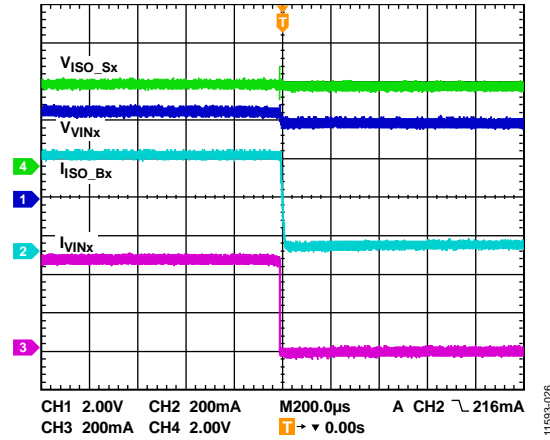


Figure 23. USB VBUS Disconnection

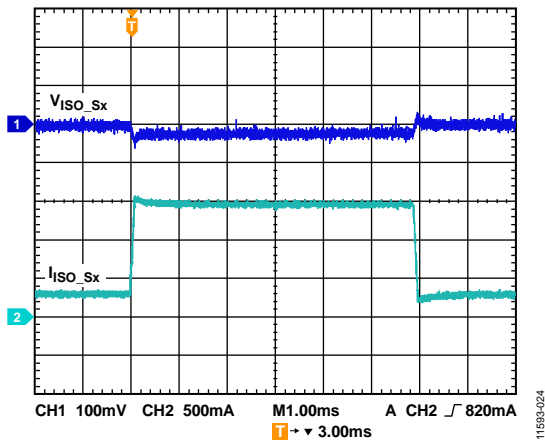


Figure 21. Load Transient,  $I_{ISO\_Sx}$  Load = 300 mA to 1500 mA to 300 mA

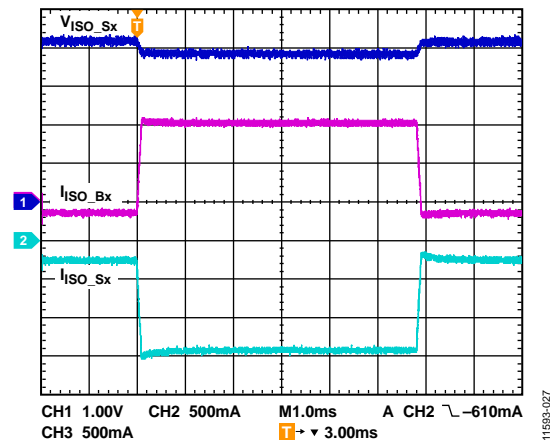


Figure 24. Load Transient,  $I_{ISO\_Sx}$  Load = 300 mA to 1500 mA to 300 mA,  $EN\_CHG = \text{High}$ ,  $ILIM[3:0] = 0110$  (Binary) = 500 mA

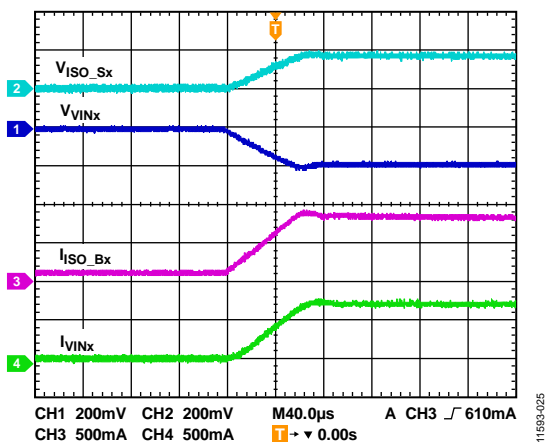


Figure 22. Input Current-Limit Transition from 100 mA to 900 mA,  $ISO\_Sx$  Load = 66  $\Omega$ , Charging = 750 mA

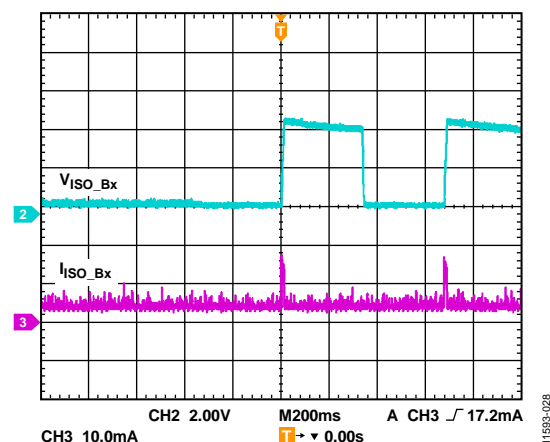


Figure 25. Battery Detection Waveform,  $VSYSTEM[2:0] = 000$  (Binary) = 4.3 V, No Battery

## THEORY OF OPERATION

### SUMMARY OF OPERATION MODES

Table 7. Summary of Operation Modes

Mode Name	V <sub>VINx</sub> Condition	Battery Condition	Trickle Charge	LDO FET State	Battery Isolation FET	System Voltage ISO_Sx	Additional Conditions <sup>1</sup>
IC Off, Standby	0 V	Any battery condition	Off	Off	On	Battery voltage or 0 V	
IC Off, Suspend	5 V	Any battery condition	Off	Off	On	Battery voltage	DIS_LDO = high
LDO Mode Off, Isolation FET On	5 V	Any battery condition	Off	Off	On	Battery voltage	Disable LDO and enable isolation FET
LDO Mode Off, Isolation FET Off (System Off)	5 V	Any battery condition	Off	Off	Off	0 V	Enable battery charging
LDO Mode, Charger Off	5 V	Any battery condition	Off	LDO	Off	4.3 V	Enable battery charging
Trickle Charge Mode	5 V	Battery < V <sub>TRK_DEAD</sub>	On	LDO	Off	4.3 V	Enable battery charging
Weak Charge Mode	5 V	V <sub>TRK_DEAD</sub> ≤ battery < V <sub>WEAK</sub>	On	CHG	CHG	3.4 V	Enable battery charging
Fast Charge Mode	5 V	Battery ≥ V <sub>WEAK</sub>	Off	CHG	CHG	3.4 V (minimum)	Enable battery charging
Charge Mode, No Battery	5 V	Open	Off	LDO	Off	4.3 V	Enable battery charging
Charge Mode, Battery (ISO_Bx) Shorted	5 V	Shorted	On	LDO	Off	4.3 V	Enable battery charging

<sup>1</sup> See Table 8 for details.

Table 8. Operation Mode Controls

Pin Configuration	DIG_IOx	Equivalent I <sup>2</sup> C Address, Data Bit(s)	Description
Enable Battery Charging	DIG_IO3	0x07, D0	Low = all charging modes disabled (fast, weak, trickle). High = all charging modes enabled (fast, weak, trickle).
Disable LDO and Enable Isolation FET	Not applicable	0x07, D3, D0	Low = LDO enabled. High = LDO disabled. In addition, when EN_CHG = low, the battery isolation FET is on; when EN_CHG = high, the battery isolation FET is off.

## INTRODUCTION

The [ADP5063](#) is a fully programmable I<sup>2</sup>C charger for single cell lithium ion or lithium polymer batteries, suitable for a wide range of portable applications.

The linear charger architecture enables up to 2.1 A output current at 4.3 V to 5.0 V (I<sup>2</sup>C programmable) on the system power supply, and up to 1.3 A of charge current into the battery from a dedicated charger.

The [ADP5063](#) operates from an input voltage of 4 V up to 6.7 V but is tolerant of voltages of up to 20 V. The 20 V voltage tolerance alleviates the concerns of the USB bus spiking during disconnection or connection scenarios.

The [ADP5063](#) features an internal FET between the linear charger output and the battery. This feature permits battery isolation and, therefore, system powering under a dead battery or no battery scenario, which allows immediate system function upon connection to a USB power supply.

The [ADP5063](#) is fully compliant with USB 3.0 and the USB Battery Charging 1.2 Compliance Plan Specification. The [ADP5063](#) is chargeable via the mini USB VBUS pin from a wall charger, car charger, or USB host port. Based on the type of USB source, which

is detected by an external USB detection device, the [ADP5063](#) can be set to apply the correct current limit for optimal charging and USB compliance. The USB charger permits correct operation under all USB compliant sources such as wall chargers, host chargers, hub chargers, and standard host and hubs.

A processor can control the USB charger using the I<sup>2</sup>C interface to program the charging current and numerous other parameters, including

- Trickle charge current level
- Trickle charge voltage threshold
- Weak charge (constant current) current level
- Fast charge (constant current) current level
- Fast charge (constant voltage) voltage level
- Fast charge safety timer period
- Watchdog safety timer parameters
- Weak battery threshold detection
- Charging complete threshold
- Recharge threshold
- Charging enable/disable
- Battery pack temperature detection and automatic charger shutdown

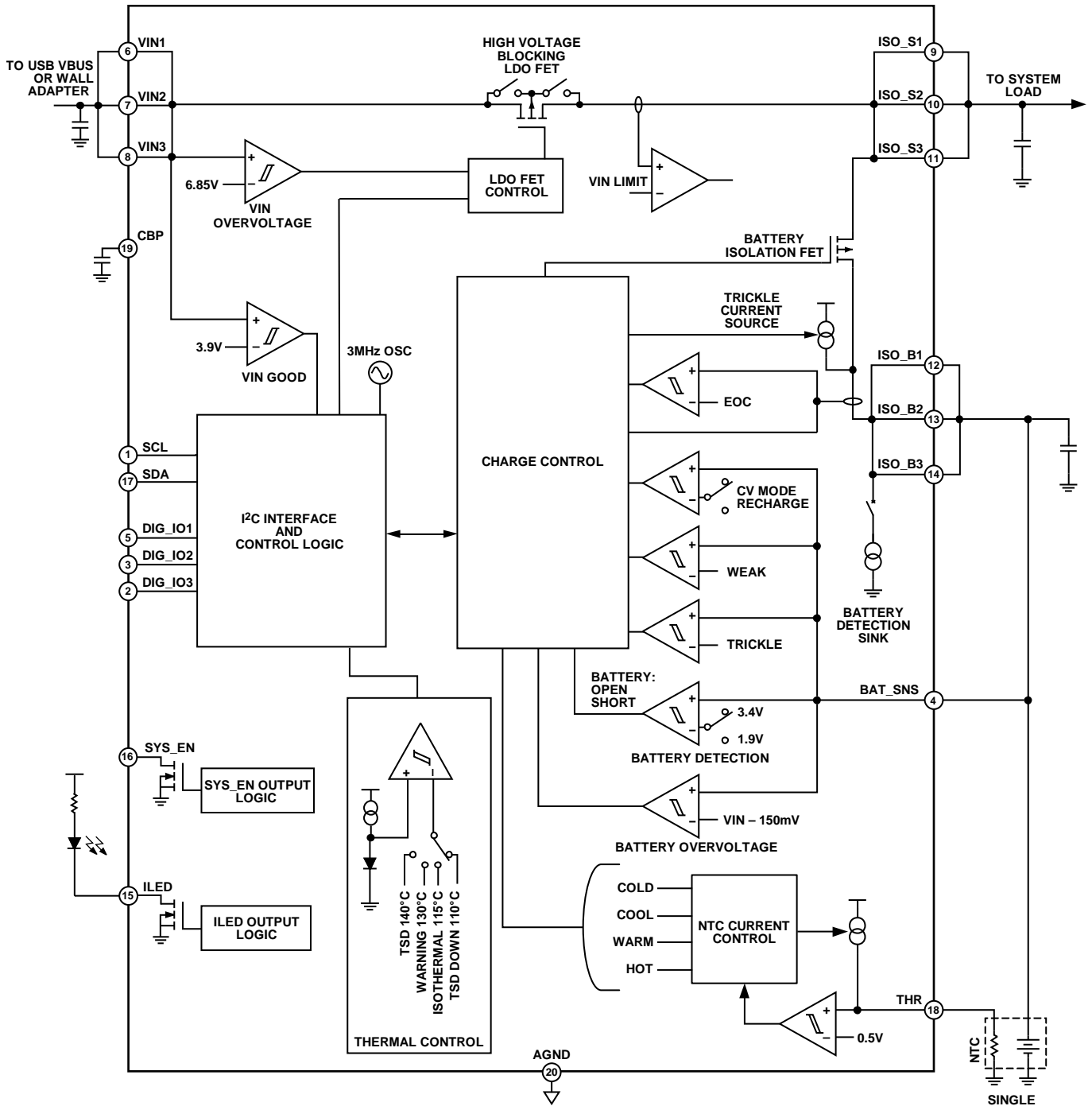


Figure 26. Block Diagram

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The [ADP5063](#) includes a number of significant features to optimize charging and functionality, including

- Thermal regulation for maximum performance.
- USB host current limits.
- Termination voltage accuracy:  $\pm 1.7\%$ .
- Battery thermistor input with automatic charger shutdown in the event that the battery temperature exceeds limits (compliant with the JEITA Li-Ion battery charging temperature specification).
- Three external pins (DIG\_IO1, DIG\_IO2, and DIG\_IO3) that directly control a number of parameters. These pins are factory programmable for maximum flexibility. They can be factory programmed for functions such as
  - Enable/disable charging.
  - Control of the 100 mA or 500 mA input current limit.
  - Control of the 1500 mA input current limit.
  - Control of the battery charge current.
  - An interrupt output pin.

