



**THE DATASHEET OF
TLV1851QDBVRQ1**



TLV185x-Q1 and TLV186x-Q1 Family of 40 V, Nanopower Comparators

1 Features

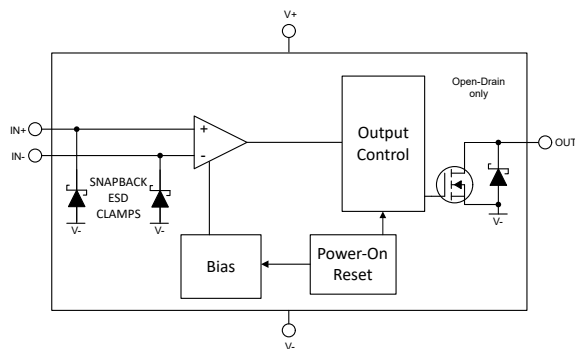
- Qualified for automotive applications
- AEC-Q100 qualified with the following results:
 - Device temperature grade 1: -40°C to 125°C ambient operating temperature range
 - Device HBM ESD classification level H1C
 - Device CDM ESD classification level C6
- Low supply current: 440 nA per channel
- Wide supply range: 1.8 V to 40 V
- Over-the-rail inputs: common-mode range extends 40 V above (V-) independent of (V+)
- Fail-safe: high impedance inputs with no supply
- Power-on-reset provides a known startup condition
- No phase reversal for overdriven inputs
- Reverse battery protection up to 40 V
- Push-pull output option (TLV185x-Q1)
- Open-drain output option (TLV186x-Q1)

2 Applications

- [Telematics eCall](#)
- [Automotive head unit](#)
- [Instrument Cluster](#)
- [On-board \(OBC\) & wireless chargers](#)

3 Description

The TLV185x-Q1 and TLV186x-Q1 are a family of Automotive grade nanopower, 40 Volt comparators with single, dual and quad channel options. The family offers fail-safe (FS) inputs with push-pull and open-drain output options. These features coupled with nanopower operation over the wide supply range of 1.8 V to 40 V make this family well-suited for house-keeping functions such as voltage and temperature monitoring in low-power, always-on systems.



Block Diagram

All devices include a Power-On Reset (POR) feature that ensures the output is in a known state until the minimum supply voltage has been reached before the output responds to the inputs, thus preventing false outputs during system power-up and power-down.

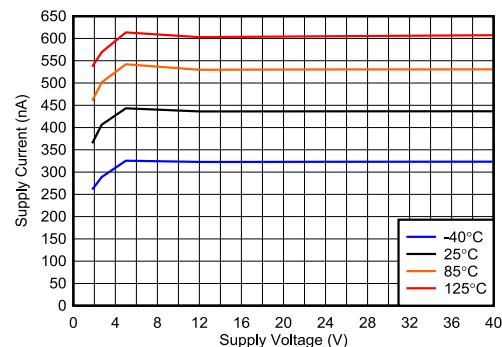
The inputs have over-the-rail capability where both inputs can exceed the supply voltage up to 40 V and still operate properly. This makes the comparators well suited for both high and low supply voltage systems without limiting the range of input voltages that can be compared. Likewise, the internal reverse battery protection feature prevents damage to the comparator in the event of improper battery installation to the supply pins.

The TLV185x-Q1 comparators have a push-pull output stage whereas the TLV186x-Q1 comparators have an open-drain output stage, making it appropriate for level translation.

Device Information

| PART NUMBER | PACKAGE ⁽¹⁾ | BODY SIZE (NOM) |
|---------------------------|------------------------|-------------------|
| TLV1851-Q1, TLV1861-Q1 | SOT-23 (5) | 1.60 mm x 2.90 mm |
| TLV1852-Q1, TLV1862-Q1 | SOIC (8) (Preview) | 3.91 mm x 4.90 mm |
| | VSSOP (8) (Preview) | 3.00 mm x 3.00 mm |
| TLV1854-Q1, TLV1864-Q1 | SOIC (14) (Preview) | 3.91 mm x 8.65 mm |
| | TSSOP (14) (Preview) | 4.40 mm x 5.00 mm |

- (1) For all available packages, see the orderable addendum at the end of the data sheet.



Supply Current vs. Supply Voltage



Table of Contents

| | | | |
|---|----|--|----|
| 1 Features | 1 | 7.3 Feature Description..... | 16 |
| 2 Applications | 1 | 7.4 Device Functional Modes..... | 16 |
| 3 Description | 1 | 8 Application and Implementation | 20 |
| 4 Revision History | 2 | 8.1 Application Information..... | 20 |
| 5 Pin Configuration and Functions | 3 | 8.2 Typical Applications..... | 23 |
| Pin Configuration: TLV1851-Q1 and TLV1861-Q1 | 3 | 8.3 Power Supply Recommendations..... | 25 |
| Pin Configurations: TLV1852-Q1 and TLV1862-Q1 | 4 | 9 Layout | 27 |
| Pin Configurations: TLV1854-Q1 and TLV1864-Q1 | 5 | 9.1 Layout Guidelines..... | 27 |
| 6 Specifications | 6 | 9.2 Layout Example..... | 27 |
| 6.1 Absolute Maximum Ratings..... | 6 | 10 Device and Documentation Support | 28 |
| 6.2 ESD Ratings..... | 6 | 10.1 Documentation Support..... | 28 |
| 6.3 Thermal Information..... | 6 | 10.2 Receiving Notification of Documentation Updates.. | 28 |
| 6.4 Recommended Operating Conditions..... | 7 | 10.3 Support Resources..... | 28 |
| 6.5 Electrical Characteristics..... | 8 | 10.4 Trademarks..... | 28 |
| 6.6 Switching Characteristics..... | 9 | 10.5 Electrostatic Discharge Caution..... | 28 |
| 6.7 Typical Characteristics..... | 10 | 10.6 Glossary..... | 28 |
| 7 Detailed Description | 16 | 11 Mechanical, Packaging, and Orderable Information | 28 |
| 7.1 Overview..... | 16 | | |
| 7.2 Functional Block Diagrams..... | 16 | | |

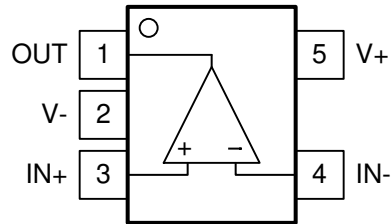
4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

| DATE | REVISION | NOTES |
|-------------|----------|-----------------|
| August 2023 | * | Initial Release |

5 Pin Configuration and Functions

Pin Configuration: TLV1851-Q1 and TLV1861-Q1



DBV Package
SOT-23-5
Top View
(Standard "north west" pinout)

Table 5-1. Pin Functions: TLV1851-Q1 and TLV1861-Q1

| PIN | | I/O | DESCRIPTION |
|------|-----|-----|-------------------------|
| NAME | NO. | | |
| OUT | 1 | O | Output |
| V- | 2 | - | Negative supply voltage |
| IN+ | 3 | I | Non-inverting (+) input |
| IN- | 4 | I | Inverting (-) input |
| V+ | 5 | - | Positive supply voltage |

Pin Configurations: TLV1852-Q1 and TLV1862-Q1

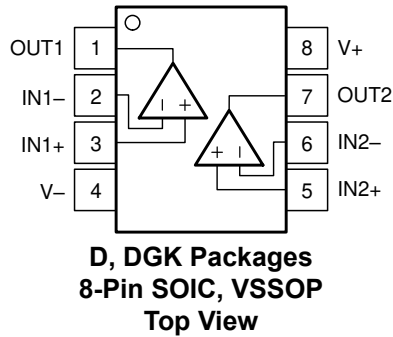
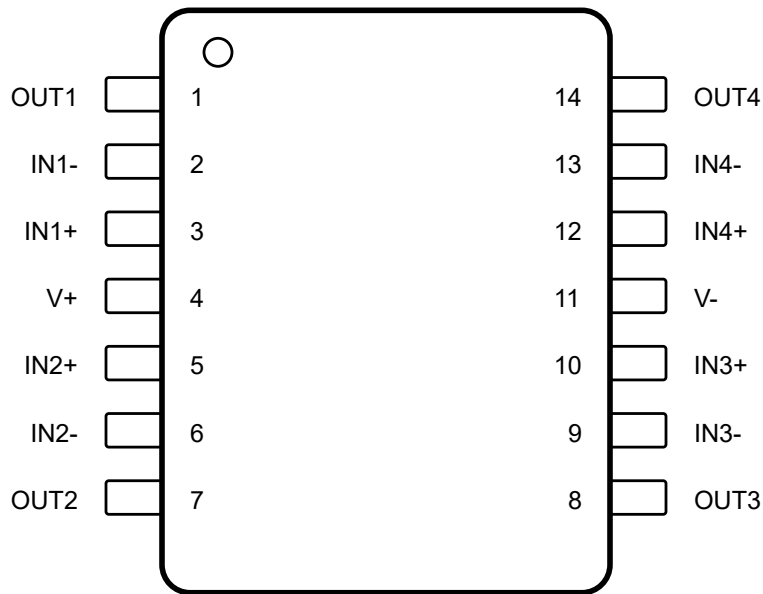


Table 5-2. Pin Functions: TLV1852-Q1 and TLV1862-Q1

| PIN | | I/O | DESCRIPTION |
|------|-----|-----|--|
| NAME | NO. | | |
| OUT1 | 1 | O | Output pin of the comparator 1 |
| IN1- | 2 | I | Inverting input pin of comparator 1 |
| IN1+ | 3 | I | Noninverting input pin of comparator 1 |
| V- | 4 | — | Negative supply voltage |
| IN2+ | 5 | I | Noninverting input pin of comparator 2 |
| IN2- | 6 | I | Inverting input pin of comparator 2 |
| OUT2 | 7 | O | Output pin of the comparator 2 |
| V+ | 8 | — | Positive supply voltage |

Pin Configurations: TLV1854-Q1 and TLV1864-Q1



Not to scale

**D, PW Packages
14-Pin SOIC, TSSOP
Top View**

Table 5-3. Pin Functions: TLV1854-Q1 and TLV1864-Q1

| PIN | | I/O | DESCRIPTION |
|------|-----|-----|--|
| NAME | NO. | | |
| OUT1 | 1 | O | Output pin of the comparator 1 |
| IN1- | 2 | I | Negative input pin of the comparator 1 |
| IN1+ | 3 | I | Positive input pin of the comparator 1 |
| V+ | 4 | - | Positive supply voltage |
| IN2+ | 5 | I | Positive input pin of the comparator 2 |
| IN2- | 6 | I | Negative input pin of the comparator 2 |
| OUT2 | 7 | O | Output pin of the comparator 2 |
| OUT3 | 8 | O | Output pin of the comparator 3 |
| IN3- | 9 | I | Negative input pin of the comparator 3 |
| IN3+ | 10 | I | Positive input pin of the comparator 3 |
| V- | 11 | - | Negative supply voltage |
| IN4+ | 12 | I | Positive input pin of the comparator 4 |
| IN4- | 13 | I | Negative input pin of the comparator 4 |
| OUT4 | 14 | O | Output pin of the comparator 4 |

6 Specifications

6.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)⁽¹⁾

| | MIN | MAX | UNIT |
|---|------|------------|------|
| Supply voltage: $V_S = (V+) - (V-)$ | | 42 | V |
| Differential Input Voltage, VID | -42 | 42 | V |
| Input pins (IN+, IN-) from (V-) ⁽²⁾ | -0.3 | 42 | V |
| Current into Input pins (IN+, IN-) ⁽³⁾ | -10 | 10 | mA |
| Output (Open-drain version only) from (V-) ⁽⁴⁾ | -0.3 | 42 | V |
| Output (OUT) (Push-Pull) from (V-) | -0.3 | (V+) + 0.3 | V |
| Output short circuit current ⁽⁵⁾ | -10 | 10 | mA |
| Junction temperature, T_J | | 150 | °C |
| Storage temperature, T_{stg} | -65 | 150 | °C |

- Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- Input terminals are diode-clamped to (V-). Inputs (IN+, IN-) can be greater than (V+) and OUT as long as it is within the -0.3 V to 42 V range
- Input terminals are diode-clamped to (V-). Input signals that swing more than 0.3 V below (V-) must be current-limited to 10 mA or less.
- Output (OUT) for open drain can be greater than (V+) and inputs (IN+, IN-) as long as it is within the -0.3 V to 42 V range
- Short-circuit to (V-) or (V+).

