



**THE DATASHEET OF
LT1097CN8**



FEATURES

- Offset Voltage 50 μ V Max
- Offset Voltage Drift 1 μ V/ $^{\circ}$ C Max
- Bias Current 250pA Max
- Offset Current 250pA Max
- Bias and Offset Current Drift 4pA/ $^{\circ}$ C Max
- Supply Current 560 μ A Max
- 0.1Hz to 10Hz Noise 0.5 μ Vp-p, 2.2pAp-p
- CMRR 115dB Min
- Voltage Gain 117dB Min
- PSRR 114dB Min
- Guaranteed Operation on Two NiCad Batteries

APPLICATIONS

- Replaces OP-07/OP-77/OP-97/OP-177/AD707/LT1001 with Improved Price/Performance
- High Impedance Difference Amplifiers
- Logarithmic Amplifiers (Wide Dynamic Range)
- Thermocouple Amplifiers
- Precision Instrumentation
- Active Filters (with Small Capacitors)

DESCRIPTION


LT[®]1097 achieves a new standard in combining low price and outstanding precision performance.

On all operational amplifier data sheets, the specifications listed on the front page are for highly selected, expensive grades, while the specs for the low cost grades are buried deep in the data sheet.

The LT1097 does not have any selected grades, the outstanding specifications shown in the Features section are for its only grade.

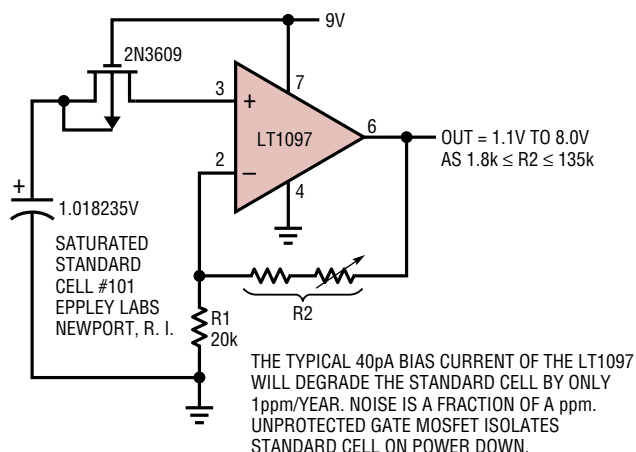
The design effort of the LT1097 concentrated on optimizing the performance of all precision specs—at only 350 μ A of supply current. Typical values are 10 μ V offset voltage, 40pA bias and offset currents, 0.2 μ V/ $^{\circ}$ C and 0.4pA/ $^{\circ}$ C drift. Common mode and power supply rejections, voltage gain are typically in excess of 128dB.

All parameters that are important for precision, low power op amps have been optimized. Consequently, using the LT1097 error budget calculations in most applications is unnecessary.

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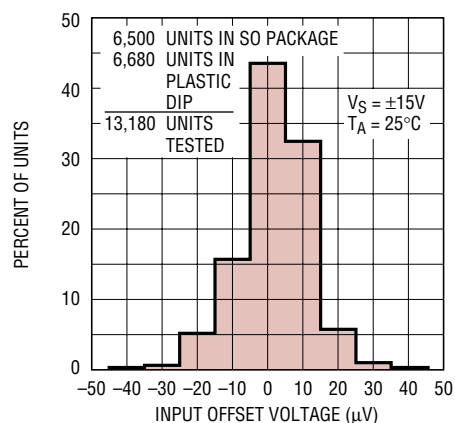
TYPICAL APPLICATION

Saturated Standard Cell Amplifier



LT1097*TA01

Input Offset Voltage Distribution



LT1097*G01

ABSOLUTE MAXIMUM RATINGS

Supply Voltage $\pm 20V$
 Differential Input Current (Note 1) $\pm 10mA$
 Input Voltage $\pm 20V$
 Output Short Circuit Duration Indefinite

Operating Temperature Range $-40^{\circ}C$ to $85^{\circ}C$
 Storage Temperature Range $-65^{\circ}C$ to $150^{\circ}C$
 Lead Temperature (Soldering, 10 sec) $300^{\circ}C$

PACKAGE/ORDER INFORMATION

| | | | |
|--|---|---|--|
| <p>TOP VIEW</p> <p>N8 PACKAGE 8-LEAD PLASTIC DIP</p> | <p>ORDER PART NUMBER</p> <p>LT1097CN8</p> | <p>TOP VIEW</p> <p>S8 PACKAGE 8-LEAD PLASTIC SO</p> | <p>ORDER PART NUMBER</p> <p>LT1097S8</p> |
|--|---|---|--|

ELECTRICAL CHARACTERISTICS $V_S = \pm 15V, V_{CM} = 0V, T_A = 25^{\circ}C$, unless otherwise noted.

