



**THE DATASHEET OF  
OPA4322AIPWR**



## OPAx322x 20-MHz, Low-Noise, 1.8-V, RRI/O, CMOS Operational Amplifier With Shutdown

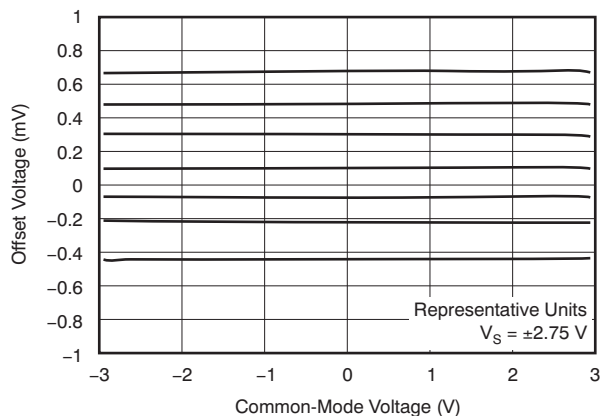
### 1 Features

- Gain Bandwidth: 20 MHz
- Low Noise: 8.5 nV $\sqrt{\text{Hz}}$  at 1 kHz
- Slew Rate: 10 V/ $\mu\text{s}$
- Low THD+N: 0.0005%
- Rail-to-Rail I/O
- Offset Voltage: 2 mV (maximum)
- Supply Voltage: 1.8 V to 5.5 V
- Supply Current: 1.5 mA/ch
  - Shutdown: 0.1  $\mu\text{A}/\text{ch}$
- Unity-Gain Stable
- Small Packages:
  - SOT-23, SON, VSSOP, TSSOP

### 2 Applications

- Sensor Signal Conditioning
- Consumer Audio
- Multi-Pole Active Filters
- Control-Loop Amplifiers
- Communications
- Security
- Scanners

#### Zero-Crossover Rail-to-Rail Input Stage Eliminates Distortion



### 3 Description

The OPAx322x series consists of single, dual, and quad-channel CMOS operational amplifiers featuring low noise and rail-to-rail inputs and outputs optimized for low-power, single-supply applications. Specified over a wide supply range of 1.8 V to 5.5 V, the low quiescent current of only 1.5 mA per channel makes these devices well-suited for power-sensitive applications.

The combination of very low noise (8.5 nV $\sqrt{\text{Hz}}$  at 1 kHz), high-gain bandwidth (20 MHz), and fast slew rate (10 V/ $\mu\text{s}$ ) make the OPAx322x family ideal for a wide range of applications, including signal conditioning and sensor amplification requiring high gains. Featuring low THD+N, the OPAx322x series is also excellent for consumer audio applications, particularly for single-supply systems.

The OPAx322S models include a shutdown mode that allow the amplifiers to be switched from normal operation to a standby current that is typically less than 0.1  $\mu\text{A}$ .

The OPA322 (single version) is available in 5-pin SOT-23 and 6-pin SOT-23, while the OPA2322 (dual version) is offered in 8-pin VSSOP, 10-pin VSSOP, 8-pin SOIC, and 8-pin SON packages. The quad version OPA4322 comes in 14-pin TSSOP and 16-pin TSSOP packages. All versions are specified for operation from  $-40^\circ\text{C}$  to  $+125^\circ\text{C}$ .

#### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
OPA322	SOT-23 (5)	2.90 mm × 1.60 mm
OPA322S	SOT-23 (6)	2.90 mm × 1.60 mm
OPA2322	SOIC (8)	4.90 mm × 3.91 mm
	VSSOP (8)	3.00 mm × 3.00 mm
	SON (8)	3.00 mm × 3.00 mm
OPA2322S	VSSOP (10)	3.00 mm × 3.00 mm
OPA4322	TSSOP (14)	5.00 mm × 4.40 mm
OPA4322S	TSSOP (16)	5.00 mm × 4.40 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.



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## 4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision E (June 2012) to Revision F	Page
• Added <i>ESD Ratings</i> table, <i>Feature Description</i> section, <i>Device Functional Modes</i> , <i>Application and Implementation</i> section, <i>Power Supply Recommendations</i> section, <i>Layout</i> section, <i>Device and Documentation Support</i> section, and <i>Mechanical, Packaging, and Orderable Information</i> section .....	1
• Changed device package descriptions to current standards .....	1
• Deleted <i>Package/Ordering Information</i> table; see <i>Package Option Addendum</i> at the end of this data sheet .....	7

Changes from Revision D (March 2012) to Revision E	Page
• Changed product status from Production Data to Mixed Status .....	1
• Updated D and DGK pinout drawing.....	5
• Added <a href="#">Figure 26</a> to <a href="#">Figure 29</a> .....	15
• Added <i>Shutdown Function</i> section .....	20

Changes from Revision C (November 2011) to Revision D	Page
• Changed product status from Mixed Status to Production Data .....	1
• Added OPA4322, OPA4322S to the Input Bias Current, <i>Input bias current</i> , <i>Over temperature</i> parameter in Electrical Characteristics table .....	9
• Changed Power Supply, OPA4322, OPA4322S <i>Over temperature</i> parameter maximum specification in the Electrical Characteristics table .....	10

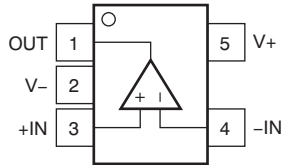
Changes from Revision B (July 2011) to Revision C	Page
• Changed status of OPA2322 SO-8 (D) to production data from product preview .....	1

<b>Changes from Revision A (May 2011) to Revision B</b>	<b>Page</b>
• Added OPA322S thermal information to <i>OPA322 Thermal Information</i> table .....	8
• Added OPA2322S thermal information to <i>OPA2322 Thermal Information</i> table .....	8
• Added OPA4322S thermal information to <i>OPA4322 Thermal Information</i> table .....	8
• Changed Input Bias Current <i>Input bias current, Over temperature</i> parameter in Electrical Characteristics table .....	9
• Changed Open-Loop Gain, <i>Open-loop voltage gain</i> parameter typical specification in the Electrical Characteristics table .	9
• Changed Open-Loop Gain, <i>Phase margin</i> parameter test conditions in the Electrical Characteristics table .....	9
• Changed Power Supply, Quiescent current per amplifier OPA322/S parameter maximum specification in the Electrical Characteristics .....	10
• Changed Power Supply, OPA322 <i>Over temperature</i> parameter maximum specification in the Electrical Characteristics table .....	10
• Changed Power Supply, Quiescent current per amplifier OPA4322/S parameter typical specification in the Electrical Characteristics .....	10
• Added test conditions to <i>Power Supply</i> section in Electrical Characteristics table .....	10
• Changed Shutdown, <i>Quiescent current, per amplifier</i> parameter maximum specification in Electrical Characteristics table .....	10
• Updated <a href="#">Figure 1</a> .....	11
• Added <a href="#">Figure 25</a> .....	15
• Changed <i>Overload Recovery Time</i> section .....	19

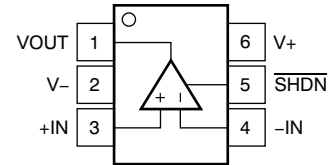
<b>Changes from Original (January 2011) to Revision A</b>	<b>Page</b>
• Changed document status to <i>Production Data</i> .....	1

## 5 Pin Configuration and Functions

OPA322: DBV Package  
 5-Pin SOT-23  
 Top View



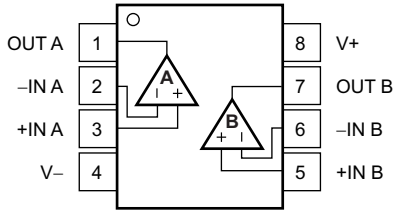
OPA322S: DBV Package  
 6-Pin SOT-23  
 Top View



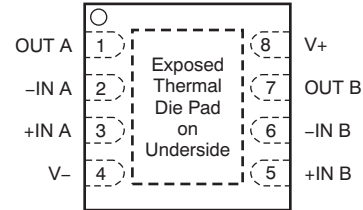
Pin Functions: OPA322, OPA322S

NAME	PIN		I/O	DESCRIPTION
	OPA322 SOT-23	OPA322S SOT-23		
-IN	4	4	I	Inverting input
+IN	3	3	I	Noninverting input
OUT	1	1	O	Output
$\overline{\text{SHDN}}$	—	5	I	Shutdown control (active low)
V-	2	2	—	Negative (lowest) power supply
V+	5	6	—	Positive (highest) power supply

**OPA2322: D and DGK Packages  
8-Pin SOIC and VSSOP  
Top View**

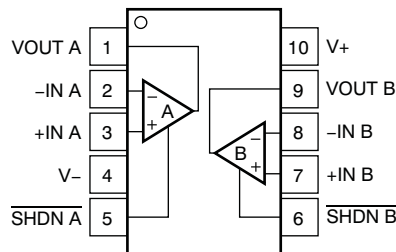


**OPA2322: DRG Package  
8-Pin SON  
Top View**



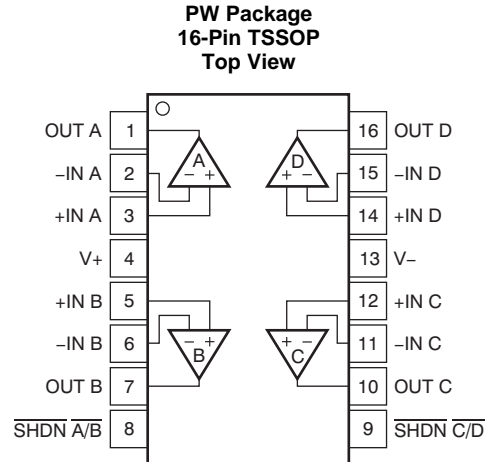
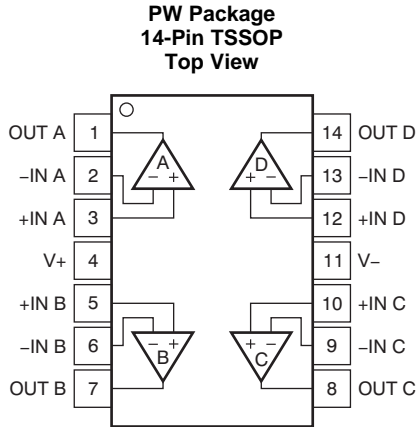
- (1) Connect thermal pad to V-.
- (2) Pad size: 2 mm × 1.2 mm.

**OPA2322S: DGS Package  
10-Pin VSSOP  
Top View**



**Pin Functions: OPA2322, OPA2322S**

NAME	PIN			I/O	DESCRIPTION
	OPA2322		OPA2322S		
	SOIC, VSSOP	SON	VSSOP		
-IN A	2	2	2	I	Inverting input, channel A
+IN A	3	3	3	I	Noninverting input, channel A
-IN B	6	6	8	I	Inverting input, channel B
+IN B	5	5	7	I	Noninverting input, channel B
OUT A	1	1	—	O	Output, channel A
OUT B	7	7	—	O	Output, channel B
$\overline{\text{SHDN A}}$	—	—	5	I	Shutdown control, channel A (active low)
$\overline{\text{SHDN B}}$	—	—	6	I	Shutdown control, channel B (active low)
V-	4	4	4	—	Negative (lowest) power supply
V+	8	8	10	—	Positive (highest) power supply
VOUT A	—	—	1	O	Output, channel A
VOUT B	—	—	9	O	Output, channel B



**Pin Functions: OPA4322, OPA4322S**

NAME	PIN		I/O	DESCRIPTION
	OPA4322 TSSOP	OPA4322S TSSOP		
-IN A	2	2	I	Inverting input, channel A
+IN A	3	3	I	Noninverting input, channel A
-IN B	6	6	I	Inverting input, channel B
+IN B	5	5	I	Noninverting input, channel B
-IN C	9	11	I	Inverting input, channel C
+IN C	10	12	I	Noninverting input, channel C
-IN D	13	15	I	Inverting input, channel D
+IN D	12	14	I	Noninverting input, channel D
OUT A	1	1	O	Output, channel A
OUT B	7	7	O	Output, channel B
OUT C	8	10	O	Output, channel C
OUT D	14	16	O	Output, channel D
$\overline{\text{SHDN A/B}}$	—	8	I	Shutdown control, channels A and B (active low)
$\overline{\text{SHDN C/D}}$		9	I	Shutdown control, channels C and D (active low)
V-	11	13	—	Negative (lowest) power supply
V+	4	4	—	Positive (highest) power supply

## 6 Specifications

### 6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
Voltage	Supply voltage, $V_S = (V+) - (V-)$		6	V
	Signal input pins <sup>(2)</sup>	(V-) – 0.5	(V+) + 0.5	V
Current	Signal input pins <sup>(2)</sup>	–10	10	mA
	Output short-circuit <sup>(3)</sup>	Continuous		
Temperature	Operating, $T_A$	–40	150	°C
	Junction, $T_J$		150	°C
	Storage, $T_{stg}$	–65	150	°C

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5 V beyond the supply rails must be current limited to 10 mA or less.
- (3) Short-circuit to ground, one amplifier per package.

