



# THE DATASHEET OF ADV7180BCPZ-REEL





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**1/06—Revision 0: Initial Version**



FUNCTIONAL BLOCK DIAGRAMS

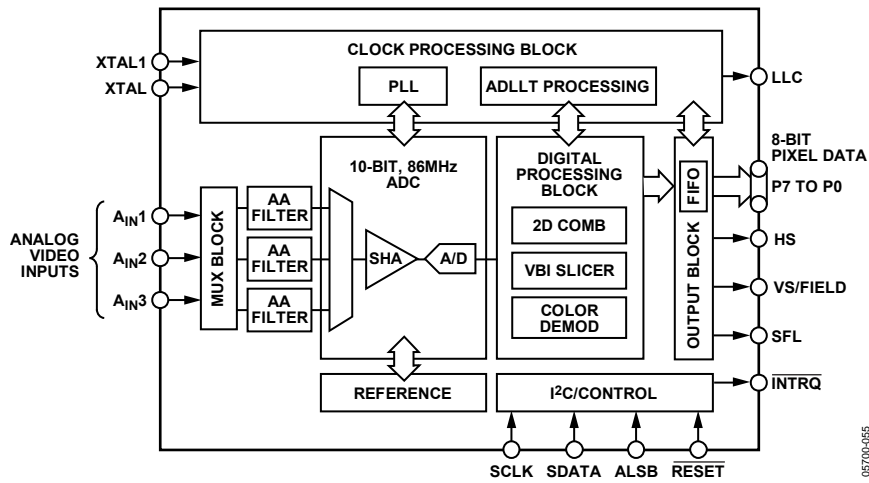


Figure 2. 32-Lead LFCSP Functional Diagram

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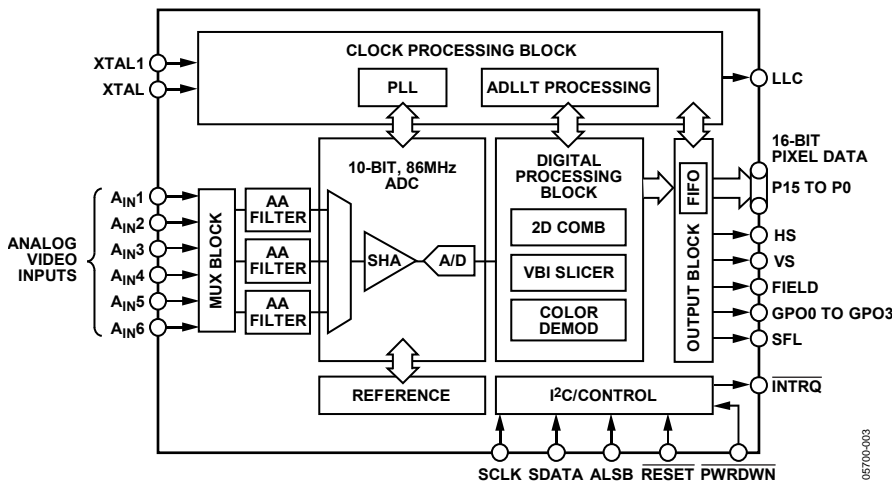


Figure 3. 64-Lead LQFP Functional Block Diagram

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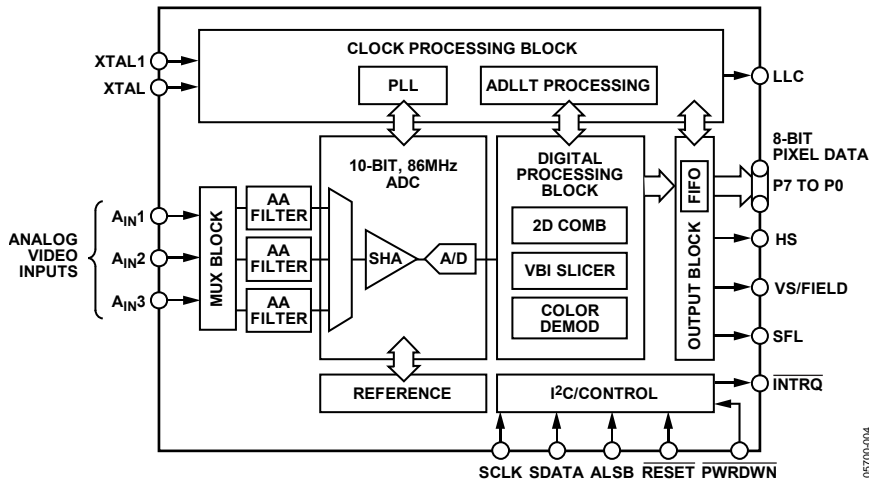


Figure 4. 40-Lead LFCSP Functional Block Diagram

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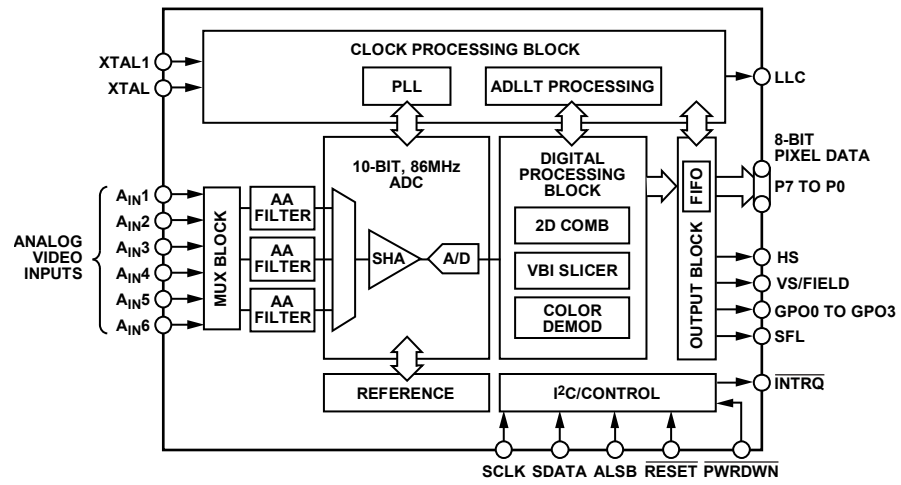


Figure 5. 48-Lead LQFP Functional Block Diagram

05700-980



**VIDEO SPECIFICATIONS**

Guaranteed by characterization.  $A_{VDD} = 1.71\text{ V to }1.89\text{ V}$ ,  $D_{VDD} = 1.65\text{ V to }2.0\text{ V}$ ,  $D_{VDDIO} = 1.62\text{ V to }3.6\text{ V}$ ,  $P_{VDD} = 1.65\text{ V to }2.0\text{ V}$ , specified at operating temperature range, unless otherwise noted.

**Table 4.**

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
<b>NONLINEAR SPECIFICATIONS</b>						
Differential Phase	DP	CVBS input, modulate five-step [NTSC]		0.6		Degrees
Differential Gain	DG	CVBS input, modulate five-step [NTSC]		0.5		%
Luma Nonlinearity	LNL	CVBS input, five-step [NTSC]		2.0		%
<b>NOISE SPECIFICATIONS</b>						
SNR Unweighted		Luma ramp		57.1		dB
		Luma flat field		58		dB
Analog Front-End Crosstalk				60		dB
<b>LOCK TIME SPECIFICATIONS</b>						
Horizontal Lock Range			-5		+5	%
Vertical Lock Range			40		70	Hz
$f_{sc}$ Subcarrier Lock Range				$\pm 1.3$		kHz
Color Lock-In Time				60		Lines
Sync Depth Range			20		200	%
Color Burst Range			5		200	%
Vertical Lock Time				2		Fields
Autodetection Switch Speed				100		Lines
Chroma Luma Gain Delay	CVBS			2.9		ns
	Y/C			5.6		ns
	YPrPb			-3.0		ns
<b>LUMA SPECIFICATIONS</b>						
Luma Brightness Accuracy		CVBS, 1 V input		1		%
Luma Contrast Accuracy		CVBS, 1 V input		1		%

**TIMING SPECIFICATIONS**

Guaranteed by characterization.  $A_{VDD} = 1.71\text{ V to }1.89\text{ V}$ ,  $D_{VDD} = 1.65\text{ V to }2.0\text{ V}$ ,  $D_{VDDIO} = 1.62\text{ V to }3.6\text{ V}$ ,  $P_{VDD} = 1.65\text{ V to }2.0\text{ V}$ , specified at operating temperature range, unless otherwise noted.

Table 5.

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
SYSTEM CLOCK AND CRYSTAL						
Nominal Frequency				28.6363		MHz
Frequency Stability					±50	ppm
I <sup>2</sup> C PORT						
SCLK Frequency					400	kHz
SCLK Minimum Pulse Width High	$t_1$		0.6			µs
SCLK Minimum Pulse Width Low	$t_2$		1.3			µs
Hold Time (Start Condition)	$t_3$		0.6			µs
Setup Time (Start Condition)	$t_4$		0.6			µs
SDA Setup Time	$t_5$		100			ns
SCLK and SDA Rise Times	$t_6$				300	ns
SCLK and SDA Fall Times	$t_7$				300	ns
Setup Time for Stop Condition	$t_8$			0.6		µs
RESET FEATURE						
Reset Pulse Width			5			ms
CLOCK OUTPUTS						
LLC Mark Space Ratio	$t_9:t_{10}$		45:55		55:45	% duty cycle
DATA AND CONTROL OUTPUTS						
Data Output Transitional Time	$t_{11}$	Negative clock edge to start of valid data ( $t_{SETUP} = t_{10} - t_{11}$ )			3.6	ns
Data Output Transitional Time	$t_{12}$	End of valid data to negative clock edge ( $t_{HOLD} = t_9 - t_{12}$ )			2.4	ns

**Timing Diagrams**

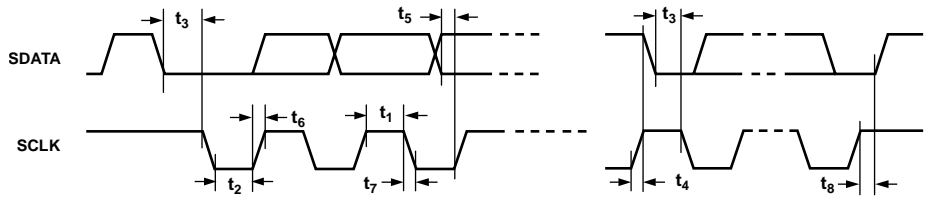


Figure 6. I<sup>2</sup>C Timing

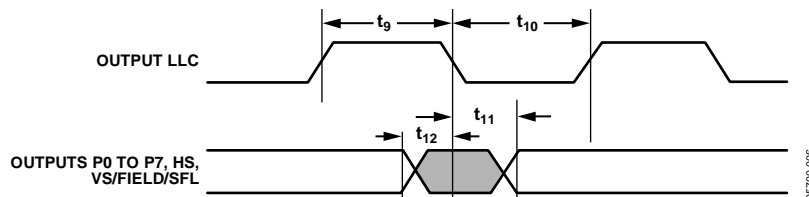


Figure 7. Pixel Port and Control Output Timing

**ANALOG SPECIFICATIONS**

Guaranteed by characterization.  $A_{VDD} = 1.71\text{ V to }1.89\text{ V}$ ,  $D_{VDD} = 1.65\text{ V to }2.0\text{ V}$ ,  $D_{VDDIO} = 1.62\text{ V to }3.6\text{ V}$ ,  $P_{VDD} = 1.65\text{ V to }2.0\text{ V}$ , specified at operating temperature range, unless otherwise noted.

Table 6.

Parameter	Test Conditions	Min	Typ	Max	Unit
CLAMP CIRCUITRY					
External Clamp Capacitor	Clamps switched off		0.1		$\mu\text{F}$
Input Impedance			10		$\text{M}\Omega$
Large-Clamp Source Current			0.4		$\text{mA}$
Large-Clamp Sink Current			0.4		$\text{mA}$
Fine Clamp Source Current			10		$\mu\text{A}$
Fine Clamp Sink Current			10		$\mu\text{A}$

**THERMAL SPECIFICATIONS**

Table 7.

Parameter	Symbol	Test Conditions	Min	Typ	Max	Unit
THERMAL CHARACTERISTICS						
Junction-to-Ambient Thermal Resistance (Still Air)	$\theta_{JA}$	4-layer PCB with solid ground plane, 32-lead LFCSP		32.5		$^{\circ}\text{C/W}$
Junction-to-Case Thermal Resistance	$\theta_{JC}$	4-layer PCB with solid ground plane, 32-lead LFCSP		2.3		$^{\circ}\text{C/W}$
Junction-to-Ambient Thermal Resistance (Still Air)	$\theta_{JA}$	4-layer PCB with solid ground plane, 40-lead LFCSP		30		$^{\circ}\text{C/W}$
Junction-to-Case Thermal Resistance	$\theta_{JC}$	4-layer PCB with solid ground plane, 40-lead LFCSP		3		$^{\circ}\text{C/W}$
Junction-to-Ambient Thermal Resistance (Still Air)	$\theta_{JA}$	4-layer PCB with solid ground plane, 64-lead LQFP		47		$^{\circ}\text{C/W}$
Junction-to-Case Thermal Resistance	$\theta_{JC}$	4-layer PCB with solid ground plane, 64-lead LQFP		11.1		$^{\circ}\text{C/W}$
Junction-to-Ambient Thermal Resistance (Still Air)	$\theta_{JA}$	4-layer PCB with solid ground plane, 48-lead LQFP		50		$^{\circ}\text{C/W}$
Junction-to-Case Thermal Resistance	$\theta_{JC}$	4-layer PCB with solid ground plane, 48-lead LQFP		20		$^{\circ}\text{C/W}$

## ABSOLUTE MAXIMUM RATINGS

Table 8.

Parameter	Rating
$A_{VDD}$ to AGND	2.2 V
$D_{VDD}$ to DGND	2.2 V
$P_{VDD}$ to AGND	2.2 V
$D_{VDDIO}$ to DGND	4 V
$D_{VDDIO}$ to $A_{VDD}$	-0.3 V to +4 V
$P_{VDD}$ to $D_{VDD}$	-0.3 V to +0.9 V
$D_{VDDIO}$ to $P_{VDD}$	-0.3 V to +4 V
$D_{VDDIO}$ to $D_{VDD}$	-0.3 V to +4 V
$A_{VDD}$ to $P_{VDD}$	-0.3 V to +0.3 V
$A_{VDD}$ to $D_{VDD}$	-0.3 V to +0.9 V
Digital Inputs Voltage	DGND - 0.3 V to $D_{VDDIO} + 0.3$ V
Digital Outputs Voltage	DGND - 0.3 V to $D_{VDDIO} + 0.3$ V
Analog Inputs to AGND	AGND - 0.3 V to $A_{VDD} + 0.3$ V
Maximum Junction Temperature ( $T_J$ max)	140°C
Storage Temperature Range	-65°C to +150°C
Infrared Reflow Soldering (20 sec)	260°C

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

This device is a high performance integrated circuit with an ESD rating of <2 kV, and it is ESD sensitive. Proper precautions should be taken for handling and assembly.

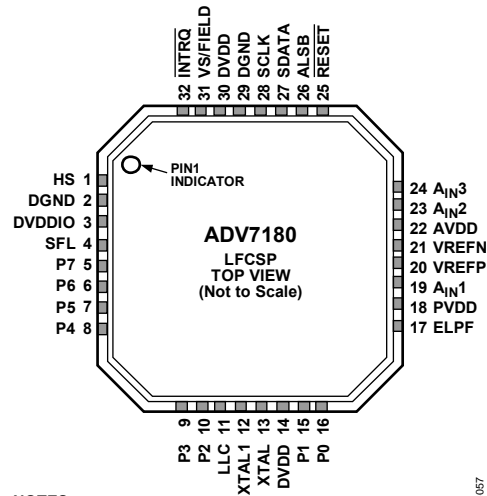
### ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

### 32-LEAD LFCSP



NOTES  
1. THE EXPOSED PAD MUST BE CONNECTED TO GND.

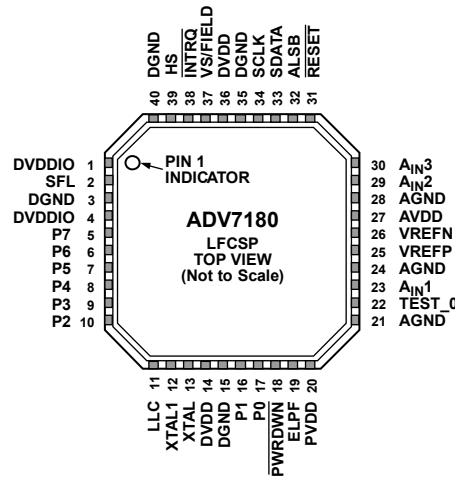
Figure 8. 32-Lead LFCSP Pin Configuration

05700-057

Table 9. 32-Lead LFCSP Pin Function Descriptions

Pin No.	Mnemonic	Type	Description
1	HS	O	Horizontal Synchronization Output Signal.
2, 29	DGND	G	Ground for Digital Supply.
3	DVDDIO	P	Digital I/O Supply Voltage (1.8 V to 3.3 V).
4	SFL	O	Subcarrier Frequency Lock. This pin contains a serial output stream that can be used to lock the subcarrier frequency when this decoder is connected to any Analog Devices digital video encoder.
5 to 10, 15, 16	P7 to P2, P1, P0	O	Video Pixel Output Port.
11	LLC	O	Line-Locked Output Clock for the Output Pixel Data. Nominally 27 MHz but varies up or down according to video line length.
12	XTAL1	O	This pin should be connected to the 28.6363 MHz crystal or not connected if an external 1.8 V, 28.6363 MHz clock oscillator source is used to clock the <a href="#">ADV7180</a> . In crystal mode, the crystal must be a fundamental crystal.
13	XTAL	I	Input Pin for the 28.6363 MHz Crystal. This pin can be overdriven by an external 1.8 V, 28.6363 MHz clock oscillator source. In crystal mode, the crystal must be a fundamental crystal.
14, 30	DVDD	P	Digital Supply Voltage (1.8 V).
17	ELPF	I	The recommended external loop filter must be connected to this ELPF pin, as shown in Figure 60.
18	PVDD	P	PLL Supply Voltage (1.8 V).
19, 23, 24	A <sub>IN1</sub> to A <sub>IN3</sub>	I	Analog Video Input Channels.
20	VREFP	O	Internal Voltage Reference Output. See Figure 60 for recommended output circuitry.
21	VREFN	O	Internal Voltage Reference Output. See Figure 60 for recommended output circuitry.
22	AVDD	P	Analog Supply Voltage (1.8 V).
25	RESET	I	System Reset Input. Active low. A minimum low reset pulse width of 5 ms is required to reset the <a href="#">ADV7180</a> circuitry.
26	ALSB	I	This pin selects the I <sup>2</sup> C address for the <a href="#">ADV7180</a> . For ALSB set to Logic 0, the address selected for a write is Address 0x40; for ALSB set to Logic 1, the address selected is Address 0x42.
27	SDATA	I/O	I <sup>2</sup> C Port Serial Data Input/Output Pin.
28	SCLK	I	I <sup>2</sup> C Port Serial Clock Input. The maximum clock rate is 400 kHz.
31	VS/FIELD	O	Vertical Synchronization Output Signal/Field Synchronization Output Signal.
32	INTRQ	O	Interrupt Request Output. Interrupt occurs when certain signals are detected on the input video (see Table 108).
	EPAD (EP)		The exposed pad must be connected to GND.

40-LEAD LFCSP



NOTES  
1. THE EXPOSED PAD MUST BE CONNECTED TO GND.

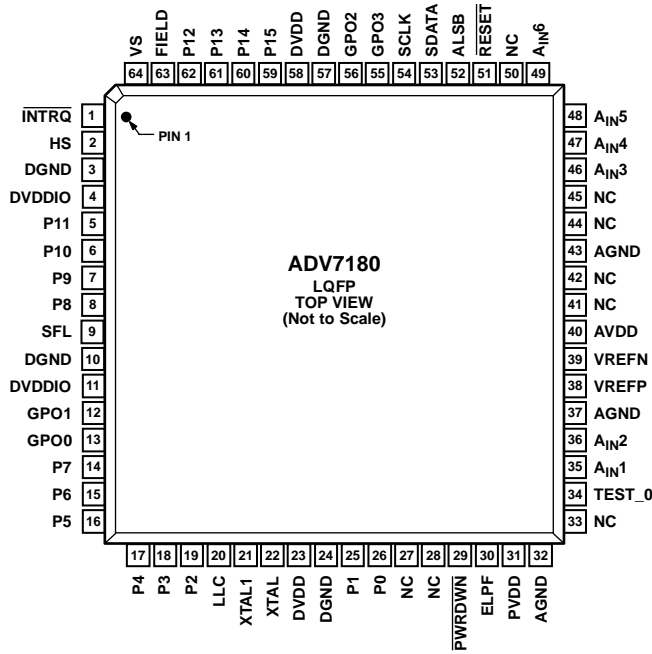
05700-007

Figure 9. 40-Lead LFCSP Pin Configuration

Table 10. 40-Lead LFCSP Pin Function Descriptions

Pin No.	Mnemonic	Type	Description
1, 4	DVDDIO	P	Digital I/O Supply Voltage (1.8 V to 3.3 V).
2	SFL	O	Subcarrier Frequency Lock. This pin contains a serial output stream that can be used to lock the subcarrier frequency when this decoder is connected to any Analog Devices digital video encoder.
3, 15, 35, 40	DGND	G	Ground for Digital Supply.
5 to 10, 16, 17	P7 to P2, P1, P0	O	Video Pixel Output Port.
11	LLC	O	Line-Locked Output Clock for the Output Pixel Data. Nominally 27 MHz but varies up or down according to video line length.
12	XTAL1	O	This pin should be connected to the 28.6363 MHz crystal or not connected if an external 1.8 V, 28.6363 MHz clock oscillator source is used to clock the <b>ADV7180</b> . In crystal mode, the crystal must be a fundamental crystal.
13	XTAL	I	Input Pin for the 28.6363 MHz Crystal. This pin can be overdriven by an external 1.8 V, 28.6363 MHz clock oscillator source. In crystal mode, the crystal must be a fundamental crystal.
14, 36	DVDD	P	Digital Supply Voltage (1.8 V).
18	PWRDWN	I	A logic low on this pin places the <b>ADV7180</b> into power-down mode.
19	ELPF	I	The recommended external loop filter must be connected to this ELPF pin, as shown in Figure 57.
20	PVDD	P	PLL Supply Voltage (1.8 V).
21, 24, 28	AGND	G	Ground for Analog Supply.
22	TEST_0	I	This pin must be tied to DGND.
23, 29, 30	A <sub>IN1</sub> to A <sub>IN3</sub>	I	Analog Video Input Channels.
25	VREFP	O	Internal Voltage Reference Output. See Figure 57 for recommended output circuitry.
26	VREFN	O	Internal Voltage Reference Output. See Figure 57 for recommended output circuitry.
27	AVDD	P	Analog Supply Voltage (1.8 V).
31	RESET	I	System Reset Input. Active low. A minimum low reset pulse width of 5 ms is required to reset the <b>ADV7180</b> circuitry.
32	ALSB	I	This pin selects the I <sup>2</sup> C address for the <b>ADV7180</b> . For ALSB set to Logic 0, the address selected for a write is Address 0x40; for ALSB set to Logic 1, the address selected is Address 0x42.
33	SDATA	I/O	I <sup>2</sup> C Port Serial Data Input/Output Pin.
34	SCLK	I	I <sup>2</sup> C Port Serial Clock Input. The maximum clock rate is 400 kHz.
37	VS/FIELD	O	Vertical Synchronization Output Signal/Field Synchronization Output Signal.
38	INTRQ	O	Interrupt Request Output. Interrupt occurs when certain signals are detected on the input video (see Table 108).
39	HS	O	Horizontal Synchronization Output Signal.
	EPAD (EP)		The exposed pad must be connected to GND.

