



MIC3289

1.2MHz PWM White LED Driver with Internal Schottky Diode and True 1-Wire Digital Control

General Description

The MIC3289 is a PWM boost-switching regulator that is optimized for constant-current white LED driver applications. The MIC3289 features an internal Schottky diode, allowing an efficient DC/DC solution that requires only 4 external components.

The MIC3289 allows for a single wire simple digital interface to control the dimming over 16 steps with a log scale to give better resolution at the lower currents and to better match the sensitivity of the human eye. The pre-programming feature allows the user to select any one of the 16 WLED current levels as the start-up brightness level.

The feedback voltage of the MIC3289 is only 250mV, allowing high efficiency while retaining excellent accuracy for the white LED current.

The MIC3289 implements a constant frequency 1.2MHz PWM control scheme. The high frequency PWM operation saves board space by reducing external component sizes. The 1.2MHz PWM scheme also reduces switching noise and ripple to the input power source.

The 2.5V to 6.5V input voltage range of MIC3289 allows direct operation from single cell Li Ion as well as 3- to 4-cell NiCad/NiMH/Alkaline batteries. Battery life is preserved with a low 1 μ A shutdown current.

The MIC3289 is available in a low profile Thin SOT23 6-lead package and a 2mm x 2mm MLF[®]-8L package and has a junction temperature range of -40°C to +125°C.

Data sheets and support documentation can be found on Micrel's web site at www.micrel.com.

Features

- Single wire combines 16 level logarithmic brightness & shutdown control
- 16V / 24V OVP options supports up to 4 & 6 WLEDs
- Start-up in any one of 16 brightness levels
- Internal Schottky diode
- 2.5V to 6.5V input voltage
- 1.2 MHz PWM operation
- Over 500mA switch current
- 250mV reference voltage
- $\pm 5\%$ LED current accuracy
- <1 μ A shutdown current
- Over temperature protection
- UVLO
- Thin SOT23-6L package option
- 2mm x 2mm leadless MLF[®]-8L package option
- -40°C to +125°C junction temperature range

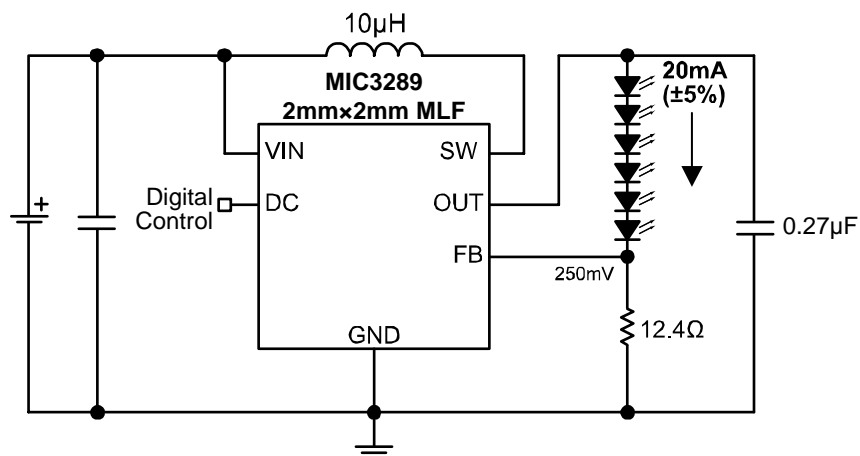
Applications

- White/Blue LED driver for backlighting
 - Cell phones
 - PDAs
 - GPS systems
 - Digital cameras
 - Multimedia / MP3 players
- LED flashlights
- Constant current power supplies

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Typical Application

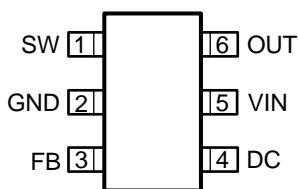


White LED Driver with OVP and Digital Control

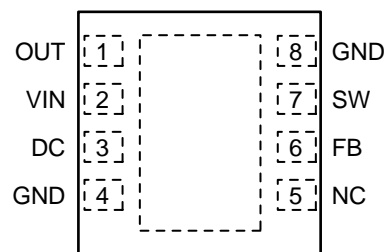
Ordering Information

Part Number	Marking Code	Output Voltage	Over Voltage Protection	Junction Temp. Range	Package
MIC3289-16YD6	WF16	Adjustable	16V	-40°C to 125°C	TSOT23-6
MIC3289-24YD6	WF24	Adjustable	24V	-40°C to 125°C	TSOT23-6
MIC3289-16YML	WFA	Adjustable	16V	-40°C to 125°C	2x2 MLF [®] -8L
MIC3289-24YML	WFB	Adjustable	24V	-40°C to 125°C	2x2 MLF [®] -8L

Pin Configuration



TSOT23-6 (D)



2mm x 2mm 8-pin MLF[®] (ML)

Pin Description

Pin Number SOT23-6	Pin Number 8-pin MLF [®]	Pin Name	Pin Name
6	1	OUT	Output and Over Voltage Protection (output)
5	2	VIN	Supply (Input): 2.5V to 6.5V for internal circuitry.
4	3	DC	Single pin digital control. See diagrams.
-	5	N/C	No connect (no internal connection to die)
3	6	FB	Feedback (Input): Output voltage sense node. Connect the cathode of the LED to this pin.
1	7	SW	Switch Node (Input): Internal power BIPOLAR collector.
2	4,8	GND	Ground (Return): Ground.
-	Pad	GND	Ground (Return): Backside pad.