### 6.2 ESD Ratings

		VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±4000
		Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±1000
		Machine model (MM)	±200

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
$V_S$	Specified voltage	1.8	5.5	V
$T_A$	Specified temperature	–40	125	°C

#### 6.4 Thermal Information: OPA322, OPA322S

THERMAL METRIC <sup>(1)</sup>		OPA322	OPA322S	UNITS
		DBV (SOT-23)	DBV (SOT-23)	
		5 PINS	6 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	219.3	177.5	°C/W
$R_{\theta JC(top)}$	Junction-to-case(top) thermal resistance	107.5	108.9	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	57.5	27.4	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	7.4	13.3	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	56.9	26.9	°C/W
$R_{\theta JC(bot)}$	Junction-to-case(bottom) thermal resistance	—	—	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

#### 6.5 Thermal Information: OPA2322, OPA2322S

THERMAL METRIC <sup>(1)</sup>		OPA2322			OPA2322S	UNITS
		D (SOIC)	DRG (SON)	DGK (VSSOP)	DGS (VSSOP)	
		8 PINS	8 PINS	8 PINS	10 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	122.6	50.6	174.8	171.5	°C/W
$R_{\theta JC(top)}$	Junction-to-case(top) thermal resistance	67.1	54.9	43.9	43	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	64	25.2	95	91.4	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	13.2	0.6	2	1.9	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	63.4	25.3	93.5	89.9	°C/W
$R_{\theta JC(bot)}$	Junction-to-case(bottom) thermal resistance	—	5.7	—	—	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

#### 6.6 Thermal Information: OPA4322, OPA4322S

THERMAL METRIC <sup>(1)</sup>		OPA4322	OPA4322S	UNITS
		PW (TSSOP)	PW (TSSOP)	
		14 PINS	16 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	109.8	105.9	°C/W
$R_{\theta JC(top)}$	Junction-to-case(top) thermal resistance	34.9	28.1	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	52.5	51.1	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	2.2	0.8	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	51.8	50.4	°C/W
$R_{\theta JC(bot)}$	Junction-to-case(bottom) thermal resistance	—	—	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 6.7 Electrical Characteristics

At  $V_S = 1.8\text{ V}$  to  $5.5\text{ V}$ , or  $\pm 0.9\text{ V}$  to  $\pm 2.75\text{ V}$ ,  $T_A = 25^\circ\text{C}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S/2$ ,  $V_{CM} = V_S/2$ ,  $V_{OUT} = V_S/2$ , and  $\text{SHDN}_X = V_S+$  (unless otherwise noted).

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
<b>OFFSET VOLTAGE</b>							
$V_{OS}$	Input offset voltage				0.5	2	mV
$dV_{OS}/dT$	vs temperature	$V_S = 5.5\text{ V}$			1.8	6	$\mu\text{V}/^\circ\text{C}$
PSR	vs power supply	$V_S = 1.8\text{ V}$ to $5.5\text{ V}$	$T_A = 25^\circ\text{C}$		10	50	$\mu\text{V}/\text{V}$
			$T_A = -40^\circ\text{C}$ to $125^\circ\text{C}$		20	65	
	Channel separation	At 1 kHz			130		dB
<b>INPUT VOLTAGE</b>							
$V_{CM}$	Common-mode voltage range			$(V-) - 0.1$		$(V+) + 0.1$	V
CMRR	Common-mode rejection ratio	$(V-) - 0.1\text{ V} < V_{CM} < (V+) + 0.1\text{ V}$	$T_A = 25^\circ\text{C}$		90	100	dB
			$T_A = -40^\circ\text{C}$ to $125^\circ\text{C}$		90		
<b>INPUT BIAS CURRENT</b>							
$I_B$	Input bias current	$T_A = 25^\circ\text{C}$			$\pm 0.2$	$\pm 10$	pA
		$T_A = -40^\circ\text{C}$ to $85^\circ\text{C}$				$\pm 50$	
		OPA322 and OPA322S, $T_A = -40^\circ\text{C}$ to $125^\circ\text{C}$				$\pm 800$	
		OPA2322 and OPA2322S, $T_A = -40^\circ\text{C}$ to $125^\circ\text{C}$				$\pm 400$	
		OPA4322 and OPA4322S, $T_A = -40^\circ\text{C}$ to $125^\circ\text{C}$				$\pm 400$	
$I_{OS}$	Input offset current	$T_A = 25^\circ\text{C}$			$\pm 0.2$	$\pm 10$	pA
		$T_A = -40^\circ\text{C}$ to $85^\circ\text{C}$				$\pm 50$	
		$T_A = -40^\circ\text{C}$ to $125^\circ\text{C}$				$\pm 400$	
<b>NOISE</b>							
	Input voltage noise	$f = 0.1\text{ Hz}$ to $10\text{ Hz}$			2.8		$\mu\text{V}_{PP}$
$e_n$	Input voltage noise density	$f = 1\text{ kHz}$			8.5		$\text{nV}/\sqrt{\text{Hz}}$
		$f = 10\text{ kHz}$			7		
$i_n$	Input current noise density	$f = 1\text{ kHz}$			0.6		$\text{fA}/\sqrt{\text{Hz}}$
<b>INPUT CAPACITANCE</b>							
	Differential				5		pF
	Common-mode				4		pF
<b>OPEN-LOOP GAIN</b>							
$A_{OL}$	Open-loop voltage gain	$0.1\text{ V} < V_O < (V+) - 0.1\text{ V}$ , $R_L = 10\text{ k}\Omega$		100	130		dB
		$0.1\text{ V} < V_O < (V+) - 0.1\text{ V}$ , $R_L = 10\text{ k}\Omega$		94			
PM	Phase margin	$V_S = 5\text{ V}$ , $C_L = 50\text{ pF}$			47		$^\circ$
<b>FREQUENCY RESPONSE</b>							
GBP	Gain bandwidth product	$V_S = 5\text{ V}$ , $C_L = 50\text{ pF}$ , unity gain			20		MHz
SR	Slew rate	$V_S = 5\text{ V}$ , $C_L = 50\text{ pF}$ , $G = +1$			10		$\text{V}/\mu\text{s}$
$t_S$	Settling time	$V_S = 5\text{ V}$ , $C_L = 50\text{ pF}$ , to 0.1%, 2-V step, $G = +1$			0.25		$\mu\text{s}$
		$V_S = 5\text{ V}$ , $C_L = 50\text{ pF}$ , to 0.01%, 2-V step, $G = +1$			0.32		
	Overload recovery time	$V_S = 5\text{ V}$ , $C_L = 50\text{ pF}$ , $V_{IN} \times G > V_S$			100		ns
THD+N	Total harmonic distortion + noise <sup>(1)</sup>	$V_S = 5\text{ V}$ , $C_L = 50\text{ pF}$ , $V_O = 4\text{ V}_{PP}$ , $G = +1$ , $f = 10\text{ kHz}$ , $R_L = 10\text{ k}\Omega$			0.0005%		
		$V_S = 5\text{ V}$ , $C_L = 50\text{ pF}$ , $V_O = 2\text{ V}_{PP}$ , $G = +1$ , $f = 10\text{ kHz}$ , $R_L = 600\ \Omega$			0.0011%		

(1) Third-order filter; bandwidth = 80 kHz at -3 dB

## Electrical Characteristics (continued)

At  $V_S = 1.8\text{ V to }5.5\text{ V}$ , or  $\pm 0.9\text{ V to } \pm 2.75\text{ V}$ ,  $T_A = 25^\circ\text{C}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S/2$ ,  $V_{CM} = V_S/2$ ,  $V_{OUT} = V_S/2$ , and  $\overline{\text{SHDN}}_X = V_{S+}$  (unless otherwise noted).

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
<b>OUTPUT</b>							
$V_O$	Voltage output (swing from both rails)	$R_L = 10\text{ k}\Omega$	$T_A = 25^\circ\text{C}$		10	20	mV
			$T_A = -40^\circ\text{C to }125^\circ\text{C}$			30	
$I_{SC}$	Short-circuit current	$V_S = 5.5\text{ V}$			$\pm 65$		mA
$C_L$	Capacitive load drive			See <a href="#">Typical Characteristics</a>			
$R_O$	Open-loop output resistance	$I_O = 0\text{ mA}$ , $f = 1\text{ MHz}$			90		$\Omega$
<b>POWER SUPPLY</b>							
$V_S$	Specified voltage range			1.8		5.5	V
$I_Q$	Quiescent current per amplifier	OPA322 and OPA322S, $I_O = 0\text{ mA}$ , $V_S = 5.5\text{ V}$	$T_A = 25^\circ\text{C}$		1.6	1.9	mA
			$T_A = -40^\circ\text{C to }125^\circ\text{C}$			2	
		OPA2322 and OPA2322S, $I_O = 0\text{ mA}$ , $V_S = 5.5\text{ V}$	$T_A = 25^\circ\text{C}$		1.5	1.75	
			$T_A = -40^\circ\text{C to }125^\circ\text{C}$			1.85	
OPA4322 and OPA4322S, $I_O = 0\text{ mA}$ , $V_S = 5.5\text{ V}$	$T_A = 25^\circ\text{C}$		1.4	1.65			
	$T_A = -40^\circ\text{C to }125^\circ\text{C}$			1.75			
	Power-on time	$V_{S+} = 0\text{ V to }5\text{ V}$ , to 90% $I_Q$ level			28		$\mu\text{s}$
<b>SHUTDOWN<sup>(2)</sup></b>							
$I_{QSD}$	Quiescent current (per amplifier)	$V_S = 1.8\text{ V to }5.5\text{ V}$ , all amplifiers disabled, $\overline{\text{SHDN}} = V_{S-}$			0.1	0.5	$\mu\text{A}$
$V_{IH}$	High voltage (enabled)	$V_S = 1.8\text{ V to }5.5\text{ V}$ , amplifier enabled		$(V+) - 0.1$			V
$V_{IL}$	Low voltage (disabled)	$V_S = 1.8\text{ V to }5.5\text{ V}$ , amplifier disabled				$(V-) + 0.1$	V
$t_{ON}$	Amplifier enable time (full shutdown) <sup>(3)</sup>	$V_S = 1.8\text{ V to }5.5\text{ V}$ , full shutdown; $G = 1$ , $V_{OUT} = 0.9 \times V_S/2$ <sup>(4)</sup>			10		$\mu\text{s}$
	Amplifier enable time (partial shutdown) <sup>(3)</sup>	$V_S = 1.8\text{ V to }5.5\text{ V}$ , partial shutdown; $G = 1$ , $V_{OUT} = 0.9 \times V_S/2$ <sup>(4)</sup>			6		$\mu\text{s}$
$t_{OFF}$	Amplifier disable time <sup>(3)</sup>	$V_S = 1.8\text{ V to }5.5\text{ V}$ , $G = 1$ , $V_{OUT} = 0.1 \times V_S/2$			3		$\mu\text{s}$
	$\overline{\text{SHDN}}$ pin input bias current (per pin)	$V_S = 1.8\text{ V to }5.5\text{ V}$ , $V_{IH} = 5\text{ V}$			0.13		$\mu\text{A}$
		$V_S = 1.8\text{ V to }5.5\text{ V}$ , $V_{IL} = 0\text{ V}$			0.04		

(2) Ensured by design and characterization; not production tested.

(3) Disable time ( $t_{OFF}$ ) and enable time ( $t_{ON}$ ) are defined as the time interval between the 50% point of the signal applied to the  $\overline{\text{SHDN}}$  pin and the point at which the output voltage reaches the 10% (disable) or 90% (enable) level.

(4) Full shutdown refers to the dual OPA2322S having both channels A and B disabled ( $\overline{\text{SHDN}}_A = \overline{\text{SHDN}}_B = V_{S-}$ ) and the quad OPA4322S having all channels A to D disabled ( $\overline{\text{SHDN}}_A/B = \overline{\text{SHDN}}_C/D = V_{S-}$ ). For partial shutdown, only one  $\overline{\text{SHDN}}$  pin is exercised; in this mode, the internal biasing and oscillator remain operational and the enable time is shorter.

## 6.8 Typical Characteristics

At  $T_A = 25^\circ\text{C}$ ,  $V_{CM} = V_{OUT} = \text{mid-supply}$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted).