See the Digital Input and Output Options section for details.

## CHARGER MODES

### Input Current Limit

The VINx input current limit is controlled via the internal I<sup>2</sup>C ILIM bits. The input current limit can also be controlled via the DIG\_IO1 pin (if factory programmed to do so) as outlined in Table 9. Any change from the 100 mA I<sup>2</sup>C default takes precedence over the pin setting.

**Table 10. Input Current Compatibility with Standard USB Limits**

Mode	Standard USB Limit	ADP5063 Function
USB (China Only)	100 mA limit for standard USB host or hub 300 mA limit for Chinese USB specification	100 mA input current limit or I <sup>2</sup> C programmed value 300 mA input current limit or I <sup>2</sup> C programmed value
USB 2.0	100 mA limit for standard USB host or hub 500 mA limit for standard USB host or hub	100 mA input current limit or I <sup>2</sup> C programmed value 500 mA input current limit or I <sup>2</sup> C programmed value
USB 3.0	150 mA limit for superspeed USB 3.0 host or hub 900 mA limit for superspeed, high speed USB host or hub charger	150 mA input current limit or I <sup>2</sup> C programmed value 900 mA input current limit or I <sup>2</sup> C programmed value
Dedicated Charger	1500 mA limit for dedicated charger or low/full speed USB host or hub charger	1500 mA input current limit or I <sup>2</sup> C programmed value

**Table 9. DIG\_IO1 Operation**

DIG_IO1	Function
0	100 mA input current limit or I <sup>2</sup> C programmed value
1	500 mA input current limit or I <sup>2</sup> C programmed value (or reprogrammed I <sup>2</sup> C value from 100 mA default)

### USB Compatibility

The [ADP5063](#) features an I<sup>2</sup>C-programmable input current limit to ensure compatibility with the requirements listed in Table 10. The current limit defaults to 100 mA to allow compatibility with a USB host or hub that is not configured.

The I<sup>2</sup>C register default is 100 mA. An I<sup>2</sup>C write command to the ILIM bits overrides the DIG\_IOx pins, and the I<sup>2</sup>C register default value can be reprogrammed for alternative requirements.

When the input current-limit feature is used, the available input current may be too low for the charger to meet the programmed charging current, I<sub>CHG</sub>, thereby reducing the rate of charge and setting the VIN\_ILIM flag.

When connecting voltage to VINx without the proper voltage level on the battery side, the high voltage blocking mechanism is in a state wherein it draws a current of <1 mA until V<sub>VINx</sub> reaches the VIN\_OK level.

The [ADP5063](#) charger provides support for the following connections through the single connector VINx pin, as shown in Table 10.

### Trickle Charge Mode

A deeply discharged Li-Ion cell can exhibit a very low cell voltage, making it unsafe to charge the cell at high current rates. The ADP5063 charger uses a trickle charge mode to reset the battery pack protection circuit and lift the cell voltage to a safe level for fast charging. A cell with a voltage below  $V_{TRK\_DEAD}$  is charged with the trickle mode current,  $I_{TRK\_DEAD}$ . During trickle charging mode, the CHARGER\_STATUS[2:0] bits are set.

During trickle charging, the ISO\_Sx node is regulated to  $V_{ISO\_STRK}$  by the LDO and the battery isolation FET is off, which means that the battery is isolated from the system power supply.

### Trickle Charge Mode Timer

The duration of trickle charge mode is monitored to ensure that the battery is revived from its deeply discharged state. If trickle charge mode runs for longer than 60 minutes without the cell voltage reaching  $V_{TRK\_DEAD}$ , a fault condition is assumed and charging stops. The fault condition is asserted on the CHARGER\_STATUS[2:0] bits, allowing the user to initiate the fault recovery procedure specified in the Fault Recovery section.

### Weak Charge Mode (Constant Current)

When the battery voltage exceeds  $V_{TRK\_DEAD}$  but is less than  $V_{WEAK}$ , the charger switches to intermediate charge mode.

During the weak charge mode, the battery voltage is too low to allow the full system to power up. Because of the low battery level, the USB transceiver cannot be powered and, therefore, cannot enumerate for more current from a USB host. Consequently, the USB limit remains at 100 mA.

The system microcontroller may or may not be powered by the charger output voltage ( $V_{ISO\_SFC}$ ), depending upon the amount of current that the microcontroller and/or the system architecture requires. When the ISO\_Sx pins power the microcontroller, the battery charge current ( $I_{CHG\_WEAK}$ ) cannot be increased above 20 mA to ensure microcontroller operation (if doing so), nor can  $I_{CHG\_WEAK}$  be increased above the 100 mA USB limit. Therefore, set the battery charging current as follows:

- Set the default 20 mA via the linear trickle charger branch (to ensure that the microprocessor remains alive if powered by the main charger output, ISO\_Sx). Any residual current on the main charger output, ISO\_Sx, is used to charge the battery.
- During weak current mode, other features may prevent the weak charging current from reaching its full programmed value. Isothermal charging mode or input current limiting for USB compatibility can affect the programmed weak charging current value under certain operating conditions. During weak charging, the ISO\_Sx node is regulated to  $V_{ISO\_SFC}$  by the battery isolation FET.

### Fast Charge Mode (Constant Current)

When the battery voltage exceeds  $V_{TRK\_DEAD}$  and  $V_{WEAK}$ , the charger switches to fast charge mode, charging the battery with the constant current,  $I_{CHG}$ . During fast charge mode (constant current), the CHARGER\_STATUS[2:0] bits are set to 010.

During constant current mode, other features may prevent the current,  $I_{CHG}$ , from reaching its full programmed value. Isothermal charging mode or input current limiting for USB compatibility can affect the value of  $I_{CHG}$  under certain operating conditions. The voltage on ISO\_Sx is regulated to stay at  $V_{ISO\_SFC}$  by the battery isolation FET when  $V_{ISO\_BX} < V_{ISO\_SFC}$ .

### Fast Charge Mode (Constant Voltage)

As the battery charges, its voltage rises and approaches the termination voltage,  $V_{TRM}$ . The ADP5063 charger monitors the voltage on the BAT\_SNS pin to determine when charging should end. However, the internal ESR of the battery pack combined with the printed circuit board (PCB) and other parasitic series resistances creates a voltage drop between the sense point at the BAT\_SNS pin and the cell terminal. To compensate for this and to ensure a fully charged cell, the ADP5063 enters a constant voltage charging mode when the termination voltage is detected on the BAT\_SNS pin. The ADP5063 reduces charge current gradually as the cell continues to charge, maintaining a voltage of  $V_{TRM}$  on the BAT\_SNS pin. During fast charge mode (constant voltage), the CHARGER\_STATUS bits are set to 011.

### Fast Charge Mode Timer

The duration of fast charge mode is monitored to ensure that the battery is charging correctly. If the fast charge mode runs for longer than  $t_{CHG}$  without the voltage at the BAT\_SNS pin reaching  $V_{TRM}$ , a fault condition is assumed and charging stops. The fault condition is asserted on the CHARGER\_STATUS[2:0] bits, allowing the user to initiate the fault recovery procedure as specified in the Fault Recovery section.

If the fast charge mode runs for longer than  $t_{CHG}$ , and  $V_{TRM}$  has been reached on the BAT\_SNS pin but the charge current has not yet fallen below  $I_{END}$ , charging stops. No fault condition is asserted in this circumstance, and charging resumes as normal if the recharge threshold is breached.

### Watchdog Timer

The ADP5063 charger features a programmable watchdog timer function to ensure that charging is under the control of the processor. The watchdog timer starts running when the ADP5063 charger determines that the processor should be operational, that is, when the processor sets the RESET\_WD bit for the first time or when the battery voltage is greater than the weak battery threshold,  $V_{WEAK}$ . When the watchdog timer has been triggered, it must be reset regularly within the watchdog timer period,  $t_{WD}$ .

While in charger mode, if the watchdog timer expires without being reset, the ADP5063 charger assumes that there is a software problem and triggers the safety timer,  $t_{SAFE}$ . For more information see the Safety Timer section.

### Safety Timer

While in charger mode, if the watchdog timer expires, the ADP5063 charger initiates the safety timer,  $t_{SAFE}$  (see the Watchdog Timer section). If the processor has programmed charging parameters by the time the charger initiates the safety timer,  $I_{LIM}$  is set to the default value. Charging continues for a period of  $t_{SAFE}$ , and then the charger switches off and sets the CHARGER\_STATUS[2:0] bits.

### Charge Complete

The ADP5063 charger monitors the charging current while in fast charge constant voltage mode. If the current falls below  $I_{END}$  and remains below  $I_{END}$  for  $t_{END}$ , charging stops and the CHDONE flag is set. If the charging current falls below  $I_{END}$  for less than  $t_{END}$  and then rises above  $I_{END}$  again, the  $t_{END}$  timer resets.

### Recharge

After the detection of charge complete and the cessation of charging, the ADP5063 charger monitors the BAT\_SNS pin as the battery discharges through normal use. If the BAT\_SNS pin voltage falls to  $V_{RCH}$ , the charger reactivates charging. Under most circumstances, triggering the recharge threshold results in the charger starting directly in fast charge constant voltage mode.

The recharge function can be disabled in the I<sup>2</sup>C interface, but a status bit (Register Address 0x0C, Bit 3) informs the system that a recharge cycle is required.

### IC Enable/Disable

The ADP5063 IC can be disabled by the DIG\_IO2 digital input pin (if factory programmed to do so) or by the I<sup>2</sup>C registers. All internal control circuits are disabled when the IC is disabled. Disabling the IC1 option can also control the states of the LDO FET and the battery isolation FET.

It is critical to note that during the disable IC1 mode, a high voltage at VINx passes to the internal supply voltage because all of the internal control circuits are disabled. The VINx supply voltage must fulfill the following condition:

$$V_{ISO\_Bx} < VINx < 5.5 \text{ V}$$

### Battery Charging Enable/Disable

The ADP5063 charging function can be disabled by setting the I<sup>2</sup>C EN\_CHG bit to low. The LDO to the system still operates under this circumstance and can be set in I<sup>2</sup>C to the default or I<sup>2</sup>C programmed system voltage from 4.3 V to 5.0 V (see Table 26 for details).

The ADP5063 charging function can also be controlled via one of the external DIG\_IOx pins (if factory programmed to do so). Any change in the I<sup>2</sup>C EN\_CHG bit takes precedence over the pin setting.

### Battery Voltage Limit to Prevent Charging

The battery monitor of the ADP5063 charger can be configured to monitor battery voltage and prevent charging when the battery voltage is higher than  $V_{CHG\_VLIM}$  (typically 3.2 V) during charging start-up (enabled by EN\_CHG or DIG\_IO3). This function can prevent unnecessary charging of a half discharged battery and, as such, can extend the lifetime of the Li-Ion battery cell. Charging starts automatically when the battery voltage drops below  $V_{CHG\_VLIM}$  and continues through full charge cycle until the battery voltage reaches  $V_{TRM}$  (typically 3.6 V).

By default, the charging voltage limit is disabled, and it can be enabled via I<sup>2</sup>C Register Address 0x08, Bit 5 (EN\_CHG\_VLIM).

### SYS\_EN Output

The ADP5063 features a SYS\_EN open-drain FET to enable the system until the battery is at the minimum required level for guaranteed system startup. When there are minimum battery voltage and/or minimum battery charge level requirements, the operation of SYS\_EN can be set by I<sup>2</sup>C programming. The SYS\_EN operation can be factory programmed to four different operating conditions, as described in Table 11.