| SYMBOL | PARAMETER | CONDITIONS | LT1097CN8 | | | LT1097S8 | | | UNITS |
|-------------------------------------|--|----------------------------------|------------|-----------------|-----------|------------|-------------------------|-----------|----------------------------------|
| | | | MIN | TYP | MAX | MIN | TYP | MAX | |
| V_{OS} | Input Offset Voltage | | | 10 | 50 | | 10 | 60 | μV |
| $\frac{\Delta V_{OS}}{\Delta TIME}$ | Long Term Input Offset Voltage Stability | | | 0.3 | | | 0.3 | | $\mu V/Mo$ |
| I_{OS} | Input Offset Current | | | 40 | 250 | | 60 | 350 | μA |
| I_B | Input Bias Current | | | ± 40 | ± 250 | | ± 50 | ± 350 | μA |
| e_n | Input Noise Voltage | 0.1Hz to 10Hz | | 0.5 | | | 0.5 | | μV_{p-p} |
| | Input Noise Voltage Density | $f_0 = 10Hz$ $f_0 = 1000Hz$ | | 16 14 | | | 16 14 | | nV/\sqrt{Hz} nV/\sqrt{Hz} |
| i_n | Input Noise Current | 0.1Hz to 10Hz | | 2.2 | | | 2.4 | | pA_{p-p} |
| | Input Noise Current Density | $f_0 = 10Hz$ $f_0 = 1000Hz$ | | 0.03 0.008 | | | 0.035 0.008 | | pA/\sqrt{Hz} pA/\sqrt{Hz} |
| | Input Resistance Differential Mode Common Mode | (Note 2) | 30 | 80 10^{12} | | 25 | 70 $8 \cdot 10^{11}$ | | $M\Omega$ Ω |
| | Input Voltage Range | | ± 13.5 | ± 14.3 | | ± 13.5 | ± 14.3 | | V |
| CMRR | Common Mode Rejection Ratio | $V_{CM} = \pm 13.5V$ | 115 | 130 | | 115 | 130 | | dB |
| PSRR | Power Supply Rejection Ratio | $V_S = \pm 1.2V$ to $\pm 20V$ | 114 | 130 | | 114 | 130 | | dB |
| A_{VOL} | Large Signal Voltage Gain | $V_O = \pm 12V, R_L = 10k$ | 700 | 2500 | | 700 | 2500 | | V/mV |
| | | $V_O = \pm 10V, R_L = 2k$ | 250 | 1000 | | 250 | 1000 | | V/mV |
| V_{OUT} | Output Voltage Swing | $R_L = 10k$ | ± 13 | ± 13.8 | | ± 13 | ± 13.8 | | V |
| | | $R_L = 2k$ | ± 11.5 | ± 13 | | ± 11.5 | ± 13 | | V |
| SR | Slew Rate | | 0.1 | 0.2 | | 0.1 | 0.2 | | $V/\mu s$ |
| GBW | Gain Bandwidth Product | | | 700 | | | 700 | | kHz |
| I_S | Supply Current | | | 350 | 560 | | 350 | 560 | μA |
| | Offset Adjustment Range | $R_{POT} = 10k$, Wiper to V^+ | | ± 600 | | | ± 600 | | μV |
| | Minimum Supply Voltage | (Note 3) | ± 1.2 | — | | ± 1.2 | — | | V |

ELECTRICAL CHARACTERISTICS $V_S = \pm 15V$, $V_{CM} = 0V$, $0^\circ C \leq T_A \leq 70^\circ C$, unless otherwise noted.

| SYMBOL | PARAMETER | CONDITIONS | | LT1097CN8 | | | LT1097S8 | | | UNITS |
|-----------|---|--------------------------------------|---|------------|------------|-----------|------------|------------|-----------|------------------|
| | | | | MIN | TYP | MAX | MIN | TYP | MAX | |
| V_{OS} | Input Offset Voltage | | ● | | 20 | 100 | | 20 | 130 | μV |
| | Average Temperature Coefficient of Input Offset Voltage | (Note 4) | ● | | 0.2 | 1 | | 0.2 | 1.4 | $\mu V/^\circ C$ |
| I_{OS} | Input Offset Current | | ● | | 60 | 430 | | 75 | 570 | μA |
| | Average Temperature Coefficient of Input Offset Current | (Note 4) | ● | | 0.4 | 4 | | 0.5 | 5 | $\mu A/^\circ C$ |
| I_B | Input Bias Current | | ● | | ± 60 | ± 430 | | ± 75 | ± 570 | μA |
| | Average Temperature Coefficient of Input Bias Current | (Note 4) | ● | | 0.4 | 4 | | 0.5 | 5 | $\mu A/^\circ C$ |
| A_{VOL} | Large Signal Voltage Gain | $V_{OUT} = \pm 12V$, $R_L \geq 10k$ | ● | 450 | 2000 | | 450 | 2000 | | V/mV |
| | | $V_{OUT} = \pm 10V$, $R_L \geq 2k$ | ● | 180 | 800 | | 180 | 800 | | V/mV |
| CMRR | Common Mode Rejection Ratio | $V_{CM} = \pm 13.5V$ | ● | 112 | 128 | | 112 | 128 | | dB |
| PSRR | Power Supply Rejection Ratio | $V_S = \pm 1.3V$ to $\pm 20V$ | ● | 111 | 128 | | 111 | 128 | | dB |
| | Input Voltage Range | | ● | ± 13.5 | ± 14.2 | | ± 13.5 | ± 14.2 | | V |
| V_{OUT} | Output Voltage Swing | $R_L = 10k$ | ● | ± 13 | ± 13.7 | | ± 13 | ± 13.7 | | V |
| I_S | Supply Current | | ● | | 380 | 700 | | 380 | 700 | μA |