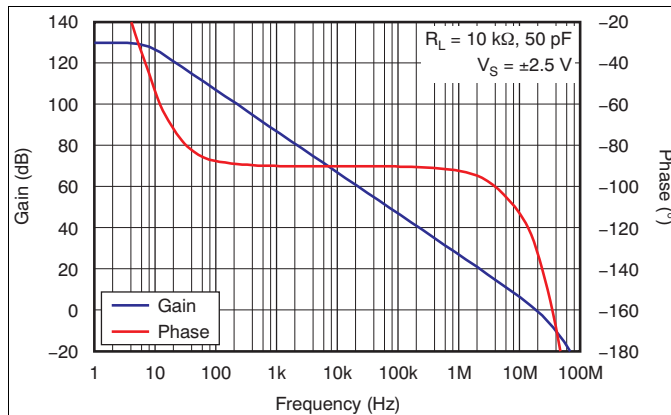


Figure 1. Open-Loop Gain and Phase vs Frequency

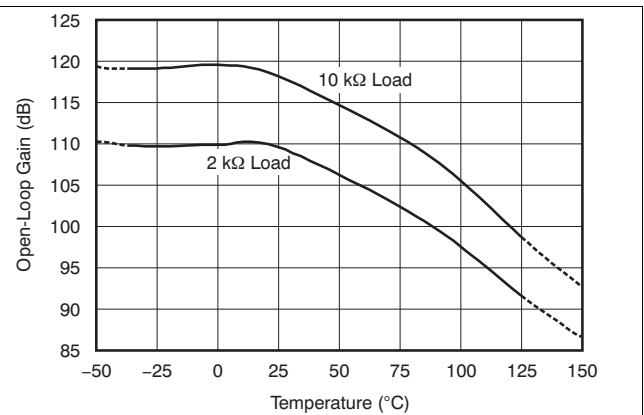


Figure 2. Open-Loop Gain vs Temperature

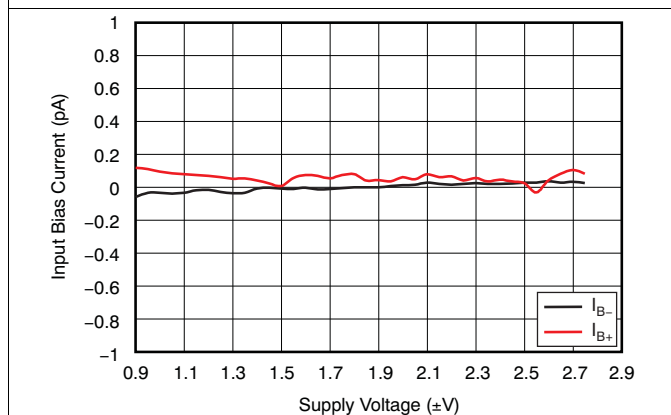


Figure 3. Input Bias Current vs Supply Voltage

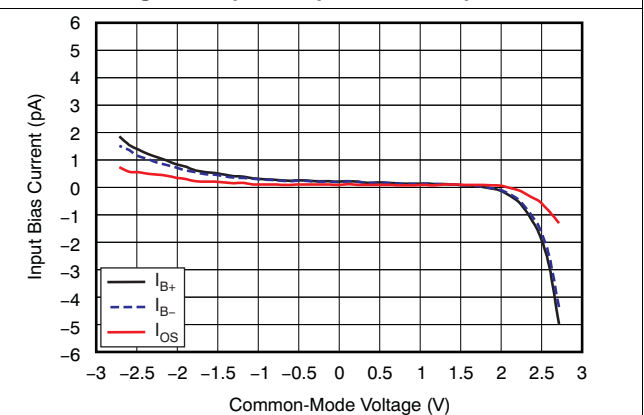


Figure 4. Input Bias Current vs Common-Mode Voltage

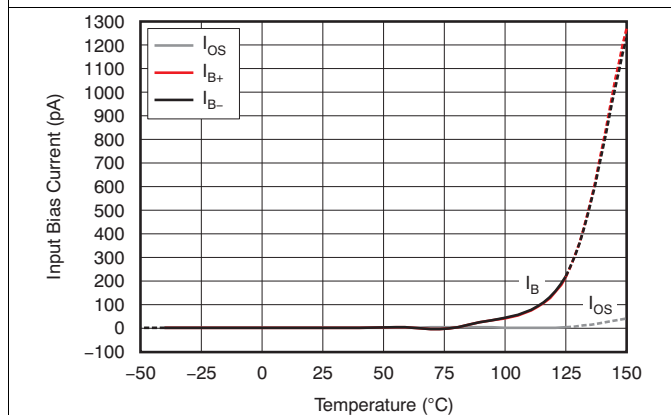


Figure 5. Input Bias Current vs Temperature

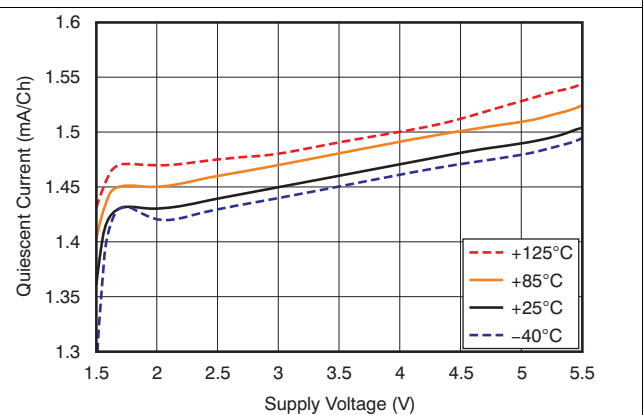


Figure 6. Quiescent Current vs Supply Voltage

## Typical Characteristics (continued)

At  $T_A = 25^\circ\text{C}$ ,  $V_{CM} = V_{OUT} = \text{mid-supply}$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted).

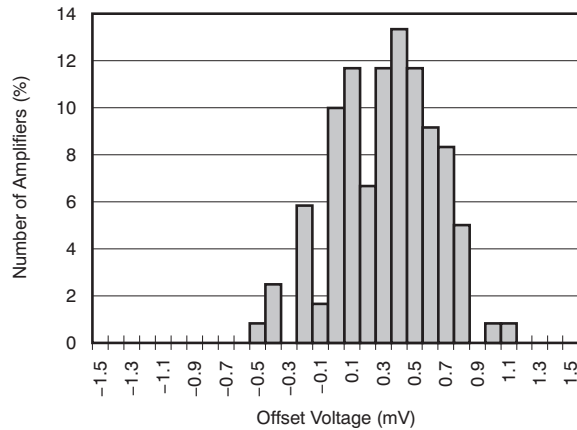


Figure 7. Offset Voltage Production Histogram

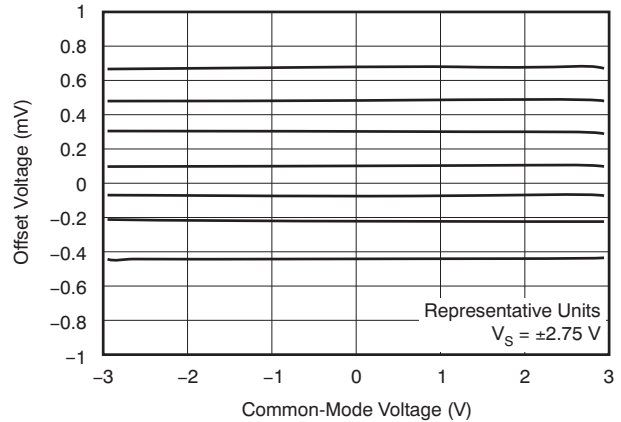


Figure 8. Offset Voltage vs Common-Mode Voltage

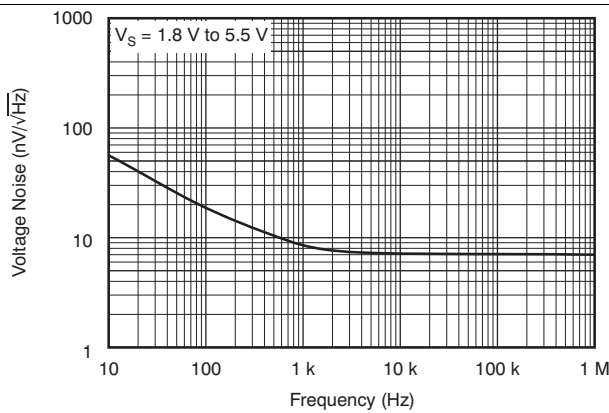


Figure 9. Input Voltage Noise Spectral Density vs Frequency

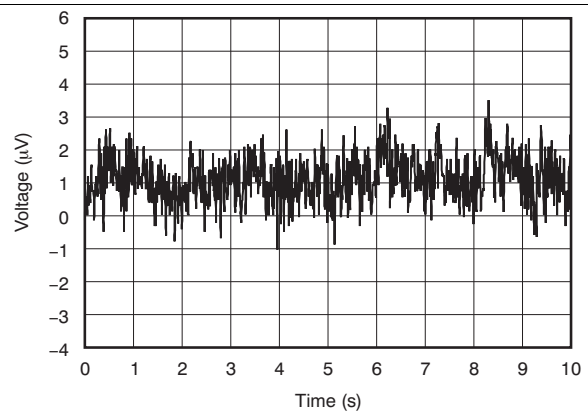


Figure 10. 0.1-Hz to 10-Hz Input Voltage Noise

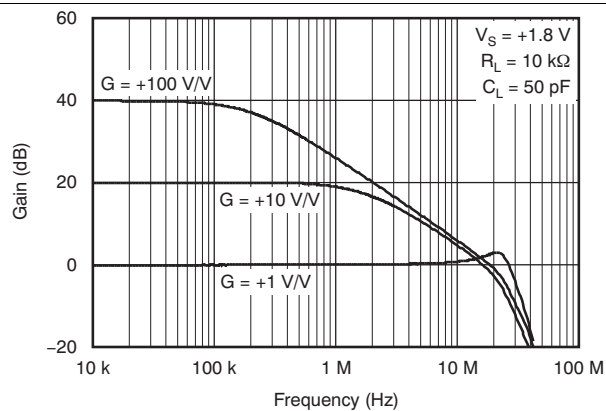


Figure 11. Closed-Loop Gain vs Frequency

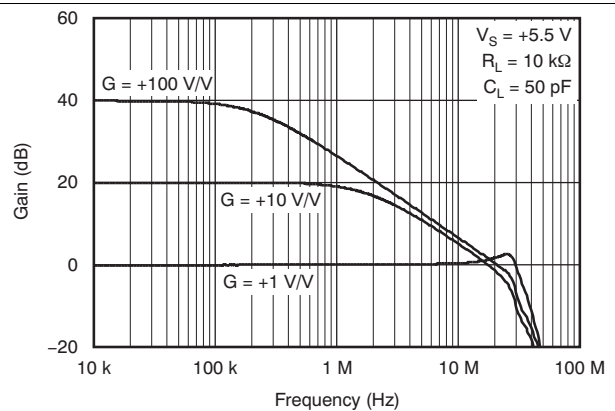


Figure 12. Closed-Loop Gain vs Frequency

## Typical Characteristics (continued)

At  $T_A = 25^\circ\text{C}$ ,  $V_{CM} = V_{OUT} = \text{mid-supply}$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted).