Table 11. SYS\_EN Mode Descriptions

SYS_EN Mode Selection	Description
00	SYS_EN is activated when LDO is active and system voltage is available.
01	SYS_EN is activated by the ISO_Bx voltage, the battery charging mode.
10	SYS_EN is activated and the isolation FET is disabled when the battery drops below $V_{WEAK}$ . This option is active when $V_{INx} = 0 \text{ V}$ and the battery monitor is activated from Register 0x07, Bit 5 (EN_BMON).
11	SYS_EN is active in LDO mode when the charger is disabled. SYS_EN is active in charging mode when $V_{ISO\_Bx} \geq V_{WEAK}$ .

### Indicator LED Output (ILED)

The ILED is an open-drain output for an indicator LED connection. Optionally, the ILED output can be used as a status output for a microcontroller. Indicator LED modes are listed in Table 12.

Table 12. Indicator LED Operation Modes

ADP5063 Mode	ILED Mode	On/Off Time
IC Off	Off	
LDO Mode Off	Off	
LDO Mode On	Off	
Charge Mode	Continuously on	
Timer Error ( $t_{TRK}$ , $t_{CHG}$ , $t_{SAFE}$ )	Blinking	167 ms/833 ms
Overtemperature ( $T_{SD}$ )	Blinking	1 sec/1 sec

## THERMAL MANAGEMENT

### ***Isothermal Charging***

The ADP5063 includes a thermal feedback loop that limits the charge current when the die temperature exceeds  $T_{LIM}$  (typically 115°C). As the on-chip power dissipation and die temperature increase, the charge current is automatically reduced to maintain the die temperature within the recommended range. As the die temperature decreases due to lower on-chip power dissipation or ambient temperature, the charge current returns to the programmed level. During isothermal charging, the THERM\_LIM I<sup>2</sup>C flag is set to high.

This thermal feedback control loop allows the user to set the programmed charge current based on typical rather than worst-case conditions.

The ADP5063 does not include a thermal feedback loop to limit ISO\_Sx load current in LDO mode. If the power dissipated on chip during LDO mode causes the die temperature to exceed 130°C, an interrupt is generated. If the die temperature continues to rise beyond 140°C, the device enters thermal shutdown.

### ***Thermal Shutdown and Thermal Early Warning***

The ADP5063 charger features a thermal shutdown threshold detector. If the die temperature exceeds  $T_{SD}$ , the ADP5063 charger is disabled, and the TSD 140°C bit is set. The ADP5063 charger can be reenabled when the die temperature drops below the  $T_{SD}$  falling limit and the TSD 140°C bit is reset. To reset the TSD 140°C bit, write to the I<sup>2</sup>C fault register, Register Address 0x0D (Bit 0) or cycle the power.

Before the die temperature reaches  $T_{SD}$ , the early warning bit is set if  $T_{SDL}$  is exceeded. This allows the system to accommodate power consumption before thermal shutdown occurs.

### ***Fault Recovery***

Before performing the following operation, it is important to ensure that the cause of the fault has been rectified.

To recover from a charger fault (when CHARGER\_STATUS[2:0] = 110), cycle the power on VINx or write high to reset the I<sup>2</sup>C fault bits in the fault register (Register Address 0x0D).

## BATTERY ISOLATION FET

The ADP5063 charger features an integrated battery isolation FET for power path control. The battery isolation FET isolates a deeply discharged Li-Ion cell from the system power supply in both trickle and fast charge modes, thereby allowing the system to be powered at all times.

When VINx is below  $V_{VIN\_OK\_RISE}$ , the battery isolation FET is in full conducting mode.

The battery isolation FET is off during trickle charge mode. When the battery voltage exceeds  $V_{TRK\_DEAD}$ , the battery isolation FET switches to the system voltage regulation mode. During system voltage regulation mode, the battery isolation FET maintains the  $V_{ISO\_SFC}$  voltage on the ISO\_Sx pins. When the battery voltage exceeds  $V_{ISO\_SFC}$ , the battery isolation FET is in full conducting mode.

The battery isolation FET supplements the battery to support high current functions on the system power supply. When the voltage on ISO\_Sx drops below  $V_{ISO\_BX}$ , the battery isolation FET enters into full conducting mode. When voltage on ISO\_Sx rises above  $V_{ISO\_BX}$ , the isolation FET enters regulating mode or full conduction mode, depending on the Li-Ion cell voltage and the linear charger mode.

## BATTERY DETECTION

### ***Battery Voltage Level Detection***

The ADP5063 charger features a battery detection mechanism to detect an absent battery. The charger actively sinks and sources current into the ISO\_Bx node, and voltage vs. time is detected. The sink phase is used to detect a charged battery, whereas the source phase is used to detect a discharged battery.

The sink phase (see Figure 27) sinks  $I_{SINK}$  current from the ISO\_Bx pins for a time period,  $t_{BATOK}$ . If ISO\_Bx is below  $V_{BATL}$  when the  $t_{BATOK}$  timer expires, the charger assumes that no battery is present and starts the source phase. If the ISO\_Bx pin exceeds the  $V_{BATL}$  voltage when the  $t_{BATOK}$  timer expires, the charger assumes that the battery is present and begins a new charge cycle.

The source phase sources  $I_{SOURCE}$  current to the ISO\_Bx pins for a time period,  $t_{BATOK}$ . If ISO\_Bx exceeds  $V_{BATH}$  before the  $t_{BATOK}$  timer expires, the charger assumes that no battery is present. If the ISO\_Bx pin does not exceed the  $V_{BATH}$  voltage when the  $t_{BATOK}$  timer expires, the charger assumes that a battery is present and begins a new charge cycle.

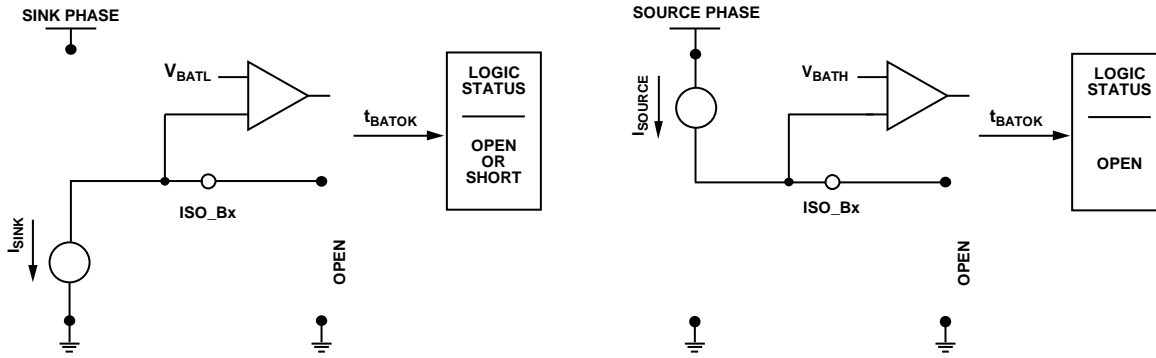


Figure 27. Sink Phase

11589-030

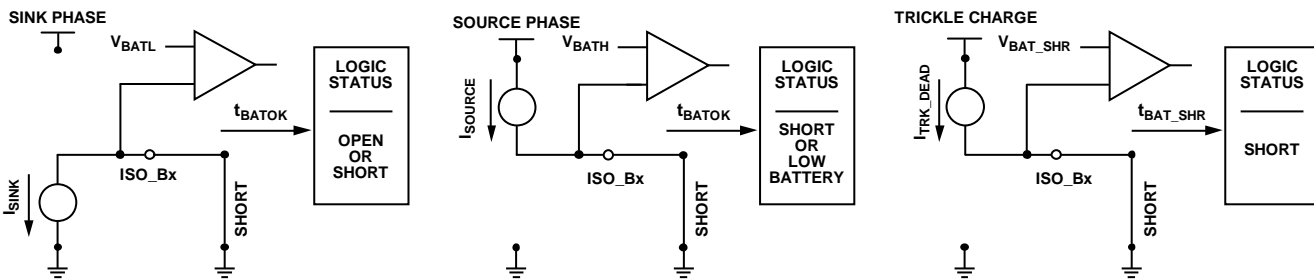


Figure 28. Trickle Charge

11589-031

**Battery (ISO\_Bx) Short Detection**

A battery short occurs under a damaged battery condition or when the battery protection circuitry is enabled.

On commencing trickle charging, the ADP5063 charger monitors the battery voltage. If this battery voltage does not exceed  $V_{BAT\_SHR}$  within the specified timeout period,  $t_{BAT\_SHR}$ , a fault is declared and the charger is stopped by turning the battery isolation FET off, but the system voltage is maintained at  $V_{ISO\_STRK}$  by the linear regulator.

After source phase, if the ISO\_Bx or BAT\_SNS level remains below  $V_{BATH}$ , either the battery voltage is low or the battery node is shorted. Because the battery voltage is low, trickle charging mode is initiated (see Figure 28). If the BAT\_SNS level remains below  $V_{BAT\_SHR}$  after  $t_{BAT\_SHR}$  has elapsed, the ADP5063 assumes that the battery node is shorted.

The trickle charge branch is active during the battery short scenario, and trickle charge current to the battery is maintained until the 60-minute trickle charge mode timer expires.

**BATTERY PACK TEMPERATURE SENSING**

**Battery Thermistor Input**

The ADP5063 charger features battery pack temperature sensing that precludes charging when the battery pack temperature is outside the specified range. The THR pin provides an on and off switching current source that must be connected directly to the battery pack thermistor terminal. The activation interval of the THR current source is 167 ms.

The battery pack temperature sensing can be controlled by I<sup>2</sup>C, using the conditions shown in Table 13. Note that the I<sup>2</sup>C register default setting for EN\_THR (Register Address 0x07) is 0 = temperature sensing off.

**Table 13. THR Input Function**

Conditions		THR Function
VINx	VISO_Bx	
Open or VIN = 0 V to 4.0 V	<2.5 V	Off
Open or VIN = 0 V to 4.0 V	>2.5 V	Off, controlled by I <sup>2</sup> C
4 V to 6.7 V	Don't care	Always on

If the battery pack thermistor is not connected directly to the THR pin, a 10 kΩ (tolerance ±20%) dummy resistor must be connected between the THR input and AGND. Leaving the THR pin open results in a false detection of the battery temperature being <0°C, and charging is disabled.

The ADP5063 charger monitors the voltage in the THR pin and suspends charging when the current is outside the range of less than 0°C or greater than 60°C.

The ADP5063 charger is designed for use with an NTC thermistor in the battery pack with a nominal room temperature value of either 10 kΩ at 25°C or 100 kΩ at 25°C, which is selected by factory programming.

The ADP5063 charger is designed for use with an NTC thermistor in the battery pack with a temperature coefficient curve (beta). Factory programming supports eight beta values covering a range from 3150 to 4400 (see Table 43).

**JEITA Li-Ion Battery Temperature Charging Specification**

The ADP5063 is compliant with the JEITA1 and JEITA2 Li-Ion battery charging temperature specifications as outlined in Table 14 and Table 16, respectively.

JEITA function can be enabled via the I<sup>2</sup>C interface and, optionally, the JEITA1 or JEITA2 function can be selected via the I<sup>2</sup>C

interface. Alternatively, the JEITA1 or JEITA2 function can be enabled as the default setting by factory programming.

When the ADP5063 identifies a hot or cold battery condition, the ADP5063 takes the following actions:

- Stops charging the battery.
- Connects or enables the battery isolation FET such that the ADP5063 continues in LDO mode.