ELECTRICAL CHARACTERISTICS $V_S = \pm 15V$, $V_{CM} = 0V$, $-40^\circ C \leq T_A \leq 85^\circ C$, unless otherwise noted. (Note 5)

| SYMBOL | PARAMETER | CONDITIONS | | LT1097CN8 | | | LT1097S8 | | | UNITS |
|-----------|---|--------------------------------------|---|------------|------------|-----------|------------|------------|-----------|------------------|
| | | | | MIN | TYP | MAX | MIN | TYP | MAX | |
| V_{OS} | Input Offset Voltage | | ● | | 25 | 130 | | 30 | 170 | μV |
| | Average Temperature Coefficient of Input Offset Voltage | | ● | | 0.3 | 1.2 | | 0.3 | 1.6 | $\mu V/^\circ C$ |
| I_{OS} | Input Offset Current | | ● | | 70 | 600 | | 85 | 750 | μA |
| | Average Temperature Coefficient of Input Offset Current | | ● | | 0.5 | 5 | | 0.6 | 6 | $\mu A/^\circ C$ |
| I_B | Input Bias Current | | ● | | ± 70 | ± 600 | | ± 85 | ± 750 | μA |
| | Average Temperature Coefficient of Input Bias Current | | ● | | 0.5 | 5 | | 0.6 | 6 | $\mu A/^\circ C$ |
| A_{VOL} | Large Signal Voltage Gain | $V_{OUT} = \pm 12V$, $R_L \geq 10k$ | ● | 300 | 1700 | | 300 | 1700 | | V/mV |
| | | $V_{OUT} = \pm 10V$, $R_L \geq 2k$ | ● | | 700 | | | 700 | | V/mV |
| CMRR | Common Mode Rejection Ratio | $V_{CM} = \pm 13.5V$ | ● | 108 | 127 | | 108 | 127 | | dB |
| PSRR | Power Supply Rejection Ratio | $V_S = \pm 1.5V$ to $\pm 20V$ | ● | 108 | 127 | | 108 | 127 | | dB |
| | Input Voltage Range | | ● | ± 13.5 | ± 14 | | ± 13.5 | ± 14 | | V |
| V_{OUT} | Output Voltage Swing | $R_L = 10k$ | ● | ± 13 | ± 13.6 | | ± 13 | ± 13.6 | | V |
| I_S | Supply Current | | ● | | 400 | 800 | | 400 | 800 | μA |

The ● denotes specifications which apply over the full operating temperature range.

Note 1: Differential input voltages greater than 1V will cause excessive current to flow through the input protection diodes unless limiting resistance is used.

Note 2: This parameter is guaranteed by design and is not tested.

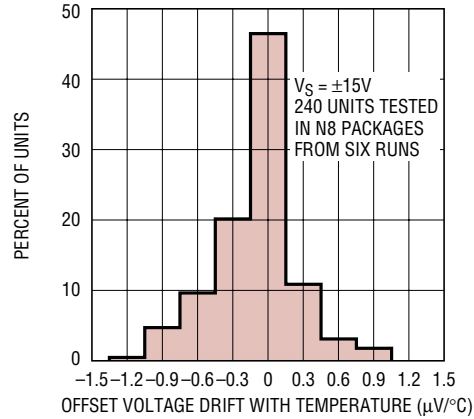
Note 3: Power supply rejection ratio is measured at the minimum supply voltage.

Note 4: This parameter is not 100% tested.

Note 5: The LT1097 is designed, characterized and expected to meet these extended temperature limits, but is not tested at $-40^\circ C$ and $85^\circ C$. Guaranteed I grade parts are available; consult factory.