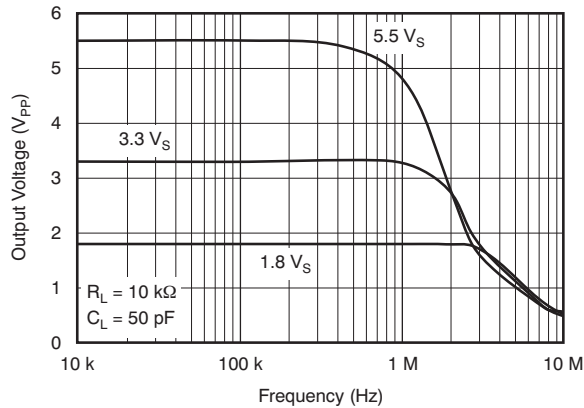


Figure 13. Maximum Output Voltage vs Frequency

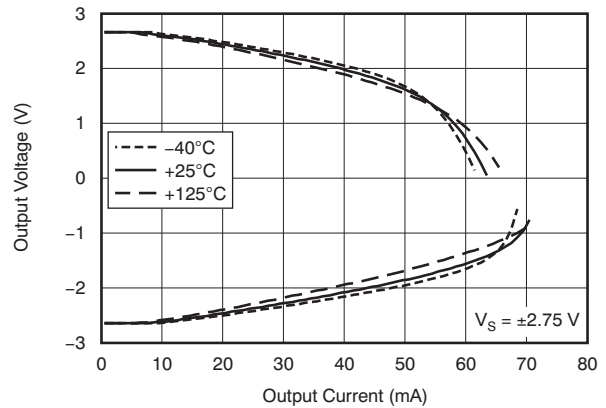


Figure 14. Output Voltage Swing vs Output Current

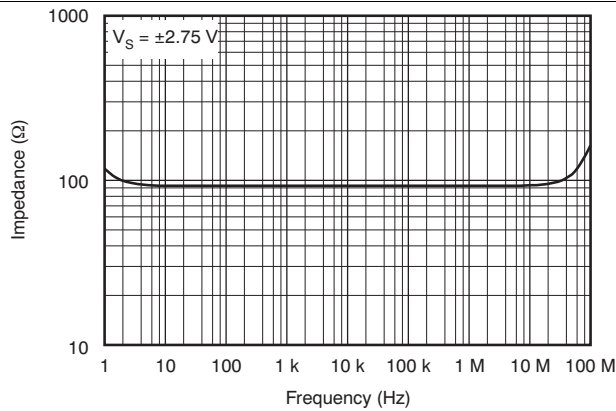


Figure 15. Open-Loop Output Impedance vs Frequency

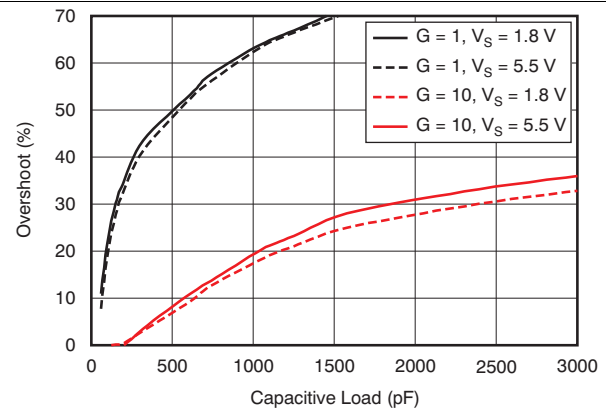


Figure 16. Small-Signal Overshoot vs Load Capacitance

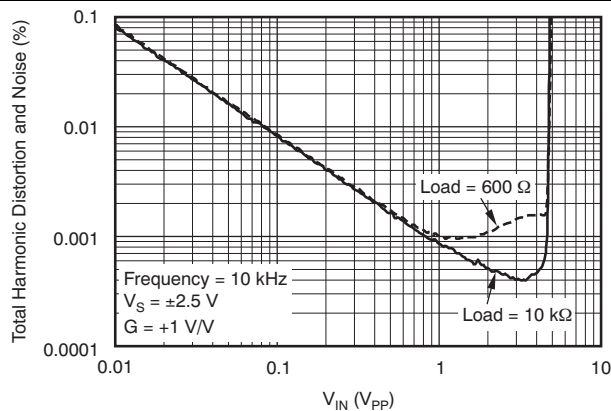


Figure 17. THD+N vs Amplitude

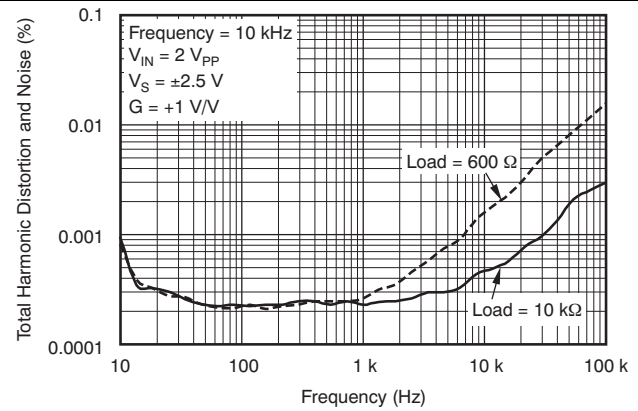
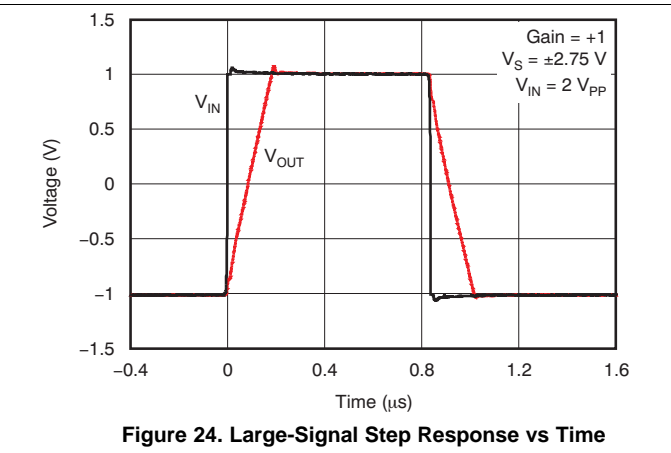
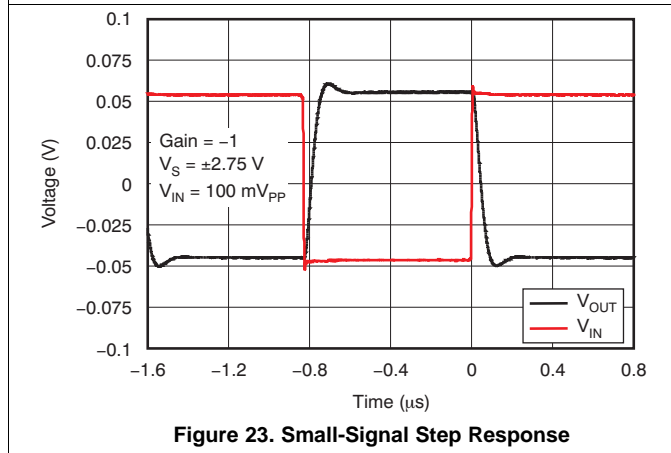
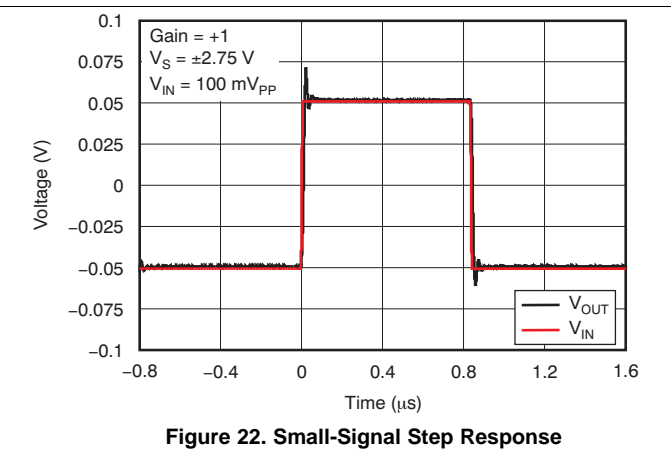
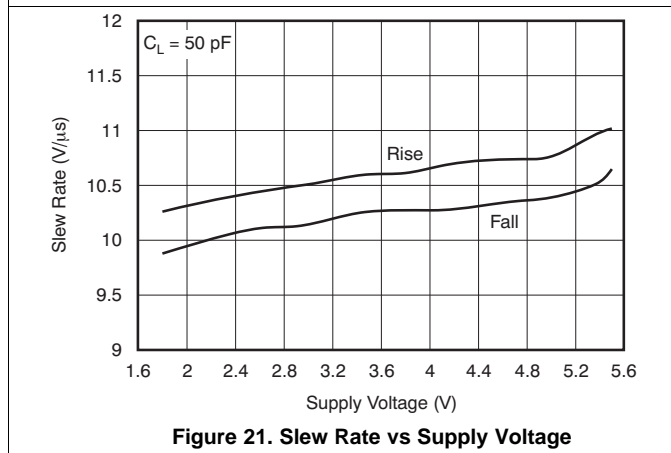
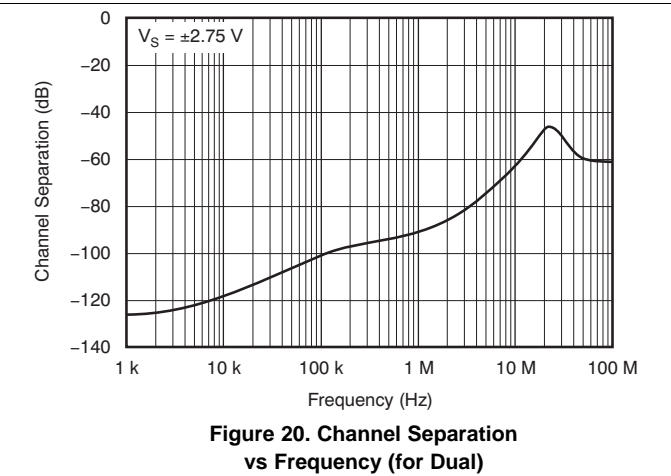
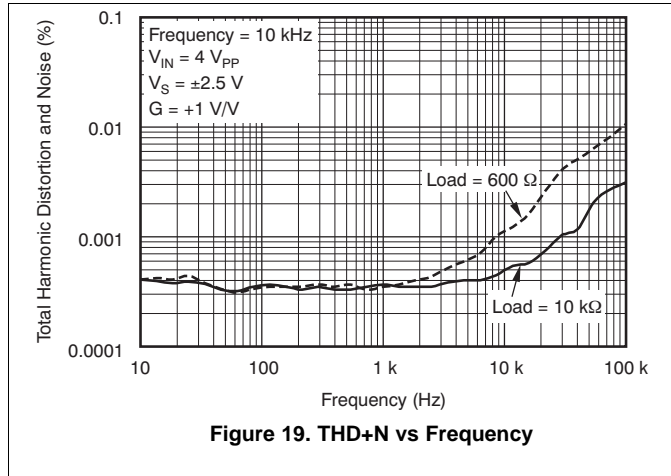


Figure 18. THD+N vs Frequency

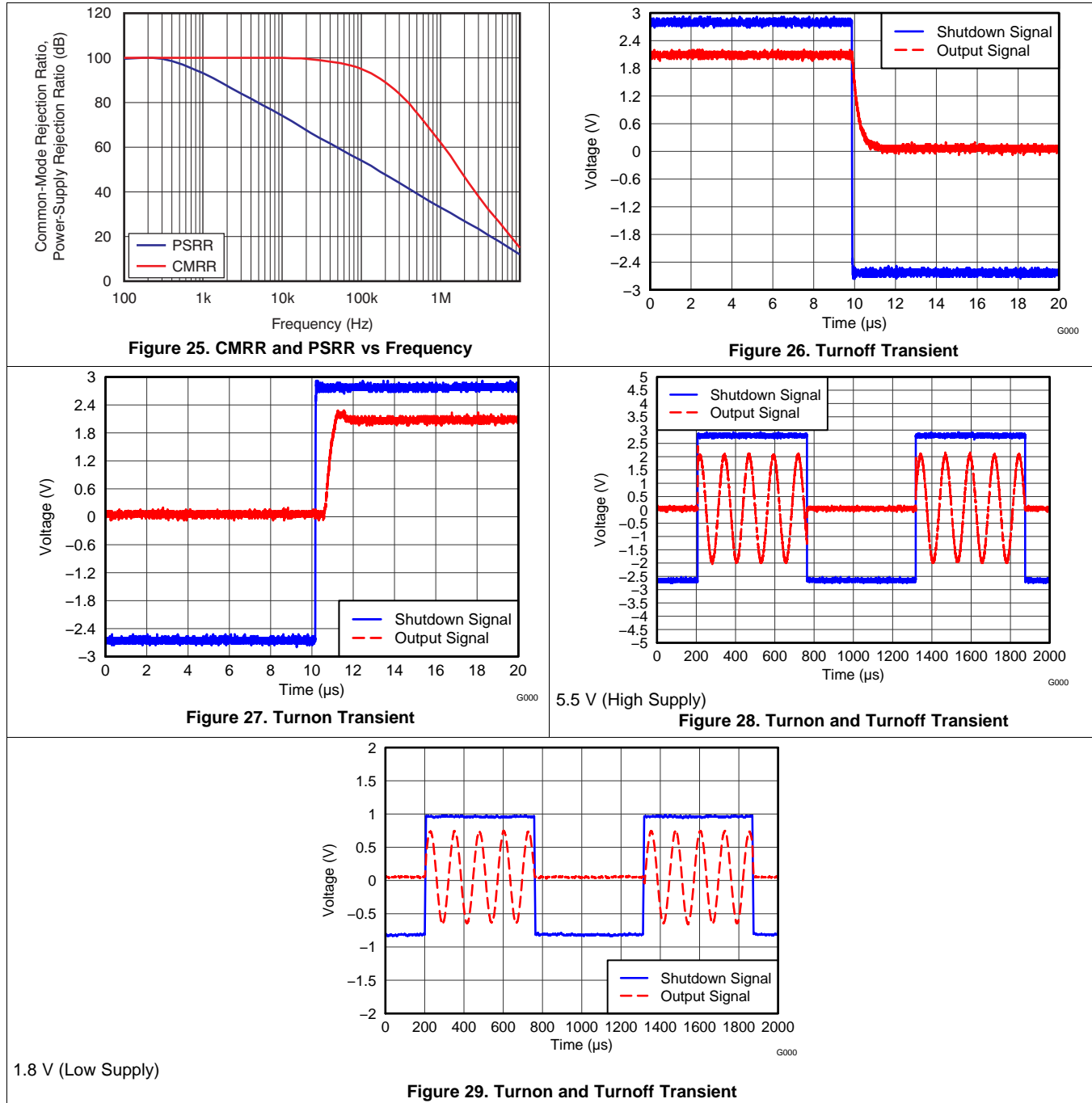
## Typical Characteristics (continued)

At  $T_A = 25^\circ\text{C}$ ,  $V_{CM} = V_{OUT} = \text{mid-supply}$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted).



### Typical Characteristics (continued)

At  $T_A = 25^\circ\text{C}$ ,  $V_{CM} = V_{OUT} = \text{mid-supply}$ , and  $R_L = 10\text{ k}\Omega$  (unless otherwise noted).

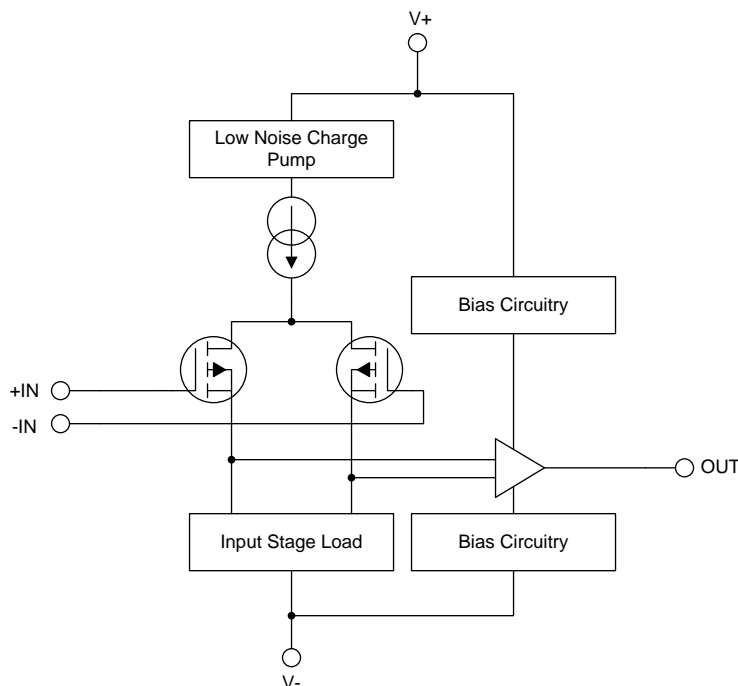


## 7 Detailed Description

### 7.1 Overview

The OPA322 family of operational amplifiers (op amps) are high-speed, precision amplifiers perfectly suited to drive 12-, 14-, and 16-bit analog-to-digital converters. Low-output impedance with flat frequency characteristics and zero-crossover distortion circuitry enable high linearity over the full input common-mode range, achieving true rail-to-rail input from a 1.8-V to 5.5-V single supply.

### 7.2 Functional Block Diagram



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### 7.3 Feature Description

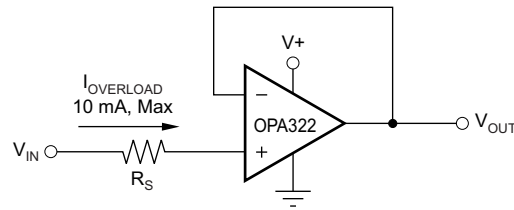
#### 7.3.1 Operating Voltage

The OPA322 series op amps are unity-gain stable and can operate on a single-supply voltage (1.8 V to 5.5 V), or a split-supply voltage ( $\pm 0.9$  V to  $\pm 2.75$  V), making them highly versatile and easy to use. The power-supply pins must have local bypass ceramic capacitors (typically 0.001  $\mu$ F to 0.1  $\mu$ F). These amplifiers are fully specified from 1.8 V to 5.5 V and over the extended temperature range of  $-40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . Parameters that can exhibit variance with regard to operating voltage or temperature are presented in the [Typical Characteristics](#).

#### 7.3.2 Input and ESD Protection

The OPA322 incorporates internal electrostatic discharge (ESD) protection circuits on all pins. In the case of input and output pins, this protection primarily consists of current-steering diodes connected between the input and power-supply pins. These ESD protection diodes also provide in-circuit input overdrive protection, as long as the current is limited to 10 mA as stated in the [Absolute Maximum Ratings](#). Many input signals are inherently current-limited to less than 10 mA; therefore, a limiting resistor is not required. [Figure 30](#) shows how a series input resistor ( $R_S$ ) may be added to the driven input to limit the input current. The added resistor contributes thermal noise at the amplifier input and the value must be kept to the minimum in noise-sensitive applications.