**Table 14. JEITA1 Specifications**

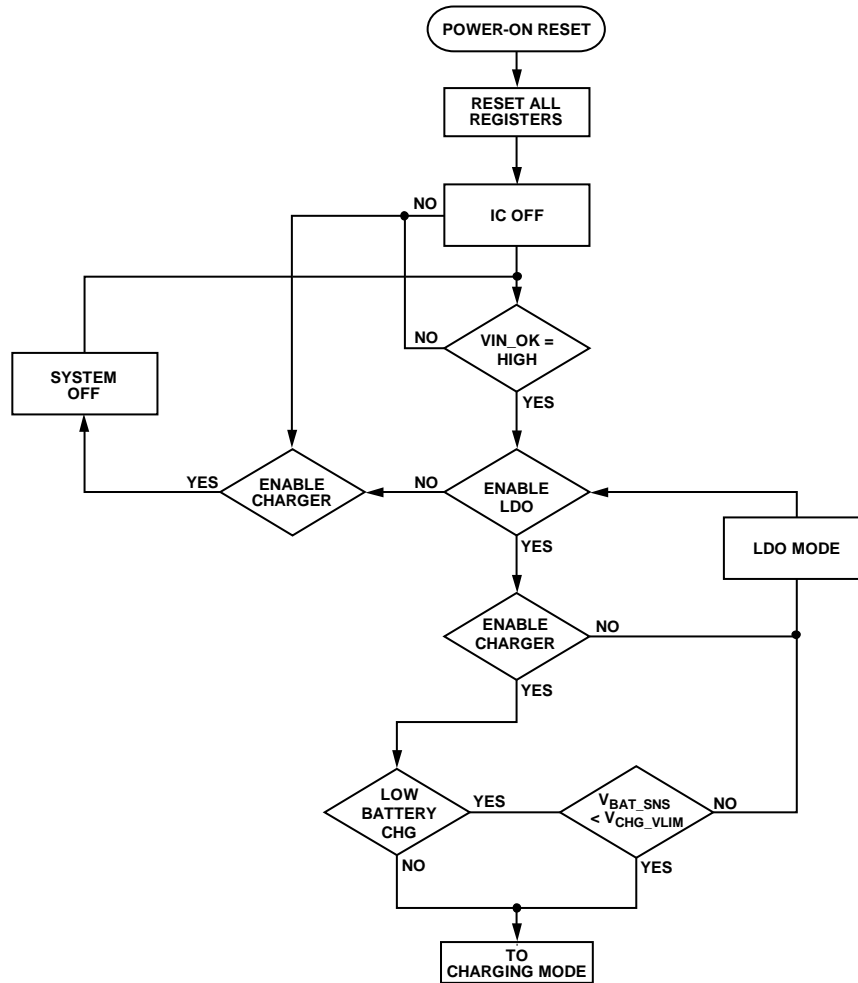
Parameter	Symbol	Conditions	Min	Max	Unit
JEITA1 Cold Temperature Limits	I <sub>JEITA_COLD</sub>	No battery charging occurs.		0	°C
JEITA1 Cool Temperature Limits	I <sub>JEITA_COOL</sub>	Battery charging occurs at approximately 50% of the programmed level. See Table 15 for specific charging current reduction levels.	0	10	°C
JEITA1 Typical Temperature Limits	I <sub>JEITA_TYP</sub>	Normal battery charging occurs at the default/programmed levels.	10	45	°C
JEITA1 Warm Temperature Limits	I <sub>JEITA_WARM</sub>	Battery termination voltage (V <sub>TRM</sub> ) is reduced by 100 mV from the programmed value.	45	60	°C
JEITA1 Hot Temperature Limits	I <sub>JEITA_HOT</sub>	No battery charging occurs.	60		°C

**Table 15. JEITA1 Reduced Charge Current Levels, Battery Cool Temperature**

ICHG[4:0] (Default)	ICHG JEITA1
00000 = 50 mA	50 mA
00001 = 100 mA	50 mA
00010 = 150 mA	50 mA
00011 = 200 mA	100 mA
00100 = 250 mA	100 mA
00101 = 300 mA	150 mA
00110 = 350 mA	150 mA
00111 = 400 mA	200 mA
01000 = 450 mA	200 mA
01001 = 500 mA	250 mA
01010 = 550 mA	250 mA
01011 = 600 mA	300 mA
01100 = 650 mA	300 mA
01101 = 700 mA	350 mA
01110 = 750 mA	350 mA
01111 = 800 mA	400 mA
10000 = 850 mA	400 mA
10001 = 900 mA	450 mA
10010 = 950 mA	450 mA
10011 = 1000 mA	500 mA
10100 = 1050 mA	500 mA
10101 = 1100 mA	550 mA
10110 = 1200 mA	600 mA
10111 to 11111 = 1300 mA	650 mA

**Table 16. JEITA2 Specifications**

Parameter	Symbol	Conditions	Min	Max	Unit
JEITA2 Cold Temperature Limits	I <sub>JEITA_COLD</sub>	No battery charging occurs.		0	°C
JEITA2 Cool Temperature Limits	I <sub>JEITA_COOL</sub>	Battery termination voltage (V <sub>TRM</sub> ) is reduced by 100 mV from the programmed value.	0	10	°C
JEITA2 Typical Temperature Limits	I <sub>JEITA_TYP</sub>	Normal battery charging occurs at the default/programmed levels.	10	45	°C
JEITA2 Warm Temperature Limits	I <sub>JEITA_WARM</sub>	Battery termination voltage (V <sub>TRM</sub> ) is reduced by 100 mV from the programmed value.	45	60	°C
JEITA2 Hot Temperature Limits	I <sub>JEITA_HOT</sub>	No battery charging occurs.	60		°C



11553-102

Figure 29. Simplified Battery and VINx Connect Flowchart

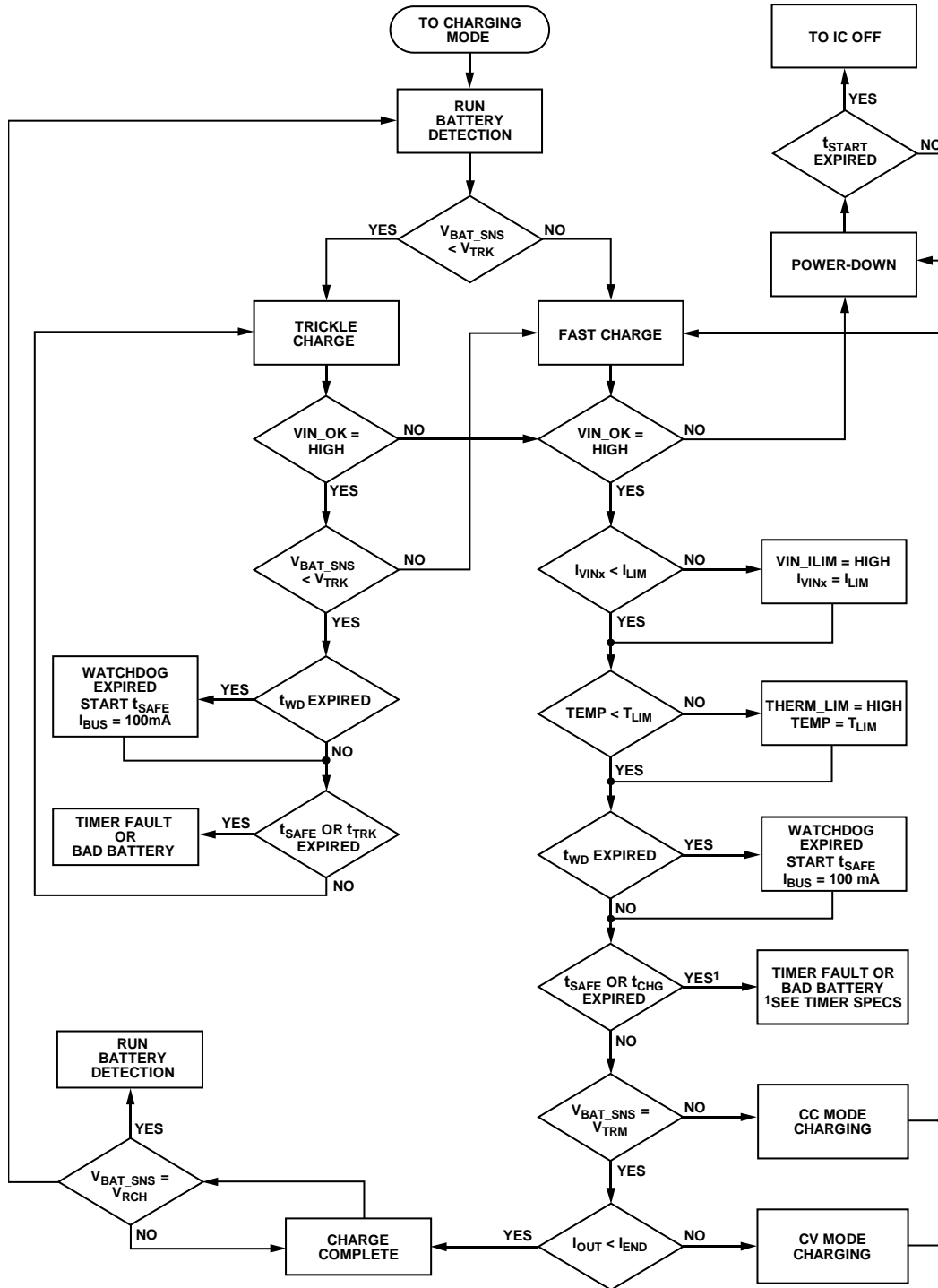


Figure 30. Simplified Charging Mode Flowchart

11593-033

**I<sup>2</sup>C INTERFACE**

The ADP5063 includes an I<sup>2</sup>C-compatible serial interface for control of the charging and LDO functions, as well as for a readback of the system status registers. The I<sup>2</sup>C chip address is 0x28 in write mode and 0x29 in read mode.

Register values are reset to the default values when the VINx supply falls below the falling voltage threshold, V<sub>VIN\_OK\_FALL</sub>. The I<sup>2</sup>C registers also reset when the battery is disconnected and V<sub>IN</sub> is 0 V.

The subaddress content selects which of the ADP5063 registers is written to first. The ADP5063 sends an acknowledgement to

the master after the 8-bit data byte has been written (see Figure 31 for an example of the I<sup>2</sup>C write sequence to a single register). The ADP5063 increments the subaddress automatically and starts receiving a data byte at the next register until the master sends an I<sup>2</sup>C stop, as shown in Figure 32.

Figure 33 shows the I<sup>2</sup>C read sequence of a single register. ADP5063 sends the data from the register denoted by the subaddress and increments the subaddress automatically, sending data from the next register until the master sends an I<sup>2</sup>C stop condition, as shown in Figure 34.

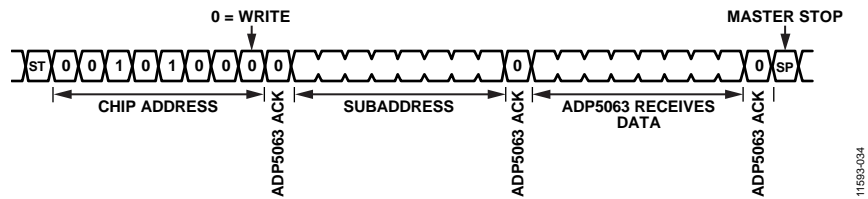


Figure 31. I<sup>2</sup>C Single Register Write Sequence

11593-034

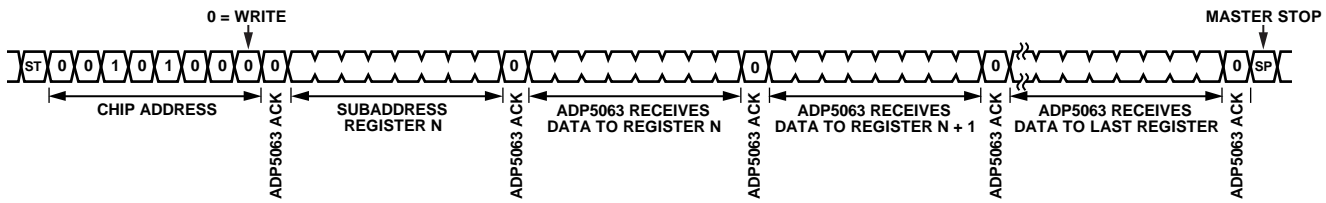


Figure 32. I<sup>2</sup>C Multiple Register Write Sequence

11593-035

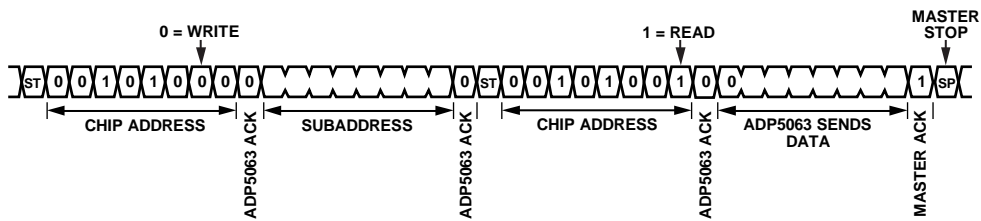


Figure 33. I<sup>2</sup>C Single Register Read Sequence

11593-036

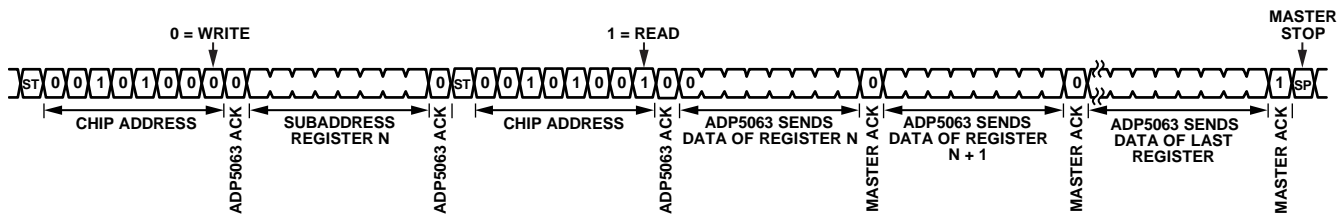


Figure 34. I<sup>2</sup>C Multiple Register Read Sequence

11593-037

I<sup>2</sup>C REGISTER MAP

See the Factory-Programmable Options section for programming option details. Note that a blank cell indicates a bit that is not used or is reserved for future use.