6.2 ESD Ratings

| | | | VALUE | UNIT |
|-------------|-------------------------|---|-------|------|
| $V_{(ESD)}$ | Electrostatic discharge | Human-body model (HBM), per AEC Q100-002 ⁽¹⁾ | ±2000 | V |
| $V_{(ESD)}$ | Electrostatic discharge | Charged-device model (CDM), per AEC Q100-0111 | ±1000 | V |

- AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Thermal Information

| THERMAL METRIC ⁽¹⁾ | | TLV185x/6x | | | | | UNIT |
|-------------------------------|--|--------------|----------|-------------|------------|----------|------|
| | | DBV (SOT-23) | D (SOIC) | DGK (VSSOP) | PW (TSSOP) | D (SOIC) | |
| | | 5 Pins | 8 Pins | 8 Pins | 14 Pins | 14 Pins | |
| R_{qJA} | Junction-to-ambient thermal resistance | 168.1 | 121.6 | 163.1 | | | °C/W |
| $R_{qJC(top)}$ | Junction-to-case (top) thermal resistance | 68.1 | 64.6 | 55.5 | | | °C/W |
| R_{qJB} | Junction-to-board thermal resistance | 37.4 | 65.1 | 84.7 | | | °C/W |
| γ_{JT} | Junction-to-top characterization parameter | 11.4 | 18.1 | 5.7 | | | °C/W |
| γ_{JB} | Junction-to-board characterization parameter | 37.1 | 64.3 | 83.1 | | | °C/W |
| $R_{qJC(bot)}$ | Junction-to-case (bottom) thermal resistance | | | | | | °C/W |

- For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

6.4 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

| | MIN | MAX | UNIT |
|---|------|-----|------|
| Supply voltage: $V_S = (V+) - (V-)$ | 1.8 | 40 | V |
| Input voltage range from (V-) | -0.1 | 40 | V |
| Common-mode input voltage range from (V-) | 0 | 40 | V |
| Output voltage for open drain from (V-) | -0.1 | 40 | V |
| Ambient temperature, T_A | -40 | 125 | °C |

6.5 Electrical Characteristics

For $V_S = (V+) - (V-) = 12V$, $V_{CM} = V_S/2$ at $T_A = 25^\circ C$ (Unless otherwise noted)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|---------------------------|--|---|------|------------|------|------------------|
| OFFSET VOLTAGE | | | | | | |
| V_{OS} | Input offset voltage | | -3.6 | ± 0.25 | 3.6 | mV |
| V_{OS} | Input offset voltage | $T_A = -40^\circ C$ to $+125^\circ C$ | -4.4 | | 4.4 | mV |
| dV_{IO}/dT | Input offset voltage drift | | | 3 | | $\mu V/^\circ C$ |
| V_{HYS} | Input hysteresis voltage | | 1 | 2.8 | 5 | mV |
| $V_{CM-Range}$ | Common-mode voltage range from (V-) | $V_S = 1.8 V$ to $40 V$ $T_A = -40^\circ C$ to $+125^\circ C$ | 0 | | 40 | V |
| POWER SUPPLY | | | | | | |
| I_Q | Quiescent current per comparator (output high) | Push Pull Output Option | | 520 | 750 | nA |
| I_Q | Quiescent current per comparator (output high) | Push Pull Output Option, $T_A = -40^\circ C$ to $125^\circ C$ | | | 1000 | nA |
| I_Q | Quiescent current per comparator (output high) | Open Drain Output option, no pull-up resistor | | 440 | 640 | nA |
| I_Q | Quiescent current per comparator (output high) | Open Drain Output option, no pull-up resistor, $T_A = -40^\circ C$ to $125^\circ C$ | | | 850 | nA |
| V_{POR} | | During power on, V_S must exceed V_{POR} for t_{ON} before the output will reflect the input. | | 1.5 | | V |
| INPUT BIAS CURRENT | | | | | | |
| I_B | Input bias current ⁽¹⁾ | | | 1 | 250 | pA |
| | | $T_A = -40^\circ C$ to $+125^\circ C$ | | | 1500 | pA |
| I_{OS} | Input offset current ⁽¹⁾ | | | 0.1 | 100 | pA |
| | | $T_A = -40^\circ C$ to $+125^\circ C$ | | | 1000 | pA |
| OUTPUT | | | | | | |
| V_{OL} | Voltage swing from (V-) | $I_{SINK} = 2 \mu A$ | | 1 | | mV |
| V_{OL} | Voltage swing from (V-) | $I_{SINK} = 50 \mu A$ | | 20 | 60 | mV |
| | | $I_{SINK} = 50 \mu A$ $T_A = -40^\circ C$ to $+125^\circ C$ | | | 100 | mV |
| V_{OH} | Voltage swing from (V+) (Push Pull only) | $I_{SOURCE} = 2 \mu A$ | | 1 | | mV |
| V_{OH} | Voltage swing from (V+) (Push Pull only) | $I_{SOURCE} = 50 \mu A$ | | 25 | 60 | mV |
| | | $I_{SOURCE} = 50 \mu A$ $T_A = -40^\circ C$ to $+125^\circ C$ | | | 100 | mV |
| I_{LKG} | Open-drain output leakage current | $V_{ID} = +0.1 V$, $V_{PULLUP} = (V+)$ | | 0.3 | | pA |
| I_{OL} | Short-circuit current | Sinking $T_A = -40^\circ C$ to $+125^\circ C$ | | 7 | | mA |

6.5 Electrical Characteristics (continued)

For $V_S = (V+) - (V-) = 12V$, $V_{CM} = V_S/2$ at $T_A = 25^\circ C$ (Unless otherwise noted)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|-----------|-----------------------|--|-----|-----|-----|------|
| I_{OH} | Short-circuit current | Sourcing (for Push-Pull only) $T_A = -40^\circ C$ to $+125^\circ C$ | | 5 | | mA |

(1) This parameter is ensured by design and/or characterization and is not tested in production.

6.6 Switching Characteristics

For $V_S = (V+) - (V-) = 12V$, $V_{CM} = V_S/2$ at $T_A = 25^\circ C$ (Unless otherwise noted)

| PARAMETER | | TEST CONDITIONS | MIN | TYP | MAX | UNIT |
|----------------------|---|---|-----|-----|-----|---------|
| OUTPUT | | | | | | |
| T_{PD-HL} | Propagation delay time, high-to-low | $V_{OD} = 10\text{ mV}$, $C_L = 25\text{ pF}$, $V_{STEP} = 100\text{ mV}$ | | 45 | | μs |
| | | $V_{OD} = 50\text{ mV}$, $C_L = 25\text{ pF}$, $V_{STEP} = 100\text{ mV}$ | | 16 | | μs |
| | | $V_{OD} = 100\text{ mV}$, $C_L = 25\text{ pF}$, $V_{STEP} = 200\text{ mV}$ | | 13 | | μs |
| T_{PD-LH} | Propagation delay time, low-to-high (Push-Pull output) | $V_{OD} = 10\text{ mV}$, $C_L = 10\text{ pF}$, $V_{STEP} = 100\text{ mV}$ | | 34 | | μs |
| T_{PD-LH} | Propagation delay time, low-to-high (Push-Pull output) | $V_{OD} = 50\text{ mV}$, $C_L = 10\text{ pF}$, $V_{STEP} = 100\text{ mV}$ | | 16 | | μs |
| T_{PD-LH} | Propagation delay time, low-to-high (Push-Pull output) | $V_{OD} = 100\text{ mV}$, $C_L = 10\text{ pF}$, $V_{STEP} = 200\text{ mV}$ | | 14 | | μs |
| T_{PD-LH} | Propagation delay time, low-to-high (Open-Drain output) | $V_{OD} = 10\text{ mV}$, $C_L = 25\text{ pF}$, $R_P = 1\text{ M}\Omega$, $V_{STEP} = 100\text{ mV}$ | | 57 | | μs |
| | | $V_{OD} = 50\text{ mV}$, $C_L = 25\text{ pF}$, $R_P = 1\text{ M}\Omega$, $V_{STEP} = 100\text{ mV}$ | | 36 | | μs |
| | | $V_{OD} = 100\text{ mV}$, $C_L = 25\text{ pF}$, $R_P = 1\text{ M}\Omega$, $V_{STEP} = 200\text{ mV}$ | | 35 | | μs |
| T_{RISE} | Output Rise Time, 20% to 80%, push-pull output | $C_L = 25\text{ pF}$ | | 0.2 | | μs |
| T_{FALL} | Output Fall Time, 80% to 20% | $C_L = 25\text{ pF}$ | | 0.2 | | μs |
| POWER ON TIME | | | | | | |
| T_{ON} | Power on-time | | | 3 | | ms |

6.7 Typical Characteristics

At $T_A = 25^\circ\text{C}$, $V_S = 12\text{ V}$, $V_{CM} = V_S/2\text{ V}$, $R_P = 1\text{ M}\Omega$ (Open Drain only), $C_L = 25\text{ pF}$, $V_{OVERDRIVE} = 100\text{ mV}$ unless otherwise noted.