## Feature Description (continued)

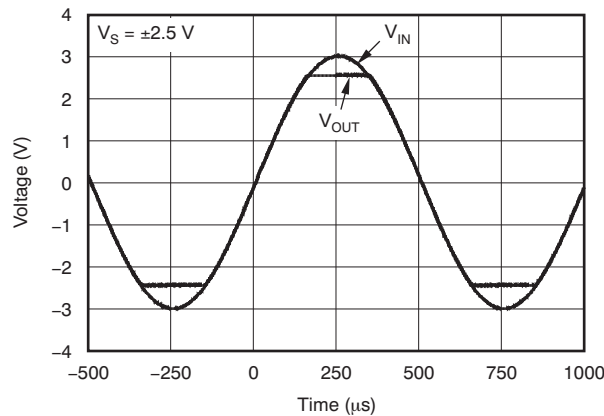


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**Figure 30. Input Current Protection**

### 7.3.3 Phase Reversal

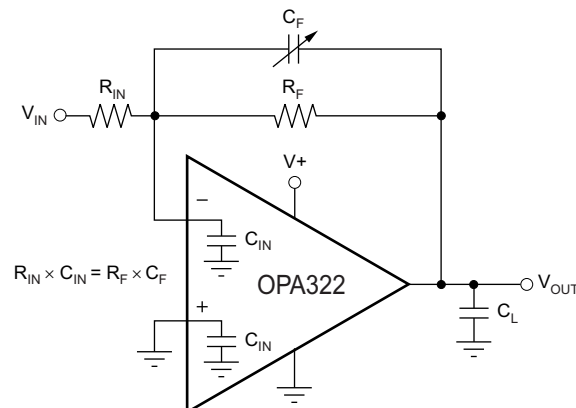
The OPA322 op amps are designed to be immune to phase reversal when the input pins exceed the supply voltages, therefore providing further in-system stability and predictability. [Figure 31](#) shows the input voltage exceeding the supply voltage without any phase reversal.



**Figure 31. No Phase Reversal**

### 7.3.4 Feedback Capacitor Improves Response

For optimum settling time and stability with high-impedance feedback networks, it may be necessary to add a feedback capacitor across the feedback resistor,  $R_F$ , as shown in [Figure 32](#). This capacitor compensates for the zero created by the feedback network impedance and the OPA322 input capacitance (and any parasitic layout capacitance). The effect becomes more significant with higher impedance networks.



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NOTE: Where  $C_{IN}$  is equal to the OPA322 input capacitance (approximately 9 pF) plus any parasitic layout capacitance.

**Figure 32. Feedback Capacitor Improves Dynamic Performance**

## Feature Description (continued)

For the circuit shown in [Figure 32](#), the value of the variable feedback capacitor must be chosen so that the input resistance times the input capacitance of the OPA322 (typically 9 pF) plus the estimated parasitic layout capacitance equals the feedback capacitor times the feedback resistor with [Equation 1](#).

$$R_{IN} \times C_{IN} = R_F \times C_F$$

where

- $C_{IN}$  is equal to the OPA322 input capacitance (sum of differential and common-mode) plus the layout capacitance (1)

The capacitor value can be adjusted until optimum performance is obtained.

### 7.3.5 EMI Susceptibility and Input Filtering

Operational amplifiers vary in susceptibility to electromagnetic interference (EMI). If conducted EMI enters the device, the DC offset observed at the amplifier output may shift from the nominal value while EMI is present. This shift is a result of signal rectification associated with the internal semiconductor junctions. While all operational amplifier pin functions can be affected by EMI, the input pins are likely to be the most susceptible. The OPA322 operational amplifier family incorporates an internal input low-pass filter that reduces the amplifier response to EMI. Both common-mode and differential mode filtering are provided by the input filter. The filter is designed for a cutoff frequency of approximately 580 MHz (–3 dB), with a roll-off of 20 dB per decade.

### 7.3.6 Output Impedance

The open-loop output impedance of the OPA322 common-source output stage is approximately 90 Ω. When the op amp is connected with feedback, this value is reduced significantly by the loop gain. For each decade rise in the closed-loop gain, the loop gain is reduced by the same amount, which results in a tenfold increase in effective output impedance. While the OPA322 output impedance remains very flat over a wide frequency range, at higher frequencies the output impedance rises as the open-loop gain of the op amp drops. However, at these frequencies the output also becomes capacitive as a result of parasitic capacitance. This characteristic, in turn, prevents the output impedance from becoming too high, which can cause stability problems when driving large capacitive loads. As mentioned previously, the OPA322 has excellent capacitive load drive capability for an op amp with its bandwidth.

### 7.3.7 Capacitive Load and Stability

The OPA322 is designed to be used in applications where driving a capacitive load is required. As with all op amps, there may be specific instances where the OPA322 can become unstable. The particular op amp circuit configuration, layout, gain, and output loading are some of the factors to consider when establishing whether an amplifier is stable in operation. An op amp in the unity-gain (+1-V/V) buffer configuration and driving a capacitive load exhibits a greater tendency to become unstable than an amplifier operated at a higher noise gain. The capacitive load, in conjunction with the op amp output resistance, creates a pole within the feedback loop that degrades the phase margin. The degradation of the phase margin increases as the capacitive loading increases. When operating in the unity-gain configuration, the OPA322 remains stable with a pure capacitive load up to approximately 1 nF.

The equivalent series resistance (ESR) of some very large capacitors ( $C_L > 1 \mu\text{F}$ ) is sufficient to alter the phase characteristics in the feedback loop such that the amplifier remains stable. Increasing the amplifier closed-loop gain allows the amplifier to drive increasingly larger capacitance. This increased capability is evident when observing the overshoot response of the amplifier at higher voltage gains, as shown in [Figure 33](#). One technique for increasing the capacitive load drive capability of the amplifier operating in unity gain is to insert a small resistor ( $R_S$ ), typically 10 Ω to 20 Ω, in series with the output, as shown in [Figure 34](#).

This resistor significantly reduces the overshoot and ringing associated with large capacitive loads. A possible problem with this technique is that a voltage divider is created with the added series resistor and any resistor connected in parallel with the capacitive load. The voltage divider introduces a gain error at the output that reduces the output swing. The error contributed by the voltage divider, however, may be insignificant. For instance, with a load resistance,  $R_L = 10 \text{ k}\Omega$  and  $R_S = 20 \Omega$ , the gain error is only about 0.2%. However, when  $R_L$  is decreased to 600 Ω, which the OPA322 is able to drive, the error increases to 7.5%.

Feature Description (continued)

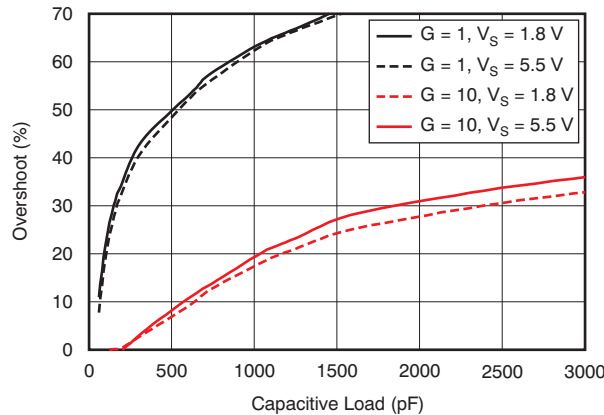
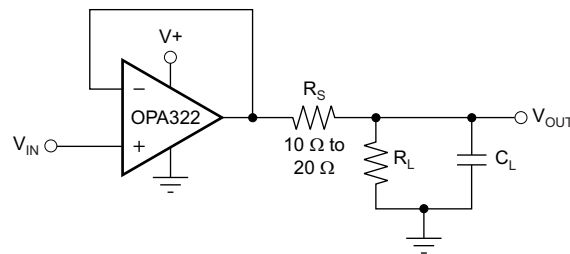


Figure 33. Small-Signal Overshoot vs Capacitive Load (100-mV<sub>PP</sub> Output Step)



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Figure 34. Improving Capacitive Load Drive

7.3.8 Overload Recovery Time

Overload recovery time is the time required for the output of the amplifier to come out of saturation and recover to the linear region. Overload recovery is particularly important in applications where small signals must be amplified in the presence of large transients. Figure 35 and Figure 36 show the positive and negative overload recovery times of the OPA322, respectively. In both cases, the time elapsed before the OPA322 comes out of saturation is less than 100 ns. In addition, the symmetry between the positive and negative recovery times allows excellent signal rectification without distortion of the output signal.

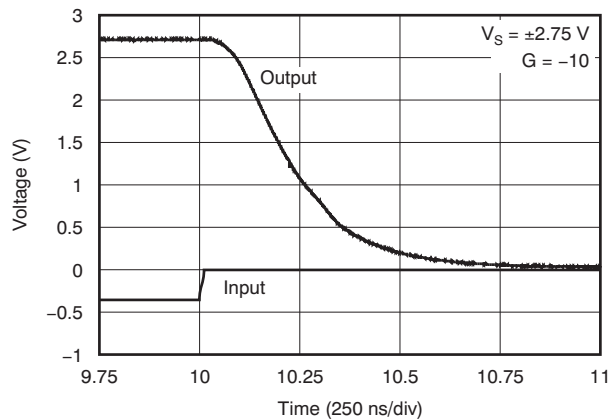


Figure 35. Positive Recovery Time

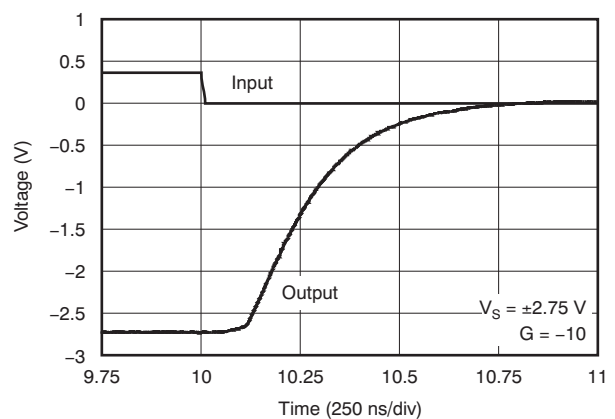


Figure 36. Negative Recovery Time

## Feature Description (continued)

### 7.3.9 Shutdown Function

The SHDN (enable) pin function of the OPAx322S is referenced to the negative supply voltage of the operational amplifier. A logic level high enables the op amp. A valid logic high is defined as voltage  $[(V+) - 0.1 \text{ V}]$ , up to  $(V+)$ , applied to the SHDN pin. A valid logic low is defined as  $[(V-) + 0.1 \text{ V}]$ , down to  $(V-)$ , applied to the enable pin. The maximum allowed voltage applied to SHDN is 5.5 V with respect to the negative supply, independent of the positive supply voltage. This pin must either be connected to a valid high or a low voltage or driven, and not left as an open circuit.

The logic input is a high-impedance CMOS input. Dual op amp versions are independently controlled and quad op amp versions are controlled in pairs with logic inputs. For battery-operated applications, this feature may be used to greatly reduce the average current and extend battery life. The enable time is 10  $\mu\text{s}$  for full shutdown of all channels; disable time is 3  $\mu\text{s}$ . When disabled, the output assumes a high-impedance state. This architecture allows the OPAx322S to be operated as a *gated* amplifier (or to have the device output multiplexed onto a common analog output bus). Shutdown time ( $t_{\text{OFF}}$ ) depends on loading conditions and increases with increased load resistance. To ensure shutdown (disable) within a specific shutdown time, the specified 10-k $\Omega$  load to mid-supply ( $V_S / 2$ ) is required. If using the OPAx322S without a load, the resulting turnoff time is significantly increased.

## 7.4 Device Functional Modes

The OPA322 family of operational amplifiers are operational when power-supply voltages between 1.8 V to 5.5 V are applied. Devices with an S suffix have a shutdown capability. For a detailed description of the shutdown function, refer to [Shutdown Function](#).

## 8 Application and Implementation

### NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 8.1 Application Information

The OPA322 family offers outstanding DC and AC performance. These devices operate up to a 5.5-V power supply and offer ultra-low input bias current and 20-MHz bandwidth. These features make the OPA322 family a robust operational amplifier for both battery-powered and industrial applications.

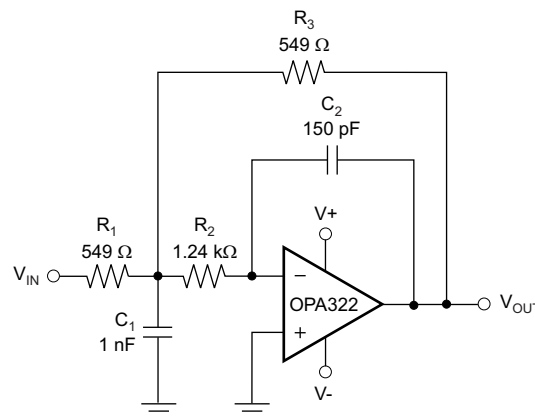
#### 8.1.1 Active Filter

The OPA322 is well-suited for active filter applications that require a wide bandwidth, fast slew rate, low-noise, single-supply operational amplifier. Figure 37 shows a 500-kHz, second-order, low-pass filter using the multiple-feedback (MFB) topology. The components have been selected to provide a maximally-flat Butterworth response. Beyond the cutoff frequency, roll-off is  $-40$  dB/dec. The Butterworth response is ideal for applications that require predictable gain characteristics, such as the anti-aliasing filter used in front of an ADC.

One point to observe when considering the MFB filter is that the output is inverted, relative to the input. If this inversion is not required, or not desired, a noninverting output can be achieved through one of these options:

1. adding an inverting amplifier;
2. adding an additional second-order MFB stage; or
3. using a noninverting filter topology, such as the Sallen-Key (shown in Figure 38).

MFB, Sallen-Key, low-pass, and high-pass filter synthesis is quickly accomplished using TI's FilterPro™ program. This software is available as a free download at [www.ti.com](http://www.ti.com).