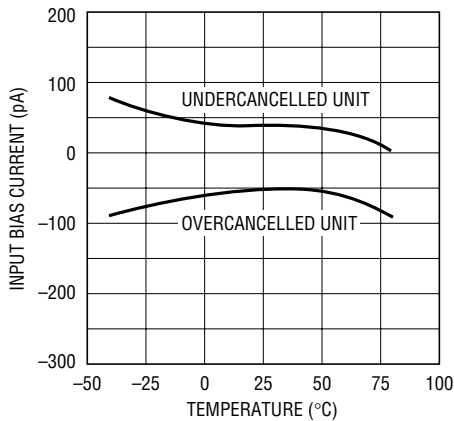
TYPICAL PERFORMANCE CHARACTERISTICS

Distribution to Offset Voltage Drift with Temperature



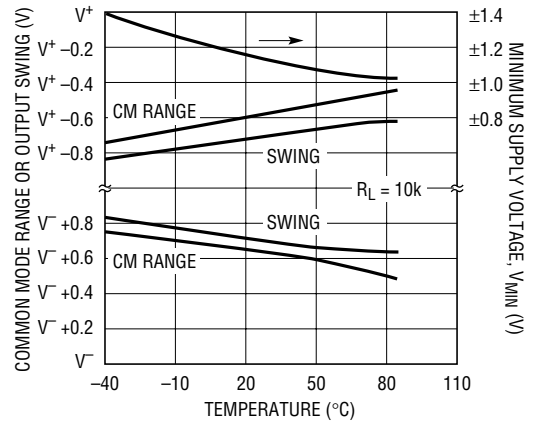
LT1097 • G02

Input Bias Current vs Temperature



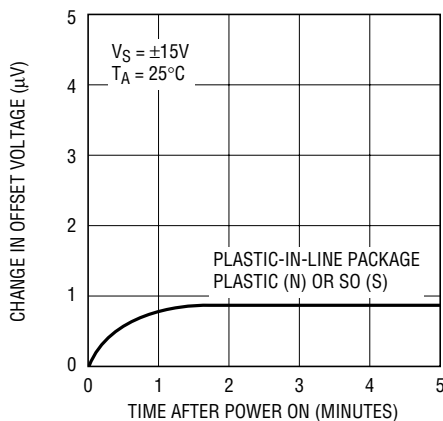
1097 • G03

Minimum Supply Voltage, Common Mode Range and Voltage Swing at V_{MIN}



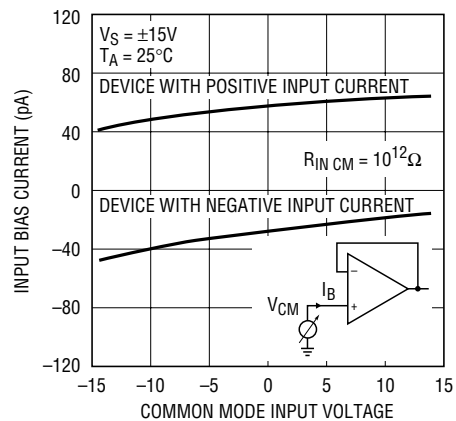
1097 • G04

Warm-Up Drift



1097 • G05

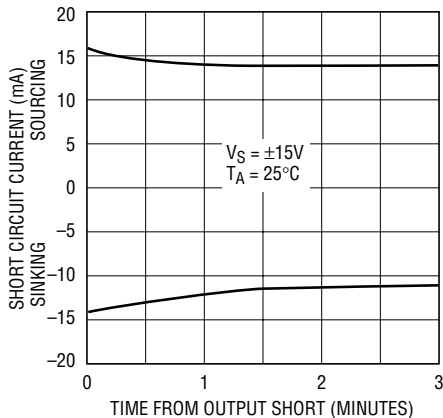
Input Bias Current Over Common Mode Range