Absolute Maximum Ratings⁽¹⁾

Supply voltage (V_{IN})	7.5V
Switch voltage (V_{SW})	-0.3V to 27V
Digital Control Voltage (V_{DC})	-0.3 to V_{IN}
FB Voltage (V_{FB})	6V
Switch Current (I_{SW})	2A
Ambient Storage Temperature (T_S)	-65°C to +150°C
ESD Rating, Note 3	2KV

Operating Ratings⁽²⁾

Supply Voltage (V_{IN})	2.5V to 6.5V
Output Voltage (V_{OUT})	(V_{IN} to V_{OVP})
Junction Temperature Range (T_J)	-40°C to +125°C
Package Thermal Impedance	
θ_{JA} 2mm x 2mm MLF [®] -8L	93°C/W
θ_{JA} TSOT23-6	235°C/W

Electrical Characteristics⁽⁴⁾

$T_A=25^\circ\text{C}$, $V_{IN} = 3.6\text{V}$, $V_{OUT} = 10\text{V}$, $I_{OUT} = 20\text{mA}$, unless otherwise noted. Bold values indicate $-40^\circ\text{C} \leq T_J \leq 125^\circ\text{C}$.

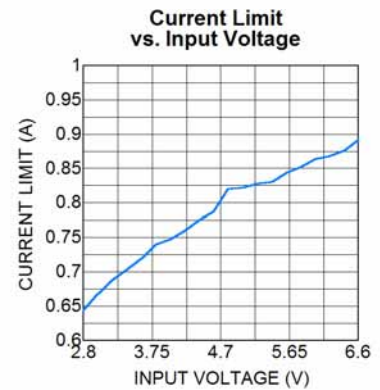
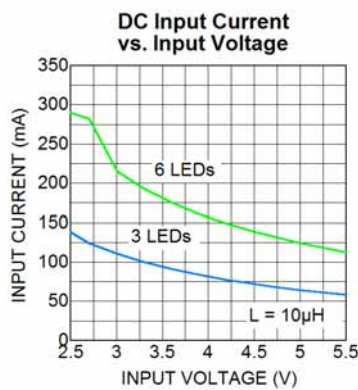
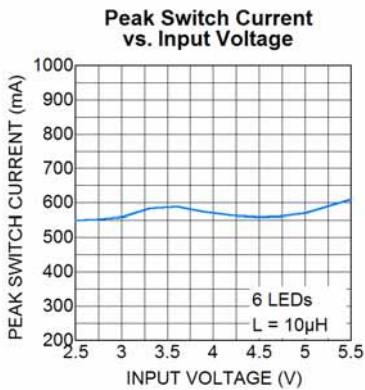
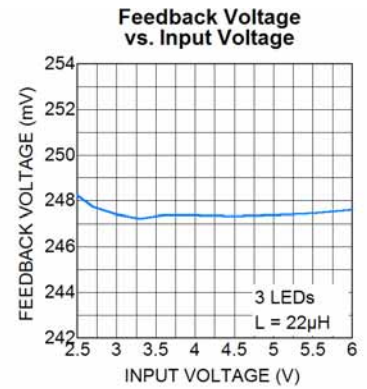
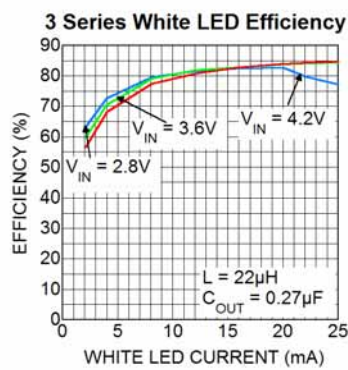
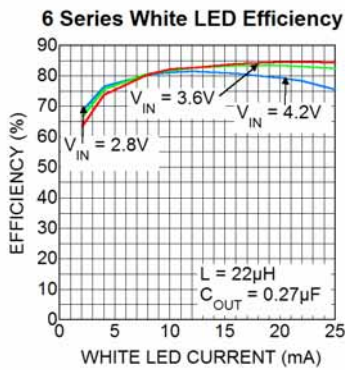
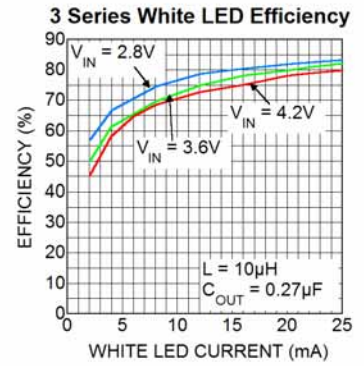
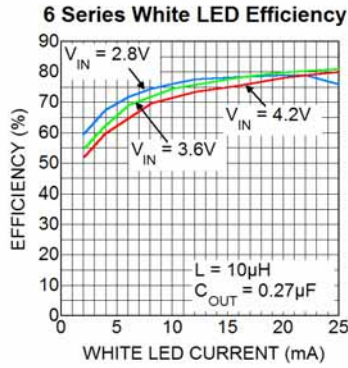
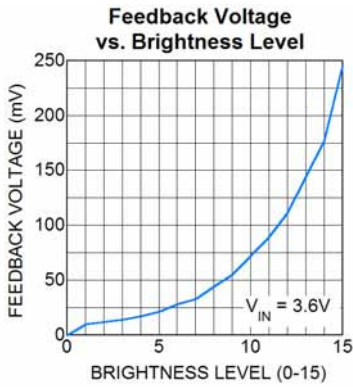
Symbol	Parameter	Condition	Min	Typ	Max	Units
V_{IN}	Supply Voltage Range		2.5		6.5	V
V_{UVLO}	Under-voltage Lockout		1.8	2.1	2.4	V
I_{VIN}	Quiescent Current	$V_{FB} > 500\text{mV}$		1.4	5	mA
I_{SD}	Shutdown Current (DC pin low)	$V_{DC} = 0\text{V}$ for $> 2\text{ms}$.		0.01	1	μA
V_{FB}	Feedback Voltage	(+/-5%)	237	250	263	mV
I_{FB}	Feedback Input Current	$V_{FB} = 250\text{mV}$		450		nA
	Line Regulation	$2.5\text{V} \leq V_{IN} \leq 4.5\text{V}$		0.5		%
	Load Regulation	$5\text{mA} \leq I_{OUT} \leq 20\text{mA}$		0.5		%
D_{MAX}	Maximum Duty Cycle		85	90		%
I_{SW}	Switch Current Limit	$V_{IN} = 3.6\text{V}$	500	750	1200	mA
V_{DC}	DC pin thresholds	High Low	1.1		0.4	V
	DC Pin Hysteresis			20		mV
I_{DC}	DC Pin Current	$V_{DC} = 3.6\text{V}$		5	10	μA
$t_{shutdown}$	Shutdown Pulse Width	$V_{IN} = 2.8\text{V}$ to 5.5V $V_{DC} = \text{Low}$	1260			μs
t_{MODE_UP}	Count UP mode pulse width	$V_{IN} = 2.8\text{V}$ to 5.5V $V_{DC} = \text{Low}$	100		160	μs
t_{MODE_DOWN}	Count Down mode pulse width	$V_{IN} = 2.8\text{V}$ to 5.5V $V_{DC} = \text{Low}$	420		500	μs
t_{start_up}	Turn-on Delay Time	$V_{IN} = 2.8\text{V}$ to 5.5V	140			μs
t_{prog_low}	Programming pulse width low	$V_{IN} = 2.8\text{V}$ to 5.5V	1		32	μs
t_{prog_high}	Programming pulse width high	$V_{IN} = 2.8\text{V}$ to 5.5V	1		32	μs
t_{delay}	Minimum Delay for mode change	$V_{IN} = 2.8\text{V}$ to 5.5V $V_{DC} = \text{High}$	140			μs
T_{prog_setup}	First Pulse Window for Preprogramming	$V_{IN} = 2.8\text{V}$ to 5.5V	35		50	μs
f_{SW}	Oscillator Frequency		1	1.2	1.35	MHz