Table 17. I<sup>2</sup>C Register Map

Register		D7	D6	D5	D4	D3	D2	D1	D0
Addr.	Name								
0x00	Manufacturer and model ID	MANUF[3:0]				Model[3:0]			
0x01	Silicon revision					REV[3:0]			
0x02	VINx pin settings					ILIM[3:0] <sup>1</sup>			
0x03	Termination settings	VTRM[5:0] <sup>1,2</sup>							CHG_VLIM[1:0] <sup>1,2</sup>
0x04	Charging current settings		ICHG[4:0] <sup>1,2</sup>						ITRK_DEAD[1:0] <sup>1</sup>
0x05	Voltage thresholds	DIS_RCH <sup>1,3</sup>	VRCH[1:0] <sup>1</sup>		VTRK_DEAD[1:0] <sup>1,3</sup>		VWEAK[2:0] <sup>1</sup>		
0x06	Timer settings			EN_TEND <sup>1</sup>	EN_CHG_TIMER <sup>1</sup>	CHG_TMR_PERIOD <sup>1</sup>	EN_WD <sup>1,3</sup>	WD_PERIOD <sup>1</sup>	RESET_WD
0x07	Functional Settings 1		DIS_IC <sup>1</sup>	EN_BMON <sup>1</sup>	EN_THR <sup>1</sup>	DIS_LDO <sup>1</sup>	EN_EOC <sup>1</sup>		EN_CHG <sup>1</sup>
0x08	Functional Settings 2	EN_JEITA <sup>1,3</sup>	JEITA_SELECT <sup>1,3</sup>	EN_CHG_VLIM <sup>1,3</sup>	IDEAL_DIODE[1:0] <sup>1,3</sup>		VSYSTEM[2:0] <sup>1,3</sup>		
0x09	Interrupt enable		EN_THERM_LIM_INT	EN_WD_INT	EN_TSD_INT	EN_THR_INT	EN_BAT_INT	EN_CHG_INT	EN_VIN_INT
0x0A	Interrupt active		THERM_LIM_INT	WD_INT	TSD_INT	THR_INT	BAT_INT	CHG_INT	VIN_INT
0x0B	Charger Status 1	VIN_OV	VIN_OK	VIN_ILIM	THERM_LIM	CHDONE	CHARGER_STATUS[2:0]		
0x0C	Charger Status 2	THR_STATUS[2:0]					RCH_LIM_INFO	BATTERY_STATUS[2:0]	
0x0D	Fault					BAT_SHR <sup>1</sup>		TSD 130°C <sup>1</sup>	TSD 140°C <sup>1</sup>
0x10	Battery short	TBAT_SHR[2:0] <sup>1</sup>						VBAT_SHR[2:0] <sup>1</sup>	
0x11	IEND	IEND[2:0] <sup>1,3</sup>			C/20 EOC <sup>1</sup>	C/10 EOC <sup>1</sup>	C/5 EOC <sup>1</sup>	SYS_EN_SET[1:0] <sup>1,3</sup>	

<sup>1</sup> These bits reset to default I<sup>2</sup>C values when VINx is connected or disconnected.

<sup>2</sup> The default I<sup>2</sup>C values of these bits are partially factory programmable.

<sup>3</sup> The default I<sup>2</sup>C values of these bits are fully factory programmable.

**REGISTER BIT DESCRIPTIONS**

In Table 18 through Table 33, the following abbreviations are used: R is read only, W is write only, R/W is read/write, and N/A means not applicable.

**Table 18. Manufacturer and Model ID, Register Address 0x00**

Bit No.	Bit Name	Access	Default	Description
[7:4]	MANUF[3:0]	R	0001	The 4-bit manufacturer identification bus
[3:0]	MODEL[3:0]	R	1001	The 4-bit model identification bus

**Table 19. Silicon Revision, Register Address 0x01**

Bit No.	Bit Name	Access	Default	Description
[7:4]	Not used	R		
[3:0]	REV[3:0]	R	0111	The 4-bit silicon revision identification bus

**Table 20. VINx Pin Settings, Register Address 0x02**

Bit No.	Bit Name	Access	Default	Description
[7:4]	Not used	R		
[3:0]	ILIM[3:0]	R/W	0000 = 100 mA	VINx input current-limit programming bus. The current into VINx can be limited to the following programmed values: 0000 = 100 mA. 0001 = 150 mA. 0010 = 200 mA. 0011 = 250 mA. 0100 = 300 mA. 0101 = 400 mA. 0110 = 500 mA. 0111 = 600 mA. 1000 = 700 mA. 1001 = 800 mA. 1010 = 900 mA. 1011 = 1000 mA. 1100 = 1200 mA. 1101 = 1500 mA. 1110 = 1800 mA. 1111 = 2100 mA.

Table 21. Termination Settings, Register Address 0x03

Bit No.	Bit Name	Access	Default	Description
[7:2]	VTRM[5:0]	R/W	000101 = 3.60 V	<p>Termination voltage programming bus. The values of the floating voltage can be programmed to the following values:</p> <p>000101 = 3.60 V.  000110 = 3.62 V.  000111 = 3.64 V.  001000 = 3.66 V.  001001 = 3.68 V.  001010 = 3.70 V.  001011 = 3.72 V.  001100 = 3.74 V.  001101 = 3.76 V.  001110 = 3.78 V.  001111 = 3.80 V.  010000 = 3.82 V.  010001 = 3.84 V.  010010 = 3.86 V.  010011 = 3.88 V.  010100 = 3.90 V.  010101 = 3.92 V.  010110 = 3.94 V.  010111 = 3.96 V.  011000 = 3.98 V.  011001 = 4.00 V.  011010 = 4.02 V.  011011 = 4.04 V.  011100 = 4.06 V.  011101 = 4.08 V.  011110 = 4.10 V.  011111 = 4.12 V.  100000 = 4.14 V.  100001 = 4.16 V.  100010 = 4.18 V.  100011 = 4.20 V.  100100 = 4.22 V.  100101 = 4.24 V.  100110 = 4.26 V.  100111 = 4.28 V.  101000 = 4.30 V.  101001 = 4.32 V.  101010 = 4.34 V.  101011 = 4.36 V.  101100 = 4.38 V.  101101 = 4.40 V.  101110 = 4.42 V.  101111 = 4.44 V.  110000 = 4.44 V.  110001 = 4.46 V.  110010 = 4.48 V.  110011 to 111111 = 4.50 V.</p>
[1:0]	CHG_VLIM[1:0]	R/W	00 = 3.2 V	<p>Charging voltage limit programming bus. The values of the charging voltage limit can be programmed to the following values:</p> <p>00 = 3.2 V.  01 = 3.4 V.  10 = 3.7 V.  11 = 3.8 V.</p>

Table 22. Charging Current Settings, Register Address 0x04

Bit No.	Bit Name	Access	Default	Description
7	Not used	R		
[6:2]	ICHG[4:0]	R/W	01110 = 750 mA	Fast charge current programming bus. The values of the constant current charge can be programmed to the the following values: 00000 = 50 mA. 00001 = 100 mA. 00010 = 150 mA. 00011 = 200 mA. 00100 = 250 mA. 00101 = 300 mA. 00110 = 350 mA. 00111 = 400 mA. 01000 = 450 mA. 01001 = 500 mA. 01010 = 550 mA. 01011 = 600 mA. 01100 = 650 mA. 01101 = 700 mA. 01110 = 750 mA. 01111 = 800 mA. 10000 = 850 mA. 10001 = 900 mA. 10010 = 950 mA. 10011 = 1000 mA. 10100 = 1050 mA. 10101 = 1100 mA. 10110 = 1200 mA. 10111 to 11111 = 1300 mA.
[1:0]	ITRK_DEAD[1:0]	R/W	10 = 20 mA	Trickle and weak charge current programming bus. The values of the trickle and weak charge currents can be programmed to the following values: 00 = 5 mA. 01 = 10 mA. 10 = 20 mA. 11 = 80 mA.

Table 23. Voltage Thresholds, Register Address 0x05

Bit No.	Bit Name	Access	Default	Description
7	DIS_RCH	R/W	0 = recharge enabled 1 = recharge disabled	0 = recharge enabled. 1 = recharge disabled.
[6:5]	VRCH[1:0]	R/W	11 = 260 mV	Recharge voltage programming bus. The values of the recharge threshold can be programmed to the following values (note that the recharge cycle can be disabled in I <sup>2</sup> C by using the DIS_RCH bit): 00 = 80 mV. 01 = 140 mV. 10 = 200 mV. 11 = 260 mV.

Bit No.	Bit Name	Access	Default	Description
[4:3]	VTRK_DEAD[1:0]	R/W	00 = 2.0 V	Trickle to fast charge dead battery voltage programming bus. The values of the trickle to fast charge threshold can be programmed to the following values: 00 = 2.0 V. 01 = 2.5 V. 10 = 2.6 V. 11 = 2.9 V.
[2:0]	VWEAK[2:0]	R/W	011 = 3.0 V	Weak battery voltage rising threshold. 000 = 2.7 V. 001 = 2.8 V. 010 = 2.9 V. 011 = 3.0 V. 100 = 3.1 V. 101 = 3.2 V. 110 = 3.3 V. 111 = 3.4 V.

Table 24. Timer Settings, Register Address 0x06

Bit No.	Bit Name	Access	Default	Description
[7:6]	Not used			
5	EN_TEND	R/W	1	0 = charge complete timer, $t_{END}$ , disabled. A 31 ms deglitch timer remains on. 1 = charge complete timer enabled.
4	EN_CHG_TIMER	R/W	1	0 = trickle/fast charge timer disabled. 1 = trickle/fast charge timer enabled.
3	CHG_TMR_PERIOD	R/W	1	Trickle and fast charge timer period. 0 = 30 sec trickle charge timer and 300-minute fast charge timer. 1 = 60 sec trickle charge timer and 600-minute fast charge timer.
2	EN_WD	R/W	0	0 = watchdog timer is disabled even when BAT_SNS exceeds $V_{WEAK}$ . 1 = watchdog timer safety timer is enabled.
1	WD_PERIOD	R/W	0	Watchdog safety timer period. 0 = 32 sec watchdog timer and 40-minute safety timer. 1 = 64 sec watchdog timer and 40-minute safety timer.
0	RESET_WD	W	0	When RESET_WD is set to logic high by I <sup>2</sup> C, the watchdog safety timer is reset.

Table 25. Functional Settings 1, Register Address 0x07

Bit No.	Bit Name	Access	Default	Description
7	Not used			
6	DIS_IC1	R/W	0	0 = normal operation. 1 = the ADP5063 is disabled; $V_{VINx}$ must be $V_{ISO\_Bx} < V_{VINx} < 5.5 V$ .
5	EN_BMON	R/W	0	0 = when $V_{VINx} < V_{VIN\_OK\_RISE}$ or $V_{VIN\_OK\_FALL}$ , the battery monitor is disabled. When $V_{VINx} = 4 V$ to 6.7 V, the battery monitor is enabled regardless of the EN_BMON state. 1 = the battery monitor is enabled even when the voltage at the VINx pins is below $V_{VIN\_OK}$ .
4	EN_THR	R/W	0	0 = when $V_{VINx} < V_{VIN\_OK\_RISE}$ or $V_{VIN\_OK\_FALL}$ , the THR current source is disabled. When $V_{VINx} = 4 V$ to 6.7 V, the THR current source is enabled regardless of the EN_THR state. 1 = THR current source is enabled even when the voltage at the VINx pins is below $V_{VIN\_OK\_RISE}$ or $V_{VIN\_OK\_FALL}$ .
3	DIS_LDO	R/W	0	0 = LDO is enabled. 1 = LDO is off. In addition, if EN_CHG = low, the battery isolation FET is on. If EN_CHG = high, the battery isolation FET is off.