1097 • G06

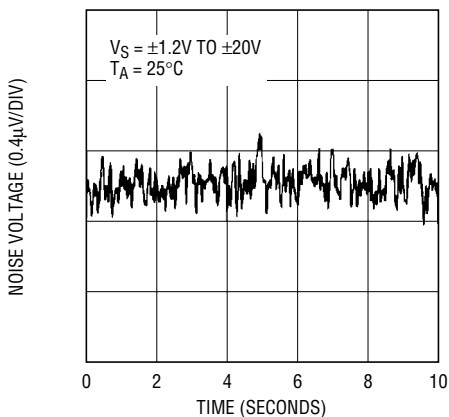
TYPICAL PERFORMANCE CHARACTERISTICS

Output Short Circuit Current vs Time



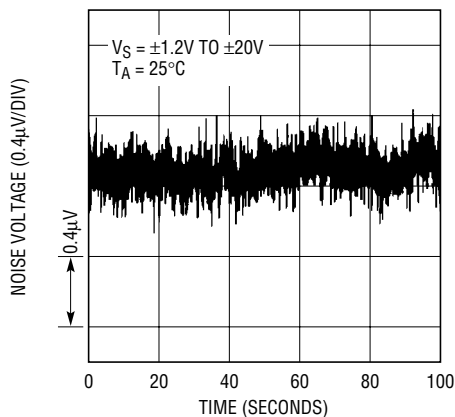
1097 • G07

0.1Hz to 10Hz Noise



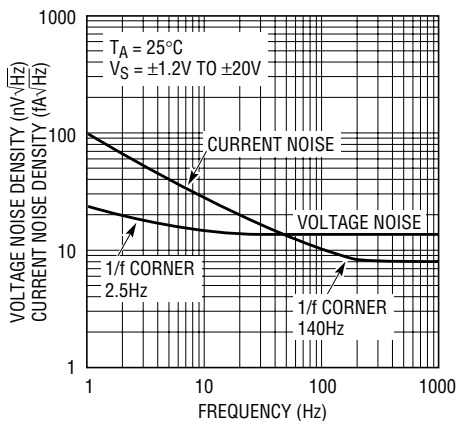
1097 • G08

0.01Hz to 10Hz Noise



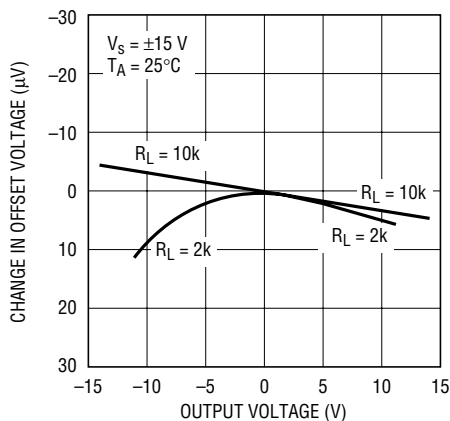
1097 • G09

Noise Spectrum



1097 • G10

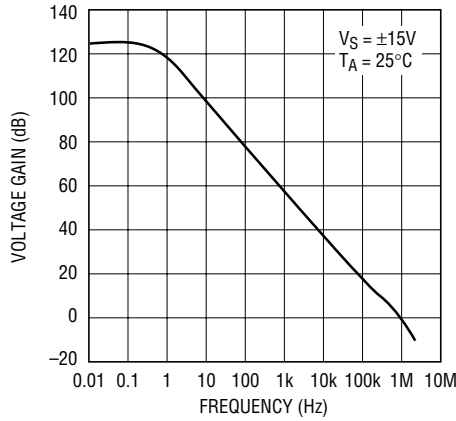
Voltage Gain



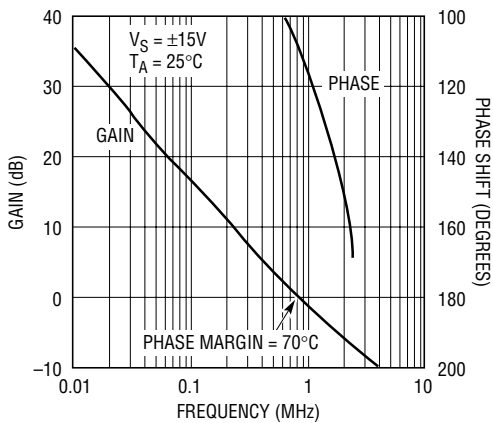
1097 • G11

TYPICAL PERFORMANCE CHARACTERISTICS

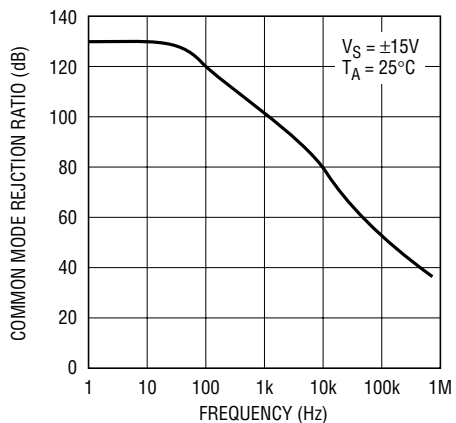
Voltage Gain vs Frequency



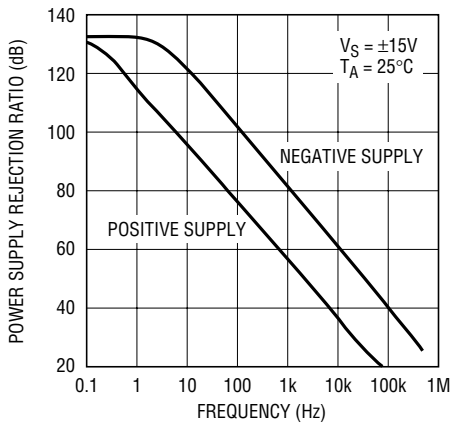
Gain, Phase Shift vs Frequency



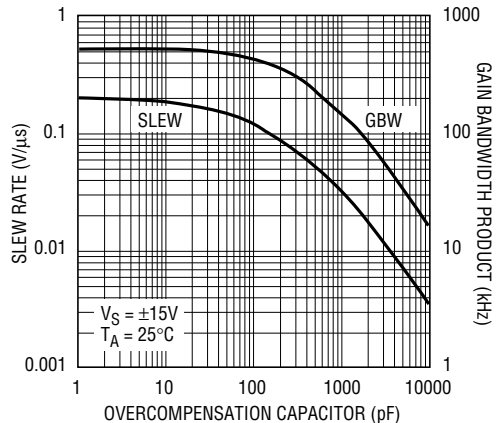
Common Mode Rejection vs Frequency



Power Supply Rejection vs Frequency

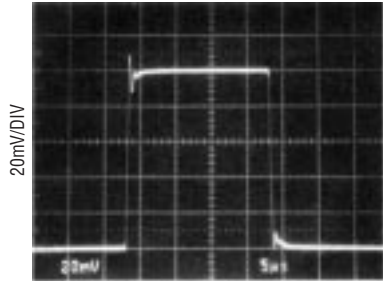


Slew Rate, Gain Bandwidth Product vs Overcompensation Capacitor



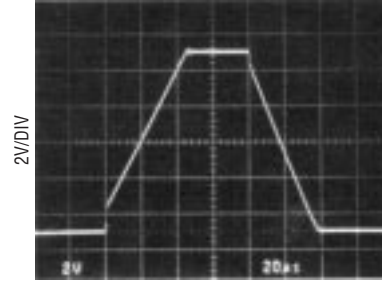
TYPICAL PERFORMANCE CHARACTERISTICS

Small Signal Transient Response



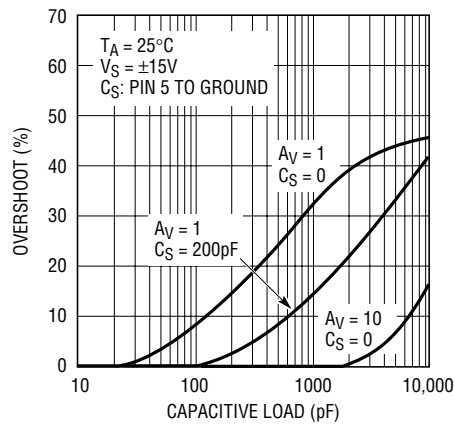
$A_V = 1$, $C_{LOAD} = 100\text{pF}$, $5\mu\text{s}/\text{DIV}$ 1097 G17