64-LEAD LQFP



NC = NO CONNECT

Figure 10. 64-Lead LQFP Pin Configuration

Table 11. 64-Lead LQFP Pin Function Description

Pin No.	Mnemonic	Type	Description
1	INTRQ	O	Interrupt Request Output. Interrupt occurs when certain signals are detected on the input video (see Table 108).
2	HS	O	Horizontal Synchronization Output Signal.
3, 10, 24, 57	DGND	G	Digital Ground.
4, 11	DVDDIO	P	Digital I/O Supply Voltage (1.8 V to 3.3 V).
5 to 8, 14 to 19, 25, 26, 59 to 62	P11 to P8, P7 to P2, P1, P0, P15 to P12	O	Video Pixel Output Port. See Table 100 for output configuration for 8-bit and 16-bit modes.
9	SFL	O	Subcarrier Frequency Lock. This pin contains a serial output stream that can be used to lock the subcarrier frequency when this decoder is connected to any Analog Devices digital video encoder.
12, 13, 55, 56	GPO0 to GPO3	O	General-Purpose Outputs. These pins can be configured via I <sup>2</sup> C to allow control of external devices.
20	LLC	O	This is a line-locked output clock for the pixel data output by the ADV7180. It is nominally 27 MHz but varies up or down according to video line length.
21	XTAL1	O	This pin should be connected to the 28.6363 MHz crystal or left as a no connect if an external 1.8 V, 28.6363 MHz clock oscillator source is used to clock the ADV7180. In crystal mode, the crystal must be a fundamental crystal.
22	XTAL	I	This is the input pin for the 28.6363 MHz crystal, or this pin can be overdriven by an external 1.8 V, 28.6363 MHz clock oscillator source. In crystal mode, the crystal must be a fundamental crystal.
23, 58	DVDD	P	Digital Supply Voltage (1.8 V).
27, 28, 33, 41, 42, 44, 45, 50	NC		No Connect. These pins are not connected internally.
29	PWRDWN	I	A logic low on this pin places the ADV7180 in power-down mode.
30	ELPF	I	The recommended external loop filter must be connected to the ELPF pin, as shown in Figure 58.
31	PVDD	P	PLL Supply Voltage (1.8 V).
32, 37, 43	AGND	G	Analog Ground.
34	TEST_0	I	This pin must be tied to DGND.
35, 36, 46 to 49	A <sub>IN</sub> 1 to A <sub>IN</sub> 6	I	Analog Video Input Channels.
38	VREFP	O	Internal Voltage Reference Output. See Figure 58 for recommended output circuitry.

Pin No.	Mnemonic	Type	Description
39	VREFN	O	Internal Voltage Reference Output. See Figure 58 for recommended output circuitry.
40	AVDD	P	Analog Supply Voltage (1.8 V).
51	RESET	I	System Reset Input. Active low. A minimum low reset pulse width of 5 ms is required to reset the <a href="#">ADV7180</a> circuitry.
52	ALSB	I	This pin selects the I <sup>2</sup> C address for the <a href="#">ADV7180</a> . For ALSB set to Logic 0, the address selected for a write is Address 0x40; for ALSB set to Logic 1, the address selected is Address 0x42.
53	SDATA	I/O	I <sup>2</sup> C Port Serial Data Input/Output Pin.
54	SCLK	I	I <sup>2</sup> C Port Serial Clock Input. The maximum clock rate is 400 kHz.
63	FIELD	O	Field Synchronization Output Signal.
64	VS	O	Vertical Synchronization Output Signal.

48-LEAD LQFP

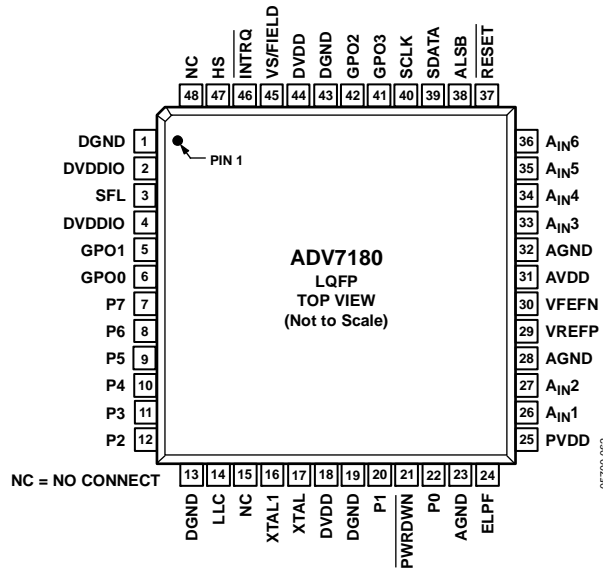


Figure 11. 48-Lead LQFP Pin Configuration

Table 12. 48-Lead LQFP Pin Function Descriptions

Pin No.	Mnemonic	Type	Description
1, 13, 19, 43	DGND	G	Digital Ground.
2, 4	DVDDIO	P	Digital I/O Supply Voltage (1.8 V to 3.3 V).
3	SFL	O	Subcarrier Frequency Lock. This pin contains a serial output stream that can be used to lock the subcarrier frequency when this decoder is connected to any Analog Devices digital video encoder.
5, 6, 41, 42	GPO0 to GPO3	O	General-Purpose Outputs. These pins can be configured via I <sup>2</sup> C to allow control of external devices.
7 to 12, 20, 22	P7 to P2, P1, P0	O	Video Pixel Output Port. See Table 100 for output configuration for 8-bit and 16-bit modes.
14	LLC	O	This is a line-locked output clock for the pixel data output by the <a href="#">ADV7180</a> . It is nominally 27 MHz but varies up or down according to video line length.
15, 48	NC		No Connect Pins. These pins are not connected internally.
16	XTAL1	O	This pin should be connected to the 28.6363 MHz crystal or left as a no connect if an external 1.8 V, 28.6363 MHz clock oscillator source is used to clock the <a href="#">ADV7180</a> . In crystal mode, the crystal must be a fundamental crystal.
17	XTAL	I	This is the input pin for the 28.6363 MHz crystal, or this pin can be overdriven by an external 1.8 V, 28.6363 MHz clock oscillator source. In crystal mode, the crystal must be a fundamental crystal.
18, 44	DVDD	P	Digital Supply Voltage (1.8 V).
21	PWRDWN	I	A logic low on this pin places the <a href="#">ADV7180</a> in power-down mode.
23, 28, 32	AGND	G	Analog Ground.
24	ELPF	I	The recommended external loop filter must be connected to the ELPF pin, as shown in Figure 59.
25	PVDD	P	PLL Supply Voltage (1.8 V).
26, 27, 33 to 36	A <sub>IN</sub> 1 to A <sub>IN</sub> 6	I	Analog Video Input Channels.
29	VREFP	O	Internal Voltage Reference Output. See Figure 59 for recommended output circuitry.
30	VREFN	O	Internal Voltage Reference Output. See Figure 59 for recommended output circuitry.
31	AVDD	P	Analog Supply Voltage (1.8 V).
37	RESET	I	System Reset Input. Active low. A minimum low reset pulse width of 5 ms is required to reset the <a href="#">ADV7180</a> circuitry.
38	ALSB	I	This pin selects the I <sup>2</sup> C address for the <a href="#">ADV7180</a> . For ALSB set to Logic 0, the address selected for a write is Address 0x40; for ALSB set to Logic 1, the address selected is Address 0x42.
39	SDATA	I/O	I <sup>2</sup> C Port Serial Data Input/Output Pin.
40	SCLK	I	I <sup>2</sup> C Port Serial Clock Input. The maximum clock rate is 400 kHz.
45	VS/FIELD	O	Vertical Synchronization Output Signal/Field Synchronization Output Signal.
46	INTRQ	O	Interrupt Request Output. Interrupt occurs when certain signals are detected on the input video (see Table 108).
47	HS	O	Horizontal Synchronization Output Signal.



ANALOG FRONT END

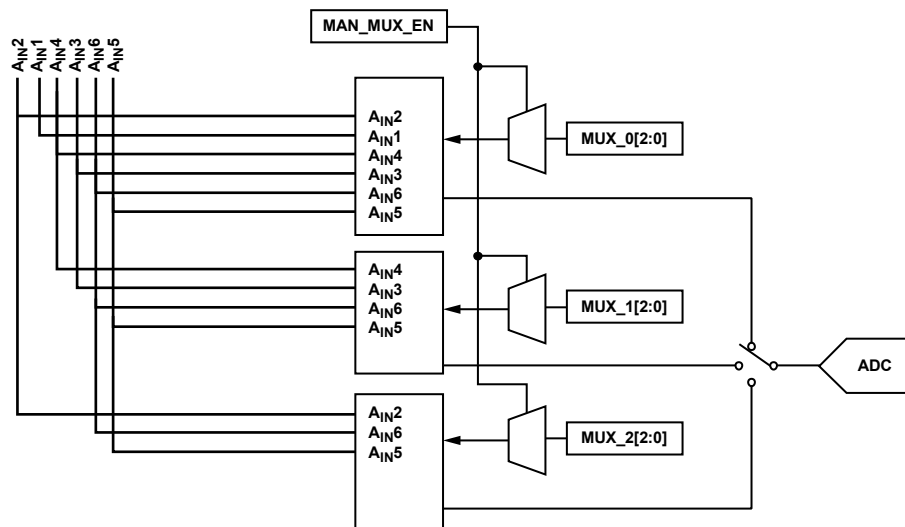


Figure 14. 64-Lead and 48-Lead LQFP Internal Pin Connections

05700-009

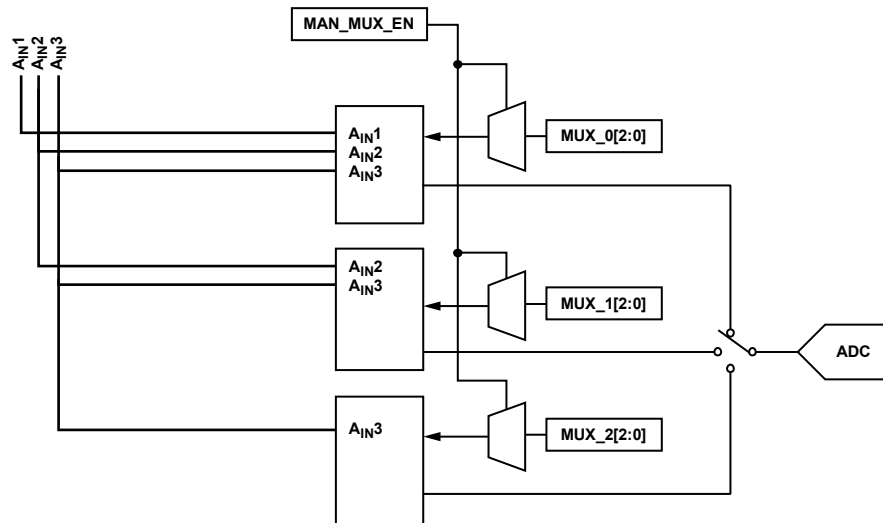


Figure 15. 40-Lead and 32-Lead LFCSP Internal Pin Connections

05700-010



## ANALOG INPUT MUXING

The **ADV7180** has an integrated analog muxing section that allows more than one source of video signal to be connected to the decoder. Figure 14 and Figure 15 outline the overall structure of the input muxing provided in the **ADV7180**.

A maximum of six CVBS inputs can be connected to and decoded by the 64-lead and 48-lead devices, and a maximum of three CVBS inputs can be connected to and decoded by the 40-lead and 32-lead LFCSP devices. As shown in the Pin Configurations and Function Descriptions section, these analog input pins lie in close proximity to one another, which requires careful design of the printed circuit board (PCB) layout. For example, route ground shielding between all signals through tracks that are physically close together. It is strongly recommended to connect any unused analog input pins to AGND to act as a shield.

### **MAN\_MUX\_EN, Manual Input Muxing Enable, Address 0xC4[7]**

To configure the **ADV7180** analog muxing section, the user must select the analog input ( $A_{IN1}$  to  $A_{IN6}$  for the 64-lead LQFP and 48-lead devices or  $A_{IN1}$  to  $A_{IN3}$  for the 40-lead and 32-lead LFCSP devices) that is to be processed by the ADC. **MAN\_MUX\_EN** must be set to 1 to enable the following muxing blocks:

- MUX0[2:0], ADC Mux Configuration, Address 0xC3[2:0]
- MUX1[2:0], ADC Mux Configuration, Address 0xC3[6:4]
- MUX2[2:0], ADC Mux Configuration, Address 0xC4[2:0]

The three mux sections are controlled by the signal buses MUX0/MUX1/MUX2[2:0]. Table 15 explains the control words used.

The input signal that contains the timing information (HS and VS) must be processed by MUX0. For example, in a Y/C input configuration, MUX0 should be connected to the Y channel and MUX1 to the C channel. When one or more muxes are not used to process video, such as the CVBS input, the idle mux and associated channel clamps and buffers should be powered down (see the description of Register 0x3A in Table 107).

**Table 15. Manual Mux Settings for the ADC (MAN\_MUX\_EN Must be Set to 1)**

MUX0[2:0]	ADC Connected To		MUX1[2:0]	ADC Connected To		MUX2[2:0]	ADC Connected To	
	LQFP-64 or LQFP-48	LFCSP-40 or LFCSP-32		LQFP-64 or LQFP-48	LFCSP-40 or LFCSP-32		LQFP-64 or LQFP-48	LFCSP-40 or LFCSP-32
000	No connect	No connect	000	No connect	No connect	000	No connect	No connect
001	$A_{IN1}$	$A_{IN1}$	001	No connect	No connect	001	No connect	No connect
010	$A_{IN2}$	No connect	010	No connect	No connect	010	$A_{IN2}$	No connect
011	$A_{IN3}$	No connect	011	$A_{IN3}$	No connect	011	No connect	No connect
100	$A_{IN4}$	$A_{IN2}$	100	$A_{IN4}$	$A_{IN2}$	100	No connect	No connect
101	$A_{IN5}$	$A_{IN3}$	101	$A_{IN5}$	$A_{IN3}$	101	$A_{IN5}$	$A_{IN3}$
110	$A_{IN6}$	No connect	110	$A_{IN6}$	No connect	110	$A_{IN6}$	No connect
111	No connect	No connect	111	No connect	No connect	111	No connect	No connect

Note the following:

- CVBS can only be processed by MUX0.
- Y/C can only be processed by MUX0 and MUX1.
- YPrPb can only be processed by MUX0, MUX1, and MUX2.

**ANTI\_ALIASING FILTERS**

The **ADV7180** has optional on-chip antialiasing (AA) filters on each of the three channels that are multiplexed to the ADC (see Figure 17). The filters are designed for standard definition video up to 10 MHz bandwidth. Figure 18 and Figure 19 show the filter magnitude and phase characteristics.

The antialiasing filters are enabled by default and the selection of INSEL[3:0] determines which filters are powered up at any given time. For example, if CVBS mode is selected, the filter circuits for the remaining input channels are powered down to conserve power. However, the antialiasing filters can be disabled or bypassed using the AA\_FILT\_MAN\_OVR control.

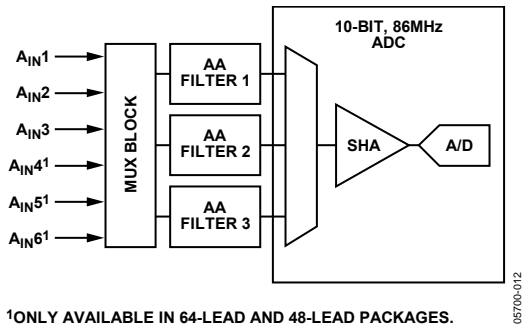


Figure 17. Antialias Filter Configuration

**AA\_FILT\_MAN\_OVR, Antialiasing Filter Override, Address 0xF3[3]**

This feature allows the user to override the antialiasing filters on/off settings, which are automatically selected by INSEL[3:0].

**AA\_FILT\_EN, Antialiasing Filter Enable, Address 0xF3[2:0]**

These bits allow the user to enable or disable the antialiasing filters on each of the three input channels multiplexed to the ADC. When disabled, the analog signal bypasses the AA filter and is routed directly to the ADC.

**AA\_FILT\_EN, Address 0xF3[0]**

When AA\_FILT\_EN[0] is 0, AA Filter 1 is bypassed.

When AA\_FILT\_EN[0] is 1, AA Filter 1 is enabled.

**AA\_FILT\_EN, Address 0xF3[1]**

When AA\_FILT\_EN[1] is 0, AA Filter 2 is bypassed.

When AA\_FILT\_EN[1] is 1, AA Filter 2 is enabled.

**AA\_FILT\_EN, Address 0xF3[2]**

When AA\_FILT\_EN[2] is 0, AA Filter 3 is bypassed.

When AA\_FILT\_EN[2] is 1, AA Filter 3 is enabled.

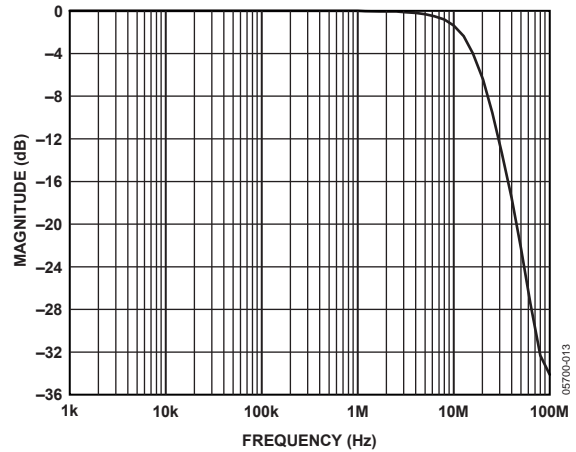


Figure 18. Antialiasing Filter Magnitude Response

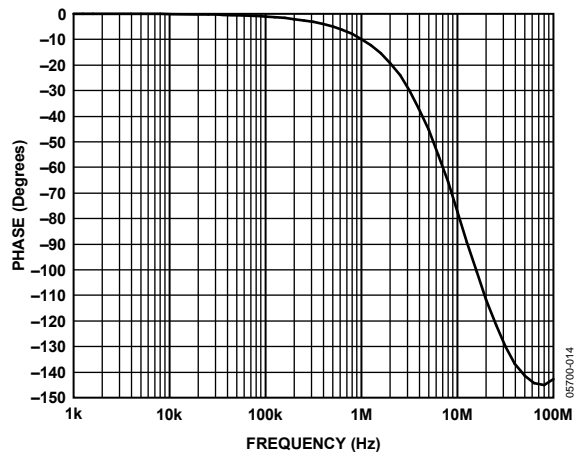


Figure 19. Antialiasing Filter Phase Response







## VIDEO PROCESSOR

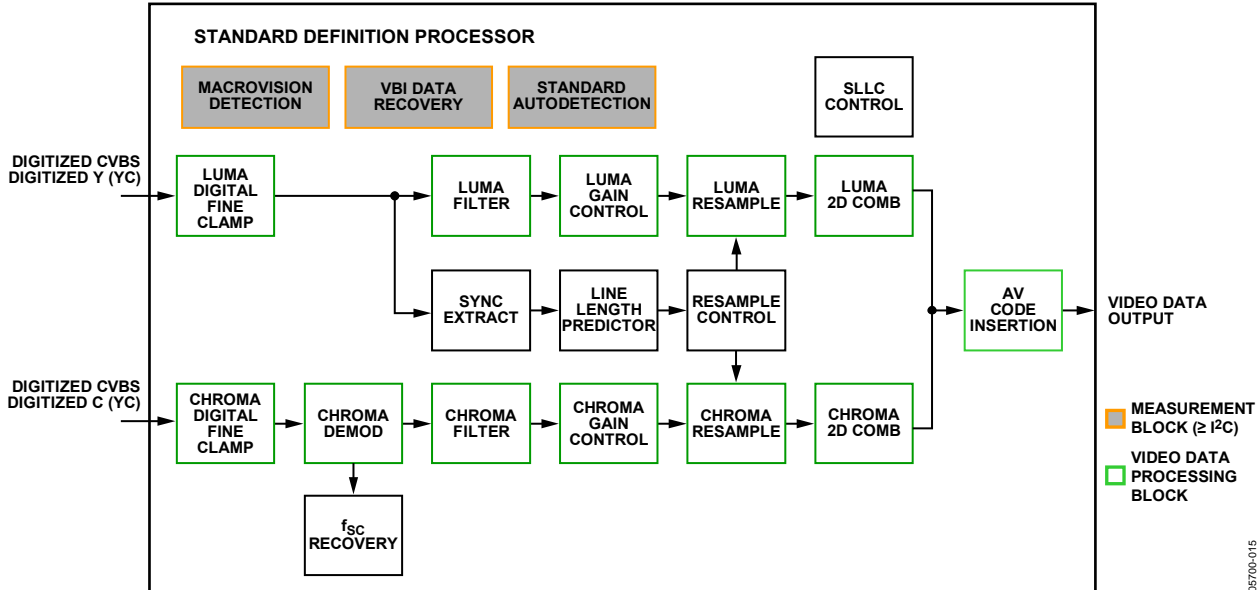


Figure 20. Block Diagram of the Video Processor

Figure 20 shows a block diagram of the [ADV7180](#) video processor. The [ADV7180](#) can handle standard definition video in CVBS, Y/C, and YPrPb formats. It can be divided into a luminance and chrominance path. If the input video is of a composite type (CVBS), both processing paths are fed with the CVBS input.