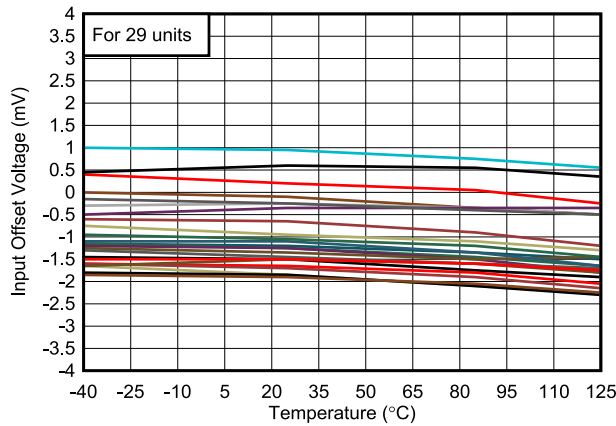


Figure 6-1. Offset vs. Temperature

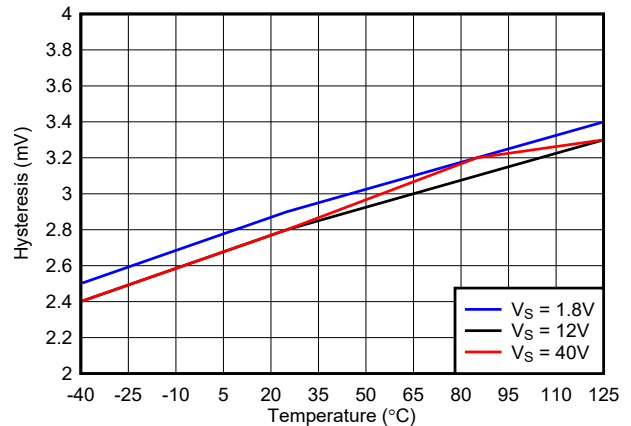


Figure 6-2. Hysteresis vs. Temperature

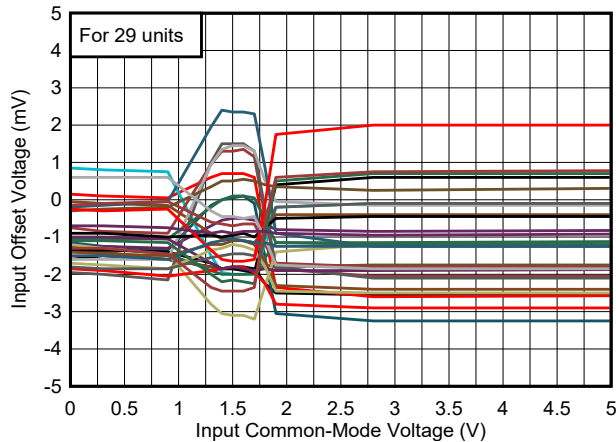


Figure 6-3. Offset vs. Common-Mode, 1.8 V

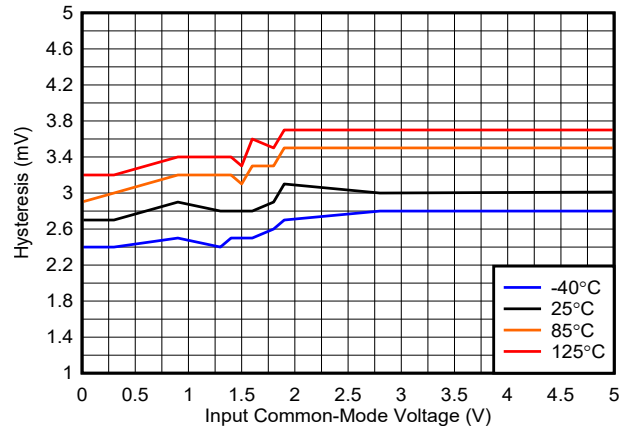


Figure 6-4. Hysteresis vs. Common-Mode, 1.8 V

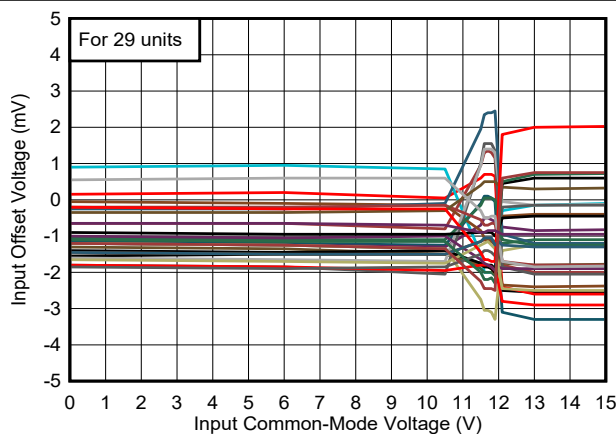


Figure 6-5. Offset vs. Common-Mode, 12 V

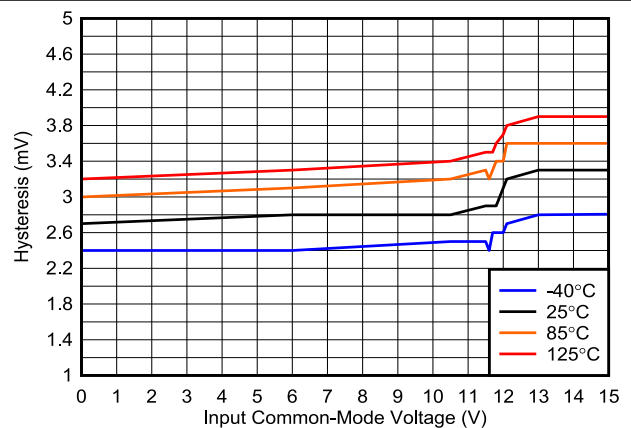


Figure 6-6. Hysteresis vs. Common-Mode, 12 V

6.7 Typical Characteristics (continued)

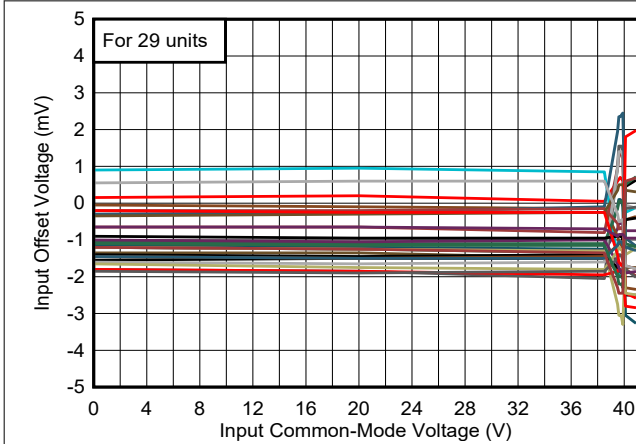


Figure 6-7. Offset vs. Common-Mode, 40 V

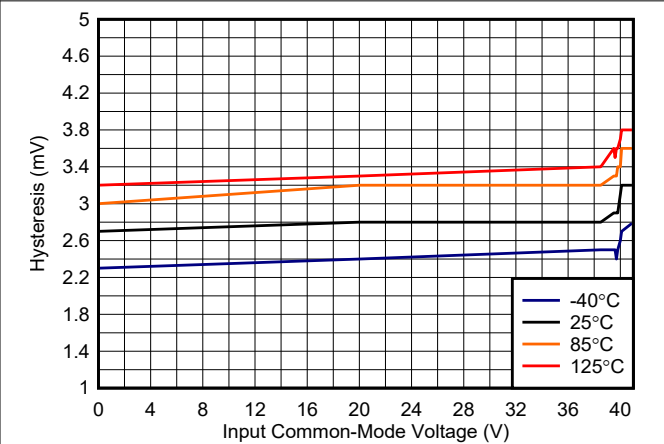


Figure 6-8. Hysteresis vs. Common-Mode, 40 V

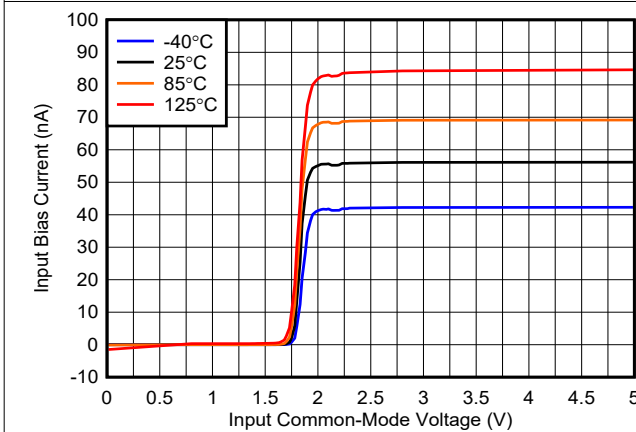


Figure 6-9. Bias Current vs. Common-Mode, 1.8 V

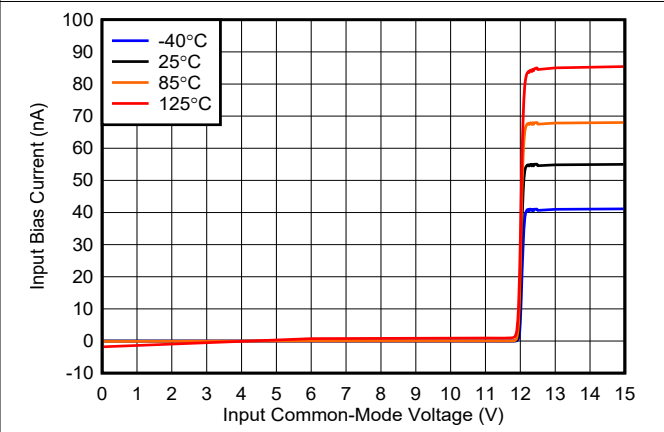


Figure 6-10. Bias Current vs. Common-Mode, 12 V

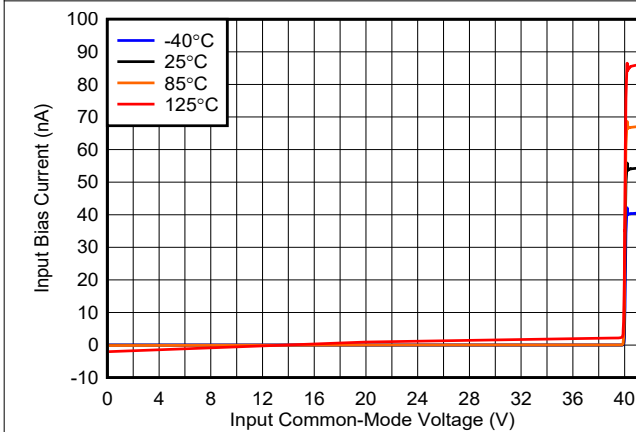


Figure 6-11. Bias Current vs. Common-Mode, 40 V

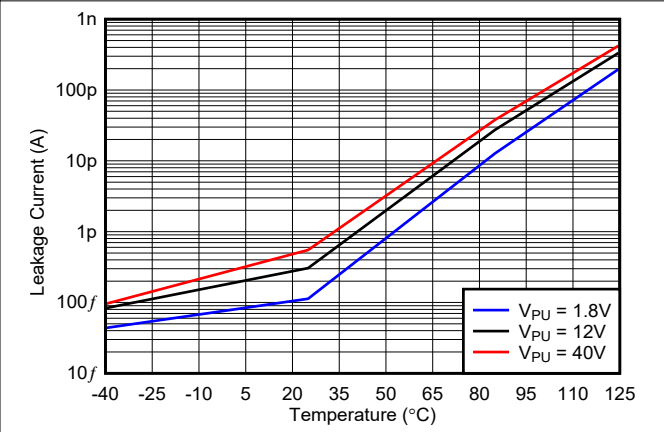


Figure 6-12. Leakage Current vs. Temperature (Open Drain only)