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Figure 37. Second-Order, Butterworth, 500-kHz Low-Pass Filter

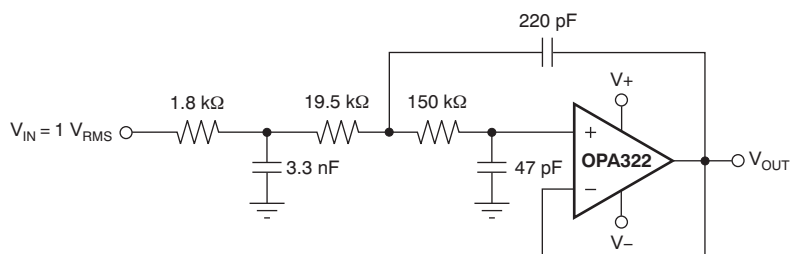
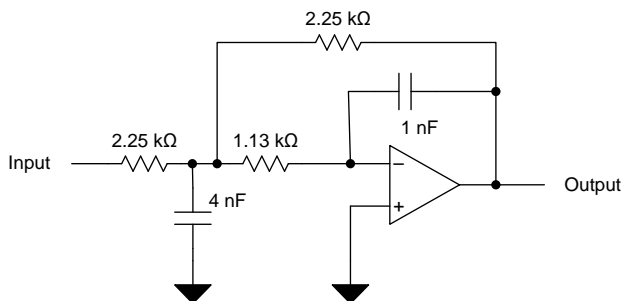


Figure 38. OPA322 Configured as a Three-Pole, 20-kHz, Sallen-Key Filter

## 8.2 Typical Application



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**Figure 39. Second-Order, Low-Pass Filter Schematic**

### 8.2.1 Design Requirements

- Gain = 1 V/V
- Low-pass cutoff frequency = 50 kHz
- –40-dB/dec filter response
- Maintain less than 3-dB gain peaking in the gain versus frequency response

### 8.2.2 Detailed Design Procedure

The infinite-gain multiple-feedback circuit for a low-pass network function is shown in. Use [Equation 2](#) to calculate the voltage transfer function.

$$\frac{\text{Output}}{\text{Input}}(s) = \frac{-1/R_1 R_3 C_2 C_5}{s^2 + (s/C_2)(1/R_1 + 1/R_3 + 1/R_4) + 1/R_3 R_4 C_2 C_5} \quad (2)$$

This circuit produces a signal inversion. For this circuit, the gain at DC and the lowpass cutoff frequency are calculated by [Equation 3](#).

$$\text{Gain} = \frac{R_4}{R_1}$$

$$f_c = \frac{1}{2\pi} \sqrt{(1/R_3 R_4 C_2 C_5)} \quad (3)$$

Software tools are readily available to simplify filter design. WEBENCH® Filter Designer is a simple, powerful, and easy-to-use active filter design program. The [WEBENCH® Filter Designer](#) lets you create optimized filter designs using a selection of TI operational amplifiers and passive components from TI's vendor partners.

Available as a web-based tool from the WEBENCH Design Center, WEBENCH Filter Designer allows you to design, optimize, and simulate complete multistage active filter solutions within minutes.

## Typical Application (continued)

### 8.2.3 Application Curve

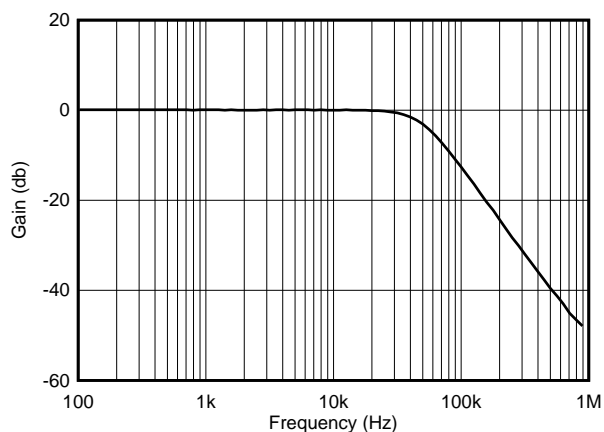


Figure 40. OPA322 Second-Order, 50-kHz, Low-Pass Filter

## 9 Power Supply Recommendations

The OPA322 family is specified for operation from 1.8 V to 5.5 V ( $\pm 0.9$  V to  $\pm 2.75$  V); many specifications apply from  $-40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ . Parameters that can exhibit significant variance with regard to operating voltage or temperature are presented in the [Typical Characteristics](#).

### CAUTION

Supply voltages larger than 6 V can permanently damage the device; see the [Absolute Maximum Ratings](#).

Place 0.1- $\mu\text{F}$  bypass capacitors close to the power-supply pins to reduce errors coupling in from noisy or high-impedance power supplies. For more detailed information on bypass capacitor placement, refer to the [Layout](#).

## 10 Layout

### 10.1 Layout Guidelines

The OPA322 is a wideband amplifier. To realize the full operational performance of the device, follow good high-frequency printed-circuit board (PCB) layout practices. The bypass capacitors must be connected between each supply pin and ground as close to the device as possible. The bypass capacitor traces must be designed for minimum inductance.

#### 10.1.1 Leadless DFN Package

The OPA2322 uses the DFN style package (also known as SON), which is a QFN with contacts on only two sides of the package bottom. This leadless package maximizes PCB space and offers enhanced thermal and electrical characteristics through an exposed pad. One of the primary advantages of the DFN package is its low height (0.8 mm).

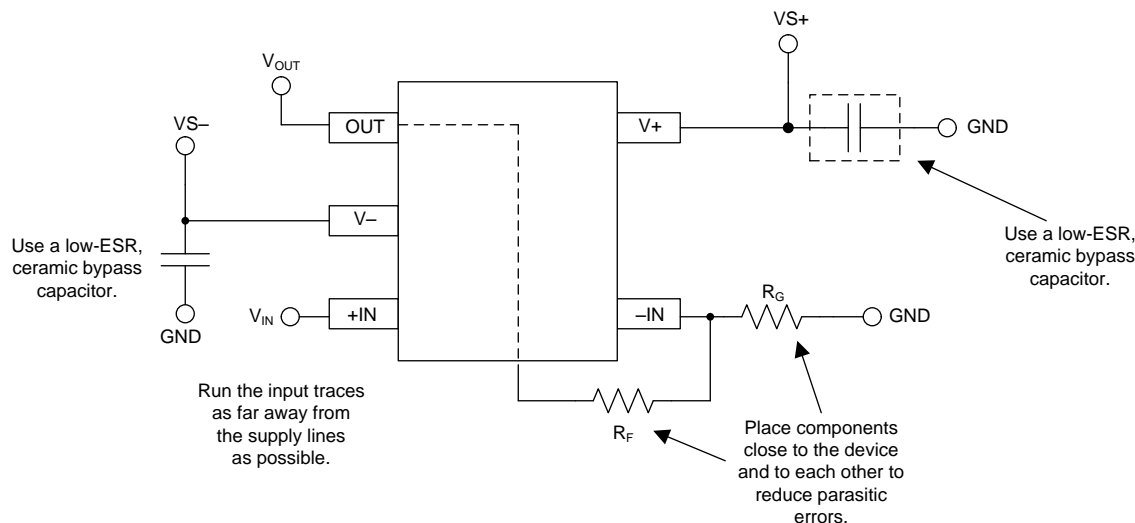
DFN packages are physically small, and have a smaller routing area. Additionally, they offer improved thermal performance, reduced electrical parasitics, and a pinout scheme that is consistent with other commonly-used packages (such as SOIC and VSSOP). The absence of external leads also eliminates bent-lead issues.

The DFN package can easily be mounted using standard PCB assembly techniques. See the application reports, [QFN/SON PCB Attachment](#) and [Quad Flatpack No-Lead Logic Packages](#). The dimension of the exposed thermal die pad is 2 mm × 1.2 mm and is centered.

#### NOTE

The exposed leadframe die pad on the bottom of the DFN package must be connected to the most negative potential ( $V^-$ ).

### 10.2 Layout Example



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**Figure 41. Layout Example**

## 11 Device and Documentation Support

### 11.1 Device Support

#### 11.1.1 Third-Party Products Disclaimer

TI'S PUBLICATION OF INFORMATION REGARDING THIRD-PARTY PRODUCTS OR SERVICES DOES NOT CONSTITUTE AN ENDORSEMENT REGARDING THE SUITABILITY OF SUCH PRODUCTS OR SERVICES OR A WARRANTY, REPRESENTATION OR ENDORSEMENT OF SUCH PRODUCTS OR SERVICES, EITHER ALONE OR IN COMBINATION WITH ANY TI PRODUCT OR SERVICE.

#### 11.1.2 Development Support

##### 11.1.2.1 TINA-TI™ (Free Software Download)

TINA™ is a simple, powerful, and easy-to-use circuit simulation program based on a SPICE engine. TINA-TI™ is a free, fully-functional version of the TINA software, preloaded with a library of macro models in addition to a range of both passive and active models. TINA-TI provides all the conventional DC, transient, and frequency domain analysis of SPICE, as well as additional design capabilities.

Available as a [free download](#) from the Analog eLab Design Center, TINA-TI offers extensive post-processing capability that allows users to format results in a variety of ways. Virtual instruments offer the ability to select input waveforms and probe circuit nodes, voltages, and waveforms, creating a dynamic quick-start tool.

---

#### NOTE

These files require that either the TINA software (from DesignSoft™) or TINA-TI software be installed. Download the free TINA-TI software from the [TINA-TI folder](#).

---

##### 11.1.2.2 DIP Adapter EVM

The [DIP Adapter EVM](#) tool provides an easy, low-cost way to prototype small surface mount ICs. The evaluation tool these TI packages: D or U (SOIC-8), PW (TSSOP-8), DGK (VSSOP-8), DBV (SOT23-6, SOT23-5 and SOT23-3), DCK (SC70-6 and SC70-5), and DRL (SOT563-6). The DIP Adapter EVM may also be used with terminal strips or may be wired directly to existing circuits.

##### 11.1.2.3 Universal Operational Amplifier EVM

The [Universal Op Amp EVM](#) is a series of general-purpose, blank circuit boards that simplify prototyping circuits for a variety of IC package types. The evaluation module board design allows many different circuits to be constructed easily and quickly. Five models are offered, with each model intended for a specific package type. PDIP, SOIC, VSSOP, TSSOP and SOT-23 packages are all supported.

---

#### NOTE

These boards are unpopulated, so users must provide their own ICs. TI recommends requesting several op amp device samples when ordering the Universal Op Amp EVM.

---

##### 11.1.2.4 TI Precision Designs

TI Precision Designs are analog solutions created by TI's precision analog applications experts and offer the theory of operation, component selection, simulation, complete PCB schematic and layout, bill of materials, and measured performance of many useful circuits. TI Precision Designs are available online at <http://www.ti.com/ww/en/analog/precision-designs/>.

##### 11.1.2.5 WEBENCH® Filter Designer

[WEBENCH® Filter Designer](#) is a simple, powerful, and easy-to-use active filter design program. The WEBENCH Filter Designer lets you create optimized filter designs using a selection of TI operational amplifiers and passive components from TI's vendor partners.

## Device Support (continued)

Available as a web-based tool from the WEBENCH® Design Center, [WEBENCH® Filter Designer](#) allows you to design, optimize, and simulate complete multistage active filter solutions within minutes.

### 11.2 Documentation Support

#### 11.2.1 Related Documentation

The following documents are relevant to using the OPA322x, and recommended for reference. All are available for download at [www.ti.com](#) unless otherwise noted.

- [QFN/SON PCB Attachment](#) (SLUA271)
- [Quad Flatpack No-Lead Logic Packages](#) (SCBA017)
- [OPA322, OPA2322, OPA4322 EMIR Immunity Performance](#) (SBOT005)
- [FilterPro™ User's Guide](#) (SBFA001)
- [AFE for Transient Recorder and Digital Fault Recorder Using High-Speed ADCs and Differential Amplifiers](#) (TIDUAT7)
- [Reference Design for Interfacing Current Output Hall Sensors and CTs With Differential ADCs/MCUs](#) (TIDUA57)
- [Single-Ended Signal Conditioning Circuit for Current and Voltage Measurement Using Fluxgate Sensors](#) (TIDU583)
- [Differential Signal Conditioning Circuit for Current and Voltage Measurement Using Fluxgate Sensors](#) (TIDU569)

### 11.3 Related Links

[Table 1](#) lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

**Table 1. Related Links**

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
OPA322	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>
OPA322S	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>
OPA2322	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>
OPA2322S	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>
OPA4322	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>
OPA4322S	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>	<a href="#">Click here</a>

### 11.4 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](#). In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

### 11.5 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

**TI E2E™ Online Community** *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At [e2e.ti.com](#), you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

## 11.6 Trademarks

FilterPro, TINA-TI, E2E are trademarks of Texas Instruments.  
WEBENCH is a registered trademark of Texas Instruments.  
TINA, DesignSoft are trademarks of DesignSoft, Inc.  
All other trademarks are the property of their respective owners.

## 11.7 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## 11.8 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
OPA2322AID	ACTIVE	SOIC	D	8	75	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	O2322A	<a href="#">Samples</a>
OPA2322AIDGKR	ACTIVE	VSSOP	DGK	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU   CU NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	OOZI	<a href="#">Samples</a>
OPA2322AIDGKT	ACTIVE	VSSOP	DGK	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU   CU NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	OOZI	<a href="#">Samples</a>
OPA2322AIDR	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	O2322A	<a href="#">Samples</a>
OPA2322AIDRGR	ACTIVE	SON	DRG	8	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPCI	<a href="#">Samples</a>
OPA2322AIDRGT	ACTIVE	SON	DRG	8	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPCI	<a href="#">Samples</a>
OPA2322SAIDGSR	ACTIVE	VSSOP	DGS	10	2500	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPBI	<a href="#">Samples</a>
OPA2322SAIDGST	ACTIVE	VSSOP	DGS	10	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	OPBI	<a href="#">Samples</a>
OPA322AIDBVR	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	RAD	<a href="#">Samples</a>
OPA322AIDBVT	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	RAD	<a href="#">Samples</a>
OPA322SAIDBVR	ACTIVE	SOT-23	DBV	6	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	RAF	<a href="#">Samples</a>
OPA322SAIDBVT	ACTIVE	SOT-23	DBV	6	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	RAF	<a href="#">Samples</a>
OPA4322AIPW	ACTIVE	TSSOP	PW	14	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	O4322A	<a href="#">Samples</a>
OPA4322AIPWR	ACTIVE	TSSOP	PW	14	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	O4322A	<a href="#">Samples</a>
OPA4322SAIPW	ACTIVE	TSSOP	PW	16	90	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	O4322SA	<a href="#">Samples</a>
OPA4322SAIPWR	ACTIVE	TSSOP	PW	16	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	O4322SA	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:  
**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSELETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

**Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

**Green (RoHS & no Sb/Br):** TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>(4)</sup> There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

<sup>(5)</sup> Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

<sup>(6)</sup> Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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**OTHER QUALIFIED VERSIONS OF OPA2322, OPA4322 :**

- Automotive: [OPA2322-Q1](#), [OPA4322-Q1](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