Symbol	Parameter	Condition	Min	Typ	Max	Units
V_D	Schottky Forward Drop	$I_D = 150\text{mA}$		0.8	1	V
I_{RD}	Schottky Leakage Current	$V_R = 30\text{V}$			4	μA
V_{OVP}	Over Voltage Protection	3289- 16 only (nominal voltage)	13	14	16	V
		3289- 24 only (nominal voltage)	21	22.5	24	V
T_J	Over-Temperature Threshold Shutdown			150		$^{\circ}\text{C}$

Notes:

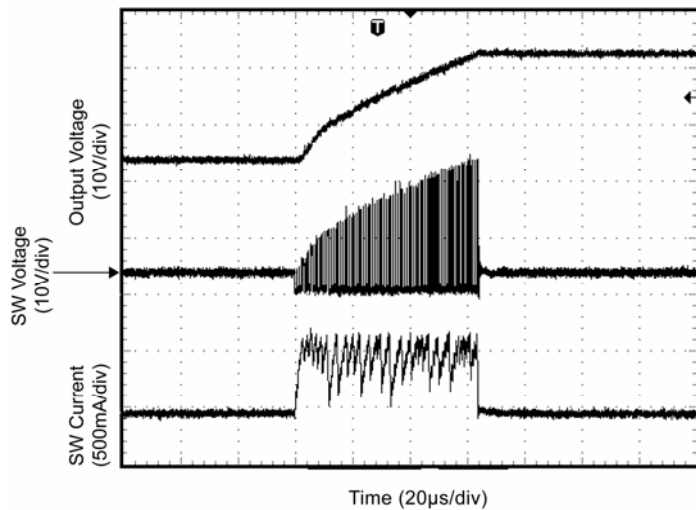
1. Absolute maximum ratings indicate limits beyond which damage to the component may occur. Electrical specifications do not apply when operating the device outside of its operating ratings. The maximum allowable power dissipation is a function of the maximum junction temperature, $T_{J(\text{Max})}$, the junction-to-ambient thermal resistance, θ_{JA} , and the ambient temperature, T_A . The maximum allowable power dissipation will result in excessive die temperature, and the regulator will go into thermal shutdown.
2. This device is not guaranteed to operate beyond its specified operating rating.
3. IC devices are inherently ESD sensitive. Handling precautions required.
4. Specification for packaged product only.