6.7 Typical Characteristics (continued)

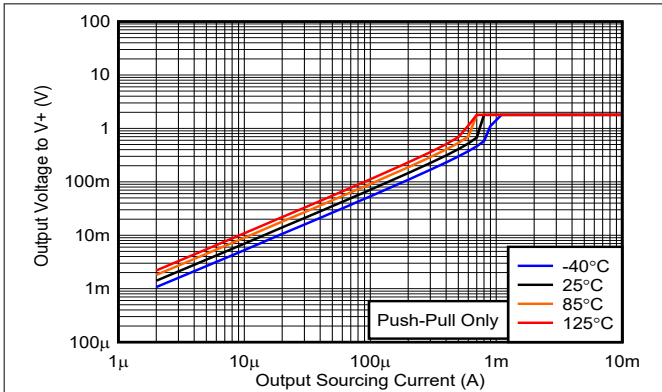


Figure 6-13. Output Voltage vs. Output Sourcing Current, 1.8V

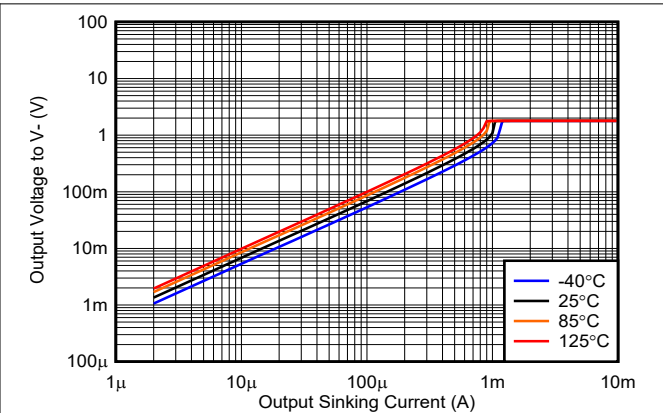


Figure 6-14. Output Voltage vs. Output Sinking Current, 1.8 V

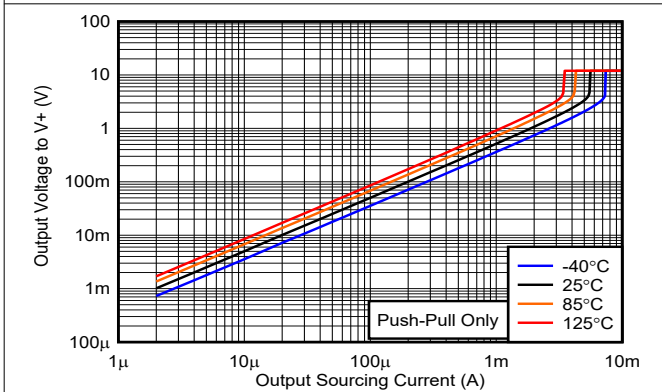


Figure 6-15. Output Voltage vs. Output Sourcing Current, 12V

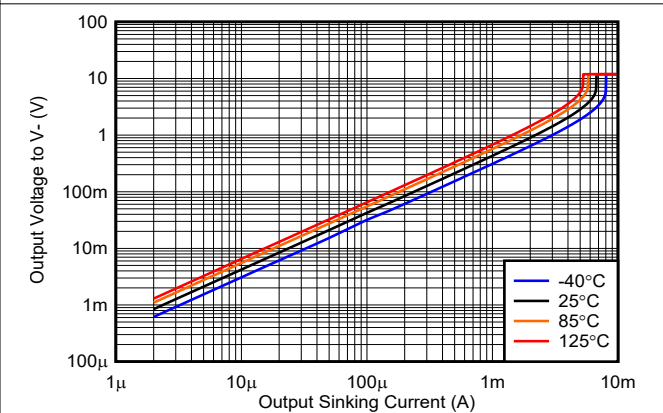


Figure 6-16. Output Voltage vs. Output Sinking Current, 12 V

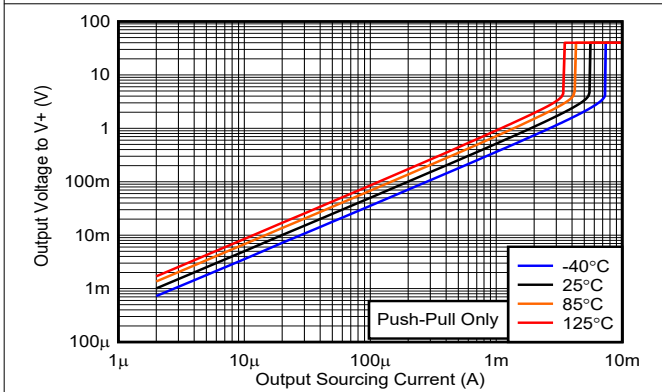


Figure 6-17. Output Voltage vs. Output Sourcing Current, 40V

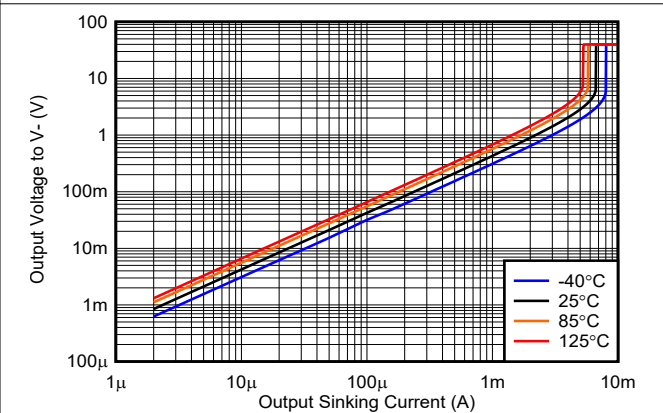


Figure 6-18. Output Voltage vs. Output Sinking Current, 40 V

6.7 Typical Characteristics (continued)

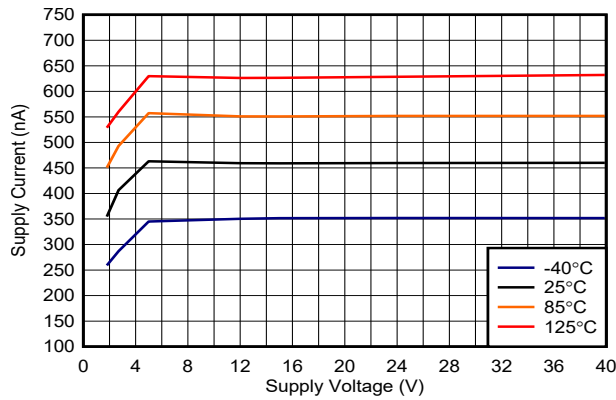


Figure 6-19. Supply Current vs. Supply Voltage (Output Low), Push-Pull

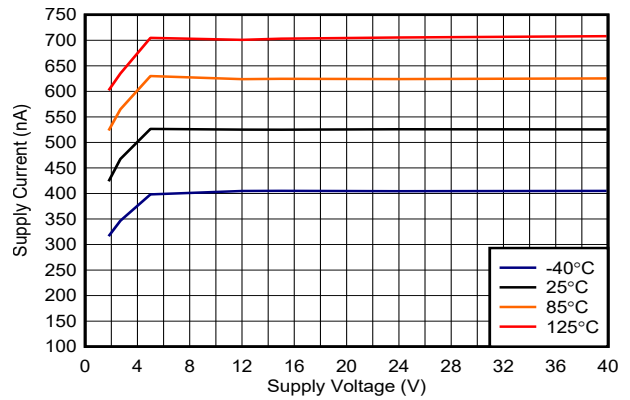


Figure 6-20. Supply Current vs. Supply Voltage (Output High), Push-Pull

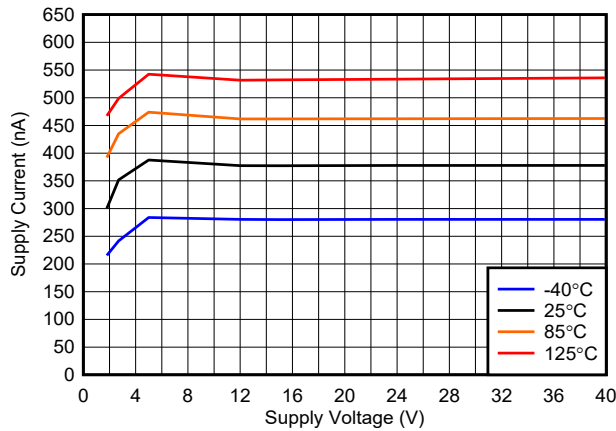


Figure 6-21. Supply Current vs. Supply Voltage (Output Low), Open Drain

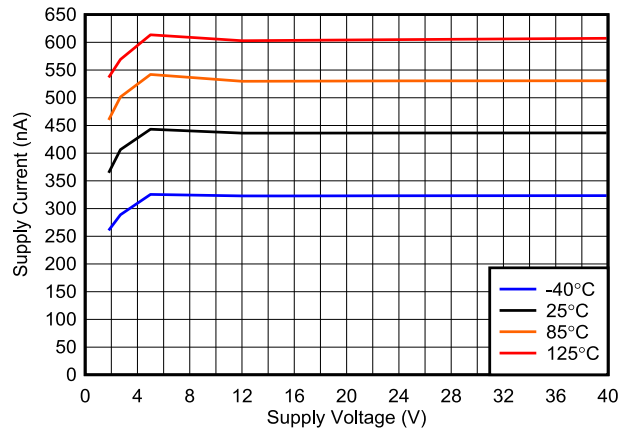


Figure 6-22. Supply Current vs. Supply Voltage (Output High), Open Drain

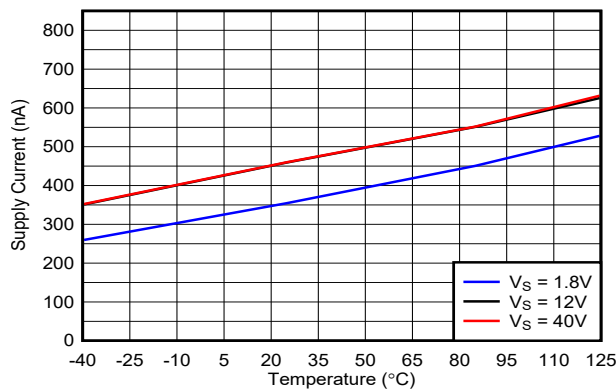


Figure 6-23. Supply Current vs. Temperature (Output Low), Push-Pull

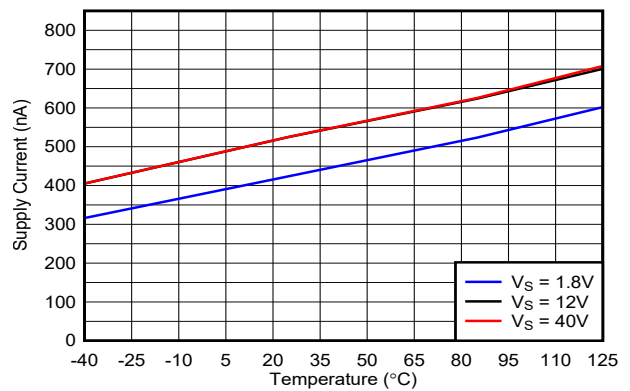


Figure 6-24. Supply Current vs. Temperature (Output High), Push-Pull

6.7 Typical Characteristics (continued)

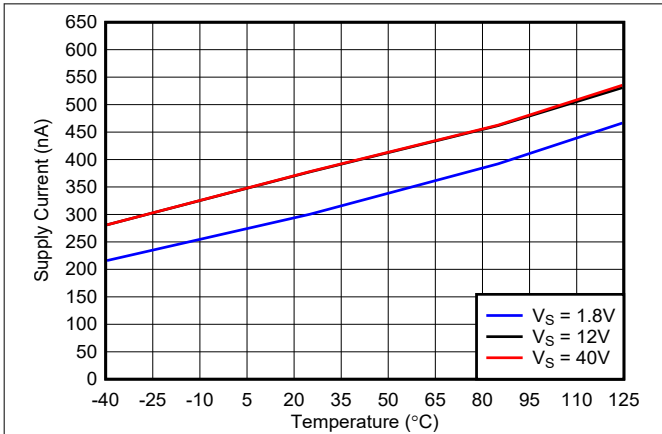


Figure 6-25. Supply Current vs. Temperature (Output Low), Open Drain

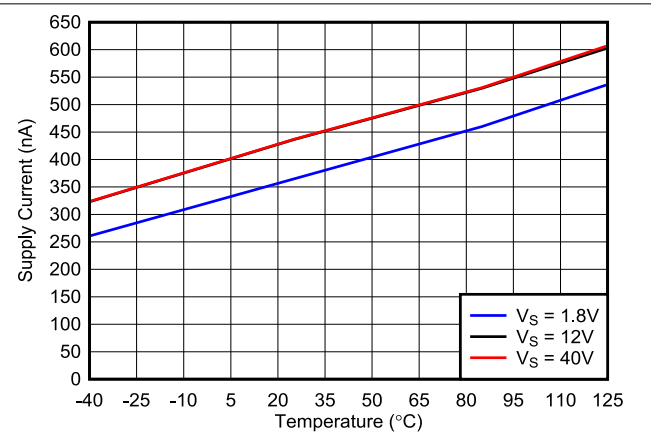


Figure 6-26. Supply Current vs. Temperature (Output High), Open Drain

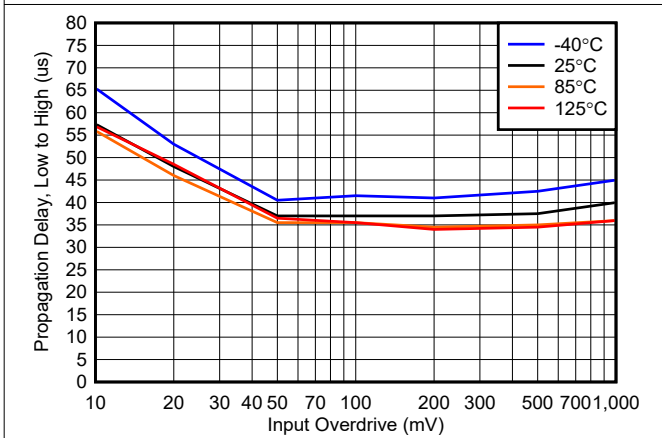


Figure 6-27. Propagation Delay, Low to High, 1.8 V, Open Drain

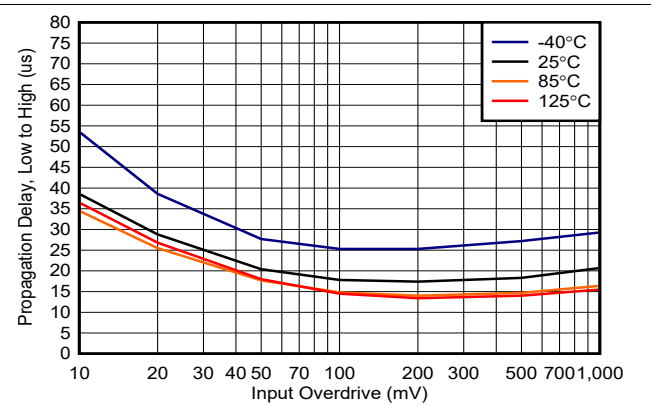


Figure 6-28. Propagation Delay, Low to High, 1.8 V, Push-Pull