## TAPE AND REEL INFORMATION



### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2322AIDGKR	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA2322AIDGKT	VSSOP	DGK	8	250	180.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA2322AIDR	SOIC	D	8	2500	330.0	12.4	6.4	5.2	2.1	8.0	12.0	Q1
OPA2322AIDRGR	SON	DRG	8	3000	330.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
OPA2322AIDRGT	SON	DRG	8	250	180.0	12.4	3.3	3.3	1.1	8.0	12.0	Q2
OPA2322SAIDGSR	VSSOP	DGS	10	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA2322SAIDGST	VSSOP	DGS	10	250	180.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA4322AIPWR	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
OPA4322SAIPWR	TSSOP	PW	16	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA2322AIDGKR	VSSOP	DGK	8	2500	367.0	367.0	35.0
OPA2322AIDGKT	VSSOP	DGK	8	250	210.0	185.0	35.0
OPA2322AIDR	SOIC	D	8	2500	367.0	367.0	35.0
OPA2322AIDRGR	SON	DRG	8	3000	367.0	367.0	35.0
OPA2322AIDRGT	SON	DRG	8	250	210.0	185.0	35.0
OPA2322SAIDGSR	VSSOP	DGS	10	2500	367.0	367.0	35.0
OPA2322SAIDGST	VSSOP	DGS	10	250	210.0	185.0	35.0
OPA4322AIPWR	TSSOP	PW	14	2000	367.0	367.0	35.0
OPA4322SAIPWR	TSSOP	PW	16	2000	367.0	367.0	35.0

# DGS0010A



# PACKAGE OUTLINE

## VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



4221984/A 05/2015

### NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
5. Reference JEDEC registration MO-187, variation BA.

# EXAMPLE BOARD LAYOUT

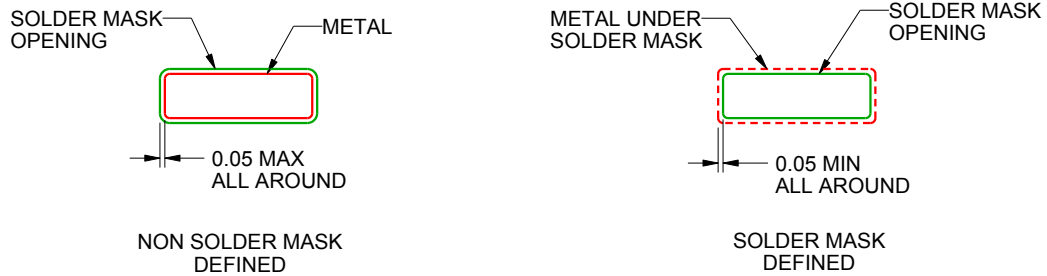
DGS0010A

VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE  
SCALE:10X



SOLDER MASK DETAILS  
NOT TO SCALE

4221984/A 05/2015

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DGS0010A

VSSOP - 1.1 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:10X

4221984/A 05/2015

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - B. This drawing is subject to change without notice.
  - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
  - D. Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
  - E. Falls within JEDEC MO-153

PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE



4211284-2/G 08/15

- NOTES:
- All linear dimensions are in millimeters.
  - This drawing is subject to change without notice.
  - Publication IPC-7351 is recommended for alternate designs.
  - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



D0008A

# PACKAGE OUTLINE

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



4214825/C 02/2019

NOTES:

- Linear dimensions are in inches [millimeters]. Dimensions in parenthesis are for reference only. Controlling dimensions are in inches. Dimensioning and tolerancing per ASME Y14.5M.
- This drawing is subject to change without notice.
- This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed .006 [0.15] per side.
- This dimension does not include interlead flash.
- Reference JEDEC registration MS-012, variation AA.

# EXAMPLE BOARD LAYOUT

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:8X



SOLDER MASK DETAILS

4214825/C 02/2019

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

D0008A

SOIC - 1.75 mm max height

SMALL OUTLINE INTEGRATED CIRCUIT



SOLDER PASTE EXAMPLE  
BASED ON .005 INCH [0.125 MM] THICK STENCIL  
SCALE:8X

4214825/C 02/2019

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.



# EXAMPLE BOARD LAYOUT

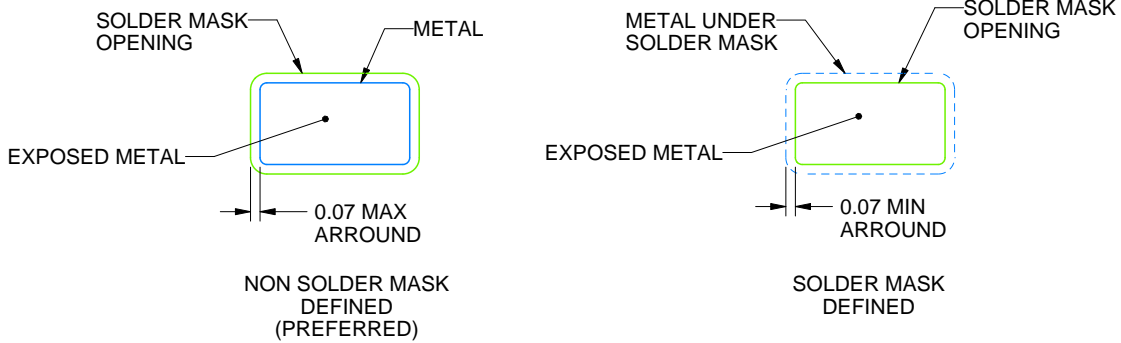
DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:15X



SOLDER MASK DETAILS

4214839/D 11/2018

NOTES: (continued)

- 5. Publication IPC-7351 may have alternate designs.
- 6. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:15X

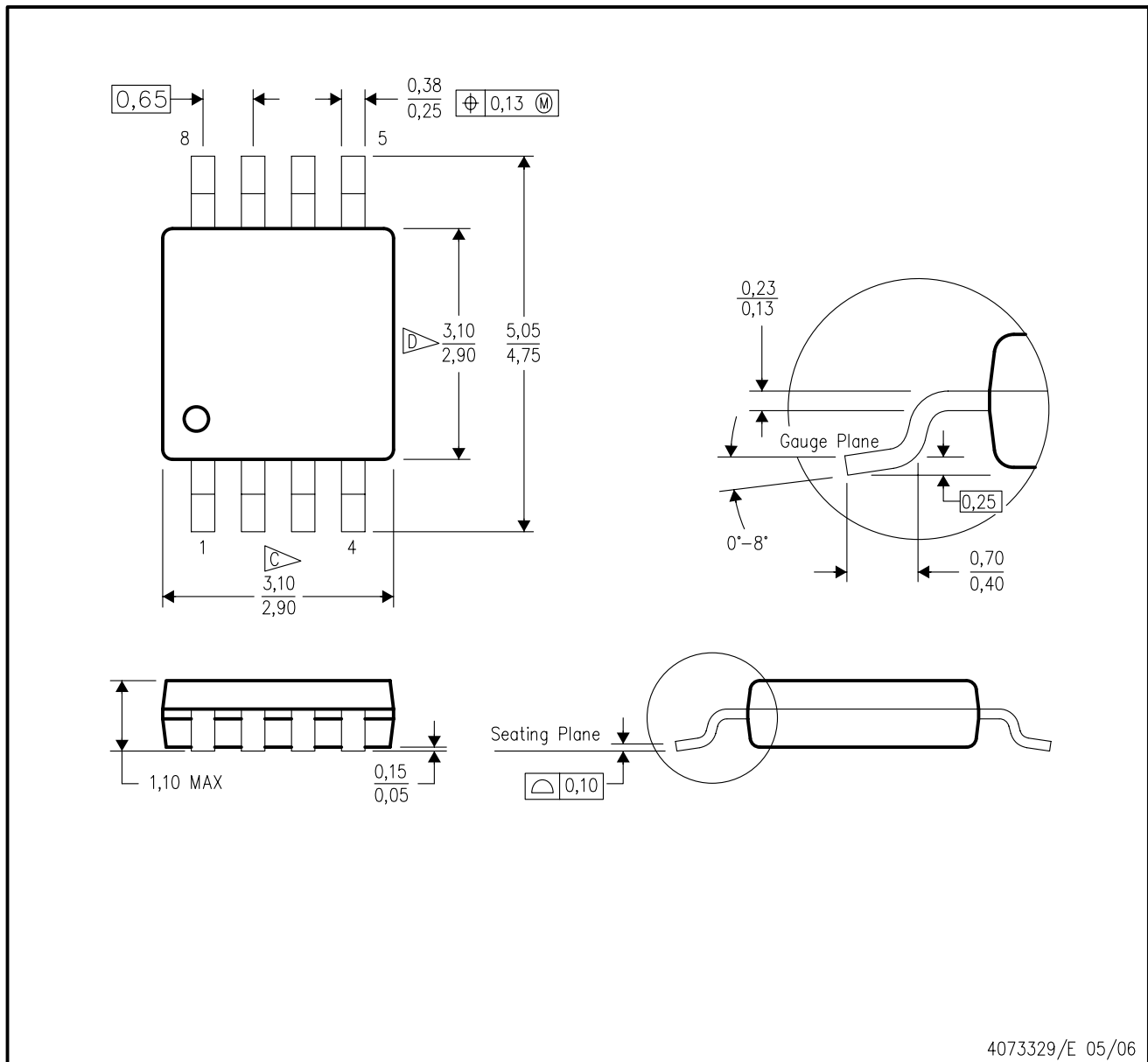
4214839/D 11/2018

NOTES: (continued)

7. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
8. Board assembly site may have different recommendations for stencil design.

DGK (S-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



4073329/E 05/06

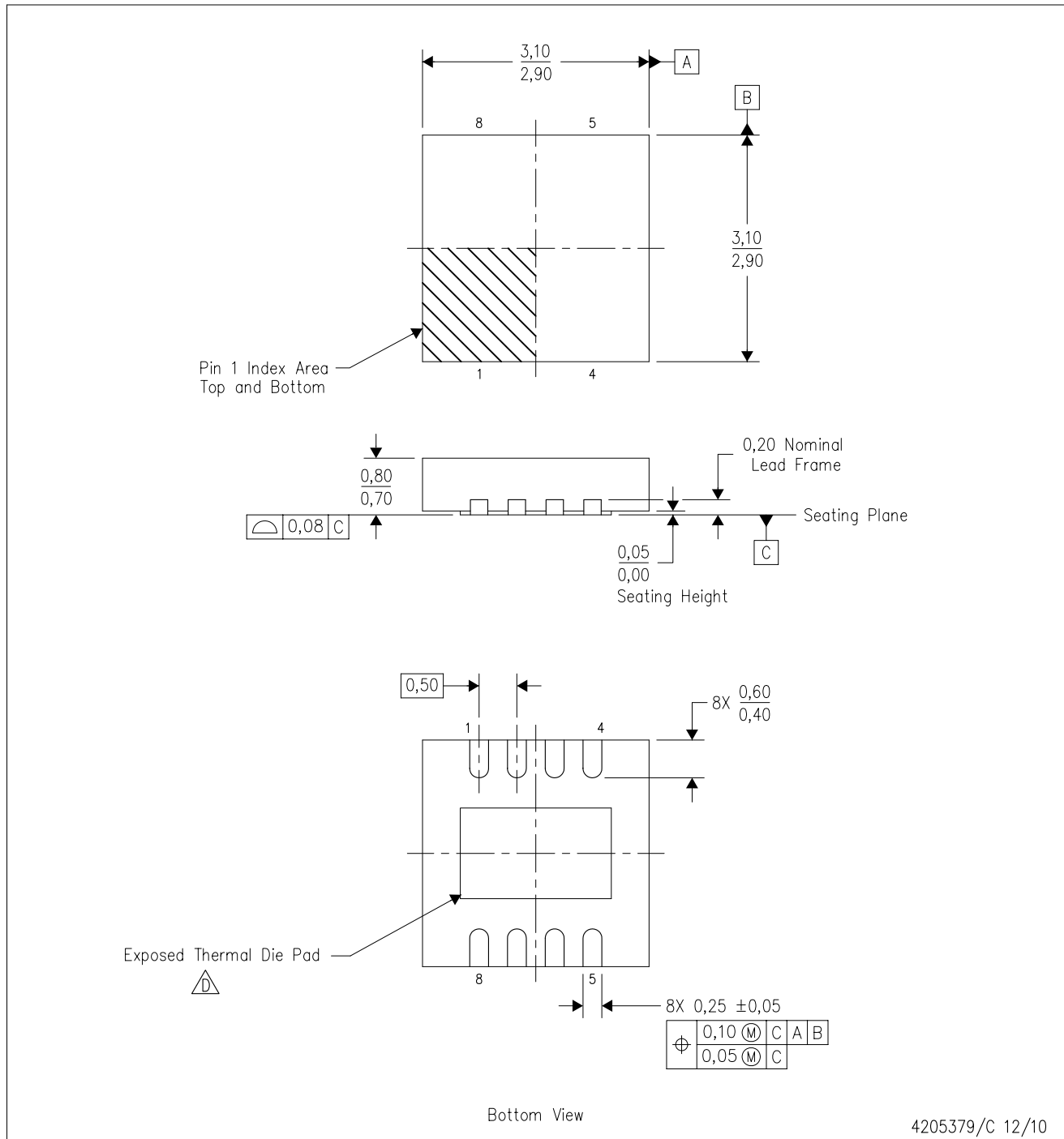
- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
  - D. Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
  - E. Falls within JEDEC MO-187 variation AA, except interlead flash.



- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

DRG (S-PWSON-N8)

PLASTIC SMALL OUTLINE NO-LEAD



- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - B. This drawing is subject to change without notice.
  - C. SON (Small Outline No-Lead) package configuration.
  - D. The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.
  - E. JEDEC MO-229 package registration pending.