**SD LUMA PATH**

The input signal is processed by the following blocks:

- Luma digital fine clamp. This block uses a high precision algorithm to clamp the video signal.
- Luma filter. This block contains a luma decimation filter (YAA) with a fixed response and some shaping filters (YSH) that have selectable responses.
- Luma gain control. The automatic gain control (AGC) can operate on a variety of different modes, including gain based on the depth of the horizontal sync pulse, peak white mode, and fixed manual gain.
- Luma resample. To correct for line length errors as well as dynamic line length changes, the data is digitally resampled.
- Luma 2D comb. The 2D comb filter provides Y/C separation.
- AV code insertion. At this point, the decoded luma (Y) signal is merged with the retrieved chroma values. AV codes can be inserted (as per ITU-R BT.656).

**SD CHROMA PATH**

The input signal is processed by the following blocks:

- Chroma digital fine clamp. This block uses a high precision algorithm to clamp the video signal.
- Chroma demodulation. This block employs a color subcarrier ( $f_{sc}$ ) recovery unit to regenerate the color subcarrier for any modulated chroma scheme. The demodulation block then performs an AM demodulation for PAL and NTSC, and an FM demodulation for SECAM.
- Chroma filter. This block contains a chroma decimation filter (CAA) with a fixed response and some shaping filters (CSH) that have selectable responses.
- Chroma gain control. AGC can operate on several different modes, including gain based on the color subcarrier amplitude, gain based on the depth of the horizontal sync pulse on the luma channel, or fixed manual gain.
- Chroma resample. The chroma data is digitally resampled to keep it perfectly aligned with the luma data. The resampling is done to correct for static and dynamic line length errors of the incoming video signal.
- Chroma 2D comb. The 2D, five line, superadaptive comb filter provides high quality Y/C separation in case the input signal is CVBS.
- AV code insertion. At this point, the demodulated chroma (Cr and Cb) signal is merged with the retrieved luma values. AV codes can be inserted (as per ITU-R BT.656).

## SYNC PROCESSING

The **ADV7180** extracts syncs embedded in the analog input video signal. There is currently no support for external HS/VS inputs. The sync extraction is optimized to support imperfect video sources, such as VCRs with head switches. The actual algorithm used employs a coarse detection based on a threshold crossing, followed by a more detailed detection using an adaptive interpolation algorithm. The raw sync information is sent to a line length measurement and prediction block. The output of this is then used to drive the digital resampling section to ensure that the **ADV7180** outputs 720 active pixels per line.

The sync processing on the **ADV7180** also includes the following specialized postprocessing blocks that filter and condition the raw sync information retrieved from the digitized analog video:

- VSYNC processor. This block provides extra filtering of the detected VSYNCS to improve vertical lock.
- HSYNC processor. The HSYNC processor is designed to filter incoming HSYNCS that have been corrupted by noise, providing much improved performance for video signals with a stable time base but poor SNR.

## VBI DATA RECOVERY

The **ADV7180** can retrieve the following information from the input video:

- Wide screen signaling (WSS)
- Copy generation management system (CGMS)
- Closed captioning (CCAP)
- Macrovision protection presence
- EDTV data
- Gemstar-compatible data slicing
- Teletext
- VITC/VPS

The **ADV7180** is also capable of automatically detecting the incoming video standard with respect to

- Color subcarrier frequency
- Field rate
- Line rate

The **ADV7180** can configure itself to support PAL B/D/I/G/H, PAL M, PAL N, PAL Combination N, NTSC M, NTSC J, SECAM 50 Hz/60 Hz, NTSC 4.43, and PAL 60.

## GENERAL SETUP

### Video Standard Selection

The VID\_SEL[3:0] bits (Address 0x00[7:4]) allow the user to force the digital core into a specific video standard. Under normal circumstances, this is not necessary. The VID\_SEL[3:0] bits default to an autodetection mode that supports PAL, NTSC, SECAM, and variants thereof.

### Autodetection of SD Modes

To guide the autodetect system of the **ADV7180**, individual enable bits are provided for each of the supported video standards. Setting the relevant bit to 0 inhibits the standard from being detected automatically. Instead, the system chooses the closest of the remaining enabled standards. The results of the autodetection block can be read back via the status registers (see the Global Status Register section for more information).

### VID\_SEL[3:0], Address 0x00[7:4]

**Table 24. VID\_SEL Function**

VID_SEL[3:0]	Description
0000 (default)	Autodetect (PAL B/G/H/I/D), NTSC J (no pedestal), SECAM
0001	Autodetect (PAL B/G/H/I/D), NTSC M (pedestal), SECAM
0010	Autodetect (PAL N) (pedestal), NTSC J (no pedestal), SECAM
0011	Autodetect (PAL N) (pedestal), NTSC M (pedestal), SECAM
0100	NTSC J
0101	NTSC M
0110	PAL 60
0111	NTSC 4.43
1000	PAL B/G/H/I/D
1001	PAL N = PAL B/G/H/I/D (with pedestal)
1010	PAL M (without pedestal)
1011	PAL M
1100	PAL Combination N
1101	PAL Combination N (with pedestal)
1110	SECAM
1111	SECAM (with pedestal)

### AD\_SEC525\_EN, Enable Autodetection of SECAM 525 Line Video, Address 0x07[7]

Setting AD\_SEC525\_EN to 0 (default) disables the autodetection of a 525-line system with a SECAM style, FM-modulated color component.

Setting AD\_SEC525\_EN to 1 enables the detection of a SECAM style, FM-modulated color component.







### DEF\_VAL\_AUTO\_EN, Default Value Automatic Enable, Address 0x0C[1]

This bit enables the automatic use of the default values for Y, Cr, and Cb when the ADV7180 cannot lock to the video signal.

Setting DEF\_VAL\_AUTO\_EN to 0 disables free-run mode. If the decoder is unlocked, it outputs noise.

Setting DEF\_VAL\_EN to 1 (default) enables free-run mode, and a colored screen set by user-programmable Y, Cr, and Cb values is displayed when the decoder loses lock.

### CLAMP OPERATION

The input video is ac-coupled into the ADV7180. Therefore, its dc value needs to be restored. This process is referred to as clamping the video. This section explains the general process of clamping on the ADV7180 and shows the different ways in which a user can configure its behavior.

The ADV7180 uses a combination of current sources and a digital processing block for clamping, as shown in Figure 22. The analog processing channel shown is replicated three times inside the IC. While only one single channel is needed for a CVBS signal, two independent channels are needed for Y/C (SVHS) type signals, and three independent channels are needed to allow component signals (YPrPb) to be processed.

The clamping can be divided into two sections:

- Clamping before the ADC (analog domain): current sources.
- Clamping after the ADC (digital domain): digital processing block.

The ADC can digitize an input signal only if it resides within the ADC 1.0 V input voltage range. An input signal with a dc level that is too large or too small is clipped at the top or bottom of the ADC range.

The primary task of the analog clamping circuits is to ensure that the video signal stays within the valid ADC input window so that the analog-to-digital conversion can take place. It is not necessary to clamp the input signal with a very high accuracy in the analog domain as long as the video signal fits within the ADC range.

After digitization, the digital fine clamp block corrects for any remaining variations in dc level. Because the dc level of an input video signal refers directly to the brightness of the picture transmitted, it is important to perform a fine clamp with high accuracy; otherwise, brightness variations may occur. Furthermore, dynamic changes in the dc level almost certainly lead to visually objectionable artifacts and must, therefore, be prohibited.

The clamping scheme has to complete two tasks. It must acquire a newly connected video signal with a completely unknown dc level, and it must maintain the dc level during normal operation.

To acquire an unknown video signal quickly, the large current clamps must be activated. It is assumed that the amplitude of the video signal at this point is of a nominal value. Control of the coarse and fine current clamp parameters is performed automatically by the decoder.

Standard definition video signals may have excessive noise on them. In particular, CVBS signals transmitted by terrestrial broadcast and demodulated using a tuner usually show very large levels of noise (>100 mV). A voltage clamp is unsuitable for this type of video signal. Instead, the ADV7180 employs a set of four current sources that can cause coarse (>0.5 mA) and fine (<0.1 mA) currents to flow into and away from the high impedance node that carries the video signal (see Figure 22).

The following sections describe the I<sup>2</sup>C signals that can be used to influence the behavior of the clamping block.

### CCLEN, Current Clamp Enable, Address 0x14[4]

The current clamp enable bit allows the user to switch off the current sources in the analog front end altogether. This may be useful if the incoming analog video signal is clamped externally.

When CCLEN is 0, the current sources are switched off.

When CCLEN is 1 (default), the current sources are enabled.

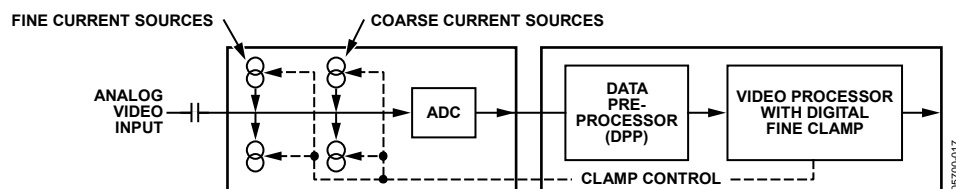


Figure 22. Clamping Overview

**DCT[1:0], Digital Clamp Timing, Address 0x15[6:5]**

The clamp timing register determines the time constant of the digital fine clamp circuitry. It is important to note that the digital fine clamp reacts quickly because it immediately corrects any residual dc level error for the active line. The time constant from the digital fine clamp must be much quicker than the one from the analog blocks.

By default, the time constant of the digital fine clamp is adjusted dynamically to suit the currently connected input signal.

**Table 34. DCT Function**

DCT[1:0]	Description
00 (default)	Slow (TC = 1 sec)
01	Medium (TC = 0.5 sec)
10	Fast (TC = 0.1 sec)
11	Determined by <a href="#">ADV7180</a> , depending on the input video parameters

**DCFE, Digital Clamp Freeze Enable, Address 0x15[4]**

This register bit allows the user to freeze the digital clamp loop at any time. It is intended for users who want to do their own clamping. To do this, disable the current sources for analog clamping via the appropriate register bits, wait until the digital clamp loop settles, and then freeze it via the DCFE bit.

When DCFE is 0 (default), the digital clamp is operational.

When DCFE is 1, the digital clamp loop is frozen.

**LUMA FILTER**

Data from the digital fine clamp block is processed by the three sets of filters that follow. Note that the data format at this point is CVBS for CVBS input or luma only for Y/C and YPrPb input formats.

- Luma antialias filter (YAA). The [ADV7180](#) receives video at a rate of 28.6363 MHz. (In the case of 4× oversampled video, the ADC samples at 57.27 MHz, and the first decimation is performed inside the DPP filters. Therefore, the data rate into the [ADV7180](#) is always 28.6363 MHz.) The ITU-R BT.601 recommends a sampling frequency of 13.5 MHz. The luma antialias filter decimates the oversampled video using a high quality linear phase, low-pass filter that preserves the luma signal while at the same time attenuating out-of-band components. The luma antialias filter (YAA) has a fixed response.

- Luma shaping filters (YSH). The shaping filter block is a programmable low-pass filter with a wide variety of responses. It can be used to selectively reduce the luma video signal bandwidth (needed prior to scaling, for example). For some video sources that contain high frequency noise, reducing the bandwidth of the luma signal improves visual picture quality. A follow-on video compression stage may work more efficiently if the video is low-pass filtered. The [ADV7180](#) has two responses for the shaping filter: one that is used for good quality composite, component, and SVHS type sources, and a second for nonstandard CVBS signals. The YSH filter responses also include a set of notches for PAL and NTSC. However, using the comb filters for Y/C separation is recommended.
- Digital resampling filter. This block allows dynamic resampling of the video signal to alter parameters such as the time base of a line of video. Fundamentally, the resampler is a set of low-pass filters. The actual response is chosen by the system with no requirement for user intervention.

Figure 24 through Figure 27 show the overall response of all filters together. Unless otherwise noted, the filters are set into a typical wideband mode.

**Y Shaping Filter**

For input signals in CVBS format, the luma shaping filters play an essential role in removing the chroma component from a composite signal. Y/C separation must aim for best possible crosstalk reduction while still retaining as much bandwidth (especially on the luma component) as possible. High quality Y/C separation can be achieved by using the internal comb filters of the [ADV7180](#). Comb filtering, however, relies on the frequency relationship of the luma component (multiples of the video line rate) and the color subcarrier (fSC). For good quality CVBS signals, this relationship is known; the comb filter algorithms can be used to separate luma and chroma with high accuracy.

In the case of nonstandard video signals, the frequency relationship may be disturbed, and the comb filters may not be able to remove all crosstalk artifacts in the best fashion without the assistance of the shaping filter block.

An automatic mode is provided that allows the ADV7180 to evaluate the quality of the incoming video signal and select the filter responses in accordance with the signal quality and video standard. YFSM, WYSFMOVR, and WYSFM allow the user to manually override the automatic decisions in part or in full.

The luma shaping filter has three control registers.

- YFSM[4:0] allows the user to manually select a shaping filter mode (applied to all video signals) or to enable an automatic selection (depending on video quality and video standard).
- WYSFMOVR allows the user to manually override the WYSFM decision.
- WYSFM[4:0] allows the user to select a different shaping filter mode for good quality composite (CVBS), component (YPrPb), and SVHS (Y/C) input signals.

In automatic mode, the system preserves the maximum possible bandwidth for good CVBS sources (because they can be successfully combed) as well as for luma components of YPrPb and Y/C sources (because they need not be combed). For poor quality signals, the system selects from a set of proprietary shaping filter responses that complements comb filter operation to reduce visual artifacts.

The decisions of the control logic are shown in Figure 23.

**YFSM[4:0], Y Shaping Filter Mode, Address 0x17[4:0]**

The Y shaping filter mode bits allow the user to select from a wide range of low-pass and notch filters. When switched in automatic mode, the filter selection is based on other register selections, such as detected video standard, as well as properties extracted from the incoming video itself, such as quality and time base stability. The automatic selection always selects the widest possible bandwidth for the video input encountered.

The Y-shaping filter mode operates as follows:

- If the YFSM settings specify a filter (that is, YFSM is set to values other than 00000 or 00001), the chosen filter is applied to all video, regardless of its quality.
- In automatic selection mode, the notch filters are only used for bad quality video signals. For all other video signals, wideband filters are used.

**WYSFMOVR, Wideband Y Shaping Filter Override, Address 0x18[7]**

Setting the WYSFMOVR bit enables the use of the WYSFM[4:0] settings for good quality video signals. For more information on luma shaping filters, see the Y Shaping Filter section and the flowchart shown in Figure 23.

When WYSFMOVR is 0, the shaping filter for good quality video signals is selected automatically.

Setting WYSFMOVR to 1 (default) enables manual override via WYSFM[4:0].

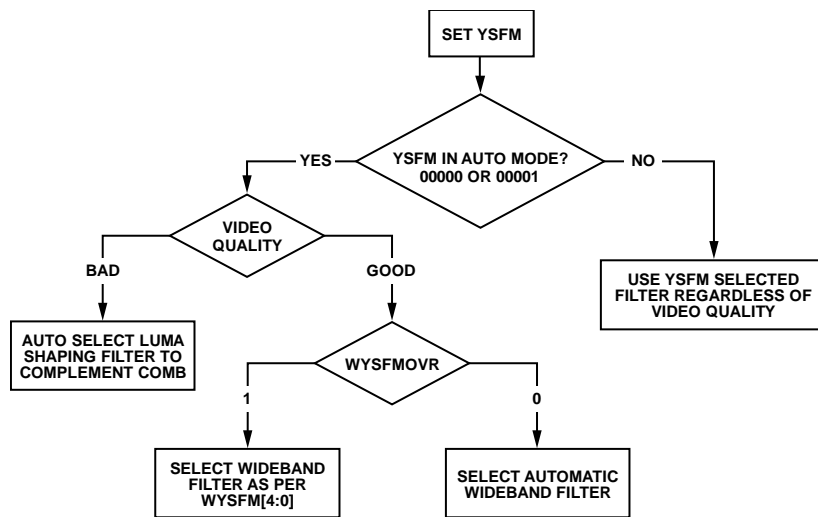


Figure 23. YFSM and WYSFM Control Flowchart

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**CHROMA FILTER**

Data from the digital fine clamp block is processed by the three sets of filters that follow. Note that the data format at this point is CVBS for CVBS inputs, chroma only for Y/C, or U/V interleaved for YPrPb input formats.

- Chroma antialias filter (CAA). The ADV7180 oversamples the CVBS by a factor of 4 and the chroma/YPrPb by a factor of 2. A decimating filter (CAA) is used to preserve the active video band and to remove any out-of-band components. The CAA filter has a fixed response.

- Chroma shaping filters (CSH). The shaping filter block (CSH) can be programmed to perform a variety of low-pass responses. It can be used to selectively reduce the bandwidth of the chroma signal for scaling or compression.
- Digital resampling filter. This block allows dynamic resampling of the video signal to alter parameters such as the time base of a line of video. Fundamentally, the resampler is a set of low-pass filters. The actual response is chosen by the system without user intervention.

Figure 28 shows the overall response of all filters together.

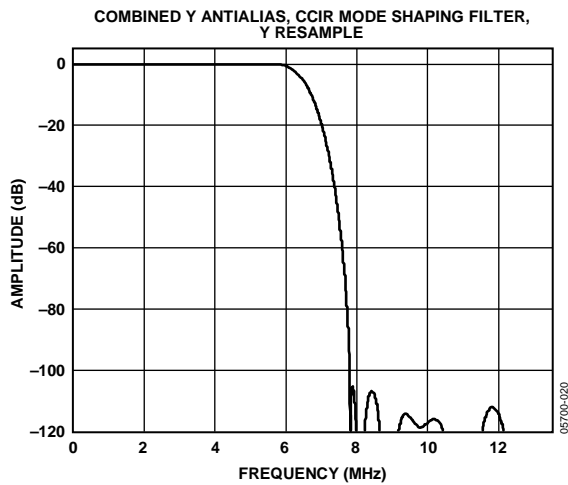


Figure 25. Combined Y Antialias, CCIR Mode Shaping Filter

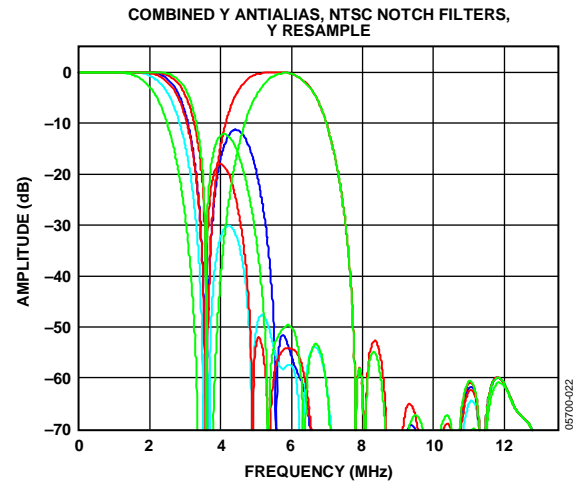


Figure 27. Combined Y Antialias Filter, NTSC Notch Filters

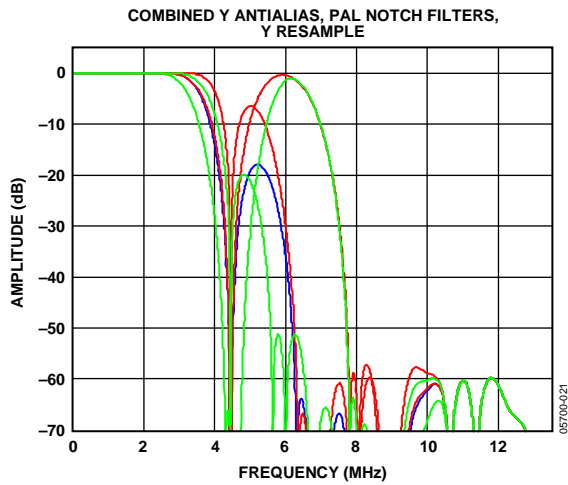


Figure 26. Combined Y Antialias, PAL Notch Filters

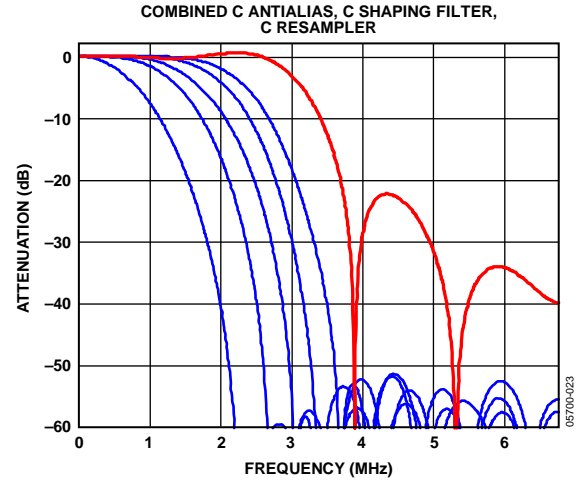


Figure 28. Chroma Shaping Filter Responses



**Luma Gain**

**LAGC[2:0], Luma Automatic Gain Control,  
Address 0x2C[6:4]**

The luma automatic gain control mode bits select the operating mode for the gain control in the luma path.