Large Signal Transient Response



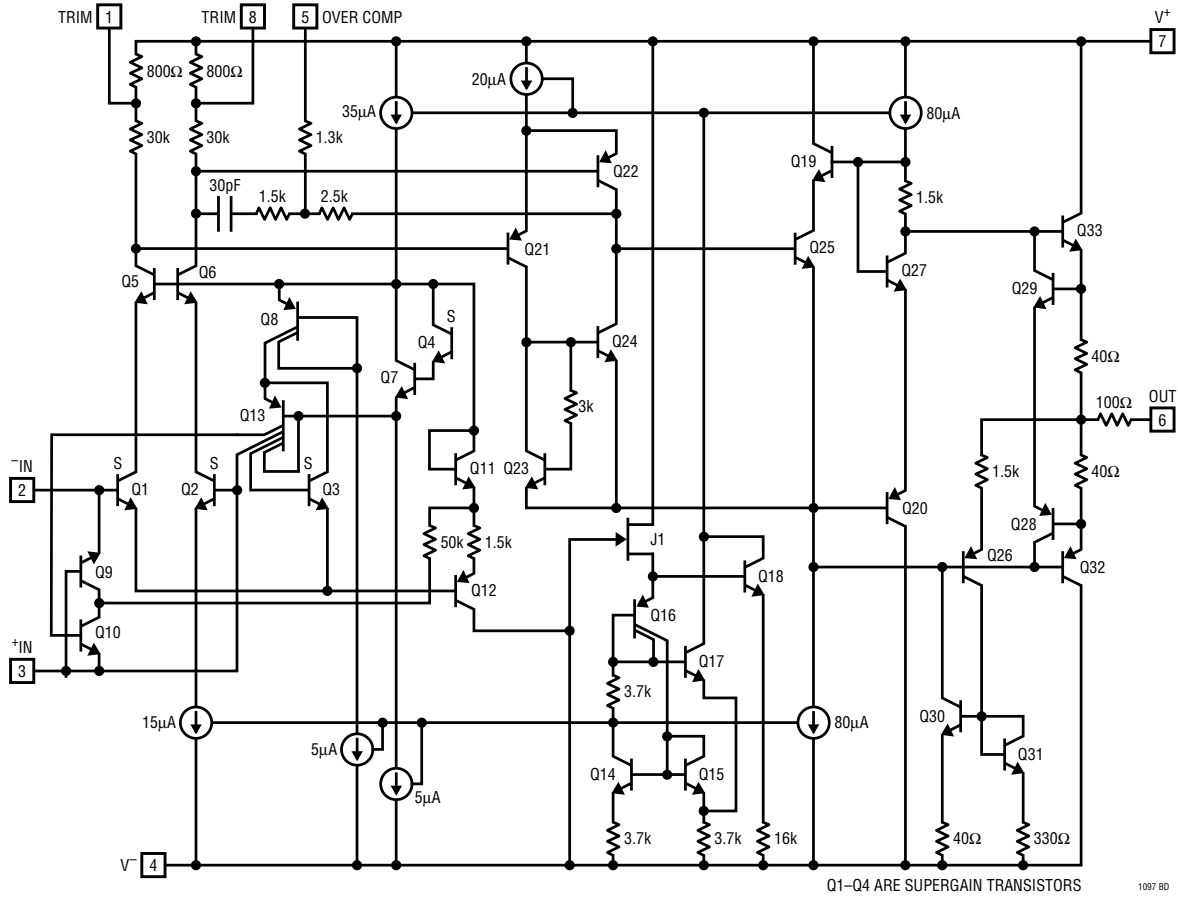
$A_V = 1$, $20\mu\text{s}/\text{DIV}$ 1097 G18

Capacitive Load Handling



1097 G19

SCHEMATIC DIAGRAM



APPLICATIONS INFORMATION

The LT1097 is pin compatible to and directly replaces such precision op amps as the OP-07, OP-77, AD707, OP-97, OP-177, LM607 and LT1001 with improved price/performance. Compatibility includes externally nulling the offset voltage, as all of the above devices are trimmed with a potentiometer between Pin 1 and Pin 8 and the wiper tied to V^+ .

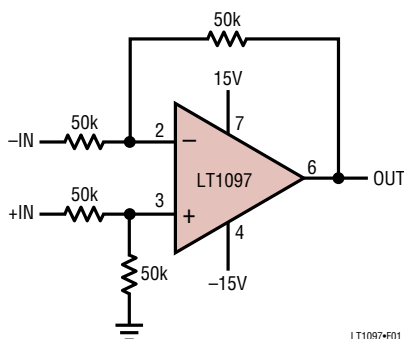
The simple difference amplifier can be used to illustrate the all-around excellence of the LT1097. The 50k input resistance is selected to be large enough compared to input signal source resistance. Simultaneously, the 50k resistors should not dominate the precision and noise error budget. Assuming perfect matching between the four resistors, the following table summarizes the input

referred performance obtained using the LT1097 and other popular, low cost precision op amps.

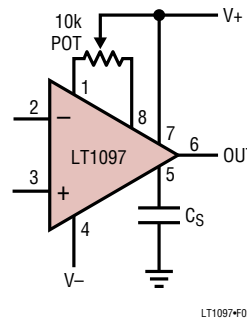
Input offset voltage can be adjusted over a $\pm 600\mu\text{V}$ range with a 10k potentiometer.

The LT1097 is internally compensated for unity gain stability. As shown on the Capacitive Load Handling plot, the LT1097 is stable with any capacitive load. However, the overcompensation capacitor, C_S , can be used to reduce overshoot with heavy capacitive loads, to narrow noise bandwidth or to stabilize circuits with gain in the feedback loop.