Bit No.	Bit Name	Access	Default	Description
2	EN_EOC	R/W	1	0 = end of charge not allowed. 1 = end of charge allowed.
1	Not used			
0	EN_CHG	R/W	0	0 = battery charging is disabled. 1 = battery charging is enabled.

Table 26. Functional Settings 2, Register Address 0x08

Bit No.	Bit Name	Access	Default	Description
7	EN_JEITA	R/W	0 = JEITA disabled	0 = JEITA compliance of the Li-Ion temperature battery charging specifications is disabled. 1 = JEITA compliance enabled.
6	JEITA_SELECT	R/W	0 = JEITA1	0 = JEITA1 is selected. 1 = JEITA2 is selected.
5	EN_CHG_VLIM	R/W	0	0 = charging voltage limit disabled. 1 = voltage limit enabled. The charger prevents charging until the battery voltage drops below the $V_{CHG\_VLIM}$ threshold.
[4:3]	IDEAL_DIODE[1:0]	R/W	00	00 = ideal diode operates constantly when $V_{ISO\_Sx} < V_{ISO\_Bx}$ . 01 = ideal diode operates when $V_{ISO\_Sx} < V_{ISO\_Bx}$ and $V_{BAT\_SNS} > V_{WEAK}$ . 10 = ideal diode is disabled. 11 = ideal diode is disabled.
[2:0]	VSYSTEM[2:0]	R/W	000 = 4.3 V	System voltage programming bus. The values of the system voltage can be programmed to the following values: 000 = 4.3 V. 001 = 4.4 V. 010 = 4.5 V. 011 = 4.6 V. 100 = 4.7 V. 101 = 4.8 V. 110 = 4.9 V. 111 = 5.0 V.

Table 27. Interrupt Enable, Register Address 0x09

Bit No.	Bit Name	Access	Default	Description
7	Not used			
6	EN_THERM_LIM_INT	R/W	0	0 = isothermal charging interrupt is disabled. 1 = isothermal charging interrupt is enabled.
5	EN_WD_INT	R/W	0	0 = watchdog alarm interrupt is disabled. 1 = watchdog alarm interrupt is enabled.
4	EN_TSD_INT	R/W	0	0 = overtemperature interrupt is disabled. 1 = overtemperature interrupt is enabled.
3	EN_THR_INT	R/W	0	0 = THR temperature thresholds interrupt is disabled. 1 = THR temperature thresholds interrupt is enabled.
2	EN_BAT_INT	R/W	0	0 = battery voltage thresholds interrupt is disabled. 1 = battery voltage thresholds interrupt is enabled.
1	EN_CHG_INT	R/W	0	0 = charger mode change interrupt is disabled. 1 = charger mode change interrupt is enabled.
0	EN_VIN_INT	R/W	0	0 = VINx pin voltage thresholds interrupt is disabled. 1 = VINx pin voltage thresholds interrupt is enabled.

Table 28. Interrupt Active, Register Address 0x0A

Bit No.	Bit Name	Access	Default	Description
7	Not used			
6	THERM_LIM_INT	R	0	0 = no interrupt. 1 = indicates an interrupt caused by isothermal charging.
5	WD_INT	R	0	0 = no interrupt. 1 = indicates an interrupt caused by the watchdog alarm. The watchdog timer expires within 2 sec or 4 sec, depending on the watchdog period setting of 32 sec or 64 sec, respectively.
4	TSD_INT	R	0	0 = no interrupt. 1 = indicates an interrupt caused by an overtemperature fault.
3	THR_INT	R	0	0 = no interrupt. 1 = indicates an interrupt caused by THR temperature thresholds.
2	BAT_INT	R	0	0 = no interrupt. 1 = indicates an interrupt caused by battery voltage thresholds.
1	CHG_INT	R	0	0 = no interrupt. 1 = indicates an interrupt caused by a charger mode change.
0	VIN_INT	R	0	0 = no interrupt. 1 = indicates an interrupt caused by VINx voltage thresholds.

Table 29. Charger Status 1, Register Address 0x0B

Bit No.	Bit Name	Access	Default	Description
7	VIN_OV	R	N/A	1 = the voltage at the VINx pins exceeds $V_{VIN\_OV}$ .
6	VIN_OK	R	N/A	1 = the voltage at the VINx pins exceeds $V_{VIN\_OK\_RISE}$ and $V_{VIN\_OK\_FALL}$ .
5	VIN_ILIM	R	N/A	1 = the current into a VINx pin is limited by the high voltage blocking FET and the charger is not running at the full programmed $I_{CHG}$ .
4	THERM_LIM	R	N/A	1 = the charger is not running at the full programmed $I_{CHG}$ but is limited by the die temperature.
3	CHDONE	R	N/A	1 = the end of a charge cycle has been reached. This bit latches on, in that it does not reset to low when the $V_{RCH}$ threshold is breached.
[2:0]	CHARGER_STATUS[2:0]	R	N/A	Charger status bus. 000 = off. 001 = trickle charge. 010 = fast charge (CC mode). 011 = fast charge (CV mode). 100 = charge complete. 101 = LDO mode. 110 = trickle or fast charge timer expired. 111 = battery detection.

Table 30. Charger Status 2, Register Address 0x0C

Bit No.	Mnemonic	Access	Default	Description
[7:5]	THR_STATUS[2:0]	R	N/A	THR pin status. 000 = off. 001 = battery cold. 010 = battery cool. 011 = battery warm. 100 = battery hot. 111 = thermistor OK.
4	Not used			
3	RCH_LIM_INFO	R	N/A	The recharge limit information function is activated when DIS_RCH is logic high and CHARGER_STATUS[2:0] = 100 (binary). The RCH_LIM_INFO bit informs the system that a recharge cycle is required. $0 = V_{BAT\_SNS} > V_{RCH}$ $1 = V_{BAT\_SNS} < V_{RCH}$
[2:0]	BATTERY_STATUS[2:0]	R		Battery status bus. 000 = battery monitor off. 001 = no battery. $010 = V_{BAT\_SNS} < V_{TRK\_DEAD}$ . $011 = V_{TRK\_DEAD} \leq V_{BAT\_SNS} < V_{WEAK}$ . $100 = V_{BAT\_SNS} \geq V_{WEAK}$ .

Table 31. Fault,<sup>1</sup> Register Address 0x0D

Bit No.	Bit Name	Access	Default	Description
[7:4]	Not used			
3	BAT_SHR	R/W	0	0 = no fault. 1 = indicates detection of a battery short.
2	Not used	R/W		
1	TSD 130°C	R/W	0	0 = no fault. 1 = indicates an overtemperature (lower) fault.
0	TSD 140°C	R/W	0	0 = no fault. 1 = indicates an overtemperature fault.

<sup>1</sup> To reset the fault bits in the fault register, cycle the power on VINx or write the corresponding I<sup>2</sup>C bit high.

Table 32. Battery Short, Register Address 0x10

Bit No.	Bit Name	Access	Default	Description
[7:5]	TBAT_SHR[2:0]	R/W	100 = 30 sec	Battery short timeout timer. 000 = 1 sec. 001 = 2 sec. 010 = 4 sec. 011 = 10 sec. 100 = 30 sec. 101 = 60 sec. 110 = 120 sec. 111 = 180 sec.
[4:3]	Not used			
[2:0]	VBAT_SHR[2:0]	R/W	100 = 2.4 V	Battery short voltage threshold level. 000 = 2.0 V. 001 = 2.1 V. 010 = 2.2 V. 011 = 2.3 V. 100 = 2.4 V. 101 = 2.5 V. 110 = 2.6 V. 111 = 2.7 V.

Table 33. IEND, Register Address 0x11

Bit No.	Bit Name	Access	Default	Description
[7:5]	IEND[2:0]	R/W	010 = 52.5 mA	Termination current programming bus. The values of the termination current can be programmed to the following values: 000 = 12.5 mA. 001 = 32.5 mA. 010 = 52.5 mA. 011 = 72.5 mA. 100 = 92.5 mA. 101 = 117.5 mA. 110 = 142.5 mA. 111 = 170.0 mA.
4	C/20 EOC	R/W	0	The C/20 EOC bit has priority over the other settings (C/5 EOC, C/10 EOC, and IEND[2:0]). 0 = not active. 1 = the termination current is ICHG[4:0] ÷ 20 with the following limitations: Minimum value = 12.5 mA. Maximum value = 170 mA.
3	C/10 EOC	R/W	0	The C/10 EOC bit has priority over the other termination current settings (C/5 EOC and IEND[2:0]), but it does not have priority over the C/20 EOC setting. 0 = not active. 1 = the termination current is ICHG[4:0] ÷ 10, unless C/20 EOC is high. The termination current is limited to the following values: Minimum value = 12.5 mA. Maximum value = 170 mA.
2	C/5 EOC	R/W	0	The C/5 EOC bit has priority over the other termination current settings (IEND[2:0]) but it does not have priority over the C/20 EOC setting or the C/10 EOC setting. 0 = not active. 1 = the termination current is ICHG[4:0] ÷ 5, unless the C/20 EOC or the C/10 EOC bit is high. The termination current is limited to the following values: Minimum value = 12.5 mA. Maximum value = 170 mA.
1:0	SYS_EN_SET[1:0]	R/W	00	Selects the operation of the system enable pin (SYS_EN). 00 = SYS_EN is activated when the LDO is active and the system voltage is available. 01 = SYS_EN is activated by the ISO_Bx voltage, the battery charging mode. 10 = SYS_EN is activated and the isolation FET is disabled when the battery drops below $V_{WEAK}$ . <sup>1</sup> 11 = SYS_EN is active in LDO mode when the charger is disabled. SYS_EN is active in charging mode when $V_{ISO\_Bx} \geq V_{WEAK}$ .

<sup>1</sup> This option is active when  $V_{INx} = 0$  V and the battery monitor is activated from Register 0x07, Bit 5 (EN\_BMON).

## APPLICATIONS INFORMATION

### EXTERNAL COMPONENTS

#### ISO\_Sx (V<sub>OUT</sub>) Capacitor Selection

To obtain stable operation of the **ADP5063** in a safe way, the combined effective capacitance of the ISO\_Sx capacitor and the system capacitance must not be less than 10 μF and must not exceed 100 μF at any point during operation.

When choosing the capacitor value, it is also important to account for the loss of capacitance caused by the output voltage dc bias. Ceramic capacitors are manufactured with a variety of dielectrics, each with a different behavior over temperature and applied voltage. Capacitors must have a dielectric that is adequate to ensure the minimum capacitance over the necessary temperature range and dc bias conditions. X5R or X7R dielectrics with a voltage rating of 6.3 V or higher are recommended for best performance. Y5V and Z5U dielectrics are not recommended for use with any dc-to-dc converter because of their poor temperature and dc bias characteristics.

The worst-case capacitance, accounting for capacitor variation over temperature, component tolerance, and voltage, is calculated using the following equation:

$$C_{EFF} = C_{OUT} \times (1 - TEMPCO) \times (1 - TOL)$$

where:

$C_{EFF}$  is the effective capacitance at the operating voltage.

$TEMPCO$  is the worst-case capacitor temperature coefficient.

$TOL$  is the worst-case component tolerance.

In this example, the worst-case temperature coefficient ( $TEMPCO$ ) over the range of  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  is assumed to be 15% for an X5R dielectric. The tolerance of the capacitor ( $TOL$ ) is assumed to be 10%, and  $C_{OUT}$  is 16 μF at 4.2 V, as shown in Figure 35.