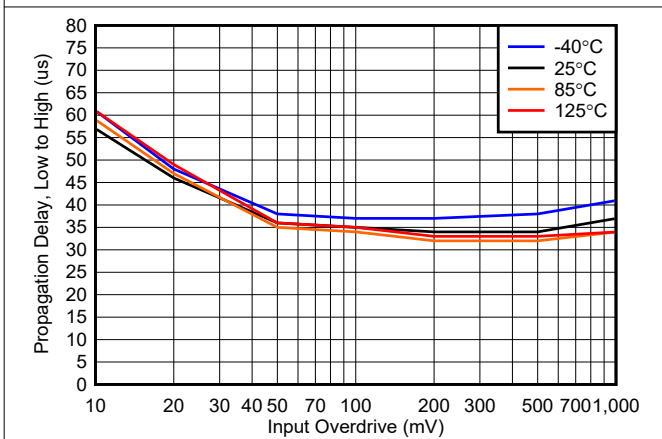


Figure 6-29. Propagation Delay, Low to High, 12 V, Open Drain

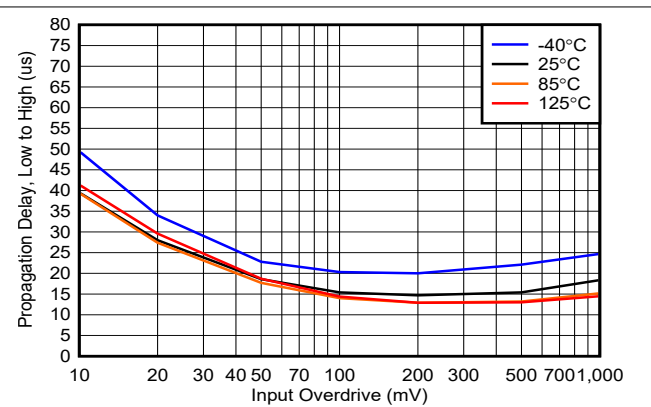


Figure 6-30. Propagation Delay, Low to High, 12 V, Push-Pull

6.7 Typical Characteristics (continued)

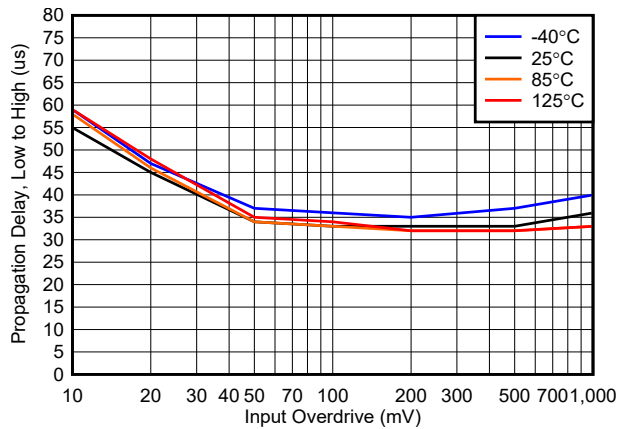


Figure 6-31. Propagation Delay, Low to High, 40 V, Open Drain

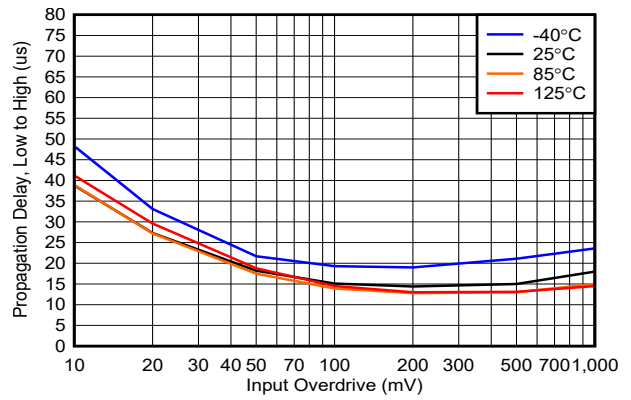


Figure 6-32. Propagation Delay, Low to High, 40 V, Push-Pull

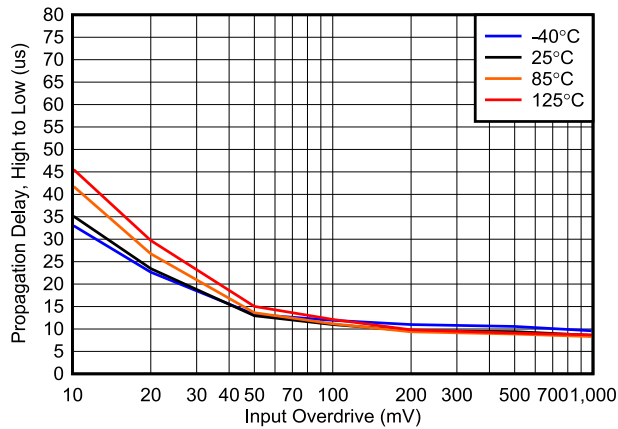


Figure 6-33. Propagation Delay, High to Low, 1.8 V

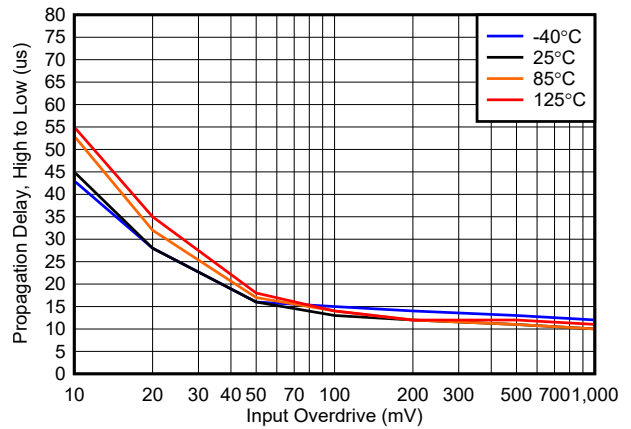


Figure 6-34. Propagation Delay, High to Low, 12 V

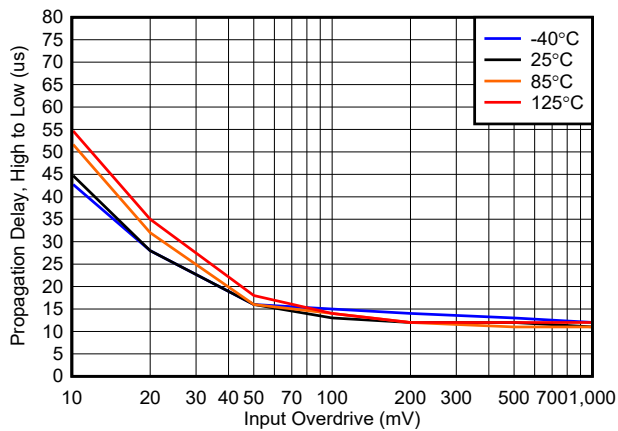


Figure 6-35. Propagation Delay, High to Low, 40 V

7 Detailed Description

7.1 Overview

The TLV185x-Q1 and TLV186x-Q1 devices are nanopower comparators with push-pull and open-drain output options. Operating down to 1.8 V while only consuming only 440 nA per channel, the TLV185x-Q1 and TLV186x-Q1 are well suited for voltage, current, and temperature sensing in low and high voltage low-power, always-on systems. An internal power-on reset circuit ensures that the output remains in a known state during power-up and power-down. Inputs have fail-safe inputs that can tolerate input transients without damage or false outputs.

7.2 Functional Block Diagrams

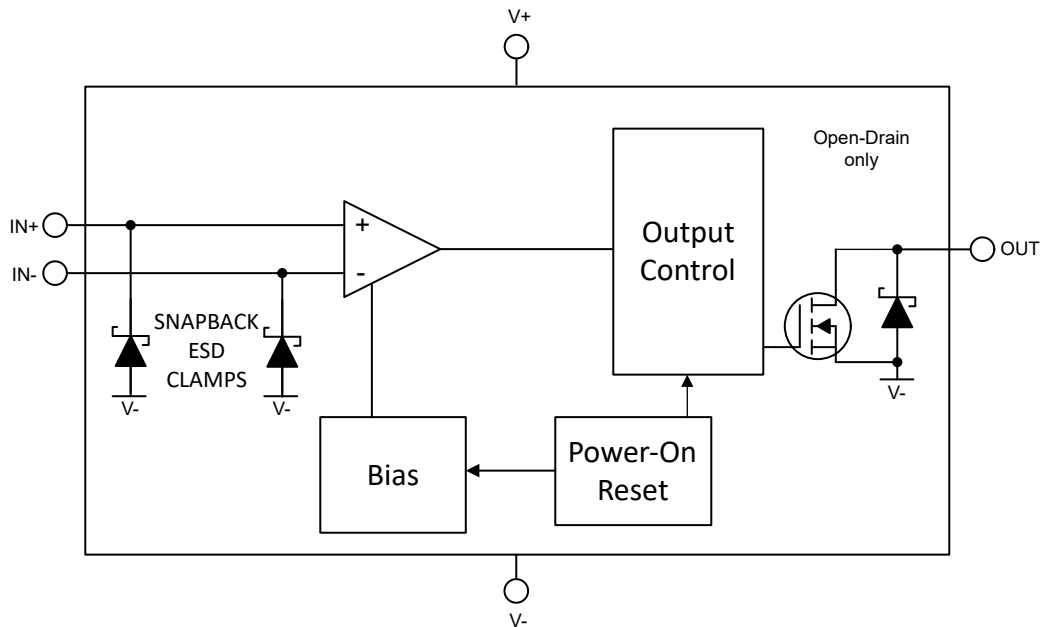


Figure 7-1. Block Diagram

7.3 Feature Description

The TLV185x-Q1 (push-pull output) and TLV186x-Q1 (open-drain output) devices are nano-power comparators that are capable of operating at high voltages. This family of comparators feature a fail safe input stage and over the rail operating condition mode capable of operating up to 40 V, independent of V_+ . The comparators also have an internal reverse battery protection feature and Power-On-Reset for known start-up conditions.