Typical Characteristics

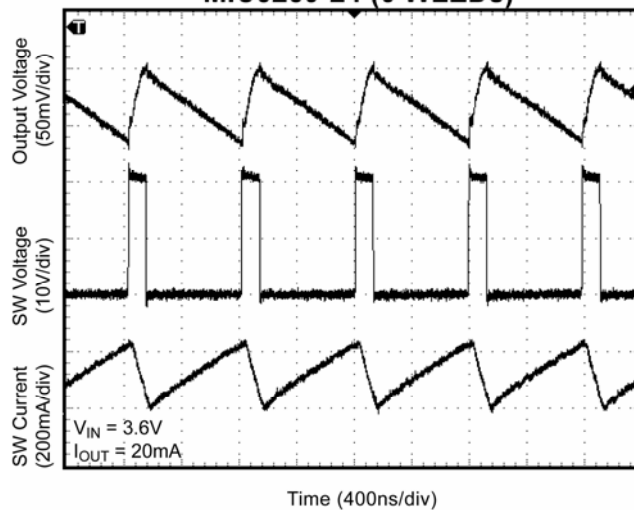


Functional Characteristics

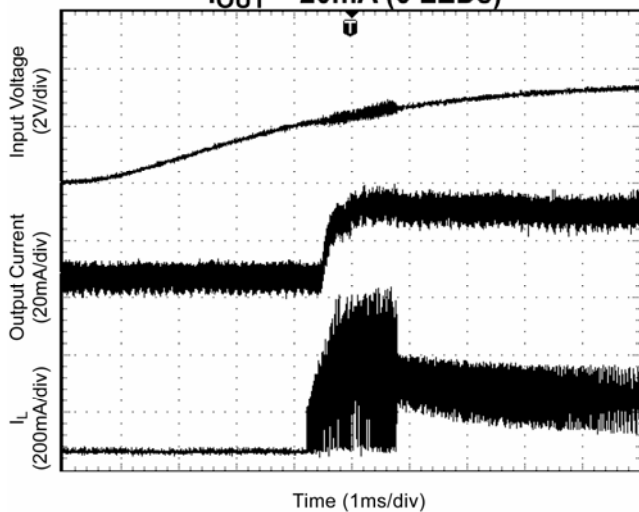
OVP Response MIC3289-24



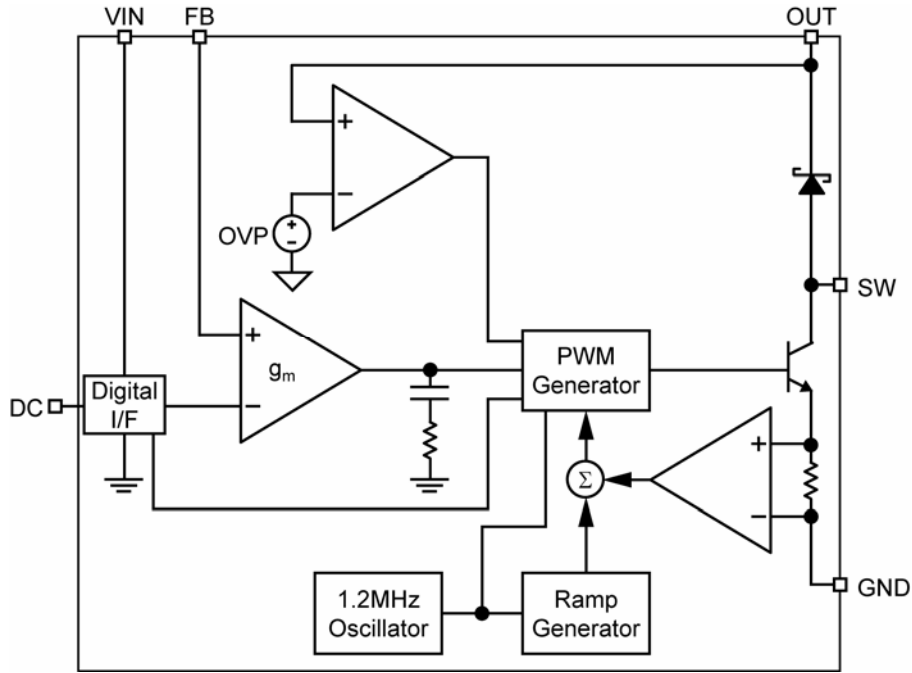
Steady State Operation MIC3289-24 (6 WLEDs)



**Start Up Waveform
 $I_{OUT} = 20mA$ (6 LEDs)**



Functional Diagram



MIC3289 Block Diagram

Functional Description

The MIC3289 is a constant frequency, PWM current mode boost regulator. It is composed of an oscillator, slope compensation ramp generator, current amplifier, gm error amplifier, PWM generator, bipolar output transistor, digital interface with D/A converter and Schottky rectifier diode. It features true one-wire digital control that may be used to vary the brightness of the output LEDs and to place the device into shutdown mode. The oscillator generates a 1.2MHz clock which triggers the PWM generator that turns on the output transistor and resets the slope compensation ramp generator. The current amplifier is used to measure the switch current by amplifying the voltage signal from the internal sense resistor. The output of the current amplifier is summed with the output of the slope compensation ramp generator. This summed current-loop signal is fed to one of the inputs of the PWM generator.

MIC3289 Block Diagram

The gm error amplifier measures the LED current through the external sense resistor and amplifies the error between the detected signal and the reference voltage indicated by the digital interface. The output of the gm error amplifier provides the voltage-loop signal that is fed to the other input of the PWM generator.

When the current-loop signal exceeds the voltage-loop signal, the PWM generator turns off the bipolar output transistor. The next clock period initiates the next switching cycle, maintaining the constant frequency current-mode PWM control. The LED current level at maximum brightness is set by the feedback resistor:

$$I_{LED} = \frac{250mV}{R_{LED}}$$

MIC3289 Digital Interface

The MIC3289 incorporates an easy to use single-wire, serial programming interface allowing users to set LED brightness to one of 16 levels spaced in a logarithmic manner. In contrast to other solutions requiring a PWM drive signal to maintain LED brightness, the MIC3289 is “set and forget”, relieving the controlling processor of the constant burden of supplying a drive signal. Additionally, brightness levels can be preset so that LEDs can be turned on at a particular brightness level.

State Diagram

The MIC3289 logic state flow is depicted in Figure 1 below. Brightness level changes are negative edge triggered while all other state changes require a logic high or low be applied to the DC pin for a specific length of time.