$\pm 27\text{V}$ Common Mode Range Difference Amplifier



Frequency Compensation and Optional Offset Nulling



Guaranteed Performance, $V_S = \pm 15\text{V}$, $T_A = 25^\circ\text{C}$

| PARAMETER | LT1097CN8 | OP-77GP | AD707JN | OP-177GP | OP-97FP | UNITS |
|--|-----------|---------|---------|----------|---------|------------------------------|
| Error Terms | | | | | | |
| V_{OS} Max | 50 | 100 | 90 | 60 | 75 | μV |
| I_{OS} Max•25k | 6 | 70 | 50 | 70 | 4 | μV |
| Gain Min, 10V Out | 14 | 5 | 3 | 5 | 50 | μV |
| CMRR, Min, $\pm 25\text{V}$ In | 22 | 20 | 13 | 22 | 39 | μV |
| PSRR, Min, $V_S = \pm 15\text{V} \pm 10\%$ | 6 | 9 | 9 | 9 | 9 | μV |
| Sum of All Error Terms | 98 | 204 | 165 | 166 | 177 | μV |
| 0.1Hz to 10Hz Noise | | | | | | |
| Voltage Noise | 0.5 | 0.38 | 0.23 | 0.38 | 0.5 | $\mu\text{Vp-p Typ}$ |
| Current Noise•50k | 0.11 | 0.75 | 0.7 | 0.75 | 0.1 | $\mu\text{Vp-p Typ}$ |
| Resistor Noise | 0.55 | 0.55 | 0.55 | 0.55 | 0.55 | $\mu\text{Vp-p Typ}$ |
| RMS sum | 0.75 | 1 | 0.92 | 1 | 0.75 | $\mu\text{Vp-p}$ |
| Drift with Temp | | | | | | |
| TCV_{OS} Max | 1 | 1.2 | 1 | 1.2 | 2 | $\mu\text{V}/^\circ\text{C}$ |
| TCI_{OS} Max•25k | 0.1 | 2.1 | 1 | 2.1 | 0.2 | $\mu\text{V}/^\circ\text{C}$ |
| Sum of Drift Terms | 1.1 | 3.3 | 2 | 3.3 | 2.2 | $\mu\text{V}/^\circ\text{C}$ |
| Supply Current Max | 0.56 | 2 | 3 | 2 | 0.6 | mA |

APPLICATIONS INFORMATION

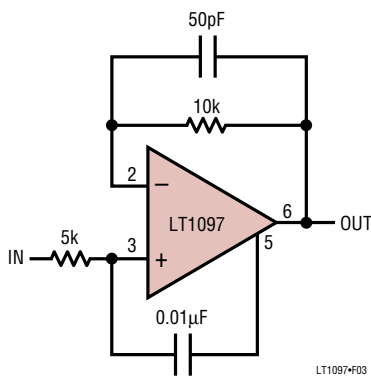
The availability of the compensation terminal permits the use of feedforward frequency compensation to enhance slew rate. The voltage follower feedforward scheme bypasses the amplifier's gain stages and slews at nearly 10V/ μ s.

The inputs of the LT1097 are protected with back-to-back diodes. In the voltage follower configuration, when the input is driven by a fast, large signal pulse (>1V), the input

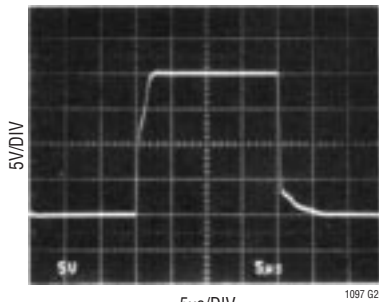
protection diodes effectively short the output to the input during slewing, and a current, limited only by the output short circuit protection will flow through the diodes.

The use of a feedback resistor, as shown in the voltage follower feedforward diagram, is recommended because this resistor keeps the current below the short circuit limit, resulting in faster recovery and settling of the output.