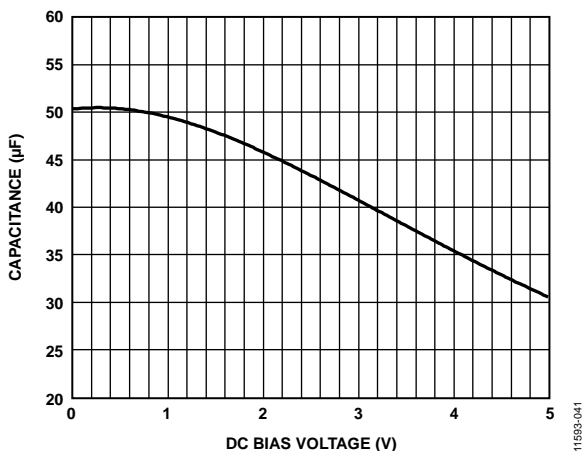


Figure 35. Murata GRM31CR61A226KE19 Capacitance vs. Bias Voltage

Substituting these values in the equation yields

$$C_{EFF} = 16 \mu\text{F} \times (1 - 0.15) \times (1 - 0.1) \approx 12.24 \mu\text{F}$$

To guarantee the performance of the charger in various operating modes, including trickle charge, constant current charge, and constant voltage charge, it is imperative that the effects of dc bias, temperature, and tolerances on the behavior of the capacitors be evaluated for each application.

#### Splitting ISO\_Sx Capacitance

In many applications, the total ISO\_Sx capacitance consists of a number of capacitors. The system voltage node (ISO\_Sx) usually supplies a single regulator or a number of ICs and regulators, each of which requires a capacitor close to its power supply input (see Figure 36).

The capacitance close to the **ADP5063** ISO\_Sx output must be at least 5 μF, as long as the total effective capacitance is at least 10 μF at any point during operation.

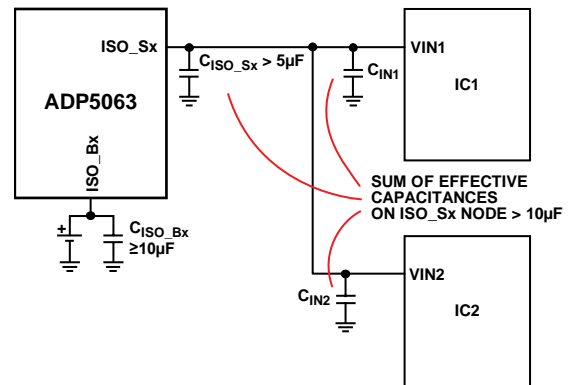


Figure 36. Splitting ISO\_Sx Capacitance

#### ISO\_Bx and ISO\_Sx Capacitor Selection

The ISO\_Bx and the ISO\_Sx effective capacitance (including temperature and dc bias effects) must not be less than 10 μF at any point during operation. Typically, a nominal capacitance of 22 μF is required to fulfill the condition at all points of operation. Suggestions for ISO\_Bx and ISO\_Sx capacitors are listed in Table 34.

#### CBP Capacitor Selection

The internal supply voltage of the **ADP5063** is equipped with a noise suppressing capacitor at the CBP terminal. Do not allow CBP capacitance to exceed 140 nF at any point during operation. Do not connect any external voltage source, any resistive load, or any other current load to the CBP terminal. Suggestions for a CBP capacitor are listed in Table 35.

**VINx Capacitor Selection**

According to the USB 2.0 specification, USB peripherals have a detectable change in capacitance on VBUS when they are attached to a USB port. The peripheral device VBUS bypass capacitance must be at least 1  $\mu\text{F}$  but not larger than 10  $\mu\text{F}$ .

The VINx input of the [ADP5063](#) is tolerant of voltages as high as 20 V; however, if an application requires exposing the VINx input to voltages of up to 20 V, the voltage range of the capacitor must also be above 20 V. Suggestions for a VINx capacitor are given in Table 36.

When using ceramic capacitors, a higher voltage range is usually achieved by selecting a component with larger physical dimensions. In applications where lower than 20 V at VINx input voltages can be guaranteed, smaller output capacitors can be used accordingly.

**Table 34. ISO\_Bx and ISO\_Sx Capacitor Suggestions**

Vendor	Part Number	Value	Voltage	Size
Murata	GRM31CR61A226KE19	22 $\mu\text{F}$	10 V	1206
Murata	GRM31CR60J226ME19	22 $\mu\text{F}$	6.3 V	1206
TDK	C3216X5R0J226M	22 $\mu\text{F}$	6.3 V	1206
Taiyo-Yuden	JMK316ABJ226KL-T	22 $\mu\text{F}$	6.3 V	1206

**Table 35. CBP Capacitor Suggestions**

Vendor	Part Number	Value	Voltage	Size
Murata	GRM155R70J104KA01	100 nF	6.3 V	0402
TDK	C1005X7R1A104K050BB	100 nF	10 V	0402

**Table 36. VINx Capacitor Suggestions**

Vendor	Part Number	Value	Voltage	Size
Murata	GRM21BR61E106MA73	10 $\mu\text{F}$	25 V	0805
TDK	C2012X5R1E106K	10 $\mu\text{F}$	25 V	0805

PCB LAYOUT GUIDELINES

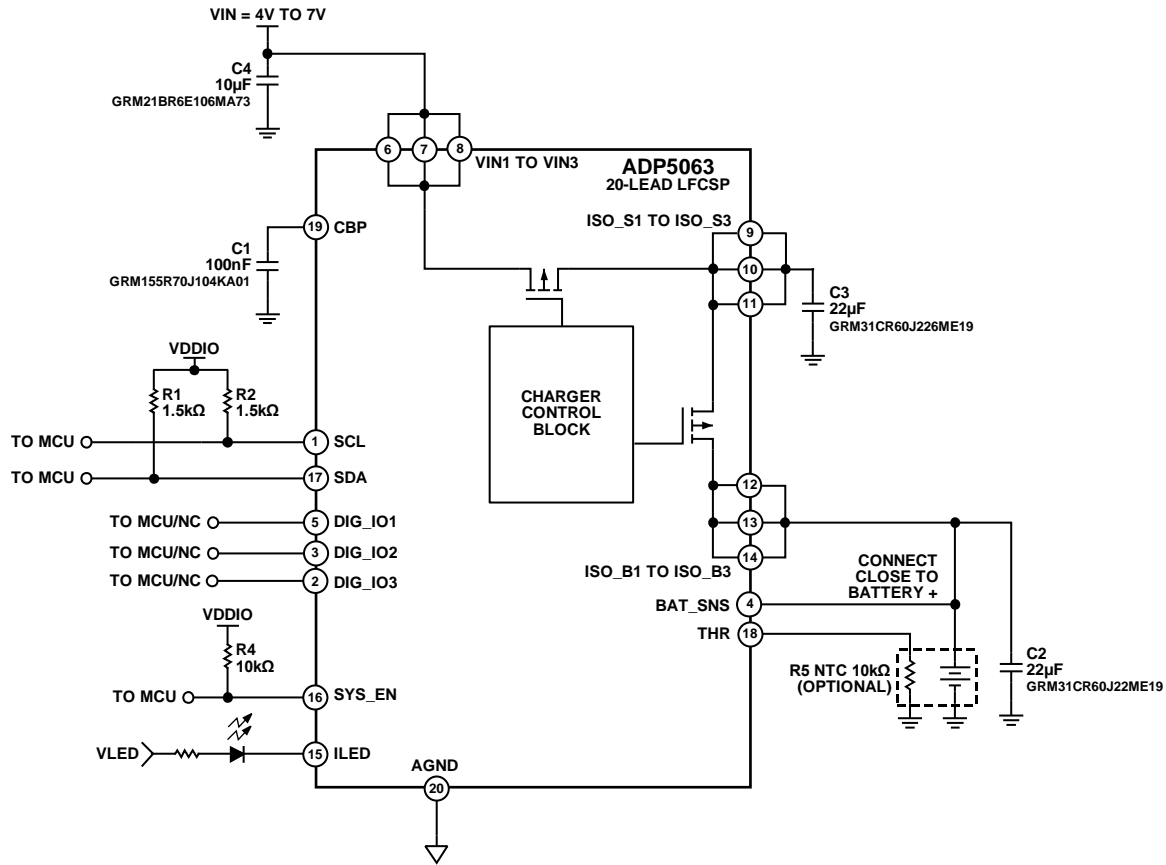


Figure 37. Reference Circuit Diagram

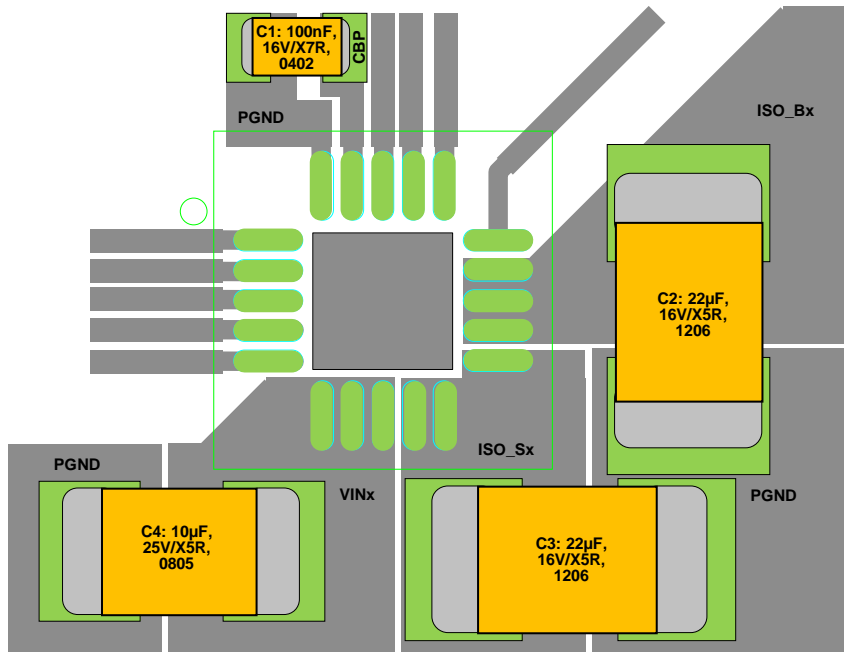


Figure 38. Reference PCB Floor Plan

## POWER DISSIPATION AND THERMAL CONSIDERATIONS

### CHARGER POWER DISSIPATION

When the **ADP5063** charger operates at high ambient temperatures and at maximum current charging and loading conditions, the junction temperature can reach the maximum allowable operating limit of 125°C.

When the junction temperature exceeds 140°C, the **ADP5063** turns off, allowing the device to cool down. When the die temperature falls below 110°C and the TSD 140°C fault bit in Register 0x0D is cleared by an I<sup>2</sup>C write, the **ADP5063** resumes normal operation.

This section provides guidelines to calculate the power dissipated in the device to ensure that the **ADP5063** operates below the maximum allowable junction temperature.

To determine the available power dissipation in different operating modes under various operating conditions, use Equation 1 through Equation 4:

$$P_D = P_{LDOFET} + P_{ISOFET} \quad (1)$$

where:

$P_{LDOFET}$  is the power dissipated in the input LDO FET.

$P_{ISOFET}$  is the power dissipated in the battery isolation FET.

Calculate the power dissipation in the LDO FET and the battery isolation FET using Equation 2 and Equation 3.

$$P_{LDOFET} = (V_{IN} - V_{ISO\_Sx}) \times (I_{CHG} + I_{LOAD}) \quad (2)$$

$$P_{ISOFET} = (V_{ISO\_Sx} - V_{ISO\_Bx}) \times I_{CHG} \quad (3)$$

where:

$V_{IN}$  is the input voltage at the VINx pins.

$V_{ISO\_Sx}$  is the system voltage at the ISO\_Sx pins.

$I_{CHG}$  is the battery charge current.

$I_{LOAD}$  is the system load current from the ISO\_Sx pins.

$V_{ISO\_Bx}$  is the battery voltage at the ISO\_Bx pins.

### LDO Mode

The system regulation voltage is user-programmable from 4.3 V to 5.0 V. In LDO mode (charging disabled, EN\_CHG = low), calculation of the total power dissipation is simplified, assuming that all current is drawn from the VINx pins and the battery is not shared with ISO\_Sx.

$$P_D = (V_{IN} - V_{ISO\_Sx}) \times I_{LOAD}$$

### Charging Mode

In charging mode, the voltage at the ISO\_Sx pins depends on the battery level. When the battery voltage is lower than  $V_{ISO\_SFC}$  (typically 3.4 V), the voltage drop over the battery isolation FET is higher and the power dissipation must be calculated using

Equation 3. When the battery voltage level reaches  $V_{ISO\_SFC}$ , the power dissipation can be calculated using Equation 4.

$$P_{ISOFET} = R_{DS(on)_ISO} \times I_{CHG} \quad (4)$$

where:

$R_{DS(on)_ISO}$  is the on resistance of the battery isolation FET (typically 110 mΩ during charging).