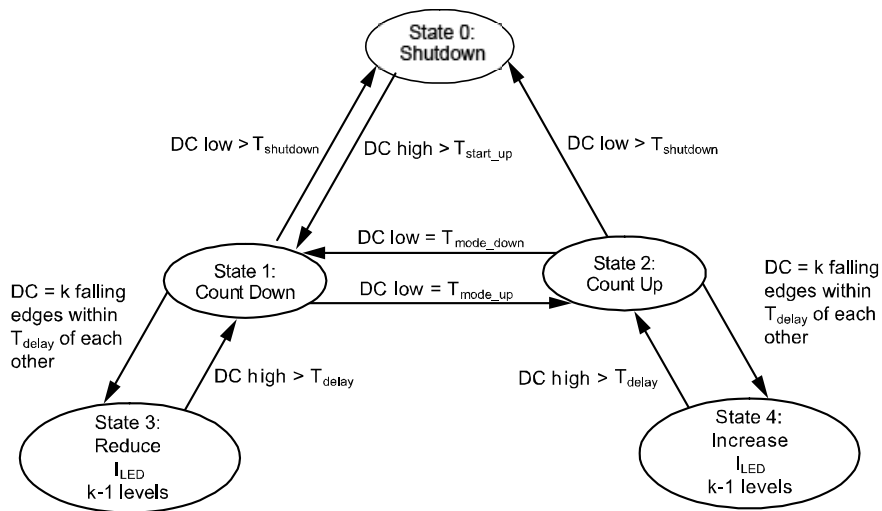


Figure 1. MIC3289 Logic State Diagram

With an input supply voltage between 2.5V and 6.5V and a logic-level LOW applied to the DC pin, the MIC3289 will enter State 0, shutdown, and remain there consuming less than 1µA.

Start Up

Presuming no presetting brightness command is issued (discussed in a later section), the MIC3289 will start-up in its default state approximately 140µs (t_{START_UP}) after a logic level high has been applied and maintained at the DC pin. In the default state the LED drive current is at the maximum brightness level of 15 and brightness counter is set to count down mode. Any falling edges during the t_{PROG_SETUP} period will cause the initial brightness level of the LEDs to be below the maximum brightness level. This is discussed in more detail in the *Presetting Brightness* section.

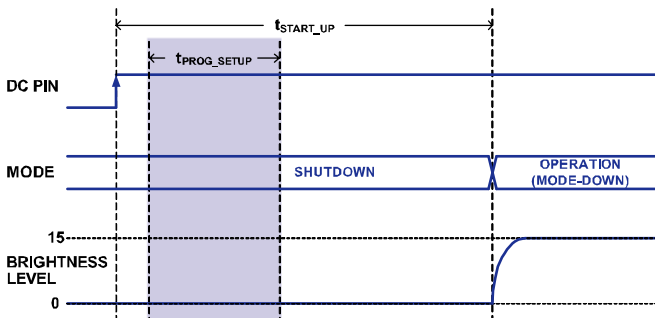


Figure 2. Typical Start-Up Timing

Shutdown

Whenever a logic-level LOW is applied to the DC input pin for a period greater than or equal to $t_{SHUTDOWN}$ (1260µs), the MIC3289 will return to State 0 entering its power saving shutdown mode.

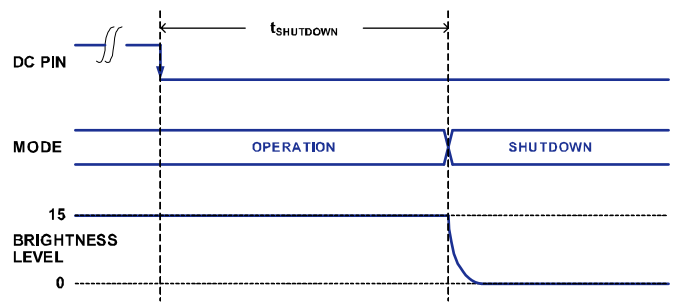


Figure 3. Shutdown Timing

Once the device is shutdown, the boost supply is disabled and the LEDs are turned off. Brightness level information stored in the MIC3289 prior to shutdown will be lost.

Programming Pulse Counter Modes

Referring to the state diagram in Figure 1, notice that there are two programming pulse counting modes. At power up the MIC3289 defaults to State 1, the Count Down Mode. The counting mode can be changed to State 2, the Count Up Mode, by pulling the DC pin low for a period equal to t_{MODE_UP} (100µs to 160µs). The device will remain in Count Up Mode until its state is changed to Count Down Mode or by disabling the MIC3289.

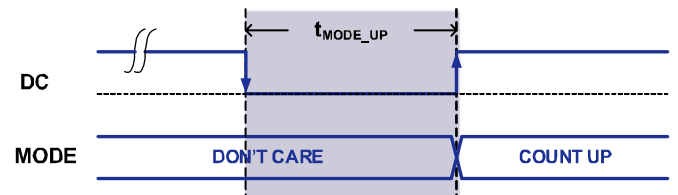


Figure 4. Mode Change to Count Up

To change the state back to Count Down Mode, pull the DC pin low for a period equal to t_{MODE_DOWN} (420 μ s to 500 μ s). Now the internal circuitry will remain in Count Down Mode until changed to Count Up as described previously.