7.4 Device Functional Modes

7.4.1 Inputs

7.4.1.1 Operating Common-Mode Ranges

The TLV185x-Q1 and TLV186x-Q1 devices have two operating common-mode ranges: within-the-rail and over-the-rail.

Within-the-Rail Operation: $IN+$ and $IN-$ are less than (V_+)

When an input pin is operating less than (V_+) , there are two operating regions defined where input voltages can be compared: low common-mode and high-common mode. In low-common mode which extends typically from 0 V to $(V_+) - 1$ V, the typical input bias current is less than 1 pA. In high common-mode which extends typically from $(V_+) - 1$ V to (V_+) , the typical input bias current is less than 14 nA.

Over-the-Rail Operation: $IN+$ and/or $IN-$ are greater than (V_+)

The TLV185x-Q1 and TLV186x-Q1 devices have a distinctive input stage that allows the input common mode range to extend from 0 V to 40 V independent of the supply voltage. This feature means that operation at low

supply voltages does not limit the range of input voltages that can be compared. When an input pin is operating over-the-rail (above (V+)), the bias current increases to a typical value of 55 nA.

See Figure 6-9 to 6-11 in the [Typical Characteristics](#) section for input bias current vs. common-mode voltages.

7.4.1.2 Fail-Safe Inputs

A feature of the TLV185x-Q1 and TLV186x-Q1 family is that the inputs are fail safe up to 40 V, independent of (V+). The inputs are maintained as high input impedance and can be of any value between -0.1 V and 40 V, even while (V+) is unpowered or below the minimum supply voltage. This feature avoids power sequencing or transient issues since the inputs are not diode clamped to (V+).

7.4.1.3 Unused Inputs

If a channel is not to be used, DO NOT tie the inputs together. Due to the high equivalent bandwidth and low offset voltage, tying the inputs directly together can cause high frequency oscillations as the device triggers on it's own internal wideband noise. Instead, the inputs should be tied to any available voltage that resides within the specified input voltage range and provides a minimum of 50mV differential voltage. For example, one input can be grounded and the other input connected to a reference voltage, or even (V+).

7.4.2 Internal Hysteresis

The device hysteresis transfer curve is shown in [Figure 7-2](#). This curve is a function of three components: V_{TH} , V_{OS} , and V_{HYST} :

- V_{TH} is the actual set voltage or threshold trip voltage.
- V_{OS} is the internal offset voltage between V_{IN+} and V_{IN-} . This voltage is added to V_{TH} to form the actual trip point at which the comparator must respond to change output states.
- V_{HYST} is the internal hysteresis (or trip window) that is designed to reduce comparator sensitivity to noise.

(2.8 mV for the TLV185x/6x family)

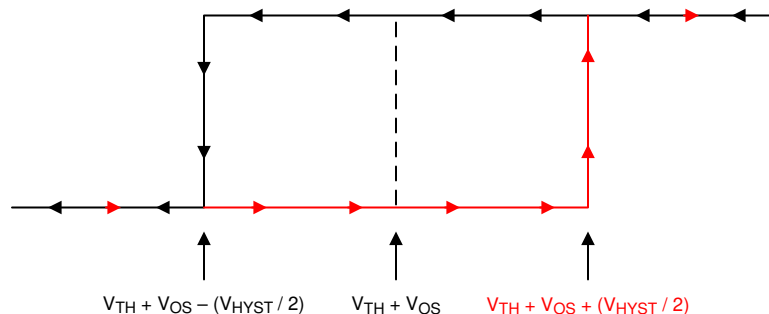


Figure 7-2. Hysteresis Transfer Curve

7.4.3 Outputs

7.4.3.1 TLV185x-Q1 Push-Pull Output

The TLV185x-Q1 features a push-pull output stage capable of both sinking and sourcing current. This allows driving loads such as LED's and MOSFET gates, as well as eliminating the need for a power-wasting external pull-up resistor. The push-pull output must never be connected to another output.

Directly shorting the output to the supply rails ((V+) when output "low" or (V-) when output "High") can result in thermal runaway and eventual device destruction at high (>12 V) supply voltages. If output shorts are possible, a series current limiting resistor is recommended to limit the power dissipation.

Unused push-pull outputs should be left floating, and never tied to a supply, ground, or another output.

7.4.3.2 TLV186x-Q1 Open-Drain Output

The TLV186x-Q1 features an open-drain (also commonly called open collector) sinking-only output stage enabling the output logic levels to be pulled up to an external voltage from 0 V up to 40 V, independent of the comparator supply voltage (V+). The open-drain output also allows logical OR'ing of multiple open drain

outputs and logic level translation. TI recommends setting the pull-up resistor current to less than 100 μ A to optimize V_{OL} logic levels. Lower pull-up resistor values will help increase the rising edge risetime, but at the expense of increasing V_{OL} and higher power dissipation. The risetime will be dependent on the time constant of the total pull-up resistance and total load capacitance. Large value pull-up resistors (>1 M Ω) will create an exponential rising edge due to the output RC time constant and increase the risetime.

Directly shorting the output to (V+) can result in thermal runaway and eventual device destruction at high (>12 V) pull-up voltages. If output shorts are possible, a series current limiting resistor is recommended to limit the power dissipation.

Unused open drain outputs should be left floating, or can be tied to the (V-) pin if floating pins are not desired.

7.4.4 ESD Protection

7.4.4.1 Inputs

The fail-safe inputs incorporates internal ESD protection circuits on all pins. The fail-safe inputs have ESD protection from each pin to (V-) which allows these pins to exceed the supply voltage (V+) up to 40 V. If input voltages are to exceed 40 V, an external clamp would be required. Likewise, negative voltages on the inputs are ESD clamped to (V-) and should be limited to less than -0.1 V.

If the inputs are to be connected to a low impedance source, such as a power supply or buffered reference line, TI recommends adding a current-limiting resistor in series with the input to limit any transient currents should the clamps conduct. The current should be limited to 10 mA or less. This series resistance can be part of any resistive input dividers or networks.

7.4.4.2 Outputs

The TLV185x-Q1 push-pull output protection also contains a conventional diode-type ESD clamps between the output and (V-), as the output should not exceed the supply rails.

The TLV186x-Q1 open-drain output ESD protection also consists of a snapback ESD clamp between the output and (V-) to allow the output to be pulled above (V+) to a maximum of 40 V.

7.4.5 Power-On Reset (POR)

The TLV185x-Q1 and TLV186x-Q1 devices have an internal Power-on-Reset (POR) circuit for known start-up or power-down conditions. While the power supply (V+) is ramping up or ramping down, the POR circuitry will be activated for up to 2 ms after the V_{POR} of 1.5 V is crossed. When the supply voltage is equal to or greater than the minimum supply voltage, and after the delay period, the comparator output reflects the state of the differential input (V_{ID}).

For the TLV185x-Q1 push-pull output devices, the output is held low during the POR period (t_{on}).

For the TLV186x-Q1 open drain output devices, the POR circuit will keep the output high impedance (Hi-Z) during the POR period (t_{on}).

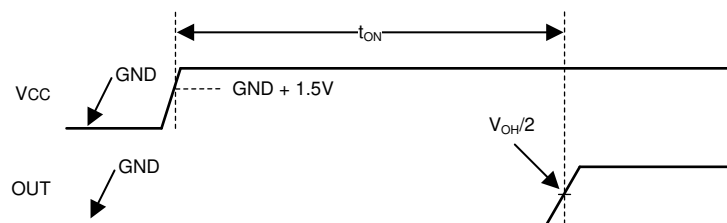


Figure 7-3. Power-On Reset Timing Diagram

Note that it the nature of an open collector output that the output will rise with the pull-up voltage during the POR period.

7.4.6 Reverse Battery Protection

The TLV185x-Q1 and TLV186x-Q1 devices have an internal reverse battery protection feature that prevents damage to the comparator in the event of improper battery installation to the supply pins. This protection feature works up to 40 V.

8 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

8.1.1 Basic Comparator Definitions

8.1.1.1 Operation

The basic comparator compares the input voltage (V_{IN}) on one input to a reference voltage (V_{REF}) on the other input. In the [Figure 8-1](#) example below, if V_{IN} is less than V_{REF} , the output voltage (V_O) is logic low (V_{OL}). If V_{IN} is greater than V_{REF} , the output voltage (V_O) is at logic high (V_{OH}). [Table 8-1](#) summarizes the output conditions. The output logic can be inverted by simply swapping the input pins.

Table 8-1. Output Conditions

| Inputs Condition | Output |
|------------------|--|
| $IN+ > IN-$ | HIGH (V_{OH}) |
| $IN+ = IN-$ | Indeterminate (chatters - see Hysteresis) |
| $IN+ < IN-$ | LOW (V_{OL}) |

8.1.1.2 Propagation Delay

There is a delay between from when the input crosses the reference voltage and the output responds. This is called the Propagation Delay. Propagation delay can be different between high-to low and low-to-high input transitions. This is shown as t_{pLH} and t_{pHL} in [Figure 8-1](#) and is measured from the mid-point of the input to the midpoint of the output.