# THERMAL PAD MECHANICAL DATA

DRG (S-PWSON-N8)

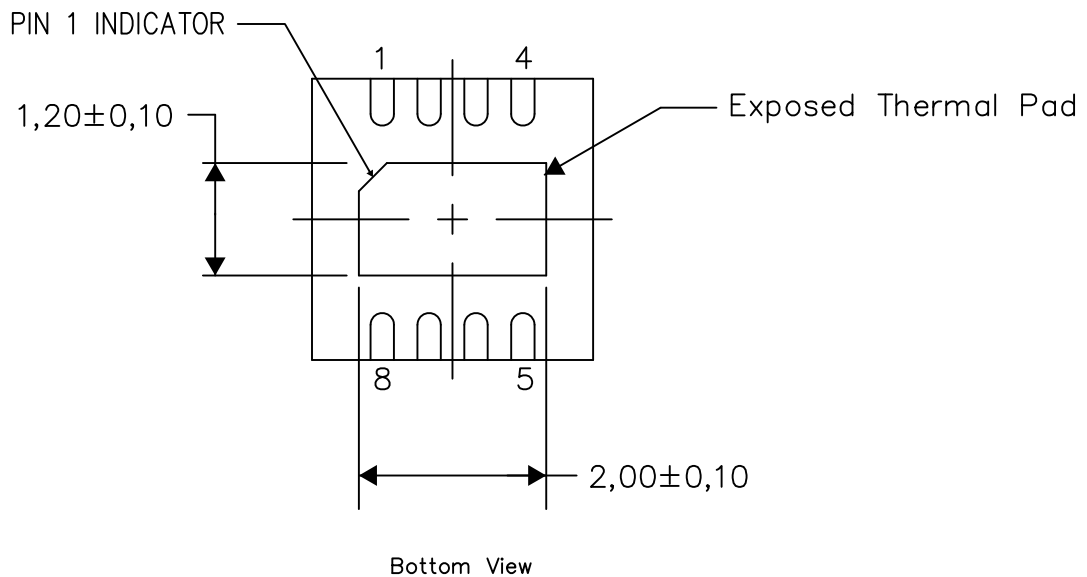
PLASTIC SMALL OUTLINE NO-LEAD

## THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at [www.ti.com](http://www.ti.com).

The exposed thermal pad dimensions for this package are shown in the following illustration.



Exposed Thermal Pad Dimensions

4206881-2/1 03/15

NOTE: All linear dimensions are in millimeters



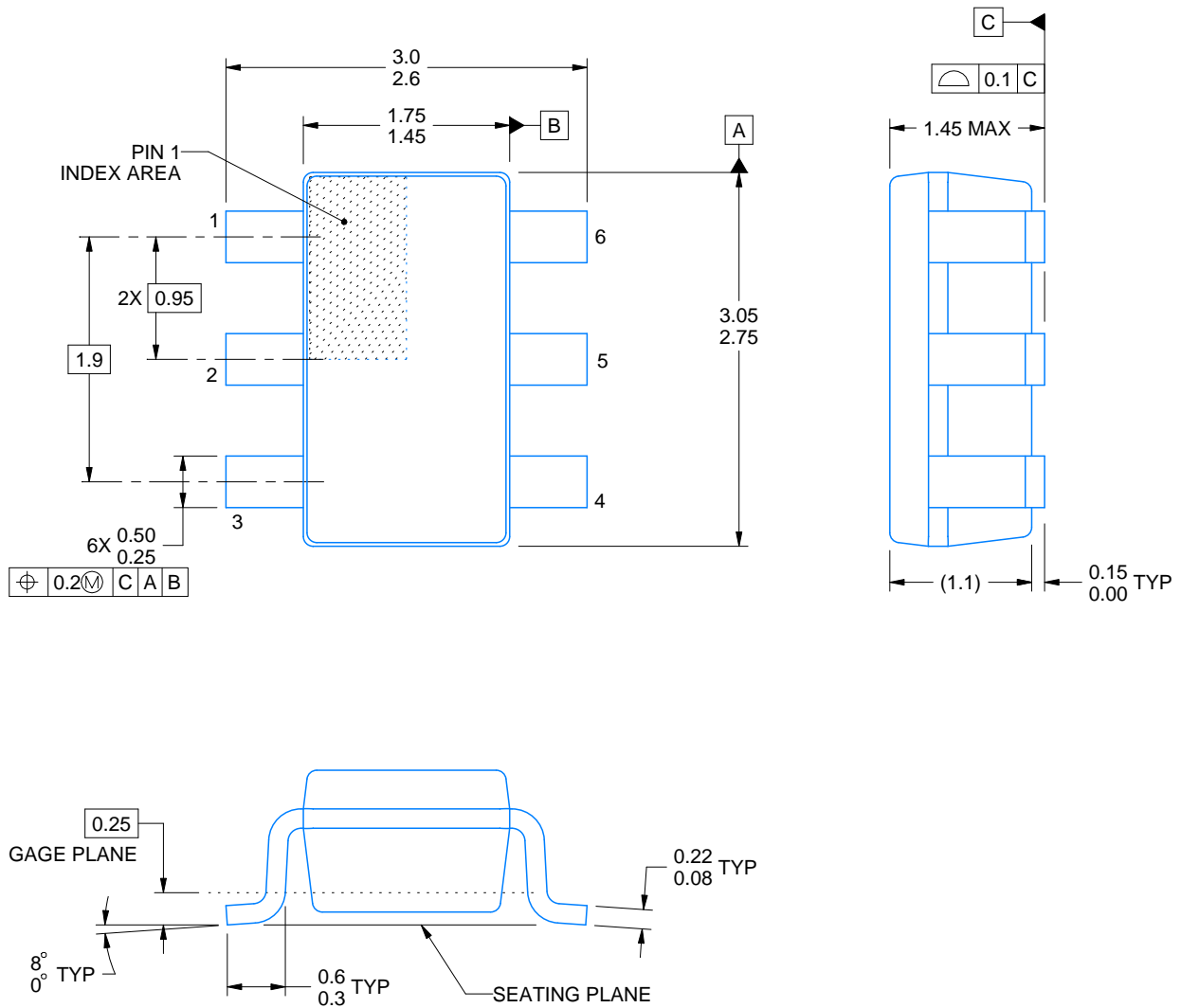
DBV0006A



# PACKAGE OUTLINE

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



4214840/B 03/2018

NOTES:

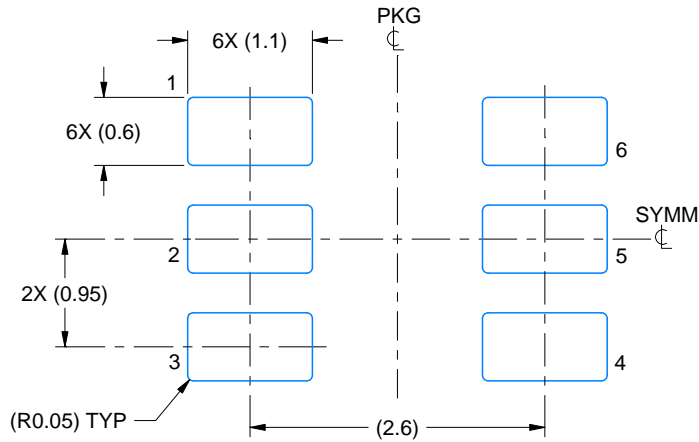
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
4. Leads 1,2,3 may be wider than leads 4,5,6 for package orientation.
5. Reference JEDEC MO-178.

# EXAMPLE BOARD LAYOUT

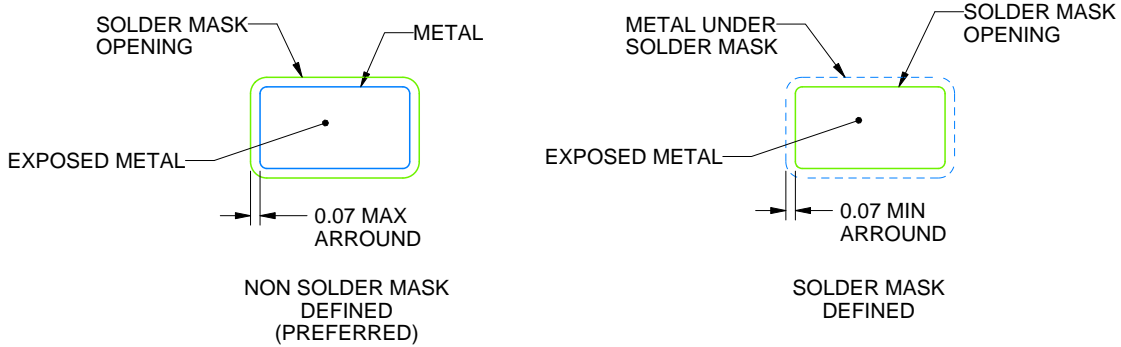
DBV0006A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:15X



SOLDER MASK DETAILS

4214840/B 03/2018

NOTES: (continued)

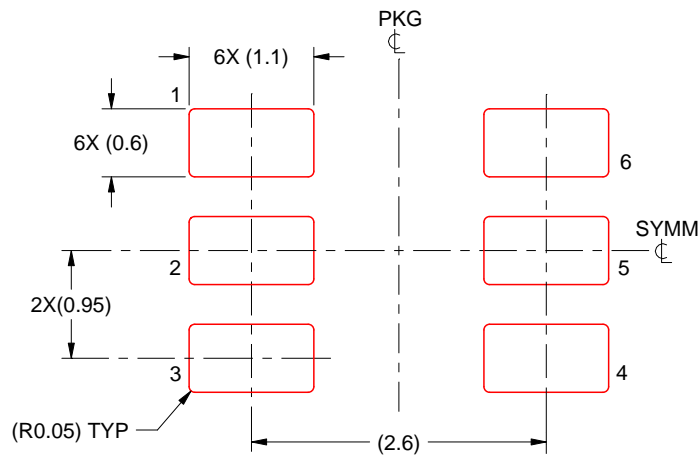
- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DBV0006A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:15X

4214840/B 03/2018

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

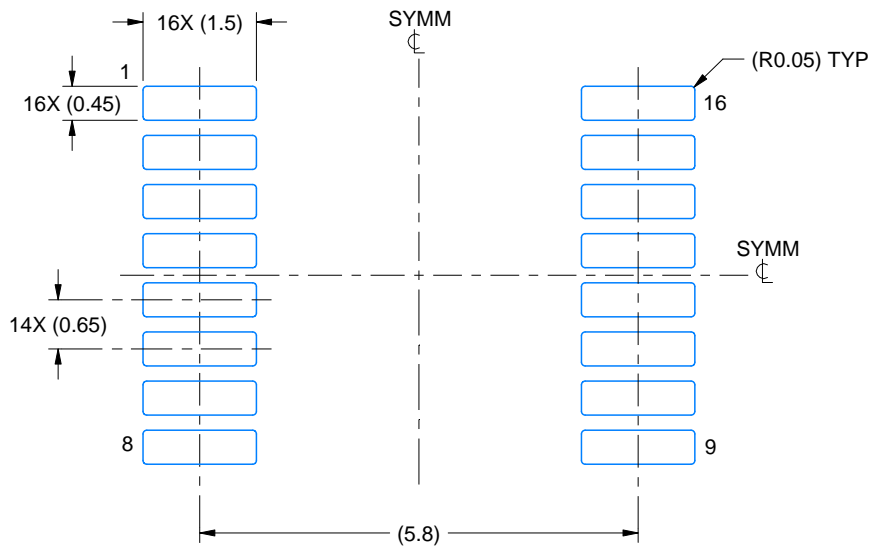


# EXAMPLE BOARD LAYOUT

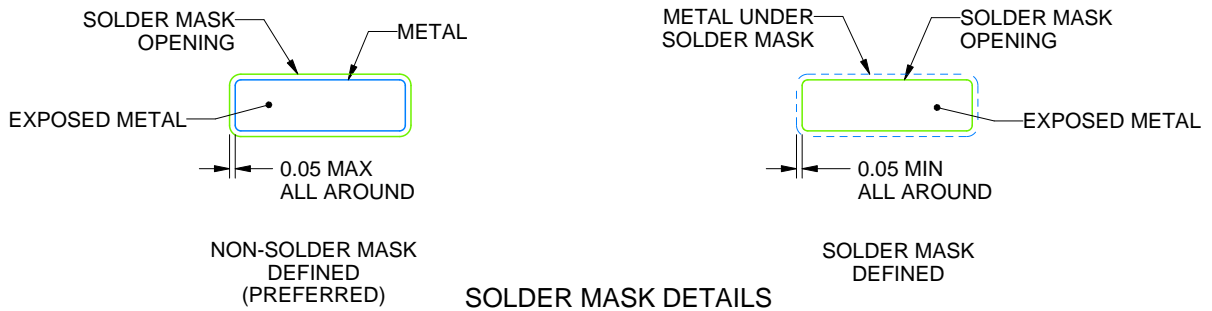
PW0016A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE: 10X



4220204/A 02/2017

NOTES: (continued)

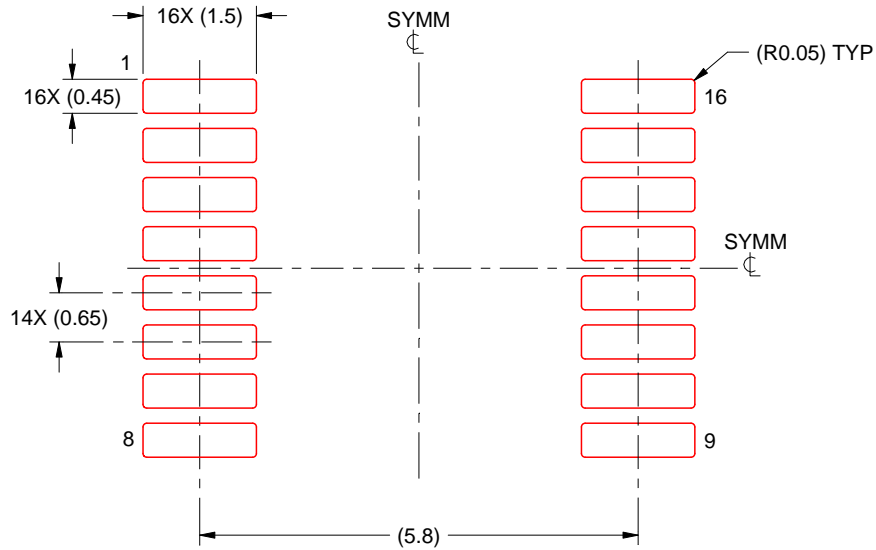
- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

PW0016A

TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE: 10X

4220204/A 02/2017

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

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