**Table 39. LAGC Function**

LAGC[2:0]	Description
000	Manual fixed gain (use LMG[11:0])
001	AGC (blank level to sync tip), peak white algorithm off
010 (default)	AGC (blank level to sync tip), peak white algorithm on
011	Reserved
100	Reserved
101	Reserved
110	Reserved
111	Freeze gain

**LAGT[1:0], Luma Automatic Gain Timing,  
Address 0x2F[7:6]**

The luma automatic gain timing register allows the user to influence the tracking speed of the luminance automatic gain control. This register only has an effect if the LAGC[2:0] register is set to 001 or 010 (automatic gain control modes).

If peak white AGC is enabled and active (see the Status 1[7:0], Address 0x10[7:0] section), the actual gain update speed is dictated by the peak white AGC loop and, as a result, the LAGT settings have no effect. As soon as the part leaves peak white AGC, LAGT becomes relevant again.

**Table 40. LAGT Function**

LAGT[1:0]	Description
00	Slow (TC = 2 sec)
01	Medium (TC = 1 sec)
10	Fast (TC = 0.2 sec)
11 (default)	Adaptive

**LG[11:0], Luma Gain, Address 0x2F[3:0],  
Address 0x30[7:0]**

**LMG[11:0], Luma Manual Gain, Address 0x2F[3:0],  
Address 0x30[7:0]**

Luma gain[11:0] is a dual-function register. If all of these registers are written to, a desired manual luma gain can be programmed. This gain becomes active if the LAGC[2:0] mode is switched to manual fixed gain. Equation 1 shows how to calculate a desired gain.

If read back, this register returns the current gain value. Depending on the setting in the LAGC[2:0] bits, the value is one of the following:

- Luma manual gain value (LAGC[2:0] set to luma manual gain mode)
- Luma automatic gain value (LAGC[2:0] set to any of the automatic modes)

**Table 41. LG/LMG Function**

LG[11:0]/LMG[11:0]	Read/Write	Description
LMG[11:0] = x	Write	Manual gain for luma path
LG[11:0] = x	Read	Actual used gain

$$Luma\ Gain = \frac{LMG[11:0]}{LumaCalibrationFactor} \quad (1)$$

where  $LMG[11:0]$  is a decimal value between 1024 and 4095.

**Calculation of the Luma Calibration Factor**

1. Using a video source, set content to a grey field and apply as a standard CVBS signal to the CVBS input of the board.
2. Using an oscilloscope, measure the signal at CVBS input to ensure that its sync depth, color burst, and luma are at the standard levels.
3. Connect the output parallel pixel bus of the [ADV7180](#) to a backend system that has unity gain and monitor output voltage.
4. Measure the luma level correctly from the black level. Turn off the Luma AGC and manually change the value of the luma gain control register, LMG[11:0], until the output luma level matches the input measured in Step 2.

This value, in decimal, is the luma calibration factor.

**BETACAM, Enable Betacam Levels, Address 0x01[5]**

If YPrPb data is routed through the [ADV7180](#), the automatic gain control modes can target different video input levels, as outlined in Table 44. The BETACAM bit is valid only if the input mode is YPrPb (component). The BETACAM bit sets the target value for AGC operation.

A review of the following sections is useful:

- The MAN\_MUX\_EN, Manual Input Muxing Enable, Address 0xC4[7] section for how component video (YPrPb) can be routed through the [ADV7180](#).
- The Video Standard Selection section to select the various standards, for example, with and without pedestal.

The AGC algorithms adjust the levels based on the setting of the BETACAM bit (see Table 42).

**Table 42. BETACAM Function**

BETACAM	Description
0 (default)	Assuming YPrPb is selected as input format: Selecting PAL with pedestal selects MII. Selecting PAL without pedestal selects SMPTE. Selecting NTSC with pedestal selects MII. Selecting NTSC without pedestal selects SMPTE.
1	Assuming YPrPb is selected as input format: Selecting PAL with pedestal selects BETACAM. Selecting PAL without pedestal selects BETACAM variant. Selecting NTSC with pedestal selects BETACAM. Selecting NTSC without pedestal selects BETACAM variant.

**Table 43. CAGC Function**

CAGC[1:0]	Description
00	Manual fixed gain (use CMG[11:0])
01	Luma gain used for chroma
10 (default)	Automatic gain (based on color burst)
11	Freeze chroma gain

**Table 44. BETACAM Levels**

Name	BETACAM (mV)	BETACAM Variant (mV)	SMPTE (mV)	III (mV)
Y	0 to +714 (including 7.5% pedestal)	0 to +714	0 to +700	0 to +700 (including 7.5% pedestal)
Pb and Pr	-467 to +467	-505 to +505	-350 to +350	-324 to +324
Sync Depth	+286	+286	+300	+300

**PW\_UPD, Peak White Update, Address 0x2B[0]**

The peak white and average video algorithms determine the gain based on measurements taken from the active video. The PW\_UPD bit determines the rate of gain change. LAGC[2:0] must be set to the appropriate mode to enable the peak white or average video mode in the first place. For more information, see the LAGC[2:0], Luma Automatic Gain Control, Address 0x2C[6:4] section.

Setting PW\_UPD to 0 updates the gain once per video line.

Setting PW\_UPD to 1 (default) updates the gain once per field.

**Chroma Gain****CAGC[1:0], Chroma Automatic Gain Control, Address 0x2C[1:0]**

The two bits of color automatic gain control mode select the basic mode of operation for automatic gain control in the chroma path.



**CHROMA TRANSIENT IMPROVEMENT (CTI)**

The signal bandwidth allocated for chroma is typically much smaller than that for luminance. In the past, this was a valid way to fit a color video signal into a given overall bandwidth because the human eye is less sensitive to chrominance than to luminance.

The uneven bandwidth, however, may lead to visual artifacts in sharp color transitions. At the border of two bars of color, both components (luma and chroma) change at the same time (see Figure 31). Due to the higher bandwidth, the signal transition of the luma component is usually much sharper than that of the chroma component. The color edge is not sharp and can be blurred, in the worst case, over several pixels.

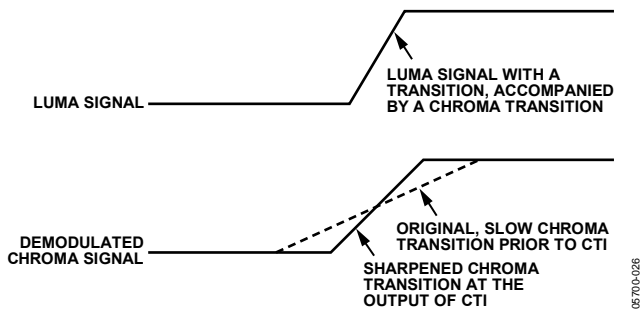


Figure 31. CTI Luma/Chroma Transition

The chroma transient improvement block examines the input video data. It detects transitions of chroma and can be programmed to create steeper chroma edges in an attempt to artificially restore lost color bandwidth. The CTI block, however, operates only on edges above a certain threshold to ensure that noise is not emphasized. Care has also been taken to ensure that edge ringing and undesirable saturation or hue distortion are avoided.

Chroma transient improvements are needed primarily for signals that have severe chroma bandwidth limitations. For those types of signals, it is strongly recommended to enable the CTI block via CTI\_EN.

**CTI\_EN, Chroma Transient Improvement Enable, Address 0x4D[0]**

Setting CTI\_EN to 0 disables the CTI block.

Setting CTI\_EN to 1 (default) enables the CTI block.

**CTI\_AB\_EN, Chroma Transient Improvement Alpha Blend Enable, Address 0x4D[1]**

The CTI\_AB\_EN bit enables an alpha blend function within the CTI block. If set to 1, the alpha blender mixes the transient improved chroma with the original signal. The sharpness of the alpha blending can be configured via the CTI\_AB[1:0] bits.

For the alpha blender to be active, the CTI block must be enabled via the CTI\_EN bit.

Setting CTI\_AB\_EN to 0 disables the CTI alpha blender.

Setting CTI\_AB\_EN to 1 (default) enables the CTI alpha-blend mixing function.

**CTI\_AB[1:0], Chroma Transient Improvement Alpha Blend, Address 0x4D[3:2]**

The CTI\_AB[1:0] controls the behavior of alpha blend circuitry that mixes the sharpened chroma signal with the original one. It thereby controls the visual impact of CTI on the output data.

For CTI\_AB[1:0] to become active, the CTI block must be enabled via the CTI\_EN bit, and the alpha blender must be switched on via CTI\_AB\_EN.

Sharp blending maximizes the effect of CTI on the picture but may also increase the visual impact of small amplitude, high frequency chroma noise.

Table 48. CTI\_AB Function

CTI_AB[1:0]	Description
00	Sharpest mixing between sharpened and original chroma signal
01	Sharp mixing
10	Smooth mixing
11 (default)	Smoothest alpha blend function

**CTI\_C\_TH[7:0], CTI Chroma Threshold, Address 0x4E[7:0]**

The CTI\_C\_TH[7:0] value is an unsigned, 8-bit number specifying how big the amplitude step in a chroma transition must be to be steepened by the CTI block. Programming a small value into this register causes even smaller edges to be steepened by the CTI block. Making CTI\_C\_TH[7:0] a large value causes the block to improve large transitions only.

The default value for CTI\_C\_TH[7:0] is 0x08, indicating the threshold for the chroma edges prior to CTI.

**DIGITAL NOISE REDUCTION (DNR) AND LUMA PEAKING FILTER**

Digital noise reduction is based on the assumption that high frequency signals with low amplitude are probably noise and that, therefore, their removal improves picture quality. The following are the two DNR blocks in the ADV7180: the DNR1 block before the luma peaking filter and the DNR2 block after the luma peaking filter, as shown in Figure 32.

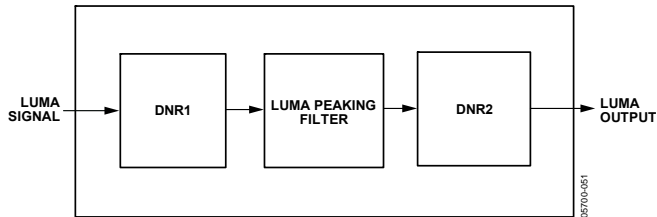


Figure 32. DNR and Peaking Block Diagram

**DNR\_EN, Digital Noise Reduction Enable, Address 0x4D[5]**

The DNR\_EN bit enables the DNR block or bypasses it.

Table 49. DNR\_EN Function

Setting	Description
0	Bypasses DNR (disable)
1 (default)	Enables digital noise reduction on the luma data

**DNR\_TH[7:0], DNR Noise Threshold, Address 0x50[7:0]**

The DNR1 block is positioned before the luma peaking block. The DNR\_TH[7:0] value is an unsigned, 8-bit number used to determine the maximum edge that is interpreted as noise and, therefore, blanked from the luma data. Programming a large value into DNR\_TH[7:0] causes the DNR block to interpret even large transients as noise and remove them. As a result, the effect on the video data is more visible. Programming a small value causes only small transients to be seen as noise and to be removed.

Table 50. DNR\_TH[7:0] Function

Setting	Description
0x08 (default)	Threshold for maximum luma edges to be interpreted as noise

**PEAKING\_GAIN[7:0], Luma Peaking Gain, Address 0xFB[7:0]**

This filter can be manually enabled. The user can select to boost or to attenuate the mid region of the Y spectrum around 3 MHz. The peaking filter can visually improve the picture by showing more definition on the picture details that contain frequency components around 3 MHz. The default value on this register passes through the luma data unaltered. A lower value attenuates the signal, and a higher value gains the luma signal. A plot of the responses of the filter is shown in Figure 33.

Table 51. PEAKING\_GAIN[7:0] Function

Setting	Description
0x40 (Default)	0 dB response

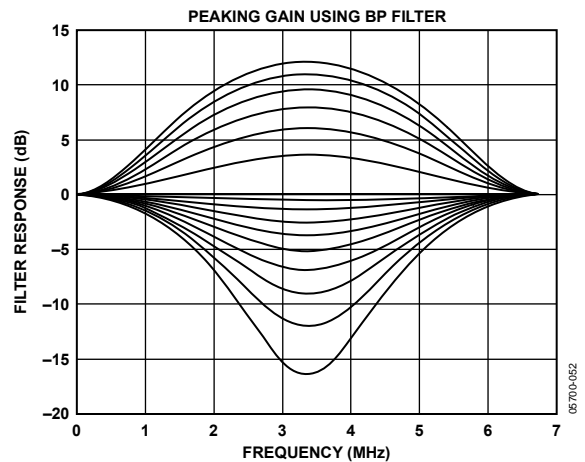


Figure 33. Peaking Filter Responses

**DNR\_TH2[7:0], DNR Noise Threshold 2, Address 0xFC[7:0]**

The DNR2 block is positioned after the luma peaking block and, therefore, affects the gained luma signal. It operates in the same way as the DNR1 block, but there is an independent threshold control, DNR\_TH2[7:0], for this block. This value is an unsigned, 8-bit number used to determine the maximum edge that is interpreted as noise and, therefore, blanked from the luma data. Programming a large value into DNR\_TH2[7:0] causes the DNR block to interpret even large transients as noise and remove them. As a result, the effect on the video data is more visible. Programming a small value causes only small transients to be seen as noise and to be removed.

Table 52. DNR\_TH2[7:0] Function

Setting	Description
0x04 (default)	Threshold for maximum luma edges to be interpreted as noise

**COMB FILTERS**

The comb filters of the ADV7180 have been greatly improved to automatically handle video of all types, standards, and levels of quality. The NTSC and PAL configuration registers allow the user to customize the comb filter operation depending on which video standard is detected (by autodetection) or selected (by manual programming).

**NTSC Comb Filter Settings**

These settings are used for NTSC M/J CVBS inputs.

**NSFSEL[1:0], Split Filter Selection NTSC, Address 0x19[3:2]**

The NSFSEL[1:0] control selects how much of the overall signal bandwidth is fed to the combs. A narrow split filter selection results in better performance on diagonal lines but more dot crawl in the final output image. The opposite is true for selecting a wide bandwidth split filter.

**Table 53. NSFSEL Function**

NSFSEL[1:0]	Description
00 (default)	Narrow
01	Medium
10	Medium
11	Wide

**CTAPSN[1:0], Chroma Comb Taps, NTSC, Address 0x38[7:6]****Table 54. CTAPSN Function**

CTAPSN[1:0]	Description
00	Do not use
01	NTSC chroma comb adapts three lines (three taps) to two lines (two taps)
10 (default)	NTSC chroma comb adapts five lines (five taps) to three lines (three taps)
11	NTSC chroma comb adapts five lines (five taps) to four lines (four taps)

**CCMN[2:0], Chroma Comb Mode, NTSC, Address 0x38[5:3]****Table 55. CCMN Function**

CCMN[2:0]	Description	Configuration
000 (default)	Adaptive comb mode	Adaptive three-line chroma comb for CTAPSN = 01 Adaptive four-line chroma comb for CTAPSN = 10 Adaptive five-line chroma comb for CTAPSN = 11
100	Disable chroma comb	
101	Fixed chroma comb (top lines of line memory)	Fixed two-line chroma comb for CTAPSN = 01 Fixed three-line chroma comb for CTAPSN = 10 Fixed four-line chroma comb for CTAPSN = 11
110	Fixed chroma comb (all lines of line memory)	Fixed three-line chroma comb for CTAPSN = 01 Fixed four-line chroma comb for CTAPSN = 10 Fixed five-line chroma comb for CTAPSN = 11
111	Fixed chroma comb (bottom lines of line memory)	Fixed two-line chroma comb for CTAPSN = 01 Fixed three-line chroma comb for CTAPSN = 10 Fixed four-line chroma comb for CTAPSN = 11

## YCMN[2:0], Luma Comb Mode NTSC, Address 0x38[2:0]

Table 56. YCMN Function

YCMN[2:0]	Description	Configuration
000 (default)	Adaptive comb mode	Adaptive three-line (three taps) luma comb
100	Disable luma comb	Use low-pass/notch filter; see the Y Shaping Filter section
101	Fixed luma comb (top lines of line memory)	Fixed two-line (two taps) luma comb
110	Fixed luma comb (all lines of line memory)	Fixed three-line (three taps) luma comb
111	Fixed luma comb (bottom lines of line memory)	Fixed two-line (two taps) luma comb

**PAL Comb Filter Settings**

These settings are used for PAL B/G/H/I/D, PAL M, PAL Combinational N, PAL 60, and NTSC 4.43 CVBS inputs.

**PSFSEL[1:0], Split Filter Selection, PAL, Address 0x19[1:0]**

The PSFSEL[1:0] control selects how much of the overall signal bandwidth is fed to the combs. A wide split filter selection eliminates dot crawl but shows imperfections on diagonal lines. The opposite is true for selecting a narrow bandwidth split filter.

Table 57. PSFSEL Function

PSFSEL[1:0]	Description
00	Narrow
01 (default)	Medium
10	Wide
11	Widest

## CTAPSP[1:0], Chroma Comb Taps PAL, Address 0x39[7:6]

Table 58. CTAPSP Function

CTAPSP[1:0]	Description
00	Do not use.
01	PAL chroma comb adapts five lines (three taps) to three lines (two taps); cancels cross luma only
10	PAL chroma comb adapts five lines (five taps) to three lines (three taps); cancels cross luma and hue error less well
11 (default)	PAL chroma comb adapts five lines (five taps) to four lines (four taps); cancels cross luma and hue error well

## CCMP[2:0], Chroma Comb Mode PAL, Address 0x39[5:3]

Table 59. CCMP Function

CCMP[2:0]	Description	Configuration
000 (default)	Adaptive comb mode	Adaptive three-line chroma comb for CTAPSN = 01 Adaptive four-line chroma comb for CTAPSN = 10 Adaptive five-line chroma comb for CTAPSN = 11
100	Disable chroma comb	
101	Fixed chroma comb (top lines of line memory)	Fixed two-line chroma comb for CTAPSN = 01 Fixed three-line chroma comb for CTAPSN = 10 Fixed four-line chroma comb for CTAPSN = 11
110	Fixed chroma comb (all lines of line memory)	Fixed three-line chroma comb for CTAPSN = 01 Fixed four-line chroma comb for CTAPSN = 10 Fixed five-line chroma comb for CTAPSN = 11
111	Fixed chroma comb (bottom lines of line memory)	Fixed two-line chroma comb for CTAPSN = 01 Fixed three-line chroma comb for CTAPSN = 10 Fixed four-line chroma comb for CTAPSN = 11

## YCMP[2:0], Luma Comb Mode PAL, Address 0x39[2:0]

Table 60. YCMP Function

YCMP[2:0]	Description	Configuration
000 (default)	Adaptive comb mode	Adaptive five lines (three taps) luma comb
100	Disable luma comb	Use low-pass/notch filter; see the Y Shaping Filter section
101	Fixed luma comb (top lines of line memory)	Fixed three lines (two taps) luma comb
110	Fixed luma comb (all lines of line memory)	Fixed five lines (three taps) luma comb
111	Fixed luma comb (bottom lines of line memory)	Fixed three lines (two taps) luma comb

**IF FILTER COMPENSATION**

**IFFILTSEL[2:0], IF Filter Select, Address 0xF8[2:0]**

The IFFILTSEL[2:0] register allows the user to compensate for SAW filter characteristics on a composite input, as would be observed on tuner outputs. Figure 34 and Figure 35 show IF filter compensation for NTSC and PAL, respectively.

The options for this feature are as follows:

- Bypass mode
- NTSC, consists of three filter characteristics
- PAL, consists of three filter characteristics

See Table 107 for programming details.

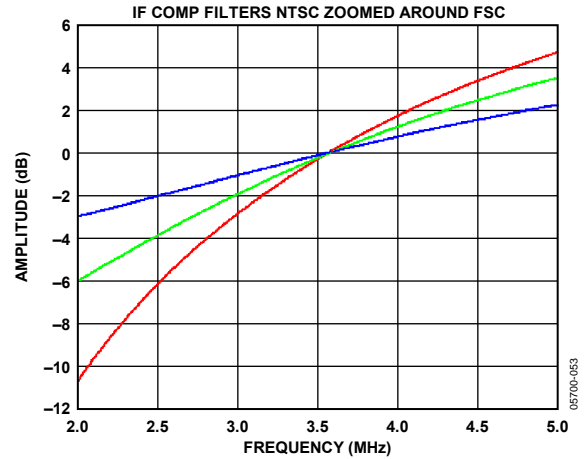


Figure 34. NTSC IF Filter Compensation

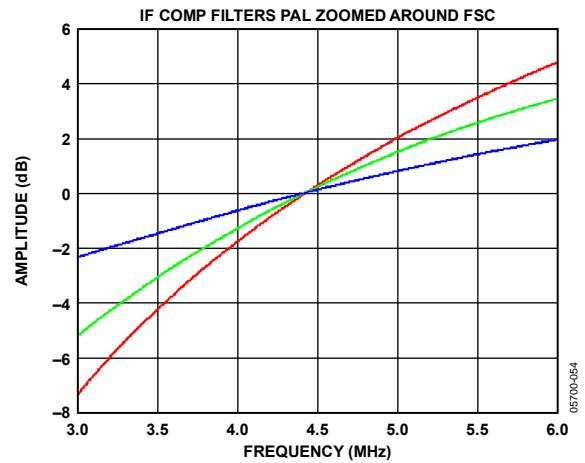


Figure 35. PAL IF Filter Compensation

### AV CODE INSERTION AND CONTROLS

This section describes the I<sup>2</sup>C-based controls that affect the following:

- Insertion of AV codes into the data stream
- Data blanking during the vertical blank interval (VBI)
- The range of data values permitted in the output data stream
- The relative delay of luma vs. chroma signals

Some of the decoded VBI data is inserted during the horizontal blanking interval. See the Gemstar Data Recovery section for more information.