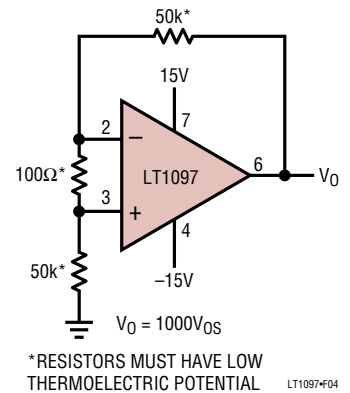
Follower Feedforward Compensation



Pulse Response of Feedforward Compensation

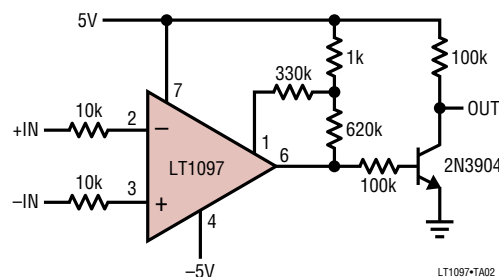


Test Circuit for Offset Voltage and its Drift with Temperature



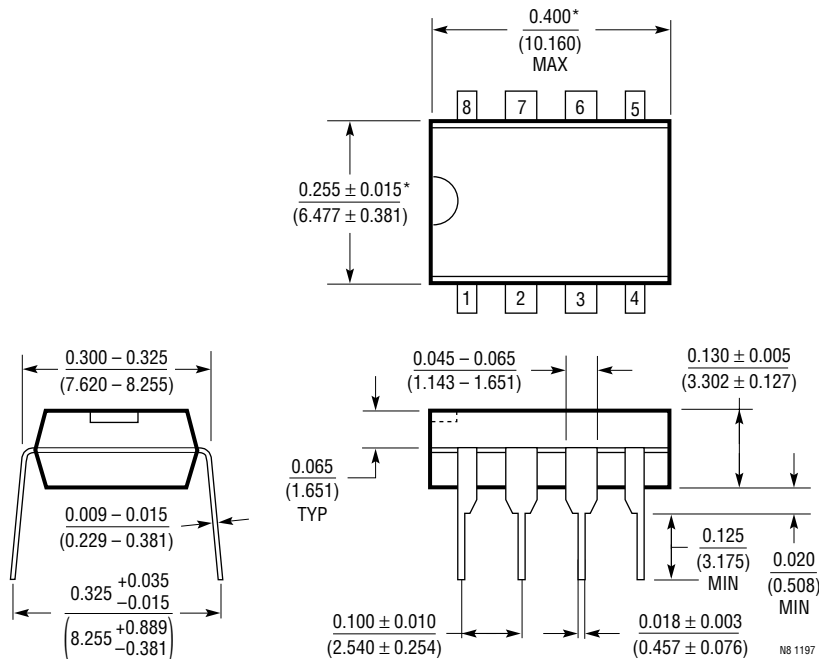
TYPICAL APPLICATION

Low Power Comparator with <10 μ V Hysteresis



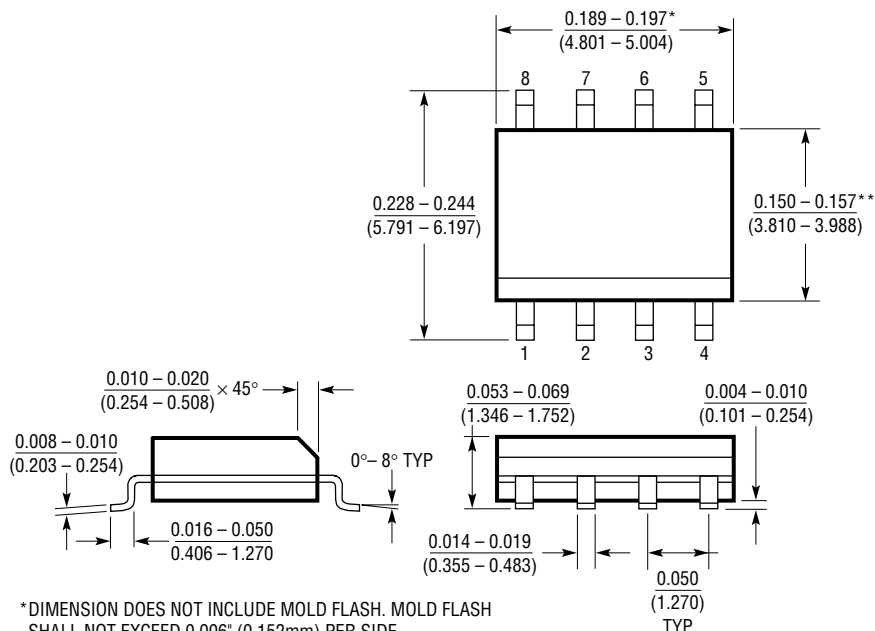
PACKAGE DESCRIPTION Dimensions in inches (millimeters) unless otherwise noted.

N8 Package
8-Lead PDIP (Narrow 0.300)
 (LTC DWG # 05-08-1510)



*THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
 MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED 0.010 INCH (0.254mm)

S8 Package
8-Lead Plastic Small Outline (Narrow 0.150)
 (LTC DWG # 05-08-1610)

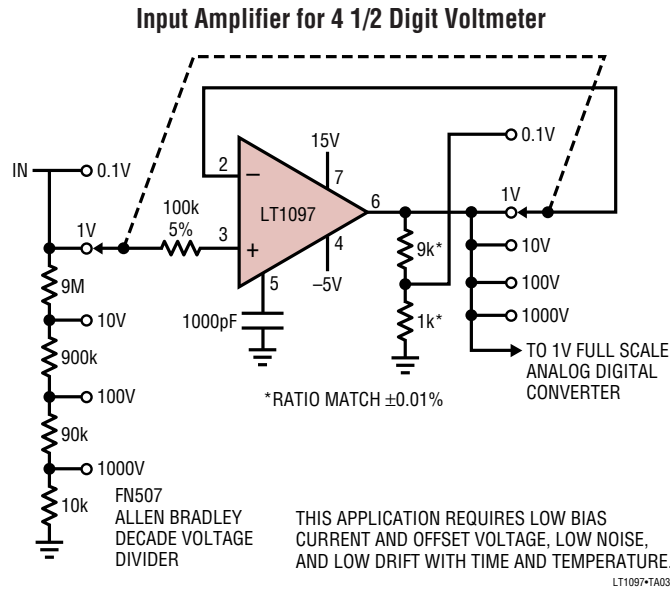


*DIMENSION DOES NOT INCLUDE MOLD FLASH. MOLD FLASH SHALL NOT EXCEED 0.006" (0.152mm) PER SIDE

**DIMENSION DOES NOT INCLUDE INTERLEAD FLASH. INTERLEAD FLASH SHALL NOT EXCEED 0.010" (0.254mm) PER SIDE

S08 0996

TYPICAL APPLICATION



RELATED PARTS

| PART NUMBER | DESCRIPTION | COMMENTS |
|---------------|---|---|
| LT1490/LT1491 | Dual/Quad General Purpose Micropower Rail-to-Rail Op Amps | Over-The-Top™ Inputs, 50µA Supply Current Per Amplifier, 2V to 44V Supply Range, 180kHz GBW |
| LT1492/LT1493 | Dual/Quad 5MHz Low Power Single Supply Op Amps | 180µV V_{OS} Max, 3V/µs Slew Rate, 550µA Supply Per Amplifier |
| LT1077 | Single Micropower Low V_{OS} Op Amp | 60µV V_{OS} Max, 68µA Supply Current, 230kHz GBW, Optimized for 5V Supplies |

Over-The-Top is a trademark of Linear Technology Corporation.

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 [Linear Technology](#) Information

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