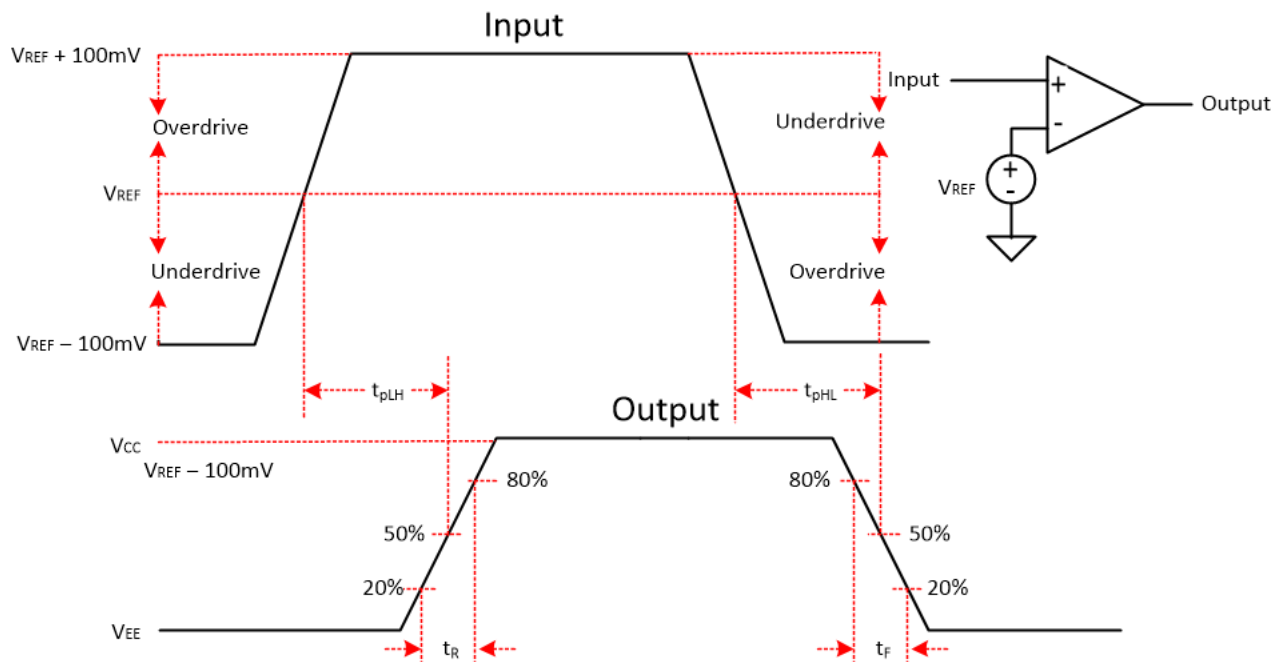


Figure 8-1. Comparator Timing Diagram

8.1.1.3 Overdrive Voltage

The overdrive voltage, V_{OD} , is the amount of input voltage beyond the reference voltage (and not the total input peak-to-peak voltage). The overdrive voltage is 100 mV as shown in the [Figure 8-1](#) example. The overdrive voltage can influence the propagation delay (t_p). The smaller the overdrive voltage, the longer the propagation delay, particularly when $<100\text{mV}$. If the fastest speeds are desired, it is recommended to apply the highest amount of overdrive possible.

The risetime (t_r) and falltime (t_f) is the time from the 20% and 80% points of the output waveform.

8.1.2 Hysteresis

The basic comparator configuration may produce a noisy "chatter" output if the applied differential input voltage is near the comparator's offset voltage. This usually occurs when the input signal is moving very slowly across the switching threshold of the comparator. This problem can be prevented by adding external hysteresis to the comparator.

Since the TLV185x-Q1 and TLV186x-Q1 devices only have a minimal amount of internal hysteresis of 2.7 mV, external hysteresis can be applied in the form of a positive feedback loop that adjusts the trip point of the comparator depending on its current output state.

The hysteresis transfer curve is shown in [Figure 8-2](#). This curve is a function of three components: V_{TH} , V_{OS} , and V_{HYST} :

- V_{TH} is the actual set voltage or threshold trip voltage.
- V_{OS} is the internal offset voltage between V_{IN+} and V_{IN-} . This voltage is added to V_{TH} to form the actual trip point at which the comparator must respond to change output states.
- V_{HYST} is the hysteresis (or trip window) that is designed to reduce comparator sensitivity to noise.

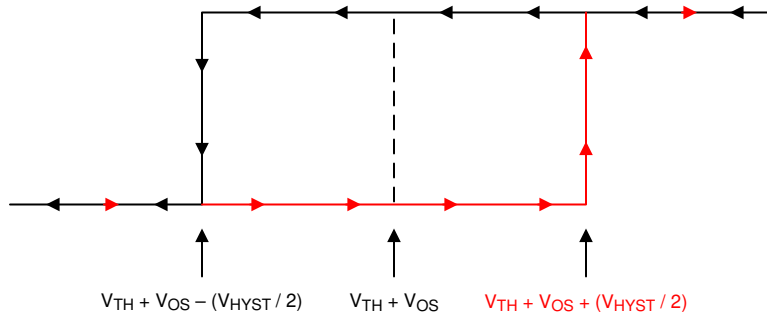


Figure 8-2. Hysteresis Transfer Curve

For more information, please see Application Note SBOA219 "[Comparator with and without hysteresis circuit](#)".

8.1.2.1 Inverting Comparator With Hysteresis

The inverting comparator with hysteresis requires a three-resistor network that is referenced to the comparator supply voltage (V_{CC}), as shown in [Figure 8-3](#).

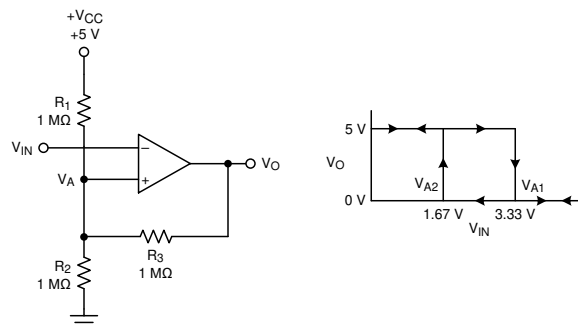


Figure 8-3. TLV185x-Q1 in an Inverting Configuration With Hysteresis

The equivalent resistor networks when the output is high and low are shown in [Figure 8-3](#).

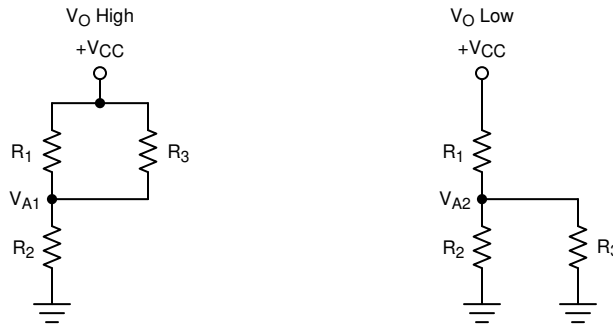


Figure 8-4. Inverting Configuration Resistor Equivalent Networks

When V_{IN} is less than V_A , the output voltage is high (for simplicity, assume V_O switches as high as V_{CC}). The three network resistors can be represented as $R1 \parallel R3$ in series with $R2$, as shown in [Figure 8-4](#).

[Equation 1](#) below defines the high-to-low trip voltage (V_{A1}).

$$V_{A1} = V_{CC} \times \frac{R2}{(R1 \parallel R3) + R2} \quad (1)$$

When V_{IN} is greater than V_A , the output voltage is low. In this case, the three network resistors can be presented as $R2 \parallel R3$ in series with $R1$, as shown in [Equation 2](#).

Use [Equation 2](#) to define the low to high trip voltage (V_{A2}).

$$V_{A2} = V_{CC} \times \frac{R2 \parallel R3}{R1 + (R2 \parallel R3)} \quad (2)$$

[Equation 3](#) defines the total hysteresis provided by the network.

$$\Delta V_A = V_{A1} - V_{A2} \quad (3)$$

8.1.2.2 Non-Inverting Comparator With Hysteresis

A non-inverting comparator with hysteresis requires a two-resistor network and a voltage reference (V_{REF}) at the inverting input, as shown in [Figure 8-5](#),

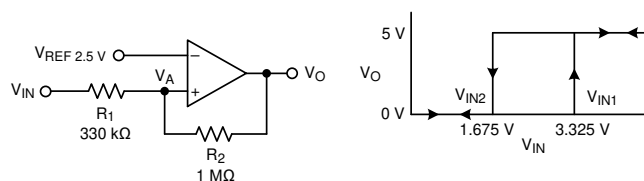


Figure 8-5. TLV185x-Q1 in a Non-Inverting Configuration With Hysteresis

The equivalent resistor networks when the output is high and low are shown in [Figure 8-6](#).

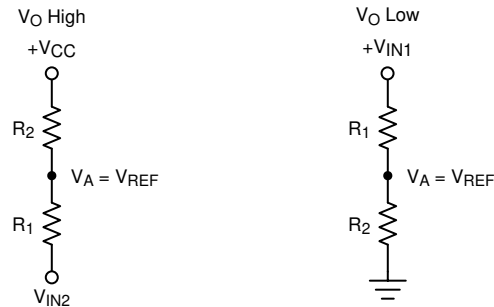


Figure 8-6. Non-Inverting Configuration Resistor Networks

When V_{IN} is less than V_{REF} , the output is low. For the output to switch from low to high, V_{IN} must rise above the V_{IN1} threshold. Use [Equation 4](#) to calculate V_{IN1} .

$$V_{IN1} = R1 \times \frac{V_{REF}}{R2} + V_{REF} \quad (4)$$

When V_{IN} is greater than V_{REF} , the output is high. For the comparator to switch back to a low state, V_{IN} must drop below V_{IN2} . Use [Equation 5](#) to calculate V_{IN2} .

$$V_{IN2} = \frac{V_{REF} (R1 + R2) - V_{CC} \times R1}{R2} \quad (5)$$

The hysteresis of this circuit is the difference between V_{IN1} and V_{IN2} , as shown in [Equation 6](#).

$$\Delta V_{IN} = V_{CC} \times \frac{R1}{R2} \quad (6)$$

For more information, please see Application Notes SNOA997 "[Inverting comparator with hysteresis circuit](#)" and SBOA313 "[Non-Inverting Comparator With Hysteresis Circuit](#)".

8.1.2.3 Inverting and Non-Inverting Hysteresis using Open-Drain Output

It is also possible to use an open drain output device, such as the TLV186x-Q1, but the output pull-up resistor must also be taken into account in the calculations. The pull-up resistor is seen in series with the feedback resistor when the output is high. Thus, the feedback resistor is actually seen as $R2 + R_{PULLUP}$. TI recommends that the pull-up resistor be at least 10 times less than the feedback resistor value.

8.2 Typical Applications

8.2.1 Window Comparator

Window comparators are commonly used to detect undervoltage and overvoltage conditions. [Figure 8-7](#) shows a simple window comparator circuit. Window comparators require open drain outputs (TLV186x-Q1 if the outputs are directly connected together).

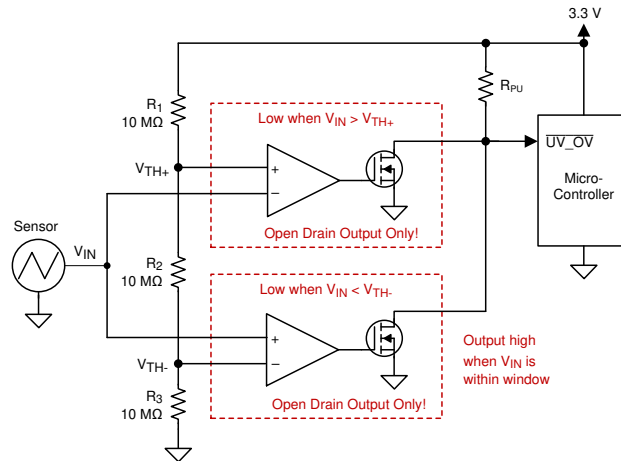


Figure 8-7. Window Comparator

8.2.1.1 Design Requirements

For this design, follow these design requirements:

- Alert (logic low output) when an input signal is less than 1.1 V
- Alert (logic low output) when an input signal is greater than 2.2 V
- Alert signal is active low
- Operate from a 3.3-V power supply

8.2.1.2 Detailed Design Procedure

Configure the circuit as shown in [Figure 8-7](#). Connect $V+$ to a 3.3-V power supply and V_{EE} to ground. Make R_1 , R_2 and R_3 each 10-M Ω resistors. These three resistors are used to create the positive and negative thresholds for the window comparator (V_{TH+} and V_{TH-}).

With each resistor being equal, V_{TH+} is 2.2 V and V_{TH-} is 1.1 V. Large resistor values such as 10-M Ω are used to minimize power consumption. The resistor values may be recalculated to provide the desired trip point values.

The sensor output voltage is applied to the inverting and noninverting inputs of the two comparators. Using two open-drain output comparators allows the two comparator outputs to be Wire-OR'ed together.