#### BT.656-4, ITU-R BT.656-4 Enable, Address 0x04[7]

Between Revision 3 and Revision 4 of the ITU-R BT.656 standards, the ITU has changed the toggling position for the V bit within the SAV EAV codes for NTSC. The ITU-R BT.656-4 standard bit allows the user to select an output mode that is compliant with either the previous or new standard. For further information, visit the International Telecommunication Union website.

Note that the standard change only affects NTSC and has no bearing on PAL.

When ITU-R BT.656-4 is 0 (default), the ITU-R BT.656-3 specification is used. The V bit goes low at EAV of Line 10 and Line 273.

When ITU-R BT.656-4 is 1, the ITU-R BT.656-4 specification is used. The V bit goes low at EAV of Line 20 and Line 283.

#### SD\_DUP\_AV, Duplicate AV Codes, Address 0x03[0]

Depending on the output interface width, it may be necessary to duplicate the AV codes from the luma path into the chroma path.

In an 8-bit wide output interface (Cb/Y/Cr/Y interleaved data), the AV codes are defined as FF/00/00/AV, with AV being the transmitted word that contains information about H/V/F.

In this output interface mode, the following assignment takes place: Cb = FF, Y = 00, Cr = 00, and Y = AV.

In a 16-bit output interface (64-lead LQFP only), where Y and Cr/Cb are delivered via separate data buses, the AV code is spread over the whole 16 bits. The SD\_DUP\_AV bit allows the user to replicate the AV codes on both buses; therefore, the full AV sequence can be found on the Y bus as well as on the Cr/Cb bus (see Figure 36).

When SD\_DUP\_AV is 0 (default), the AV codes are in single fashion (to suit 8-bit interleaved data output).

When SD\_DUP\_AV is 1, the AV codes are duplicated (for 16-bit interfaces).

#### VBI\_EN, Vertical Blanking Interval Data Enable, Address 0x03[7]

The VBI enable bit allows data such as intercast and closed caption data to be passed through the luma channel of the decoder with a minimal amount of filtering. All data for Line 1 to Line 21 is passed through and available at the output port. The ADV7180 does not blank the luma data and automatically switches all filters along the luma data path into their widest bandwidth. For active video, the filter settings for YSH and YPK are restored.

See the BL\_C\_VBI, Blank Chroma During VBI, Address 0x04[2] section for information on the chroma path.

When VBI\_EN is 0 (default), all video lines are filtered/scaled.

When VBI\_EN is 1, only the active video region is filtered/scaled.

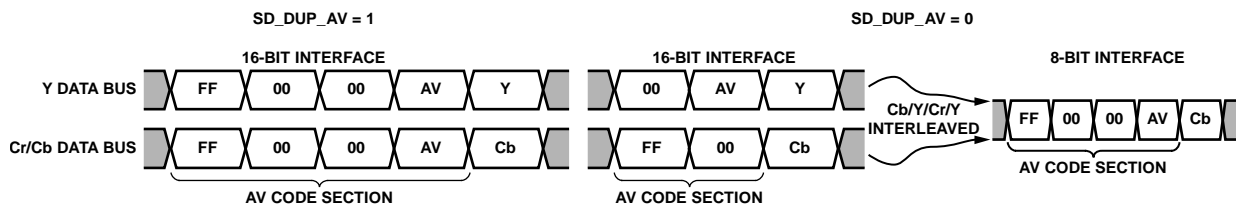


Figure 36. AV Code Duplication Control (64-Lead LQFP Only)

06700-027







**PF, Polarity FIELD, Address 0x37[3]**

The polarity of the FIELD pin for the 64-lead LQFP part can be inverted using the PF bit.

The FIELD pin can be inverted using the PF bit.

When PF is 0 (default), FIELD is active high.

When PF is 1, FIELD is active low.

**Table 65. User Settings for NTSC (See Figure 39)**

Register	Register Name	Write
0x31	VS/FIELD Control 1	0x1A
0x32	VS/FIELD Control 2	0x81
0x33	VS/FIELD Control 3	0x84
0x34	HS Position Control 1	0x00
0x35	HS Position Control 2	0x00
0x36	HS Position Control 3	0x7D
0x37	Polarity	0xA1
0xE5	NTSV V bit begin	0x41
0xE6	NTSC V bit end	0x84
0xE7	NTSC F bit toggle	0x06

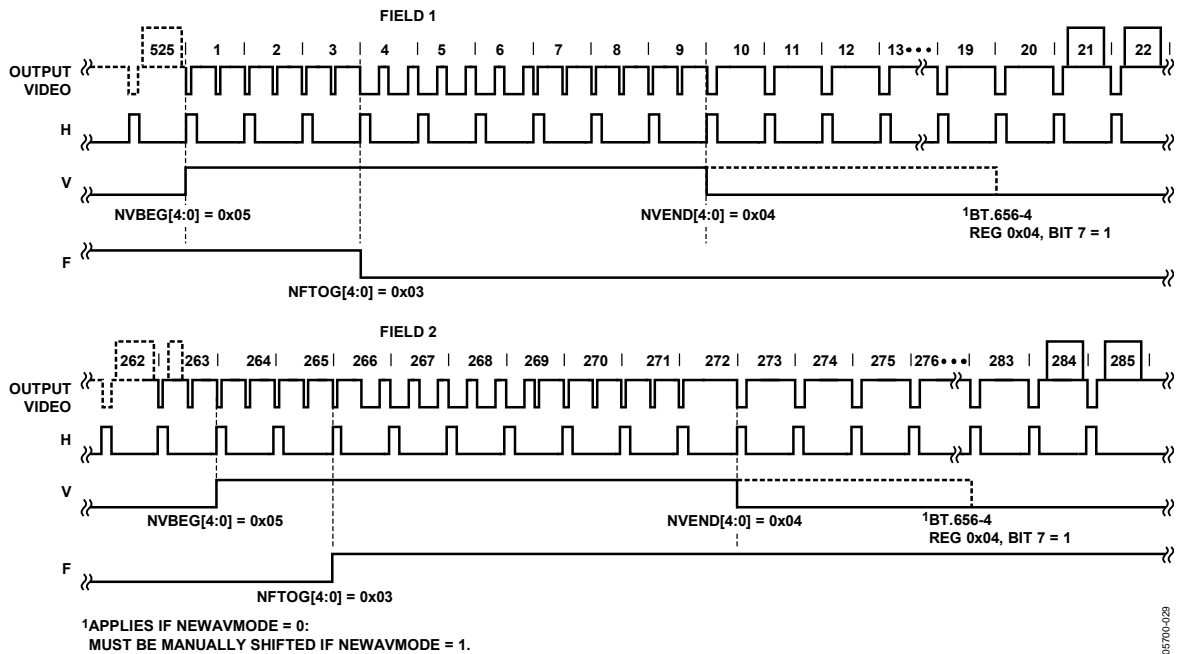


Figure 38. NTSC Default, ITU-R BT.656 (the Polarity of H, V, and F is Embedded in the Data)

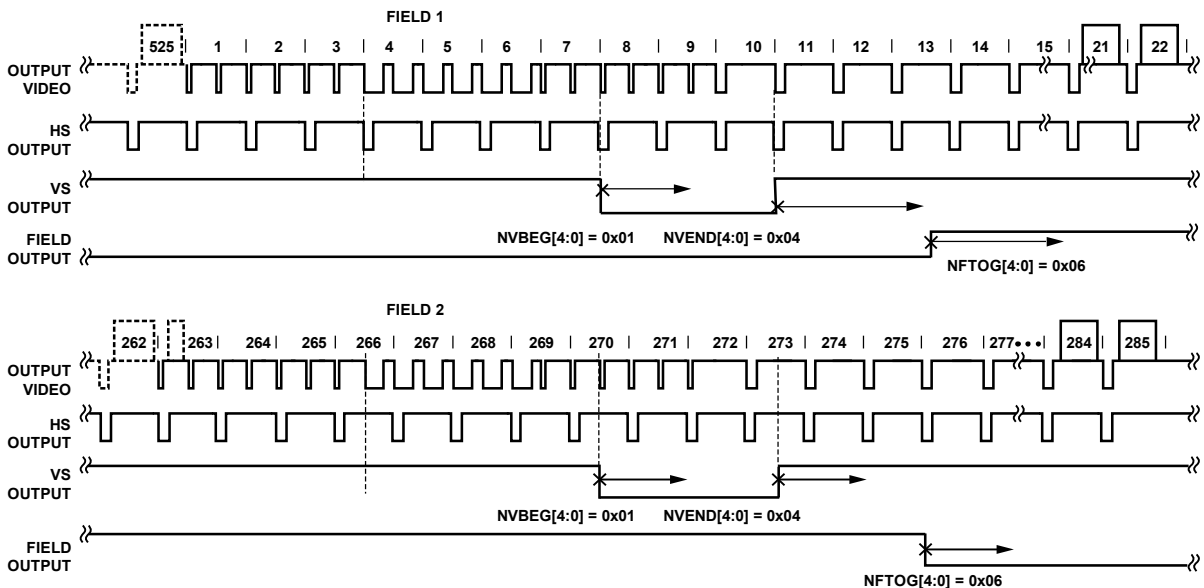


Figure 39. NTSC Typical VSYNC/FIELD Positions Using the Register Writes in Table 65

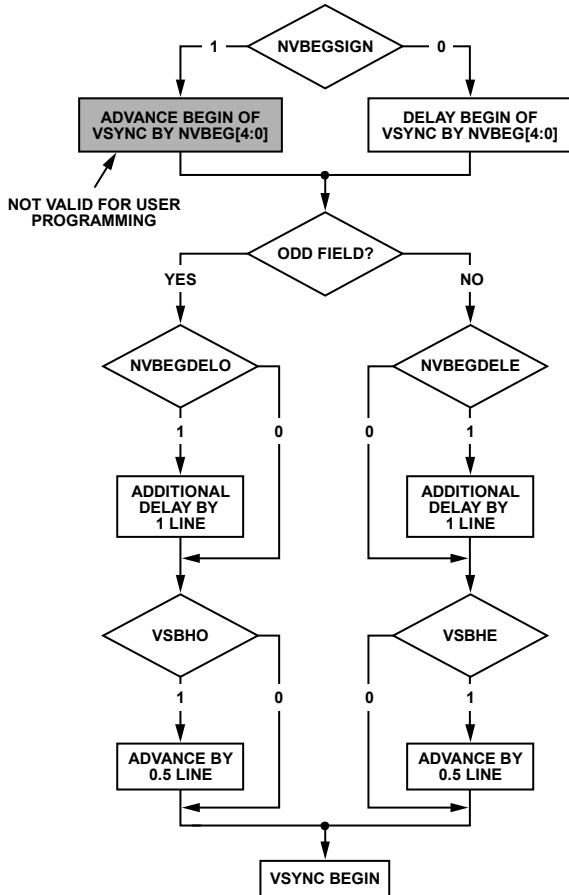


Figure 40. NTSC VSYNC Begin

**NVBEGDELO, NTSC VSYNC Begin Delay on Odd Field, Address 0xE5[7]**

When NVBEGDELO is 0 (default), there is no delay. Setting NVBEGDELO to 1 delays VSYNC going high on an odd field by a line relative to NVBEG.

**NVBEGDELE, NTSC VSYNC Begin Delay on Even Field, Address 0xE5[6]**

When NVBEGDELE is 0 (default), there is no delay. Setting NVBEGDELE to 1 delays VSYNC going high on an even field by a line relative to NVBEG.

**NVBEGSIGN, NTSC VSYNC Begin Sign, Address 0xE5[5]**

Setting NVBEGSIGN to 0 delays the start of VSYNC. Set for user manual programming.

Setting NVBEGSIGN to 1 (default) advances the start of VSYNC (not recommended for user programming).

**NVBEG[4:0], NTSC VSYNC Begin, Address 0xE5[4:0]**

The default value of NVBEG is 00101, indicating the NTSC VSYNC begin position.

For all NTSC/PAL VSYNC timing controls, both the V bit in the AV code and the VSYNC signal on the VS pin are modified.

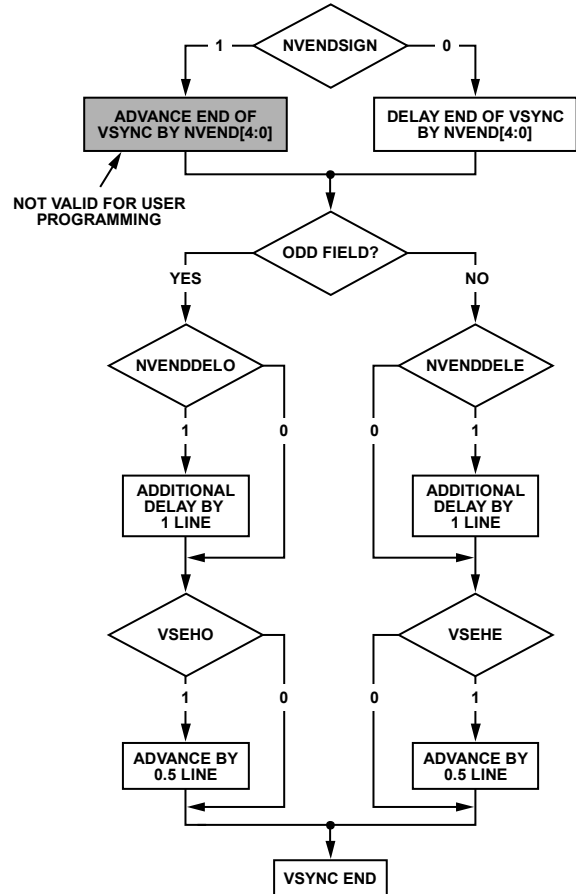


Figure 41. NTSC VSYNC End

**NVENDDELO, NTSC VSYNC End Delay on Odd Field, Address 0xE6[7]**

When NVENDDELO is 0 (default), there is no delay. Setting NVENDDELO to 1 delays VSYNC from going low on an odd field by a line relative to NVEND.

**NVENDDELE, NTSC VSYNC End Delay on Even Field, Address 0xE6[6]**

When NVENDDELE is set to 0 (default), there is no delay. Setting NVENDDELE to 1 delays VSYNC from going low on an even field by a line relative to NVEND.

**NVENDSIGN, NTSC VSYNC End Sign, Address 0xE6[5]**

Setting NVENDSIGN to 0 (default) delays the end of VSYNC. Set for user manual programming.

Setting NVENDSIGN to 1 advances the end of VSYNC (not recommended for user programming).

**NVEND[4:0], NTSC VSYNC End, Address 0xE6[4:0]**

The default value of NVEND is 00100, indicating the NTSC VSYNC end position.

For all NTSC/PAL VSYNC timing controls, both the V bit in the AV code and the VSYNC signal on the VS pin are modified.

**NFTOGDELO, NTSC FIELD Toggle Delay on Odd Field, Address 0xE7[7]**

When NFTOGDELO is 0 (default), there is no delay.

Setting NFTOGDELO to 1 delays the field toggle/transition on an odd field by a line relative to NFTOG.

**NFTOGDELE, NTSC Field Toggle Delay on Even Field, Address 0xE7[6]**

When NFTOGDELE is 0, there is no delay.

Setting NFTOGDELE to 1 (default) delays the field toggle/transition on an even field by a line relative to NFTOG.

**NFTOGSIGN, NTSC Field Toggle Sign, Address 0xE7[5]**

Setting NFTOGSIGN to 0 delays the field transition. Set for user manual programming.

Setting NFTOGSIGN to 1 (default) advances the field transition (not recommended for user programming).

**NFTOG[4:0], NTSC Field Toggle, Address 0xE7[4:0]**

The default value of NFTOG is 00011, indicating the NTSC field toggle position.

For all NTSC/PAL field timing controls, both the F bit in the AV code and the field signal on the FIELD pin are modified.

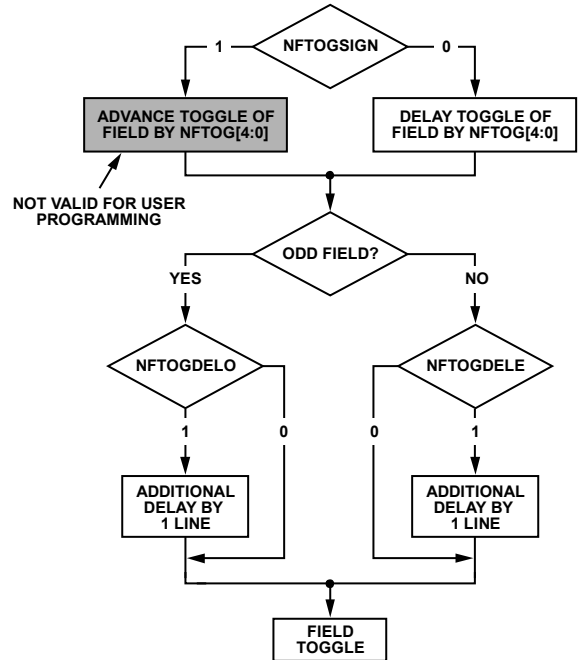


Figure 42. NTSC FIELD Toggle

05700-033

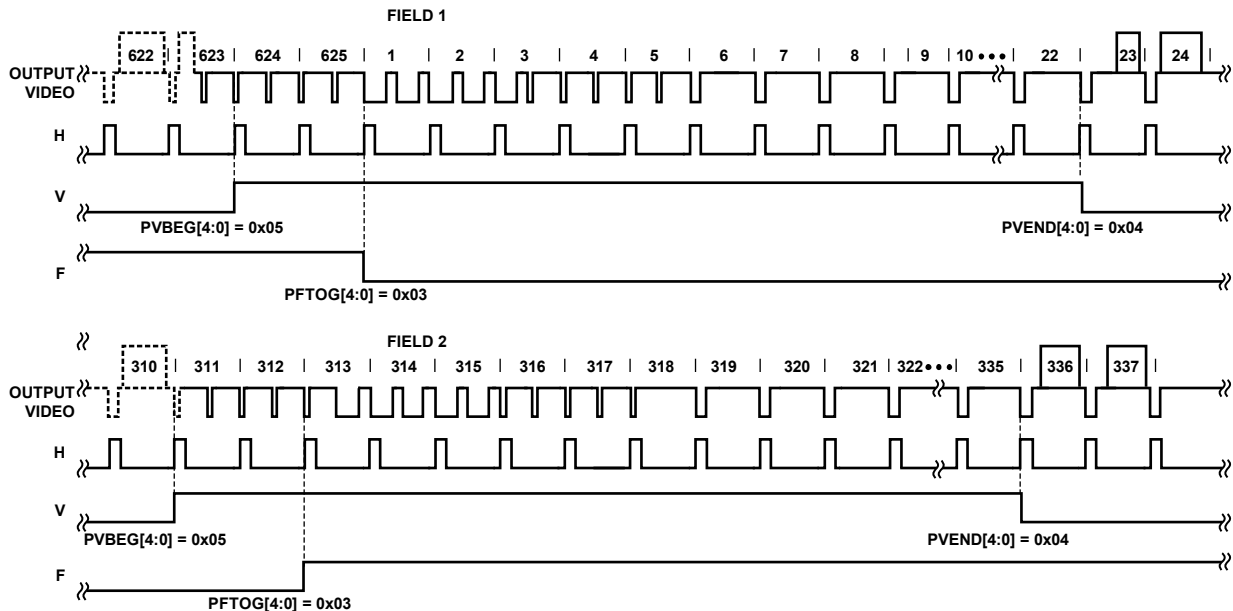


Figure 43. PAL Default, ITU-R BT.656 (the Polarity of H, V, and F Is Embedded in the Data)

05700-034

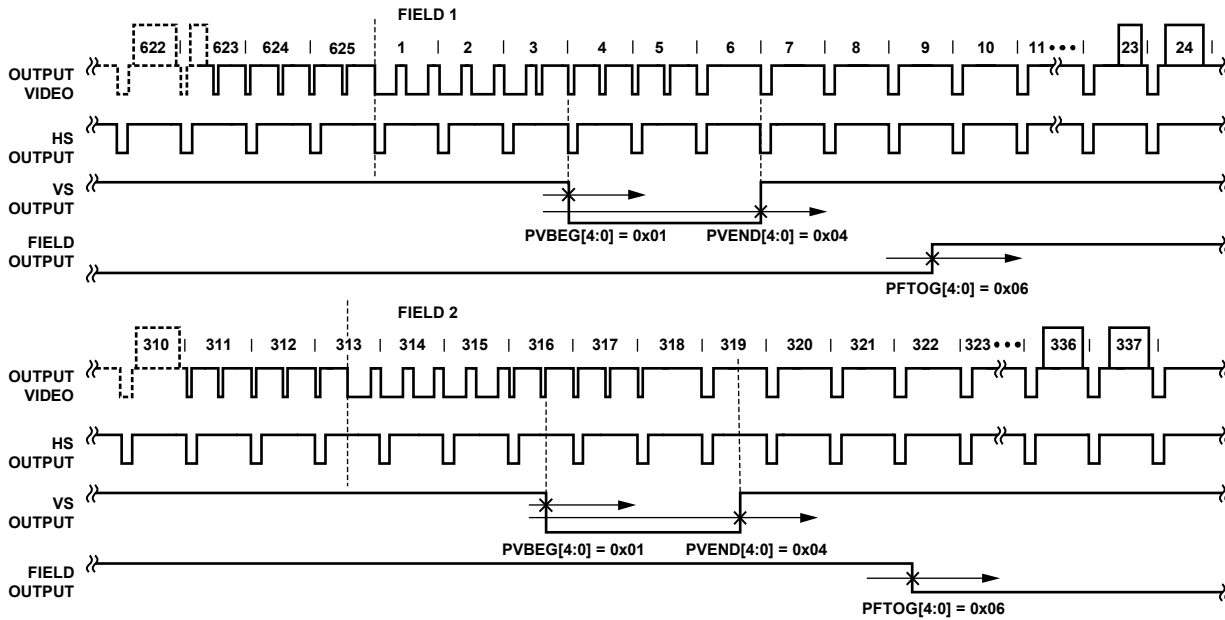


Figure 44. PAL Typical VS/FIELD Positions Using the Register Writes Shown in Table 66

Table 66. User Settings for PAL (See Figure 44)

Register	Register Name	Write
0x31	VS/FIELD Control 1	0x1A
0x32	VS/FIELD Control 2	0x81
0x33	VS/FIELD Control 3	0x84
0x34	HS Position Control 1	0x00
0x35	HS Position Control 2	0x00
0x36	HS Position Control 3	0x7D
0x37	Polarity	0xA1
0xE8	PAL V bit begin	0x41
0xE9	PAL V bit end	0x84
0xEA	PAL F bit toggle	0x06

**PVBEG[4:0], PAL VSYNC Begin, Address 0xE8[4:0]**

The default value of PVBEG is 00101, indicating the PAL VSYNC begin position. For all NTSC/PAL VSYNC timing controls, the V bit in the AV code and the VSYNC signal on the VS pin are modified.