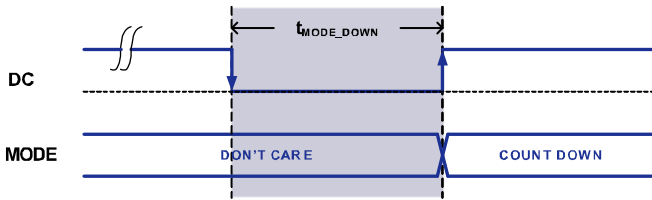


Figure 5. Mode Change to Count Down

Programming the Brightness Level

MIC3289 is designed to start driving the output LEDs (t_{START_UP}) in 140 μ s at the maximum brightness level of 15. After start up, the internal control logic is ready to decrease the LED brightness upon receiving programming pulses (negative edges applied to DC pin). Since MIC3289 starts in Count Down Mode, the brightness level is decreased one level by applying two programming pulses, as shown in Figure 6. Each programming pulse has a high (t_{PROG_HIGH}) and a low (t_{PROG_LOW}) pulse width that must be between 1 μ s to 32 μ s. Note that n+1 number of pulses are needed to decrease brightness by n level(s) since the first clock pulse is ignored. Ignoring the first clock pulse is necessary in order that Mode Change (t_{MODE_UP} , t_{MODE_DOWN}) pulses do not result in adjustments to the brightness level. The MIC3289 internal circuit can be changed from Count Down Mode to Count Up Mode and vice versa. The user may elect to send a Mode Change as shown in Figure 4 to set the MIC3289 to step up the brightness level with subsequent programming pulses. For proper operation, ensure the DC pin has remained high for at least t_{DELAY} (140 μ s) before issuing a mode change command.

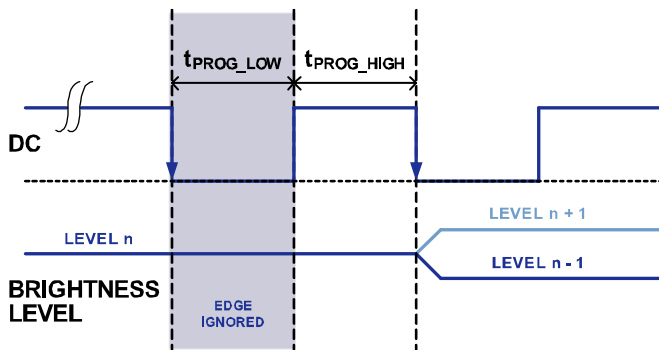


Figure 6. Brightness Programming Pulses

Brightness programming pulses are not restricted to just one pair at a time. Multiple level changes can be set as

shown in Figure 7 below. When issuing multiple brightness level adjustment commands to the DC pin, ensure both t_{PROG_LOW} and t_{PROG_HIGH} are within 1 μ s and 32 μ s.

To maintain operation at the current brightness level simply maintain a logic level high signal at the DC pin.

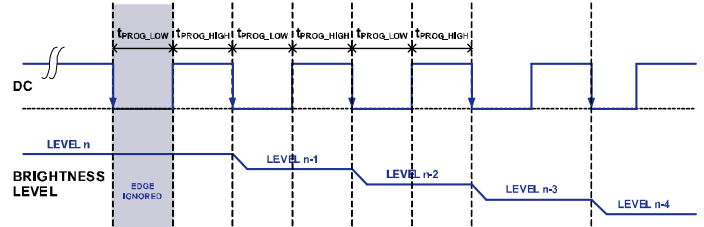


Figure 7. Decreasing Brightness Several Levels

As mentioned, MIC3289 can be programmed to set LED drive current to produce one of 16 distinct brightness levels. The internal logic keeps track of the brightness level with an Up/Down counter circuit. The following section explains how the brightness counter functions with continued programming edges.

Counter Roll-Over

The MIC3289 internal up/down counter contains registers from 0 to 15. When the brightness level is at 0 and a programming pulse forces the brightness to step down, then the counter will roll-over to level 15. This is illustrated in Figure 8 below.

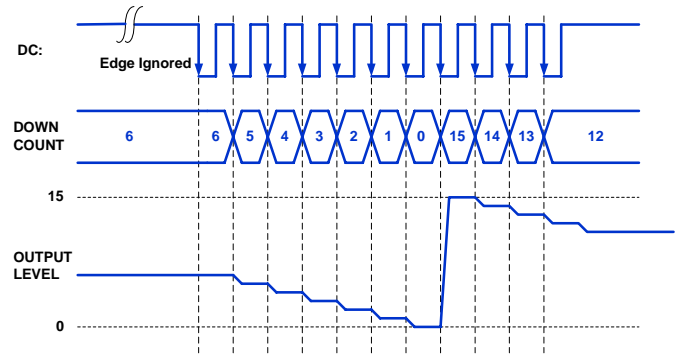


Figure 8. Down Counter Roll-over

Similarly, when the counter mode is set to Count Up and a programming pulse forces the brightness level to step up from level 15, then the counter will roll-over to level 0 as illustrated in figure 9.

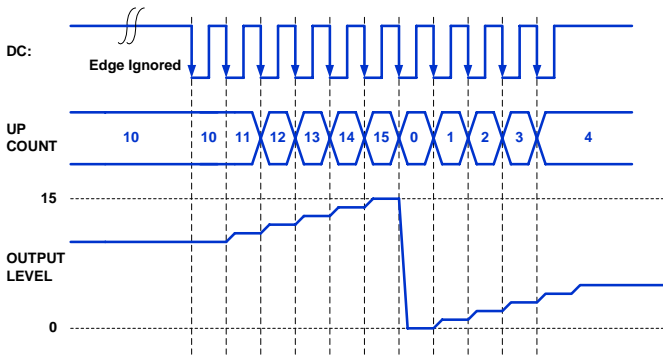


Figure 9. Up Counter Roll-over

One-Step Brightness Changes

For applications where a keypad button press is to be translated into a brightness level change, the following method of decreasing the brightness level may be useful. This “One-Step” brightness change procedure relieves the user from keeping track of the MIC3289’s up/down counter state. It combines a counter mode change with a programming pulse, therefore a one-step decrease in brightness is assured no matter what the previous up/down counter mode was.