$I_{CHG}$  is the battery charge current.

The thermal control loop of the **ADP5063** automatically limits the charge current to maintain a die temperature below  $T_{LIM}$  (typically 115°C).

The most intuitive and practical way to calculate the power dissipation in the **ADP5063** device is to measure the power dissipated at the input and all of the outputs. Perform the measurements at the worst-case conditions (voltages, currents, and temperature). The difference between input and output power is the power that is dissipated in the device.

### JUNCTION TEMPERATURE

In cases where the board temperature,  $T_A$ , is known, the thermal resistance parameter,  $\theta_{JA}$ , can be used to estimate the junction temperature rise.  $T_J$  is calculated from  $T_A$  and  $P_D$  using the formula

$$T_J = T_A + (P_D \times \theta_{JA}) \quad (5)$$

The typical  $\theta_{JA}$  value for the 20-lead LFCSP is 35.6°C/W (see Table 5). A very important factor to consider is that  $\theta_{JA}$  is based on a 4-layer, 4 in × 3 in, 2.5 oz. copper board as per JEDEC standard, and real-world applications may use different sizes and layers. It is important to maximize the copper to remove the heat from the device. Copper exposed to air dissipates heat better than copper used in the inner layers.

If the case temperature can be measured, the junction temperature is calculated by

$$T_J = T_C + (P_D \times \theta_{JC}) \quad (6)$$

where  $T_C$  is the case temperature and  $\theta_{JC}$  is the junction-to-case thermal resistance provided in Table 5.

The reliable operation of the charger can be achieved only if the estimated die junction temperature of the **ADP5063** (Equation 5) is less than 125°C. Reliability and mean time between failures (MTBF) are greatly affected by increasing the junction temperature. Additional information about product reliability can be found in the *ADI Reliability Handbook* located at the following URL: [http://www.analog.com/reliability\\_handbook](http://www.analog.com/reliability_handbook).

## FACTORY-PROGRAMMABLE OPTIONS

### CHARGER OPTIONS

Table 37 to Table 49 list the factory-programmable options of the ADP5063. In each of these tables, the selection column represents the default setting of Model ADP5063ACPZ-1-R7.

**Table 37. Default Termination Voltage**

Option	Selection
000 = 4.20 V	001 = 3.60 V
001 = 3.60 V	
010 = 3.70 V	
011 = 3.80 V	
100 = 3.90 V	
101 = 4.00 V	
110 = 4.10 V	
111 = 4.40 V	

**Table 38. Default Fast Charge Current**

Option	Selection
000 = 500 mA	100 = 750 mA
001 = 300 mA	
010 = 550 mA	
011 = 600 mA	
100 = 750 mA	
101 = 900 mA	
110 = 1300 mA	
111 = 1300 mA	

**Table 39. Default End of Charge Current**

Option	Selection
000 = 52.5 mA	000 = 52.5 mA
001 = 72.5 mA	
010 = 12.5 mA	
011 = 32.5 mA	
100 = 142.5 mA	
101 = 167.5 mA	
110 = 92.5 mA	
111 = 117.5 mA	

**Table 40. Default Trickle to Fast Charge Threshold**

Option	Selection
00 = 2.5 V	01 = 2.0 V
01 = 2.0 V	
10 = 2.9 V	
11 = 2.6 V	

**Table 41. Default System Voltage**

Option	Selection
000 = 4.3 V	000 = 4.3 V
001 = 4.4 V	
010 = 4.5 V	
011 = 4.6 V	
100 = 4.7 V	
101 = 4.8 V	
110 = 4.9 V	
111 = 5.0 V	

**Table 42. Thermistor Resistance**

Option	Selection
0 = 10 k $\Omega$	0 = 10 k $\Omega$
1 = 100 k $\Omega$	

**Table 43. Thermistor Beta Value**

Option	Selection
0100 = 3150	0100 = 3150
0101 = 3350	
0110 = 3500	
0111 = 3650	
1000 = 3850	
1001 = 4000	
1010 = 4200	
1011 = 4400	

**Table 44. DIS\_IC1 Mode Select**

Option	Selection
0 = DIC_IC1 mode select, VINx current = 280 $\mu$ A, ISO_Bx can float, no leak to ISO_Bx	0 = DIC_IC1 mode select, VINx current = 280 $\mu$ A, ISO_Bx can float, no leak to ISO_Bx
1 = DIC_IC1 mode select, VINx current = 110 $\mu$ A, supply switch leaks from VINx to ISO_Bx	

**Table 45. Trickle or Fast Charge Timer Fault Operation**

Option	Selection
0 = after timeout LDO off, charging off	1 = after timeout LDO mode active, charging off
1 = after timeout LDO mode active, charging off	

**I<sup>2</sup>C REGISTER DEFAULTS****Table 46. I<sup>2</sup>C Register Default Settings**

Bit Name	I <sup>2</sup> C Register Address, Bit Location	Option	Selection
CHG_VLIM[1:0]	Address 0x03, Bits[1:0]	0 = limit 3.2 V 1 = limit 3.7 V	0 = limit 3.2 V
DIS_RCH	Address 0x05, Bit 7	0 = recharge enabled 1 = recharge disabled	0 = recharge enabled
EN_WD	Address 0x06, Bit 2	0 = watchdog disabled 1 = watchdog enabled	0 = disabled
DIS_IC1	Address 0x07, Bit 6	0 = not activated 1 = activated	0 = not activated
EN_CHG	Address 0x07, Bit 0	0 = charging disabled 1 = charging enabled	0 = charging disabled
EN_JEITA	Address 0x08, Bit 7	0 = JEITA function disabled 1 = JEITA function enabled	0 = JEITA function disabled
JEITA_SELECT	Address 0x08, Bit 6	0 = JEITA1 charging 1 = JEITA2 charging	0 = JEITA1 charging
EN_CHG_VLIM	Address 0x08, Bit 5	0 = limit disabled 1 = limit enabled	0 = limit disabled
IDEAL_DIODE[1:0]	Address 0x08, Bits[4:3]	00 = ideal diode operates when $V_{ISO\_Sx} < V_{ISO\_Bx}$  01 = ideal diode operates when $V_{ISO\_Sx} < V_{ISO\_Bx}$ and $V_{BAT\_SNS} > V_{WEAK}$ 10 = ideal diode is disabled 11 = ideal diode is disabled	00 = ideal diode operates when $V_{ISO\_Sx} < V_{ISO\_Bx}$

**DIGITAL INPUT AND OUTPUT OPTIONS****Table 47. I<sup>2</sup>C Address 0x11, Bits[1:0], SYS\_EN\_SET Default**

Option	Selection (Default)
00 = SYS_EN is activated when LDO is active and system voltage is available.	00
01 = SYS_EN is activated by ISO_Bx voltage; battery charging mode.	
10 = SYS_EN is activated and the isolation FET is disabled when the battery drops below $V_{WEAK}$ <sup>1</sup> .	
11 = SYS_EN is active in LDO mode when the charger is disabled. SYS_EN is active in charging mode when $V_{ISO\_Bx} \geq V_{WEAK}$ .	

<sup>1</sup> This option is active when  $V_{INx} = 0$  V and the battery monitor is activated from Register 0x07, Bit D5 (EN\_BMON).

**DIG\_IO1, DIG\_IO2, and DIG\_IO3 Options****Table 48. DIG\_IO1 Polarity**

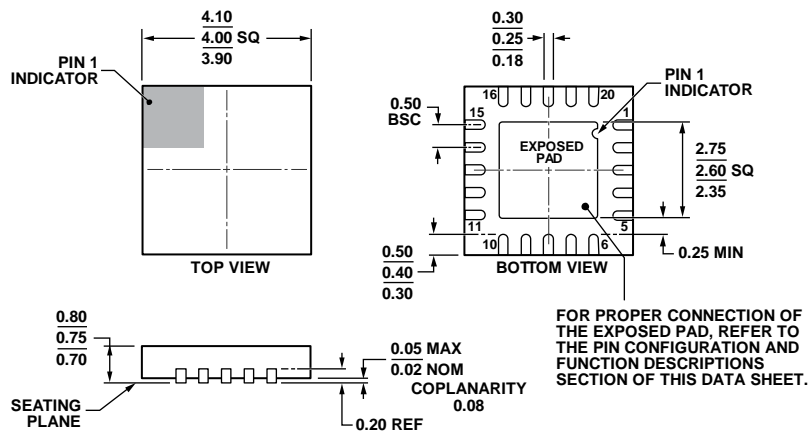
Option	Selection
0 = DIG_IO1 polarity, high active operation 1 = DIG_IO1 polarity, low active operation	0 = high active

**Table 49. DIG\_IOx Options**

Option	DIG_IO1 Function	DIG_IO2 Function	DIG_IO3 Function	Selection
0000	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	Disable IC1 Low = not activated High = activated	Charging disable/enable Low = charging disable High = charging enabled	
0010	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	I <sub>VINx</sub> limit Not applicable High = I <sub>VINx</sub> limit at 1500 mA	Disable IC1 Low = not activated High = activated	
0011	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	I <sub>VINx</sub> limit Not applicable High = I <sub>VINx</sub> limit at 1500 mA	Fast charge current Low = ICHG[4:0] High = ICHG[4:0] ÷ 2	
0100	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	I <sub>VINx</sub> limit Not applicable High = I <sub>VIN</sub> limit at 1500 mA	LDO Low = LDO active High = LDO disabled	
0101	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	I <sub>VINx</sub> limit Not applicable High = I <sub>VINx</sub> limit at 1500 mA	Charging Low = charging disabled High = charging enabled	0101
0110	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	Recharge Not applicable High = disable recharge	Charging Low = charging disabled High = charging enabled	
0111	Charging Low = charging disabled High = charging enabled	Disable IC1 Low = not activated High = activated	Recharge Not applicable High = disable recharge	
1000	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	I <sub>VINx</sub> limit Not applicable High = I <sub>VINx</sub> limit 1500 mA	Interrupt output Not applicable Not applicable	
1001	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	Charging Low = charging disabled High = charging enabled	Interrupt output Not applicable Not applicable	
1010	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	Disable IC1 Low = not activated High = activated	Interrupt output Not applicable Not applicable	
1011	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	Recharge Not applicable High = disable recharge	Interrupt output Not applicable Not applicable	
1100	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	Fast charge current Low = ICHG High = ICHG[4:0] ÷ 2	Interrupt output Not applicable Not applicable	
1101	I <sub>VINx</sub> limit Low = 100 mA High = 500 mA	LDO Low = LDO active High = LDO disabled	Interrupt output Not applicable Not applicable	
1110	I <sub>VINx</sub> limit Not applicable High = I <sub>VINx</sub> limit 1500 mA	Charging Low = charging disabled High = charging enabled	Interrupt output Not applicable Not applicable	
1111	Disable IC1 Low = not activated High = activated	Charging Low = charging disabled High = charging enabled	Interrupt output Not applicable Not applicable	

# PACKAGING AND ORDERING INFORMATION

## OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MO-220-WGGD.

Figure 39. 20-Lead Lead Frame Chip Scale Package [LFCSP\_WQ]  
 4 mm × 4 mm Body, Very Very Thin Quad  
 (CP-20-8)  
 Dimensions shown in millimeters

000509-B

## ORDERING GUIDE

Model <sup>1, 2</sup>	Temperature Range (Junction)	Package Description	Package Option
ADP5063ACPZ-1-R7	-40°C to +125°C	20-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-20-8
ADP5063CP-EVALZ		Evaluation Board	

<sup>1</sup> Z = RoHS Compliant Part.

<sup>2</sup> For additional factory-programmable options, contact an Analog Devices, Inc., local sales or distribution representative.

**NOTES**

I<sup>2</sup>C refers to a communications protocol originally developed by Philips Semiconductors (now NXP Semiconductors).

## Looking for pricing, stock, or lifecycle information?

Click below to explore more details on WIN SOURCE:

- ⊖ [View ADP5063ACPZ-1-R7 on WIN SOURCE](#)
- ⊖ [Analog Devices Inc. Information](#)

## Optimize Your Supply Chain with WIN SOURCE Solutions

- ✓ Global Sourcing Solution
- ✓ Obsolete Management
- ✓ Cost Control Management
- ✓ Shortage Management
- ✓ Alternative Solution
- ✓ Excess Inventory Management