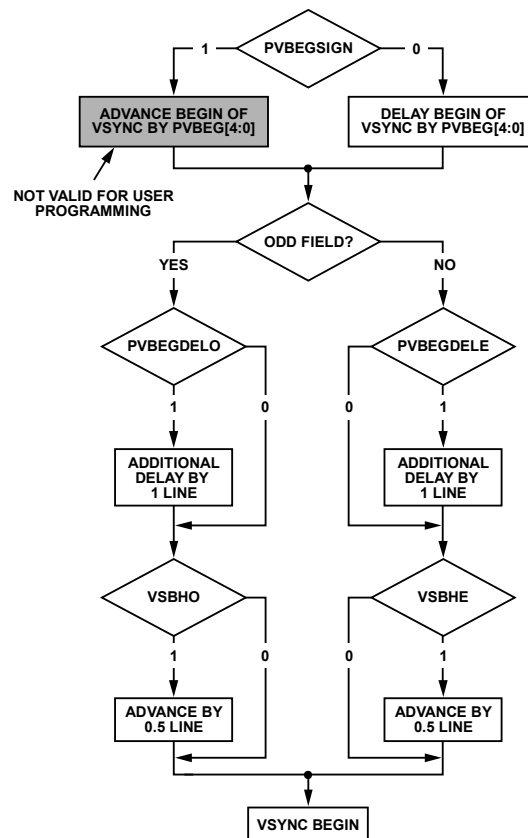


Figure 45. PAL VSYNC Begin

**PVBEGDELO, PAL VSYNC Begin Delay on Odd Field, Address 0xE8[7]**

When PVBEGDELO is 0 (default), there is no delay. Setting PVBEGDELO to 1 delays VSYNC going high on an odd field by a line relative to PVBEG.

**PVBEGDELE, PAL VSYNC Begin Delay on Even Field, Address 0xE8[6]**

When PVBEGDELE is 0, there is no delay. Setting PVBEGDELE to 1 (default) delays VSYNC going high on an even field by a line relative to PVBEG.

**PVBESIGN, PAL VSYNC Begin Sign, Address 0xE8[5]**

Setting PVBESIGN to 0 delays the beginning of VSYNC. Set for user manual programming.

Setting PVBESIGN to 1 (default) advances the beginning of VSYNC (not recommended for user programming).

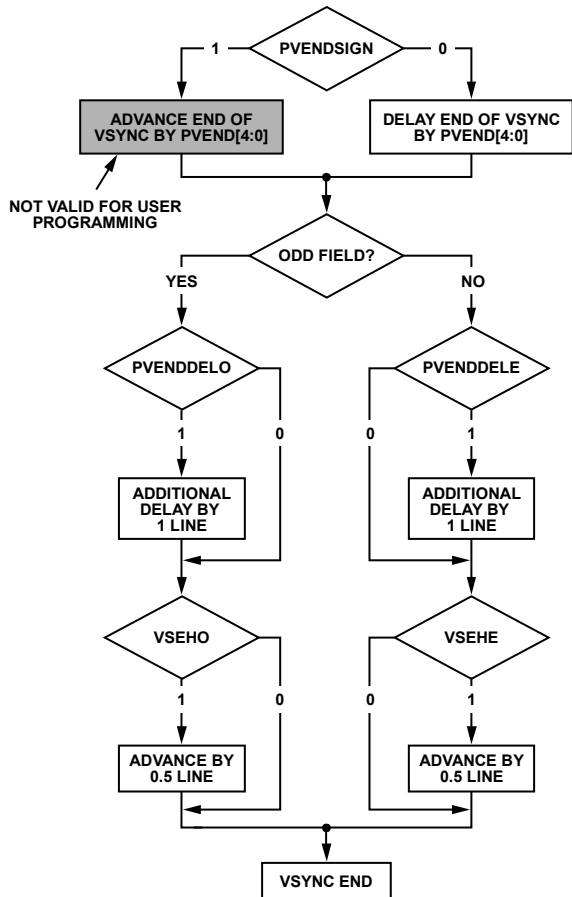


Figure 46. PAL VSYNC End

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**PVENDDELO, PAL VSYNC End Delay on Odd Field, Address 0xE9[7]**

When PVENDDELO is 0 (default), there is no delay.

Setting PVENDDELO to 1 delays VSYNC going low on an odd field by a line relative to PVEND.

**PVENDDELE, PAL VSYNC End Delay on Even Field, Address 0xE9[6]**

When PVENDDELE is 0 (default), there is no delay.

Setting PVENDDELE to 1 delays VSYNC going low on an even field by a line relative to PVEND.

**PVENDSIGN, PAL VSYNC End Sign, Address 0xE9[5]**

Setting PVENDSIGN to 0 (default) delays the end of VSYNC (set for user manual programming).

Setting PVENDSIGN to 1 advances the end of VSYNC (not recommended for user programming).

**PVEND[4:0], PAL VSYNC End, Address 0xE9[4:0]**

The default value of PVEND is 10100, indicating the PAL VSYNC end position.

For all NTSC/PAL VSYNC timing controls, both the V bit in the AV code and the VSYNC signal on the VS pin are modified.

**PFTOGDELO, PAL Field Toggle Delay on Odd Field, Address 0xEA[7]**

When PFTOGDELO is 0 (default), there is no delay.

Setting PFTOGDELO to 1 delays the F toggle/transition on an odd field by a line relative to PFTOG.

**PFTOGDELE, PAL Field Toggle Delay on Even Field, Address 0xEA[6]**

When PFTOGDELE is 0, there is no delay.

Setting PFTOGDELE to 1 (default) delays the F toggle/transition on an even field by a line relative to PFTOG.

**PFTOGSIGN, PAL Field Toggle Sign, Address 0xEA[5]**

Setting PFTOGSIGN to 0 delays the field transition. Set for user manual programming.

Setting PFTOGSIGN to 1 (default) advances the field transition (not recommended for user programming).

**PFTOG, PAL Field Toggle, Address 0xEA[4:0]**

The default value of PFTOG is 00011, indicating the PAL field toggle position.

For all NTSC/PAL field timing controls, the F bit in the AV code and the field signal on the FIELD pin are modified.

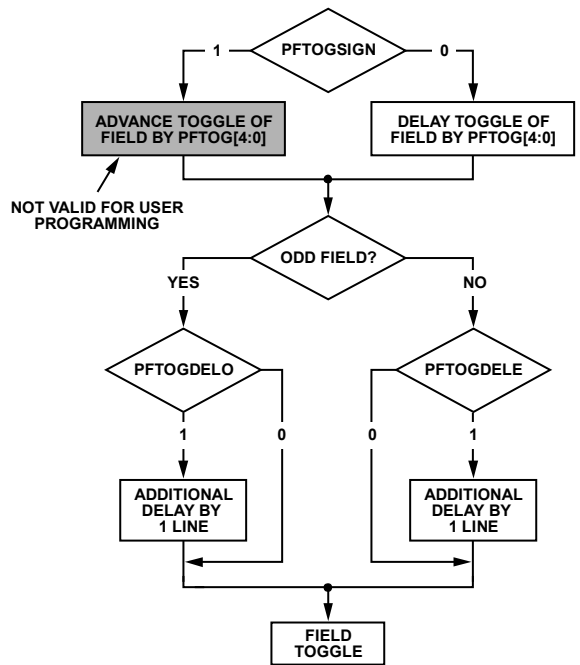


Figure 47. PAL F Toggle

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Table 69. Default Standards on Lines for PAL and NTSC

PAL—625/50				NTSC—525/60			
Line No.	Default VBI Data Decoded	Line No.	Default VBI Data Decoded	Line No.	Default VBI Data Decoded	Line No.	Default VBI Data Decoded
6	WST	318	VPS	23	Gemstar_1x	286	Gemstar_1x
7	WST	319	WST	24	Gemstar_1x	287	Gemstar_1x
8	WST	320	WST	25	Gemstar_1x	288	Gemstar_1x
9	WST	321	WST	10	NABTS	272	NABTS
10	WST	322	WST	11	NABTS	273	NABTS
11	WST	323	WST	12	NABTS	274	NABTS
12	WST	324	WST	13	NABTS	275	NABTS
13	WST	325	WST	14	VITC	276	NABTS
14	WST	326	WST	15	NABTS	277	VITC
15	WST	327	WST	16	VITC	278	NABTS
16	VPS	328	WST	17	NABTS	279	VITC
17	N/A	329	VPS	18	NABTS	280	NABTS
18	N/A	332	VITC	19	NABTS	281	NABTS
19	VITC	333	WST	20	CGMS	282	NABTS
20	WST	334	WST	21	CCAP	283	CGMS
21	WST	335	CCAP	22 + full odd field	NABTS	284	CCAP
22	CCAP	336	WST			285 + full even field	NABTS
23	WSS	337 + full even field	WST				
24 + full odd field	WST						

Table 70. VBI Data Standards for Manual Configuration

VBI_DATA_Px_Ny	625/50—PAL	525/60—NTSC
0000	Disable VDP	Disable VDP
0001	Teletext system identified by VDP_TTXT_TYPE	Teletext system identified by VDP_TTXT_TYPE
0010	VPS-ETSI EN 300 231 V 1.3.1	Reserved
0011	VITC	VITC
0100	WSS ITU-R BT.1119-1/ETSI.EN.300294	CGMS EIA-J CPR-1204/IEC 61880
0101	Reserved	Gemstar_1x
0110	Reserved	Gemstar_2x
0111	CCAP	CCAP EIA-608
1000 to 1111	Reserved	Reserved



























































































Interrupt and VDP Map			Bit (Shading Indicates Default State)								Comments	Notes	
Address	Register	Bit Description	7	6	5	4	3	2	1	0			
0x45	Raw Status 2 (read only)	CCAPD								0	No CCAPD data detected—VBI System 2	These bits are status bits only; they cannot be cleared or masked; Register 0x46 is used for this purpose	
		Reserved					x	x	x	1	CCAPD data detected—VBI System 2		
		EVEN_FIELD				0					0		Current SD field is odd numbered
		Reserved				1					1		Current SD field is even numbered
		Reserved	x	x									
		MPU_STIM_INTRQ	0								0		MPU_STIM_INTRQ = 0
			1							1	MPU_STIM_INTRQ = 1		
0x46	Interrupt Status 2 (read only)	CCAPD_Q								0	Closed captioning not detected in the input video signal—VBI System 2	These bits can be cleared or masked by Register 0x47 and Register 0x48, respectively; note that the interrupt in Register 0x46 for the CCAP, Gemstar, CGMS, and WSS data uses the Mode 1 data slicer	
		Reserved								1	Closed captioning data detected in the video input signal—VBI System 2		
		GEMD_Q								0	Gemstar data not detected in the input video signal—VBI System 2		
		Reserved								1	Gemstar data detected in the input video signal—VBI System 2		
		Reserved				x	x						
		SD_FIELD_CHNGD_Q				0					0		SD signal has not changed field from odd to even or vice versa
		Reserved				1					1		SD signal has changed Field from odd to even or vice versa
		Reserved			x								Not used
		Reserved	x										Not used
MPU_STIM_INTRQ_Q	0								0	Manual interrupt not set			
			1							1	Manual interrupt set		
0x47	Interrupt Clear 2 (write only)	CCAPD_CLR								0	Do not clear—VBI System 2	Note that interrupt in Register 0x46 for the CCAP, Gemstar, CGMS, and WSS data uses the Mode 1 data slicer	
		Reserved								1	Clears CCAPD_Q bit—VBI System 2		
		GEMD_CLR								0	Do not clear		
		Reserved								1	Clears GEMD_Q bit		
		Reserved				0	0						
		SD_FIELD_CHNGD_CLR				0					0		Do not clear
		Reserved				1					1		Clears SD_FIELD_CHNGD_Q bit
Reserved	x	x								Not used			
MPU_STIM_INTRQ_CLR	0								0	Do not clear			
			1							1	Clears MPU_STIM_INTRQ_Q bit		
0x48	Interrupt Mask 2 (read/write)	CCAPD_MSK								0	Masks CCAPD_Q bit—VBI System 2	Note that interrupt in Register 0x46 for the CCAP, Gemstar, CGMS, and WSS data uses the Mode 1 data slicer	
		Reserved								1	Unmasks CCAPD_Q bit—VBI System 2		
		GEMD_MSK								0	Masks GEMD_Q bit—VBI System 2		
		Reserved								1	Unmasks GEMD_Q bit—VBI System 2		
		Reserved				0	0						Not used
		SD_FIELD_CHNGD_MSK				0					0		Masks SD_FIELD_CHNGD_Q bit
		Reserved				1					1		Unmasks SD_FIELD_CHNGD_Q bit
Reserved			0	0						Not used			
MPU_STIM_INTRQ_MSK	0								0	Masks MPU_STIM_INTRQ_Q bit			
			1							1	Unmasks MPU_STIM_INTRQ_Q bit		
0x49	Raw Status 3 (read only)	SD_OP_50Hz; SD 60 Hz/50 Hz frame rate at output								0	SD 60 Hz signal output	These bits are status bits only; they cannot be cleared or masked; Register 0x4A is used for this purpose	
		Reserved								1	SD 50 Hz signal output		
		SD_V_LOCK								0	SD vertical sync lock not established		
		Reserved								1	SD vertical sync lock established		
		SD_H_LOCK							0		0		SD horizontal sync lock not established
		Reserved						x					Not used
		SCM_LOCK				0					0		SECAM lock not established
Reserved				1					1	SECAM lock established			
Reserved	x	x	x							Not used			

Interrupt and VDP Map			Bit (Shading Indicates Default State)								Comments	Notes		
Address	Register	Bit Description	7	6	5	4	3	2	1	0				
0x4A	Interrupt Status 3 (read only)	SD_OP_CHNG_Q; SD 60 Hz/50 Hz frame rate at output								0	No change in SD signal standard detected at the output	These bits can be cleared and masked by Register 0x4B and Register 0x4C, respectively		
											1		A change in SD signal standard is detected at the output	
		SD_V_LOCK_CHNG_Q									0		No change in SD VSYNC lock status	
													1	SD VSYNC lock status has changed
		SD_H_LOCK_CHNG_Q									0		No change in HSYNC lock status	
													1	SD HSYNC lock status has changed
		SD_AD_CHNG_Q; SD autodetect changed									0		No change in AD_RESULT[2:0] bits in Status 1 register	
													1	AD_RESULT[2:0] bits in Status 1 register have changed
		SCM_LOCK_CHNG_Q; SECAM lock									0		No change in SECAM lock status	
										1	SECAM lock status has changed			
PAL_SW_LK_CHNG_Q									0	No change in PAL swinging burst lock status				
										1	PAL swinging burst lock status has changed			
	Reserved		x	x							Not used			
0x4B	Interrupt Clear 3 (write only)	SD_OP_CHNG_CLR									0	Do not clear		
												1	Clears SD_OP_CHNG_Q bit	
		SD_V_LOCK_CHNG_CLR										0	Do not clear	
													1	Clears SD_V_LOCK_CHNG_Q bit
		SD_H_LOCK_CHNG_CLR										0	Do not clear	
													1	Clears SD_H_LOCK_CHNG_Q bit
		SD_AD_CHNG_CLR										0	Do not clear	
													1	Clears SD_AD_CHNG_Q bit
		SCM_LOCK_CHNG_CLR										0	Do not clear	
											1	Clears SCM_LOCK_CHNG_Q bit		
PAL_SW_LK_CHNG_CLR										0	Do not clear			
											1	Clears PAL_SW_LK_CHNG_Q bit		
	Reserved		x	x							Not used			
0x4C	Interrupt Mask 3 (read/write)	SD_OP_CHNG_MSK									0	Masks SD_OP_CHNG_Q bit		
												1	Unmasks SD_OP_CHNG_Q bit	
		SD_V_LOCK_CHNG_MSK										0	Masks SD_V_LOCK_CHNG_Q bit	
													1	Unmasks SD_V_LOCK_CHNG_Q bit
		SD_H_LOCK_CHNG_MSK										0	Masks SD_H_LOCK_CHNG_Q bit	
													1	Unmasks SD_H_LOCK_CHNG_Q bit
		SD_AD_CHNG_MSK										0	Masks SD_AD_CHNG_Q bit	
													1	Unmasks SD_AD_CHNG_Q bit
		SCM_LOCK_CHNG_MSK										0	Masks SCM_LOCK_CHNG_Q bit	
											1	Unmasks SCM_LOCK_CHNG_Q bit		
PAL_SW_LK_CHNG_MSK										0	Masks PAL_SW_LK_CHNG_Q bit			
											1	Unmasks PAL_SW_LK_CHNG_Q bit		
	Reserved		x	x							Not used			
0x4E	Interrupt Status 4 (read only)	VDP_CCAPD_Q									0	Closed captioning not detected	These bits can be cleared and masked by Register 0x4F and Register 0x50, respectively; note that an interrupt in Register 0x4E for the CCAP, Gemstar, CGMS, WSS, VPS, PDC, UTC, and VITC data uses the VDP data slicer	
												1		Closed captioning detected
		Reserved										x		
		VDP_CGMS_WSS_CHNGD_Q; see 0x9C Bit 4 of user sub map to determine whether interrupt is issued for a change in detected data or for when data is detected regardless of content										0		CGMS/WSS data is not changed/not available
														1
		Reserved										x		
		VDP_GS_VPS_PDC_UTC_CHNG_Q; see 0x9C Bit 5 of User Sub Map to determine whether interrupt is issued for a change in detected data or for when data is detected regardless of content										0		Gemstar/PDC/VPS/UTC data is not changed/not available
														1
		Reserved										x		
		VDP_VITC_Q										0		VITC data is not available in the VDP
											1	VITC data is available in the VDP		
	Reserved		x											

Interrupt and VDP Map			Bit (Shading Indicates Default State)								Comments	Notes							
Address	Register	Bit Description	7	6	5	4	3	2	1	0									
0x4F	Interrupt Clear 4 (write only)	VDP_CCAPD_CLR								0	Do not clear	Note that an interrupt in Register 0x4E for the CCAP, Gemstar, CGMS, WSS, VPS, PDC, UTC, and VITC data uses the VDP data slicer							
										1	Clears VDP_CCAPD_Q								
		Reserved									0								
		VDP_CGMS_WSS_CHNGD_CLR									0		Do not clear						
											1		Clears VDP_CGMS_WSS_CHNGD_Q						
		Reserved							0										
		VDP_GS_VPS_PDC_UTC_CHNG_CLR							0				Do not clear						
									1				Clears VDP_GS_VPS_PDC_UTC_CHNG_Q						
0x50	Interrupt Mask 4	VDP_CCAPD_MSK									0	Masks VDP_CCAPD_Q	Note that an interrupt in Register 0x4E for the CCAP, Gemstar, CGMS, WSS, VPS, PDC, UTC, and VITC data uses the VDP data slicer						
											1	Unmasks VDP_CCAPD_Q							
		Reserved										0							
		VDP_CGMS_WSS_CHNGD_MSK										0		Masks VDP_CGMS_WSS_CHNGD_Q					
												1		Unmasks VDP_CGMS_WSS_CHNGD_Q					
		Reserved										0							
		VDP_GS_VPS_PDC_UTC_CHNG_MSK										0		Masks VDP_GS_VPS_PDC_UTC_CHNG_Q					
												1		Unmasks VDP_GS_VPS_PDC_UTC_CHNG_Q					
0x60	VDP_Config_1	VDP_TTXT_TYPE_MAN[1:0]									0	0	PAL: Teletext-ITU-BT.653-625/50-A, NTSC: reserved						
												0	1	PAL: Teletext-ITU-BT.653-625/50-B (WST), NTSC: Teletext-ITU-BT.653-525/60-B					
												1	0	PAL: Teletext-ITU-BT.653-625/50-C, NTSC: Teletext-ITU-BT.653-525/60-C, or EIA516 (NABTS)					
												1	1	PAL: Teletext-ITU-BT.653-625/50-D, NTSC: Teletext-ITU-BT.653-525/60-D					
		VDP_TTXT_TYPE_MAN_ENABLE											0	User programming of teletext type disabled					
													1	User programming of teletext type enabled					
		WST_PKT_DECODE_DISABLE												0	Enable hamming decoding of WST packets				
														1	Disable hamming decoding of WST packets				
		Reserved											1	0	0	0			
		0x61	VDP_Config_2	Reserved						x	x		0	0					
				AUTO_DETECT_GS_TYPE												0	Disable autodetection of Gemstar type		
														1	Enable autodetection of Gemstar type				
0x62	VDP_ADF_Config_1	Reserved												0	0	0			
		ADF_DID[4:0]															1	User-specified DID sent in the ancillary data stream with VDP decoded data	
		ADF_MODE[1:0]															0	0	Nibble mode
																	0	1	Byte mode, no code restrictions
																	1	0	Byte mode with 0x00 and 0xFF prevented
																	1	1	Reserved
		ADF_ENABLE															0	Disable insertion of VBI decoded data into ancillary 656 stream	
															1	Enable insertion of VBI decoded data into ancillary 656 stream			
0x63	VDP_ADF_Config_2	ADF_SDID[5:0]															1	User-specified SDID sent in the ancillary data stream with VDP decoded data	
		Reserved															x		
		DUPLICATE_ADF															0	Ancillary data packet is spread across the Y and C data streams	
															1	Ancillary data packet is duplicated on the Y and C data streams			