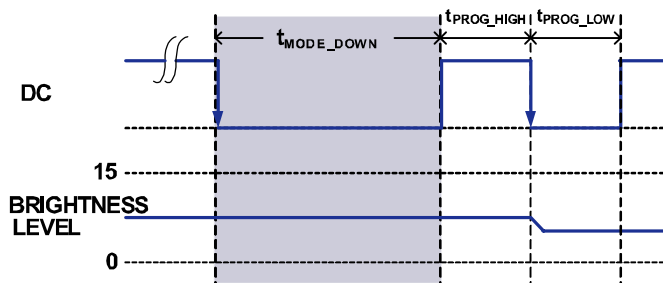


Figure 10. One-Step Brightness Decrease

This method is quite simple and the only requirement is that the first DC low period be equal to the t_{MODE_DOWN} (420µs to 500µs) and immediately followed by a falling edge within t_{PROG_HIGH} (1µs to 32µs) as shown in Figure 10 for One-Step Brightness Decrease. Similarly a one-step increase can be assured by first generating a DC down pulse whose period is equal to the t_{MODE_UP} (100µs to 160µs) and immediately followed by a falling edge within t_{PROG_HIGH} (1µs to 32µs). Figure 11 illustrates the proper timing for execution of a One-Step Brightness Increase.

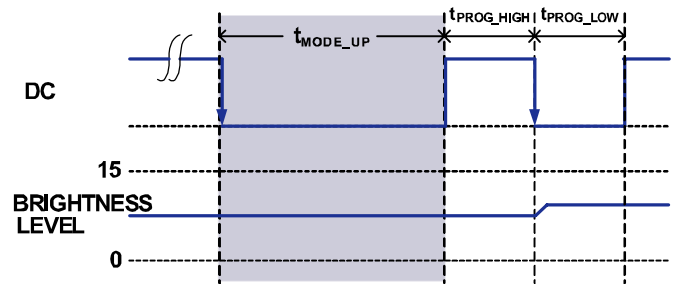


Figure 11. One-Step Brightness Increase

Presetting Brightness

The brightness level can be preset before the MIC3289 begins to drive the LEDs by sending a series of programming edges via the DC pin during the t_{START_UP} (140µs) period and within 35µs to 50µs after the DC pin is pulled high. The 15µs timeframe between 35µs and 50µs is the t_{PROG_SETUP} period. The MIC3289 does not drive current into the load until DC pin is kept high for t_{START_UP} (140µs) after presetting has concluded in order to grant the user sufficient time to preset LED brightness. The first presetting pulse edge must occur somewhere between the timeframe of 35µs to 50µs after DC pin is first pulled HIGH otherwise the MIC3289 may continue to start at the full (default) brightness level.

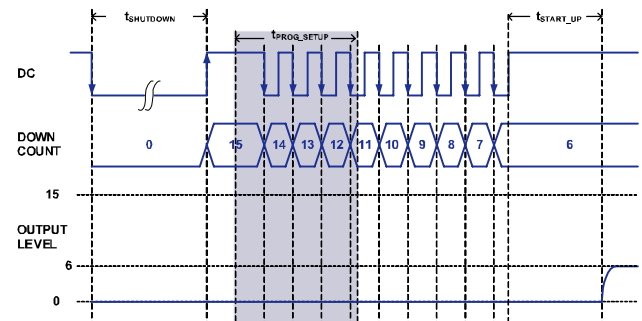


Figure 12. Presetting Timing

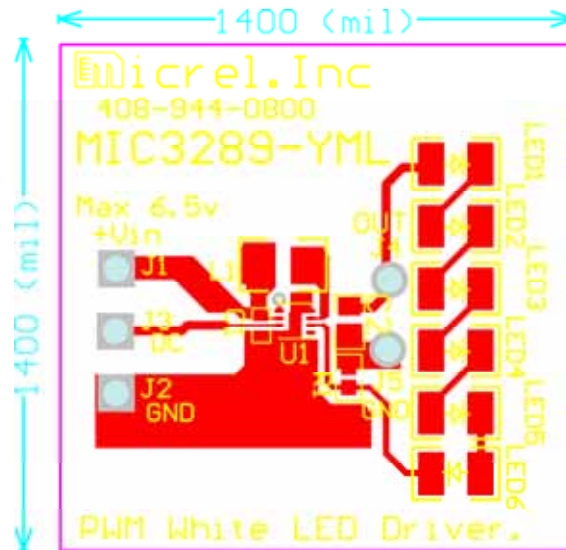
Figure 12 shows the correct presetting sequence to set the MIC3289 brightness to level 6 prior to start up. The sequence is initiated by driving the DC pin low for a period exceeding $t_{SHUTDOWN}$ (1260µs) to insure that the part has entered the power saving shutdown state erasing all brightness level state and mode setting. Then the DC pin is driven high and the first presetting pulse edge is entered within the t_{PROG_SETUP} window. Notice that when using the presetting feature the first programming pulse is not ignored. This is because the counter’s default mode is Count Down and a Mode Change cannot be performed in the presetting mode. (Note that the same timing requirements of standard brightness programming also apply during presetting brightness.)