Interrupt and VDP Map			Bit (Shading Indicates Default State)								Comments	Notes	
Address	Register	Bit Description	7	6	5	4	3	2	1	0			
0x64	VDP_LINE_00E	VBI_DATA_P318[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 318 (PAL), NTSC—N/A	If set to 1, all VBI_DATA_Px_Ny bits can be set as desired	
		Reserved		0	0	0							
		MAN_LINE_PGM	0										Decode default standards on the lines indicated in Table 69
			1										Manually program the VBI standard to be decoded on each line; see Table 70
0x65	VDP_LINE_00F	VBI_DATA_P319_N286[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 319 (PAL), Line 286 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P6_N23[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 6 (PAL), Line 23 (NTSC)		
0x66	VDP_LINE_010	VBI_DATA_P320_N287[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 320 (PAL), Line 287 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P7_N24[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 7 (PAL), Line 24 (NTSC)		
0x67	VDP_LINE_011	VBI_DATA_P321_N288[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 321 (PAL), Line 288 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P8_N25[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 8 (PAL), Line 25 (NTSC)		
0x68	VDP_LINE_012	VBI_DATA_P322[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 322 (PAL), NTSC—N/A	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P9[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 9 (PAL), NTSC—N/A		
0x69	VDP_LINE_013	VBI_DATA_P323[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 323 (PAL), NTSC—N/A	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P10[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 10 (PAL), NTSC—N/A		
0x6A	VDP_LINE_014	VBI_DATA_P324_N272[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 324 (PAL), Line 272 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P11[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 11 (PAL); NTSC—N/A		
0x6B	VDP_LINE_015	VBI_DATA_P325_N273[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 325 (PAL), Line 273 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P12_N10[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 12 (PAL), Line 10 (NTSC)		
0x6C	VDP_LINE_016	VBI_DATA_P326_N274[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 326 (PAL), Line 274 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P13_N11[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 13 (PAL), Line 11 (NTSC)		
0x6D	VDP_LINE_017	VBI_DATA_P327_N275[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 327 (PAL), Line 275 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P14_N12[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 14 (PAL), Line 12 (NTSC)		
0x6E	VDP_LINE_018	VBI_DATA_P328_N276[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 328 (PAL), Line 276 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P15_N13[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 15 (PAL), Line 13 (NTSC)		
0x6F	VDP_LINE_019	VBI_DATA_P329_N277[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 329 (PAL), Line 277 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P16_N14[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 16 (PAL), Line 14 (NTSC)		
0x70	VDP_LINE_01A	VBI_DATA_P330_N278[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 330 (PAL), Line 278 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P17_N15[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 17 (PAL), Line 15 (NTSC)		
0x71	VDP_LINE_01B	VBI_DATA_P331_N279[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 331 (PAL), Line 279 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P18_N16[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 18 (PAL), Line 16 (NTSC)		
0x72	VDP_LINE_01C	VBI_DATA_P332_N280[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 332 (PAL), Line 280 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P19_N17[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 19 (PAL), Line 17 (NTSC)		
0x73	VDP_LINE_01D	VBI_DATA_P333_N281[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 333 (PAL), Line 281 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P20_N18[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 20 (PAL), Line 18 (NTSC)		

Interrupt and VDP Map			Bit (Shading Indicates Default State)								Comments	Notes	
Address	Register	Bit Description	7	6	5	4	3	2	1	0			
0x74	VDP_LINE_01E	VBI_DATA_P334_N282[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 334 (PAL), Line 282 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P21_N19[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 21 (PAL), Line 19 (NTSC)		
0x75	VDP_LINE_01F	VBI_DATA_P335_N283[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 335 (PAL), Line 283 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P22_N20[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 22 (PAL), Line 20 (NTSC)		
0x76	VDP_LINE_020	VBI_DATA_P336_N284[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 336 (PAL), Line 284 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P23_N21[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 23 (PAL), Line 21 (NTSC)		
0x77	VDP_LINE_021	VBI_DATA_P337_N285[3:0]					0	0	0	0	Sets VBI standard to be decoded from Line 337 (PAL), Line 285 (NTSC)	MAN_LINE_PGM must be set to 1 for these bits to be effective	
		VBI_DATA_P24_N22[3:0]	0	0	0	0					Sets VBI standard to be decoded from Line 24 (PAL), Line 22 (NTSC)		
0x78	VDP_STATUS (read only)	CC_AVL								0	Closed captioning not detected	CC_CLEAR resets the CC_AVL bit	
										1	Closed captioning detected		
		CC_EVEN_FIELD								0	Closed captioning decoded from odd field		
										1	Closed captioning decoded from even field		
		CGMS_WSS_AVL							0		CGMS/WSS not detected	CGMS_WSS_CLEAR resets the CGMS_WSS_AVL bit	
									1		CGMS/WSS detected		
		Reserved						0					
		GS_PDC_VPS_UTC_AVL					0					GS/PDC/VPS/UTC not detected	GS_PDC_VPS_UTC_CLEAR resets the GS_PDC_VPS_UTC_AVL bit
						1						GS/PDC/VPS/UTC detected	
		GS_DATA_TYPE			0							Gemstar_1x detected	
				1							Gemstar_2x detected		
	VITC_AVL			0							VITC not detected	VITC_CLEAR resets the VITC_AVL bit	
				1							VITC detected		
	TTXT_AVL			0							Teletext not detected		
				1							Teletext detected		
	VDP_STATUS_CLEAR (write only)	CC_CLEAR									0	Does not reinitialize the CCAP readback registers	This is a self-clearing bit
											1	Reinitializes the CCAP readback registers	
Reserved										0			
CGMS_WSS_CLEAR										0	Does not reinitialize the CGMS/WSS readback registers	This is a self-clearing bit	
										1	Reinitializes the CGMS/WSS readback registers		
Reserved								0					
GS_PDC_VPS_UTC_CLEAR						0					Does not reinitialize the GS/PDC/VPS/UTC readback registers	This is a self-clearing bit	
						1					Refreshes the GS/PDC/VPS/UTC readback registers		
Reserved					0								
VITC_CLEAR				0						Does not reinitialize the VITC readback registers	This is a self-clearing bit		
				1						Reinitializes the VITC readback registers			
Reserved			0										
0x79	VDP_CCAP_DATA_0 (read only)	CCAP_BYTE_1[7:0]	x	x	x	x	x	x	x	x	Decoded Byte 1 of CCAP		
0x7A	VDP_CCAP_DATA_1 (read only)	CCAP_BYTE_2[7:0]	x	x	x	x	x	x	x	x	Decoded Byte 2 of CCAP		
0x7D	VDP_CGMS_WSS_DATA_0 (read only)	CGMS_CRC[5:2]					x	x	x	x	Decoded CRC sequence for CGMS		
		Reserved	0	0	0	0							
0x7E	VDP_CGMS_WSS_DATA_1 (read only)	CGMS_WSS[13:8]			x	x	x	x	x	x	Decoded CGMS/WSS data		
		CGMS_CRC[1:0]	x	x							Decoded CRC sequence for CGMS		
0x7F	VDP_CGMS_WSS_DATA_2 (read only)	CGMS_WSS[7:0]	x	x	x	x	x	x	x	x	Decoded CGMS/WSS data		
0x84	VDP_GS_VPS_PDC_UTC_0 (read only)	GS_VPS_PDC_UTC_BYTE_0[7:0]	x	x	x	x	x	x	x	x	Decoded Gemstar/VPS/PDC/UTC data		
0x85	VDP_GS_VPS_PDC_UTC_1 (read only)	GS_VPS_PDC_UTC_BYTE_1[7:0]	x	x	x	x	x	x	x	x	Decoded Gemstar/VPS/PDC/UTC data		

Interrupt and VDP Map			Bit (Shading Indicates Default State)								Comments	Notes	
Address	Register	Bit Description	7	6	5	4	3	2	1	0			
0x86	VDP_GS_VPS_PDC_UTC_2 (read only)	GS_VPS_PDC_UTC_BYTE_2[7:0]	x	x	x	x	x	x	x	x	Decoded Gemstar/VPS/PDC/UTC data		
0x87	VDP_GS_VPS_PDC_UTC_3 (read only)	GS_VPS_PDC_UTC_BYTE_3[7:0]	x	x	x	x	x	x	x	x	Decoded Gemstar/VPS/PDC/UTC data		
0x88	VDP_VPS_PDC_UTC_4 (read only)	VPS_PDC_UTC_BYTE_4[7:0]	x	x	x	x	x	x	x	x	Decoded VPS/PDC/UTC data		
0x89	VDP_VPS_PDC_UTC_5 (read only)	VPS_PDC_UTC_BYTE_5[7:0]	x	x	x	x	x	x	x	x	Decoded VPS/PDC/UTC data		
0x8A	VDP_VPS_PDC_UTC_6 (read only)	VPS_PDC_UTC_BYTE_6[7:0]	x	x	x	x	x	x	x	x	Decoded VPS/PDC/UTC data		
0x8B	VDP_VPS_PDC_UTC_7 (read only)	VPS_PDC_UTC_BYTE_7[7:0]	x	x	x	x	x	x	x	x	Decoded VPS/PDC/UTC data		
0x8C	VDP_VPS_PDC_UTC_8 (read only)	VPS_PDC_UTC_BYTE_8[7:0]	x	x	x	x	x	x	x	x	Decoded VPS/PDC/UTC data		
0x8D	VDP_VPS_PDC_UTC_9 (read only)	VPS_PDC_UTC_BYTE_9[7:0]	x	x	x	x	x	x	x	x	Decoded VPS/PDC/UTC data		
0x8E	VDP_VPS_PDC_UTC_10 (read only)	VPS_PDC_UTC_BYTE_10[7:0]	x	x	x	x	x	x	x	x	Decoded VPS/PDC/UTC data		
0x8F	VDP_VPS_PDC_UTC_11 (read only)	VPS_PDC_UTC_BYTE_11[7:0]	x	x	x	x	x	x	x	x	Decoded VPS/PDC/UTC data		
0x90	VDP_VPS_PDC_UTC_12 (read only)	VPS_PDC_UTC_BYTE_12[7:0]	x	x	x	x	x	x	x	x	Decoded VPS/PDC/UTC data		
0x92	VDP_VITC_DATA_0 (read only)	VITC_DATA_0[7:0]	x	x	x	x	x	x	x	x	Decoded VITC data		
0x93	VDP_VITC_DATA_1 (read only)	VITC_DATA_1[7:0]	x	x	x	x	x	x	x	x	Decoded VITC data		
0x94	VDP_VITC_DATA_2 (read only)	VITC_DATA_2[7:0]	x	x	x	x	x	x	x	x	Decoded VITC data		
0x95	VDP_VITC_DATA_3 (read only)	VITC_DATA_3[7:0]	x	x	x	x	x	x	x	x	Decoded VITC data		
0x96	VDP_VITC_DATA_4 (read only)	VITC_DATA_4[7:0]	x	x	x	x	x	x	x	x	Decoded VITC data		
0x97	VDP_VITC_DATA_5 (read only)	VITC_DATA_5[7:0]	x	x	x	x	x	x	x	x	Decoded VITC data		
0x98	VDP_VITC_DATA_6 (read only)	VITC_DATA_6[7:0]	x	x	x	x	x	x	x	x	Decoded VITC data		
0x99	VDP_VITC_DATA_7 (read only)	VITC_DATA_7[7:0]	x	x	x	x	x	x	x	x	Decoded VITC data		
0x9A	VDP_VITC_DATA_8 (read only)	VITC_DATA_8[7:0]	x	x	x	x	x	x	x	x	Decoded VITC data		
0x9B	VDP_VITC_CALC_CRC (read only)	VITC_CRC[7:0]	x	x	x	x	x	x	x	x	Decoded VITC CRC data		
0x9C	VDP_OUTPUT_SEL	Reserved					0	0	0	0			
		WSS_CGMS_CB_CHANGE				0						Disable content-based updating of CGMS and WSS data	The available bit shows the availability of data only when its content has changed
					1							Enable content-based updating of CGMS and WSS data	
		GS_VPS_PDC_UTC_CB_CHANGE			0							Disable content-based updating of Gemstar, VPS, PDC, and UTC data	
					1							Enable content-based updating of Gemstar, VPS, PDC, and UTC data	
		PC_GS_VPS_PDC_UTC[1:0]	0	0									Gemstar_1x/Gemstar_2x
	0	1									VPS		
	1	0									PDC		
	1	1									UTC		

<sup>1</sup> Shading indicates default values.  
<sup>2</sup> x indicates a bit that keeps the last written value.

## PCB LAYOUT RECOMMENDATIONS

The [ADV7180](#) is a high precision, high speed, mixed-signal device. To achieve the maximum performance from the part, it is important to have a well laid out PCB. The following is a guide for designing a board using the [ADV7180](#).

### ANALOG INTERFACE INPUTS

Take care when routing the inputs on the PCB. Keep track lengths to a minimum, and use  $75\ \Omega$  trace impedances when possible. In addition, trace impedances other than  $75\ \Omega$  increase the chance of reflections.

### POWER SUPPLY DECOUPLING

It is recommended to decouple each power supply pin with  $0.1\ \mu\text{F}$  and  $10\ \text{nF}$  capacitors. The fundamental idea is to have a decoupling capacitor within about  $0.5\ \text{cm}$  of each power pin. In addition, avoid placing the capacitor on the opposite side of the PCB from the [ADV7180](#) because doing so interposes inductive vias in the path. The decoupling capacitors must be located between the power plane and the power pin. Current must flow from the power plane to the capacitor and then to the power pin. Do not apply the power connection between the capacitor and the power pin. Placing a via underneath the  $100\ \text{nF}$  capacitor pads, down to the power plane, is the best approach (see Figure 56).

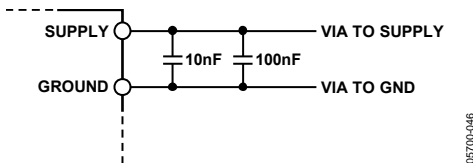


Figure 56. Recommended Power Supply Decoupling

It is particularly important to maintain low noise and good stability of  $P_{\text{VDD}}$ . Careful attention must be paid to regulation, filtering, and decoupling. It is highly desirable to provide separate regulated supplies for each of the analog circuitry groups ( $A_{\text{VDD}}$ ,  $D_{\text{VDD}}$ ,  $D_{\text{VDDIO}}$ , and  $P_{\text{VDD}}$ ).

Some graphic controllers use substantially different levels of power when active (during active picture time) and when idle (during horizontal and vertical sync periods). This can result in a measurable change in the voltage supplied to the analog supply regulator, which can in turn produce changes in the regulated analog supply voltage. This can be mitigated by regulating the analog supply, or at least  $P_{\text{VDD}}$ , from a different, cleaner power source, for example, from a  $12\ \text{V}$  supply.

Using a single ground plane for the entire board is also recommended.

Experience repeatedly shows that the noise performance is the same or better with a single ground plane. Using multiple ground planes can be detrimental because each separate ground plane is smaller, and long ground loops can result.

### PLL

Place the PLL loop filter components as close as possible to the ELPF pin. It must also be placed on the same side of the PCB as the [ADV7180](#). Do not place any digital or other high frequency traces near these components. Use the values suggested in this data sheet with tolerances of 10% or less.

### VREFN AND VREFP

Place the circuit associated with these pins as close as possible and on the same side of the PCB as the [ADV7180](#).

### DIGITAL OUTPUTS (BOTH DATA AND CLOCKS)

Try to minimize the trace length that the digital outputs have to drive. Longer traces have higher capacitance, requiring more current and, in turn, causing more internal digital noise. Shorter traces reduce the possibility of reflections.

Adding a  $30\ \Omega$  to  $50\ \Omega$  series resistor can suppress reflections, reduce EMI, and reduce the current spikes inside the [ADV7180](#). If series resistors are used, place them as close as possible to the [ADV7180](#) pins. However, try not to add vias or extra length to the output trace to place the resistors closer.

If possible, limit the capacitance that each of the digital outputs drives to less than  $15\ \text{pF}$ . This can easily be accomplished by keeping traces short and by connecting the outputs to only one device. Loading the outputs with excessive capacitance increases the current transients inside the [ADV7180](#), creating more digital noise on its power supplies.

The 40-lead and 32-lead LFCSP have an exposed metal paddle on the bottom of the package. This paddle must be soldered to PCB ground for proper heat dissipation and for noise and mechanical strength benefits.

### DIGITAL INPUTS

The digital inputs on the [ADV7180](#) are designed to work with  $1.8\ \text{V}$  to  $3.3\ \text{V}$  signals and are not tolerant of  $5\ \text{V}$  signals. Extra components are needed if  $5\ \text{V}$  logic signals are required to be applied to the decoder.

### TYPICAL CIRCUIT CONNECTION

Examples of how to connect the 40-lead LFCSP, 64-lead LQFP, 48-lead LQFP, and 32-lead LFCSP video decoders are shown in Figure 57, Figure 58, Figure 59, and Figure 60. For a detailed schematic of the [ADV7180](#) evaluation boards, contact a local Analog Devices field applications engineer or an Analog Devices distributor.