External Component Selection

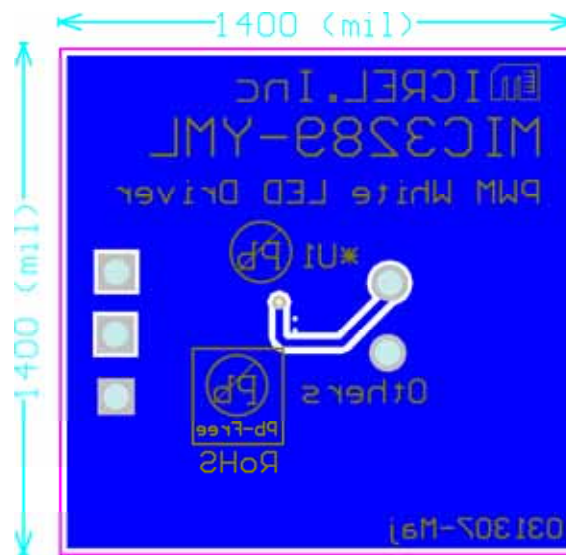
The MIC3289 can be used across a wide range of applications. The table below shows recommended inductor and output capacitor values for applications driving 3-6 LEDs in series assuming a 20mA maximum drive current from Li-Ion battery source.

Series LEDs	L	Manufacturer	Min C _{OUT}	Manufacturer
3	22μH	LQH43MN220K03 (Murata)	2.2μF	0603YD225MAT2A (AVX)
		NLC453232T-220K (TDK)		GRM188R61C225KE15D (Murata)
	10μH	LQH43MN100K03 (Murata)	0.33μF	0603YD334MAT2A (AVX)
		NLCV32T-100K-PFR (TDK)		GRM188RT1C224KA01D (Murata)
	4.7μH	LQH43MN4R7K03 (Murata)	0.22μF	06036ZD224MAT2A (AVX)
		NLCV32T-4R7M-PFR (TDK)		
4	22μH	LQH43MN220K03 (Murata)	1.0μF	0805YD105MAT (AVX)
		NLC453232T-220K (TDK)		GRM188R61E105KA12D (Murata)
	10μH	LQH43MN100K03 (Murata)	0.33μF	06033D334MAT2A (AVX)
		NLCV32T-100K-PFR (TDK)		GRM21BR71E334KA01L (Murata)
	4.7μH	LQH43MN4R7K03 (Murata)	0.27μF	VJ0805Y274KXAAT (Vishay)
		NLCV32T-4R7M-PFR (TDK)		
5.6	22μH	LQH43MN220K03 (Murata)	0.33μF	06033D334MAT2A (AVX)
		NLC453232T-220K (TDK)		GRM21BR71E334KA01L (Murata)
	10μH	LQH43MN100K03 (Murata)	0.27μF	VJ0805Y274KXAAT (Vishay)
		NLCV32T-100K-PFR (TDK)		
	4.7μH	LQH43MN4R7K03 (Murata)	0.22μF	06036ZD224MAT2A (AVX)
		NLCV32T-4R7M-PFR (TDK)		

Layout Recommendations

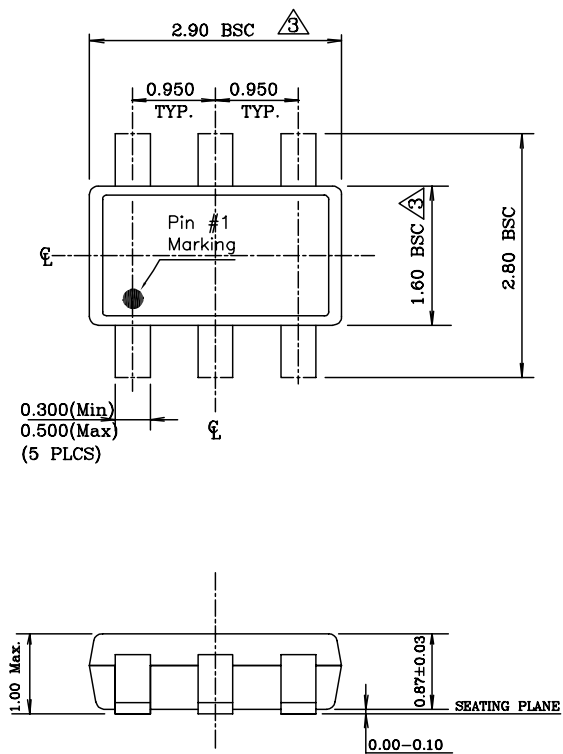


Top



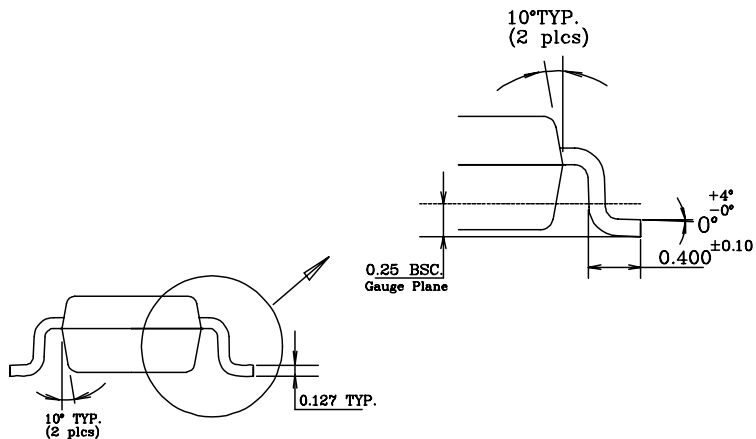
Bottom

Package Information



NOTE:

1. Dimensions and tolerances are as per ANSI Y14.5M, 1994.
2. Die is facing up for mold. Die is facing down for trim/form, ie. reverse trim/form.
- ⚠ Dimensions are exclusive of mold flash and gate burr.
4. The footlength measuring is based on the gauge plane method.
5. All specification comply to Jedec Spec M0193 Issue C.
6. All dimensions are in millimeters.



6-Pin TSOT23 (D)

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