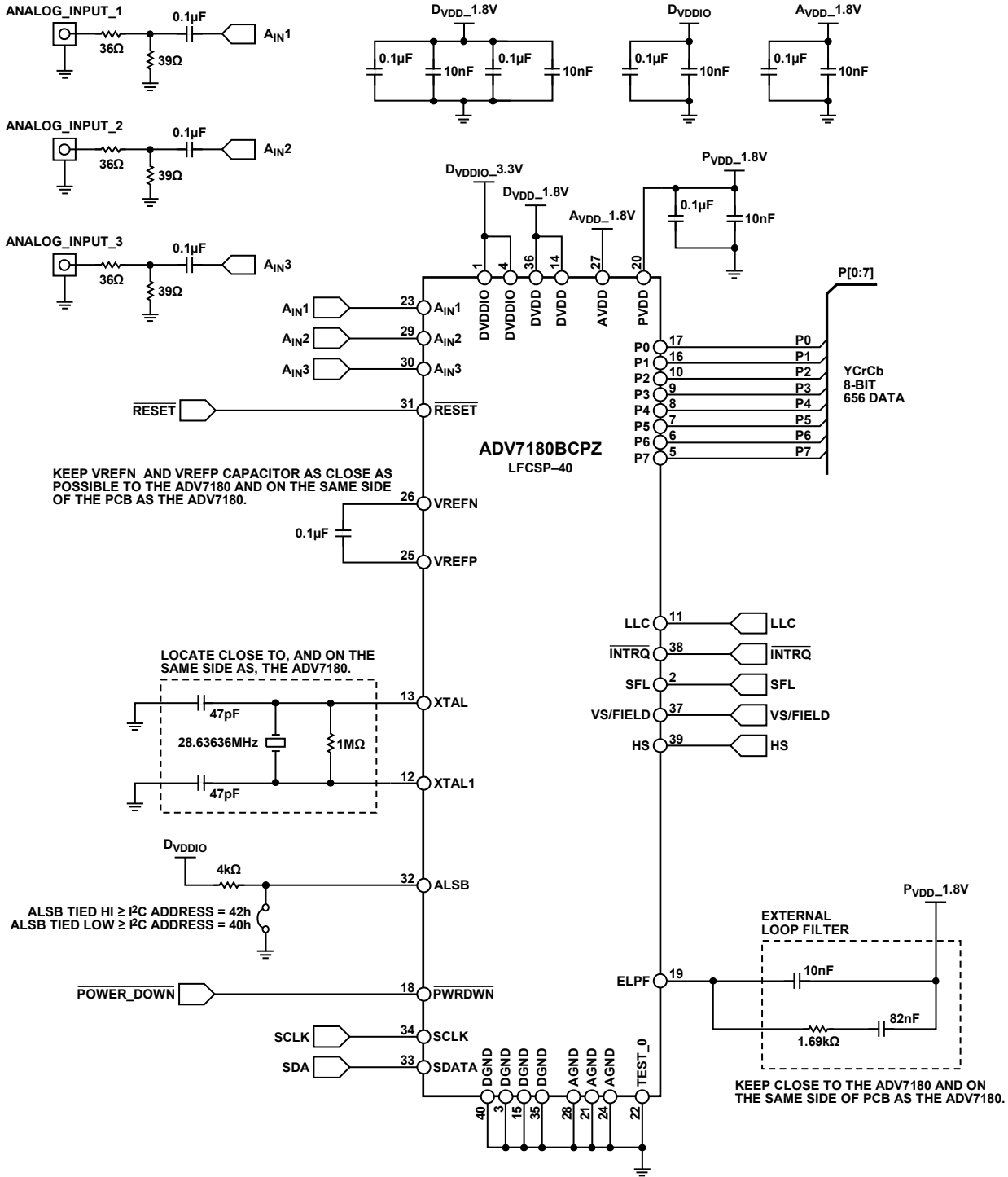


Figure 57. 40-Lead LFCSP Typical Connection Diagram

05700-048

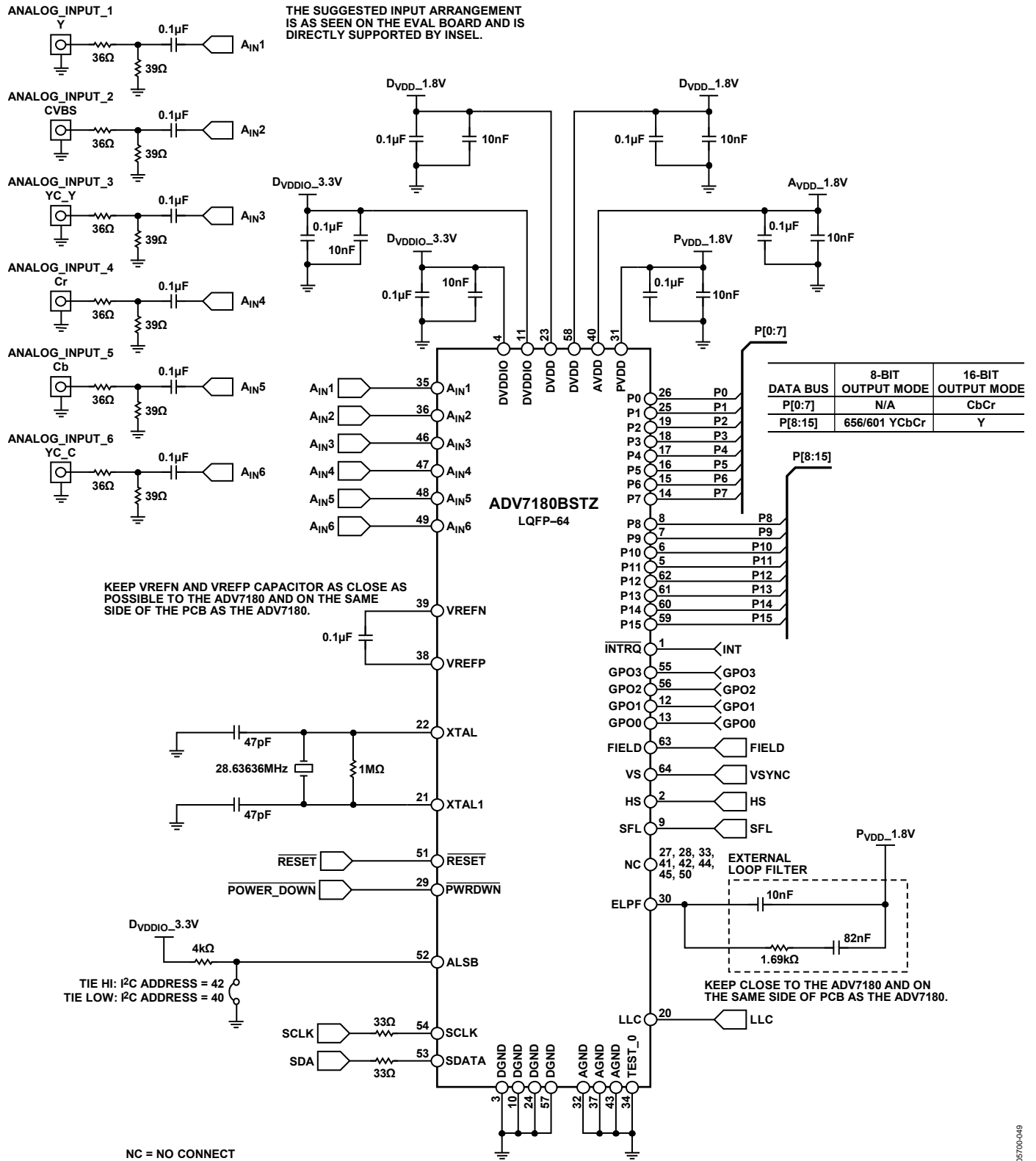


Figure 58. 64-Lead LQFP Typical Connection Diagram

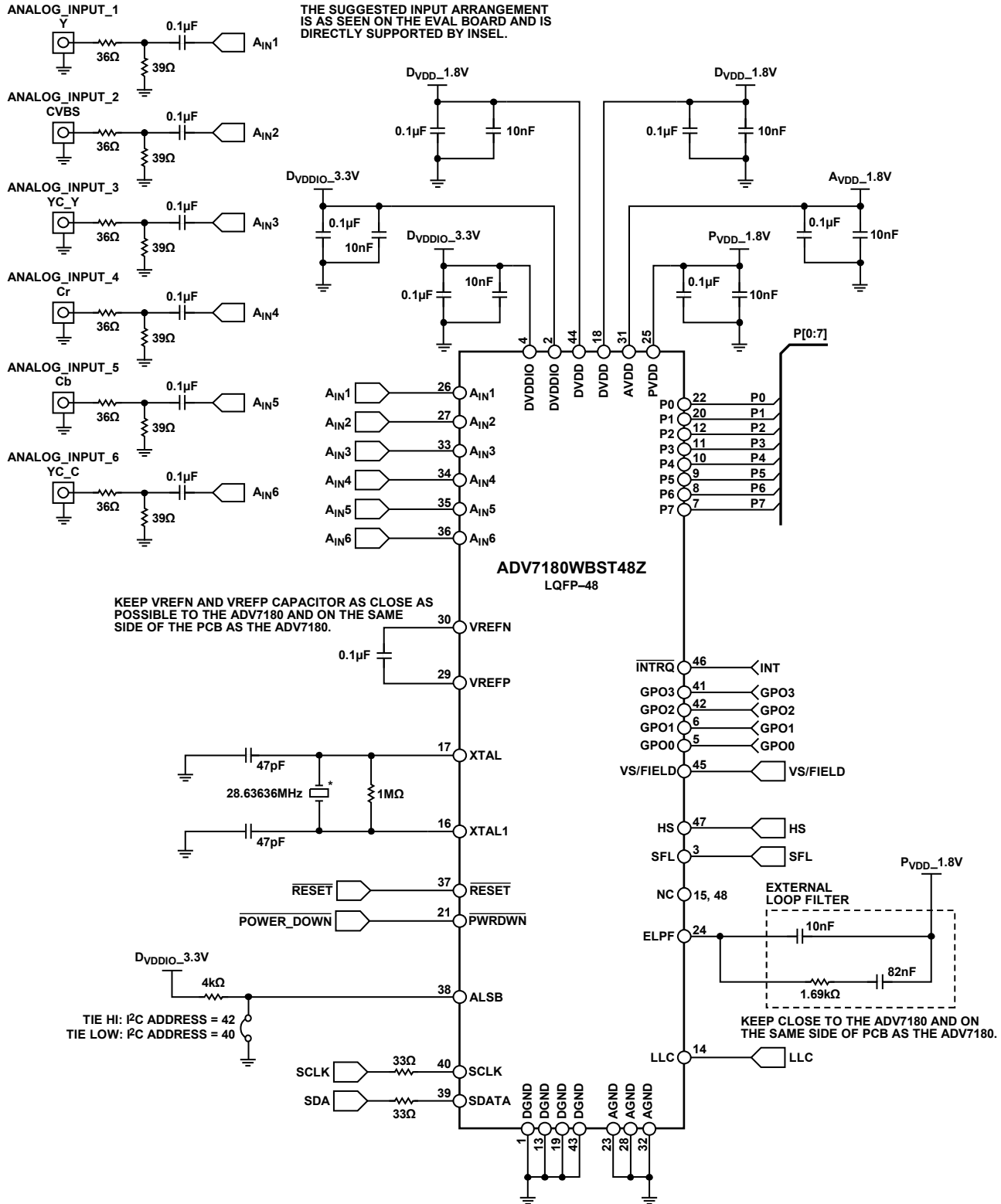


Figure 59. 48-Lead LQFP Typical Connection Diagram

05700-061

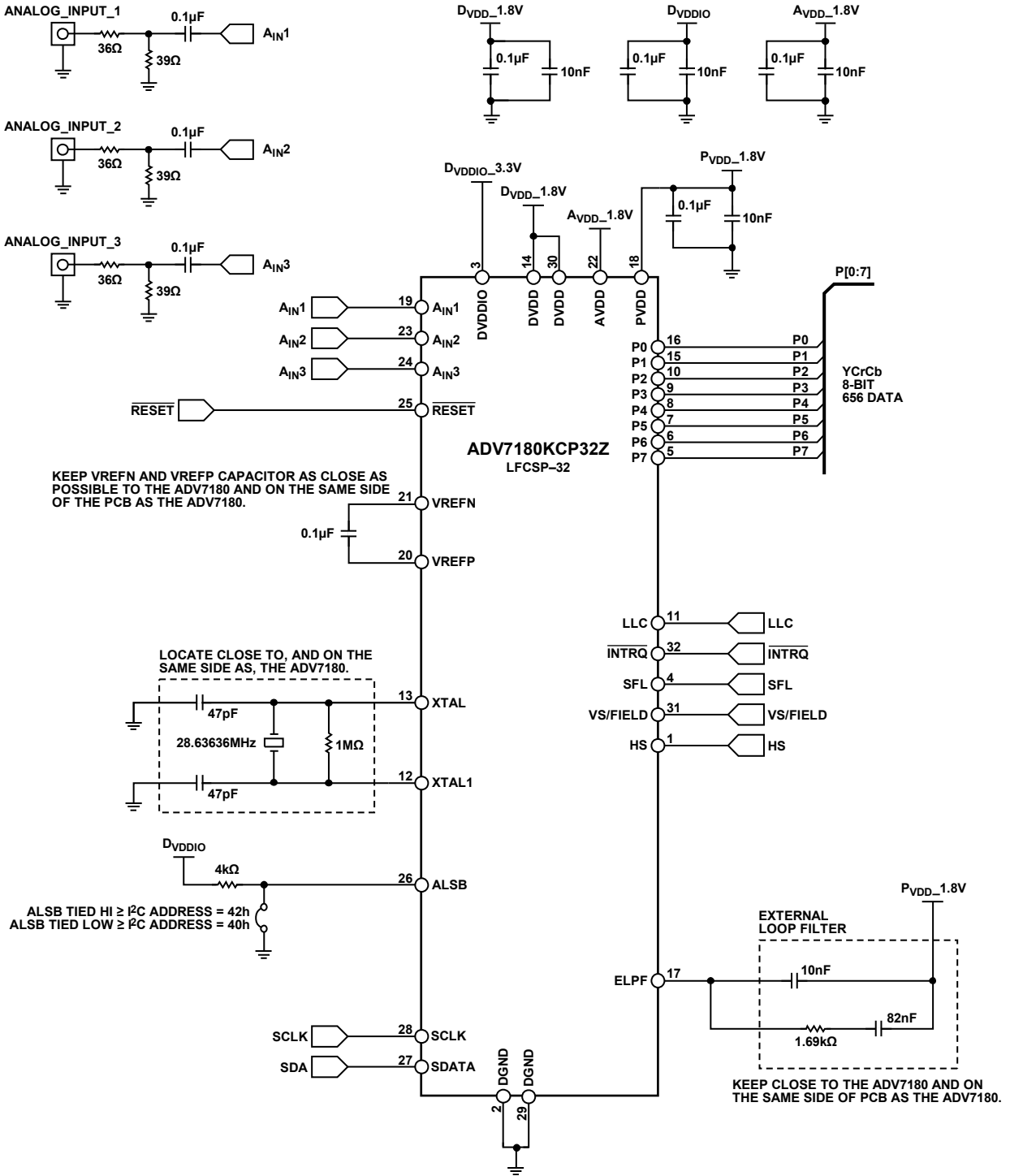
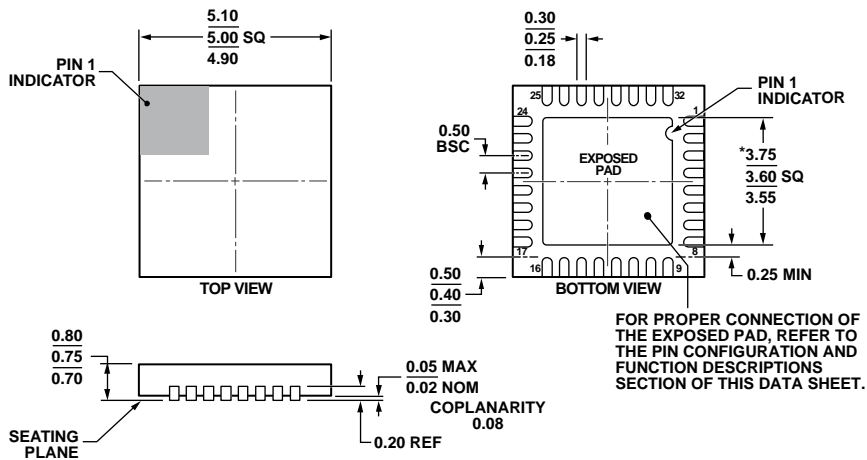


Figure 60. 32-Lead LFCSP Typical Connection Diagram

05700-056

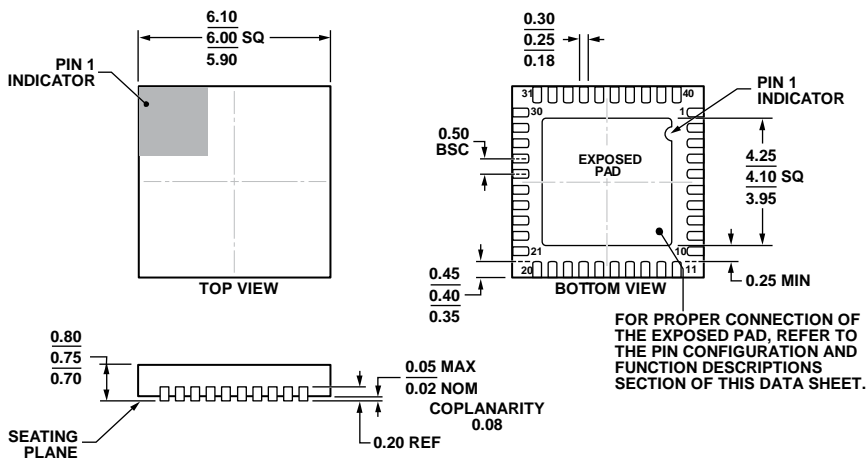
OUTLINE DIMENSIONS



\*COMPLIANT TO JEDEC STANDARDS MO-220-WHHD-5 WITH EXCEPTION TO EXPOSED PAD DIMENSION.

Figure 61. 32-Lead Lead Frame Chip Scale Package [LFCSP\_WQ]  
 5 mm × 5 mm Body, Very Very Thin Quad  
 (CP-32-12)  
 Dimensions shown in millimeters

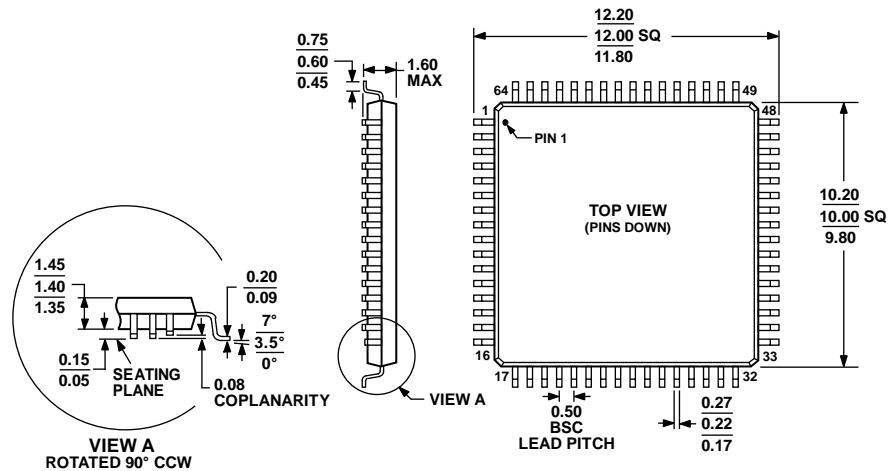
08-16-2010-B



COMPLIANT TO JEDEC STANDARDS MO-220-WJJD.

Figure 62. 40-Lead Lead Frame Chip Scale Package [LFCSP\_WQ]  
 6 mm × 6 mm Body, Very Very Thin Quad  
 (CP-40-9)  
 Dimensions shown in millimeters

05-06-2011-A

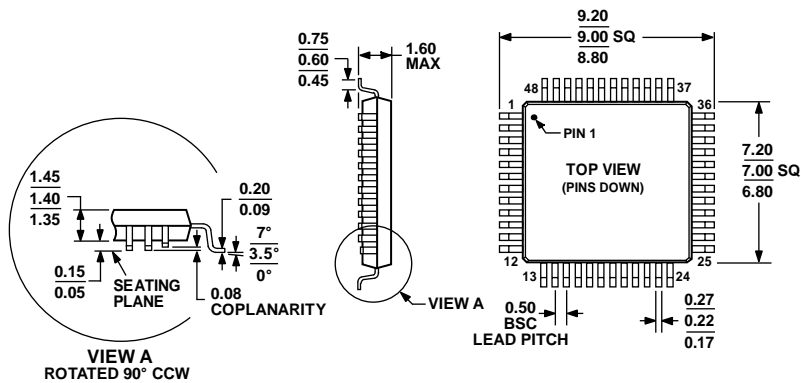


COMPLIANT TO JEDEC STANDARDS MS-026-BCD

Figure 63. 64-Lead Low Profile Quad Flat Package [LQFP]  
10 mm × 10 mm Body  
(ST-64-2)

Dimensions shown in millimeters

051706-A



COMPLIANT TO JEDEC STANDARDS MS-026-BBC

Figure 64. 48-Lead Low Profile Quad Flat Package [LQFP]  
7 mm × 7 mm Body  
(ST-48)

Dimensions shown in millimeters

051706-A

## ORDERING GUIDE

Model <sup>1,2</sup>	Temperature Range	Package Description	Package Option
ADV7180KCP32Z	–10°C to +70°C	32-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-32-12
ADV7180KCP32Z-RL	–10°C to +70°C	32-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-32-12
ADV7180BCPZ	–40°C to +85°C	40-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-40-9
ADV7180BCPZ-REEL	–40°C to +85°C	40-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-40-9
ADV7180BSTZ	–40°C to +85°C	64-Lead Low Profile Quad Flat Package [LQFP]	ST-64-2
ADV7180BSTZ-REEL	–40°C to +85°C	64-Lead Low Profile Quad Flat Package [LQFP]	ST-64-2
ADV7180WBCP32Z	–40°C to +85°C	32-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-32-12
ADV7180WBCP32Z-RL	–40°C to +85°C	32-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-32-12
ADV7180WBCPZ	–40°C to +125°C	40-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-40-9
ADV7180WBCPZ-REEL	–40°C to +125°C	40-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-40-9
ADV7180WBSTZ	–40°C to +125°C	64-Lead Low Profile Quad Flat Package [LQFP]	ST-64-2
ADV7180WBSTZ-REEL	–40°C to +125°C	64-Lead Low Profile Quad Flat Package [LQFP]	ST-64-2
ADV7180WBST48Z	–40°C to +85°C	48-Lead Low Profile Quad Flat Package [LQFP]	ST-48
ADV7180WBST48Z-RL	–40°C to +85°C	48-Lead Low Profile Quad Flat Package [LQFP]	ST-48
ADV7180KST48Z	–10°C to +70°C	48-Lead Low Profile Quad Flat Package [LQFP]	ST-48
ADV7180KST48Z-RL	–10°C to +70°C	48-Lead Low Profile Quad Flat Package [LQFP]	ST-48
ADV7180BST48Z	–40°C to +85°C	48-Lead Low Profile Quad Flat Package [LQFP]	ST-48
ADV7180BST48Z-RL	–40°C to +85°C	48-Lead Low Profile Quad Flat Package [LQFP]	ST-48
ADV7180BCP32Z	–40°C to +85°C	32-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-32-12
ADV7180BCP32Z-RL	–40°C to +85°C	32-Lead Lead Frame Chip Scale Package [LFCSP_WQ]	CP-32-12
EVAL-ADV7180LQEBZ		Evaluation Board for the 64-Lead LQFP	
EVAL-ADV7180LFEBZ		Evaluation Board for the 40-Lead LFCSP	
EVAL-ADV7180-32EBZ		Evaluation Board for the 32-Lead LFCSP	

<sup>1</sup> Z = RoHS Compliant Part.

<sup>2</sup> W = Qualified for Automotive Applications.

## AUTOMOTIVE PRODUCTS

The **ADV7180W** models are available with controlled manufacturing to support the quality and reliability requirements of automotive applications. Note that these automotive models may have specifications that differ from the commercial models, and designers should review the product Specifications section of this data sheet carefully. Only the automotive grade products shown are available for use in automotive applications. Contact your local Analog Devices account representative for specific product ordering information and to obtain the specific automotive reliability reports for these models.

Note that the **ADV7180** is a Pb-free, environmentally friendly product. It is manufactured using the most up-to-date materials and processes. The coating on the leads of each device is 100% pure Sn electroplate. The device is suitable for Pb-free applications and can withstand surface-mount soldering at up to 255°C (±5°C).

In addition, it is backward-compatible with conventional SnPb soldering processes. This means that the electroplated Sn coating can be soldered with Sn/Pb solder pastes at conventional reflow temperatures of 220°C to 235°C.

I<sup>2</sup>C refers to a communications protocol originally developed by Philips Semiconductors (now NXP Semiconductors).

## Looking for pricing, stock, or lifecycle information?

Click below to explore more details on WIN SOURCE:

- ⊖ [View ADV7180BCPZ-REEL on WIN SOURCE](#)
- ⊖ [Analog Devices Inc. Information](#)

## Optimize Your Supply Chain with WIN SOURCE Solutions

- ✓ Global Sourcing Solution
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- ✓ Cost Control Management
- ✓ Shortage Management
- ✓ Alternative Solution
- ✓ Excess Inventory Management