The respective comparator outputs will be low when the sensor is less than 1.1 V or greater than 2.2 V. The respective comparator outputs will be high when the sensor is in the range of 1.1 V to 2.2 V (within the "window"), as shown in [Figure 8-8](#).

8.2.1.3 Application Curve

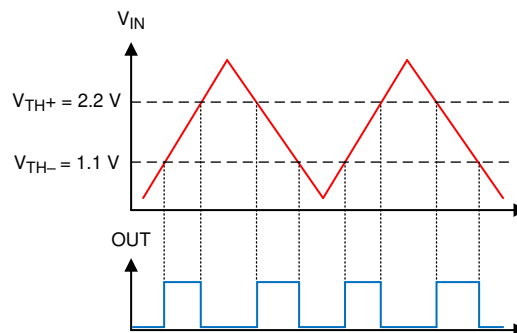


Figure 8-8. Window Comparator Results

For more information, please see Application note SBOA221 "[Window comparator circuit](#)".

8.2.2 Undervoltage Detection

Undervoltage detection is frequently required in low-power, always-on systems to alert the downstream device that a battery or supply voltage has drooped below the nominal voltage level. The TLV185x-Q1 and TLV186x-Q1 devices have an over-the-rail capability which allows the comparator to monitor high voltages up to 40 V on the inputs while operating at lower supply voltages such as 3.3 V, as shown in Figure 8-9. The high voltage battery V_{BAT} is divided down to be compared against a reference voltage, which can be set by a shunt regulator such as the TL431. When the voltage on the non-inverting input drops below the reference voltage, then the output of the comparator will toggle and send an alert signal to a MCU.

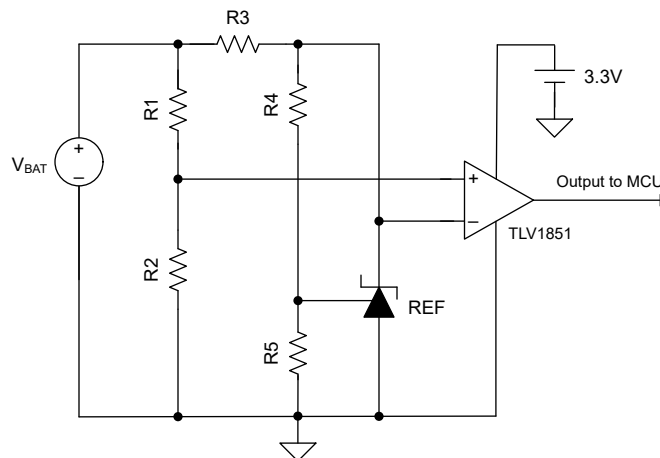


Figure 8-9. Undervoltage Detection

8.2.3 Reverse Battery and Overvoltage Protection Scheme

In battery powered applications, it is essential to have safeguards in place to protect the internal electronics against reverse battery connections. Figure 8-10 shows a protection scheme using TLV1851-Q1 to prevent damage to the overall system in the events of both improper battery installation and overvoltage conditions. Under either of these instances, Q1 and Q2 will open, thus protecting the system from ever seeing reverse battery or overvoltage conditions.

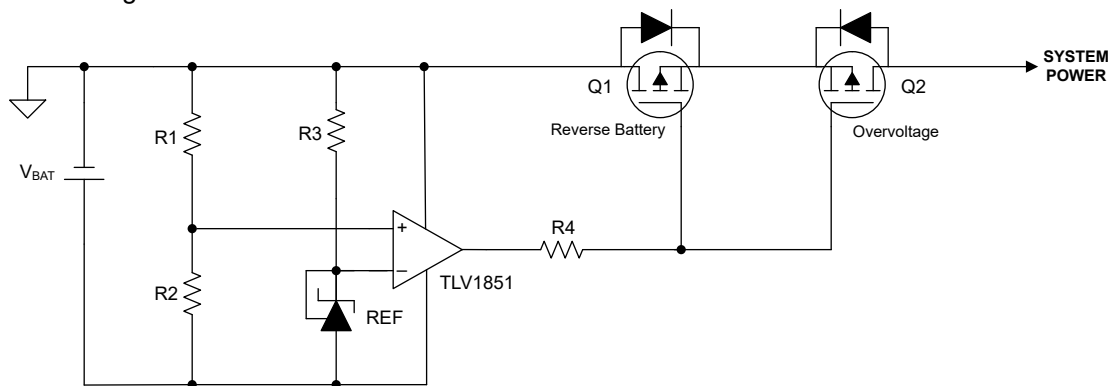


Figure 8-10. Reverse Battery and Overvoltage Protection Scheme

8.3 Power Supply Recommendations

Due to the fast output edges, it is critical to have bypass capacitors on the supply pin to prevent supply ringing and false triggers and oscillations. Bypass the supply directly at *each* device with a low ESR 0.1 μF ceramic bypass capacitor directly between the (V+) pin and ground pins. Narrow peak currents will be drawn during the output transition time, particularly for the push-pull output device. These narrow pulses can cause un-bypassed supply lines and poor grounds to ring, possibly causing variation that can eat into the input voltage range and create an inaccurate comparison or even oscillations.

The device may be powered from both "split" supplies ((V+) &(V-)), or "single" supplies ((V+) and GND), with GND applied to the (V-) pin. Input signals must stay within the recommended input range for either type. Note that with a "split" supply the output will now swing "low" (V_{OL}) to (V-) potential and not GND.

9 Layout

9.1 Layout Guidelines

For accurate comparator applications it is important maintain a stable power supply with minimized noise and glitches. Output rise and fall times are in the tens of nanoseconds, and should be treated as high speed logic devices. The bypass capacitor should be as close to the supply pin as possible and connected to a solid ground plane, and preferably directly between the (V+) and GND pins.

Minimize coupling between outputs and inputs to prevent output oscillations. Do not run output and input traces in parallel unless there is a (V+) or GND trace between output to reduce coupling. When series resistance is added to inputs, place resistor close to the device. A low value (<100 ohms) resistor may also be added in series with the output to dampen any ringing or reflections on long, non-impedance controlled traces. For best edge shapes, controlled impedance traces with back-terminations should be used when routing long distances.

9.2 Layout Example

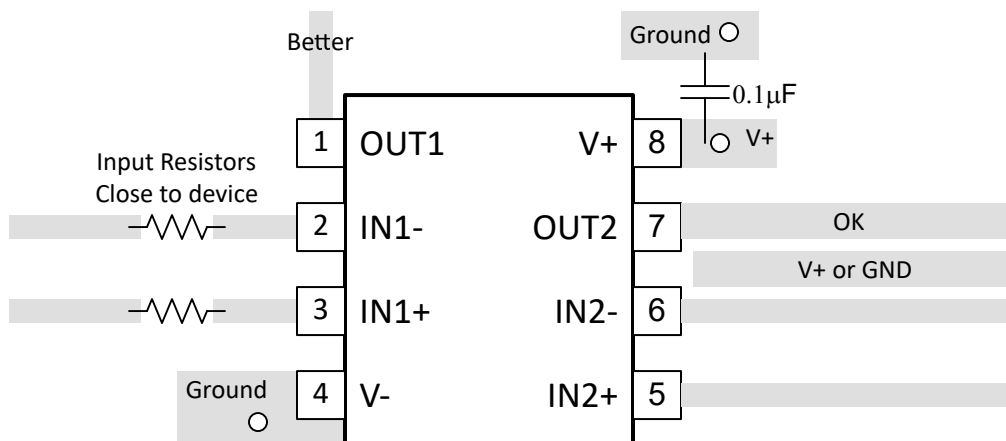


Figure 9-1. Dual Layout Example

10 Device and Documentation Support

10.1 Documentation Support

10.1.1 Related Documentation

[Analog Engineers Circuit Cookbook: Amplifiers \(See Comparators section\) - SLYY137](#)

[Precision Design, Comparator with Hysteresis Reference Design— TIDU020](#)

[Window comparator circuit - SBOA221](#)

[Reference Design, Window Comparator Reference Design— TIPD178](#)

[Comparator with and without hysteresis circuit - SBOA219](#)

[Inverting comparator with hysteresis circuit - SNOA997](#)

[Non-Inverting Comparator With Hysteresis Circuit - SBOA313](#)

[A Quad of Independently Func Comparators - SNOA654](#)

10.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. Click on [Subscribe to updates](#) to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

10.3 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

Linked content is provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

10.4 Trademarks

TI E2E™ is a trademark of Texas Instruments.
All trademarks are the property of their respective owners.

10.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

10.6 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

11 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

| Orderable part number | Status (1) | Material type (2) | Package Pins | Package qty Carrier | RoHS (3) | Lead finish/ Ball material (4) | MSL rating/ Peak reflow (5) | Op temp (°C) | Part marking (6) |
|--------------------------------|---------------|----------------------|------------------|-----------------------|-------------|--------------------------------------|-----------------------------------|--------------|---------------------|
| TLV1851QDBVRQ1 | Active | Production | SOT-23 (DBV) 5 | 3000 LARGE T&R | Yes | NIPDAU | Level-1-260C-UNLIM | -40 to 125 | 851Q |
| TLV1851QDBVRQ1.A | Active | Production | SOT-23 (DBV) 5 | 3000 LARGE T&R | Yes | NIPDAU | Level-1-260C-UNLIM | -40 to 125 | 851Q |
| TLV1861QDBVRQ1 | Active | Production | SOT-23 (DBV) 5 | 3000 LARGE T&R | Yes | NIPDAU | Level-1-260C-UNLIM | -40 to 125 | 861Q |
| TLV1861QDBVRQ1.A | Active | Production | SOT-23 (DBV) 5 | 3000 LARGE T&R | Yes | NIPDAU | Level-1-260C-UNLIM | -40 to 125 | 861Q |

(1) **Status:** For more details on status, see our [product life cycle](#).

(2) **Material type:** When designated, preproduction parts are prototypes/experimental devices, and are not yet approved or released for full production. Testing and final process, including without limitation quality assurance, reliability performance testing, and/or process qualification, may not yet be complete, and this item is subject to further changes or possible discontinuation. If available for ordering, purchases will be subject to an additional waiver at checkout, and are intended for early internal evaluation purposes only. These items are sold without warranties of any kind.

(3) **RoHS values:** Yes, No, RoHS Exempt. See the [TI RoHS Statement](#) for additional information and value definition.

(4) **Lead finish/Ball material:** Parts may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

(5) **MSL rating/Peak reflow:** The moisture sensitivity level ratings and peak solder (reflow) temperatures. In the event that a part has multiple moisture sensitivity ratings, only the lowest level per JEDEC standards is shown. Refer to the shipping label for the actual reflow temperature that will be used to mount the part to the printed circuit board.

(6) **Part marking:** There may be an additional marking, which relates to the logo, the lot trace code information, or the environmental category of the part.

Multiple part markings will be inside parentheses. Only one part marking contained in parentheses and separated by a "~" will appear on a part. If a line is indented then it is a continuation of the previous line and the two combined represent the entire part marking for that device.

Important Information and Disclaimer: The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

OTHER QUALIFIED VERSIONS OF TLV1851-Q1, TLV1861-Q1 :

- Catalog : [TLV1851](#), [TLV1861](#)

NOTE: Qualified Version Definitions:

- Catalog - TI's standard catalog product

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
|----------------|--------------|-----------------|------|------|--------------------|--------------------|---------|---------|---------|---------|--------|---------------|
| TLV1851QDBVRQ1 | SOT-23 | DBV | 5 | 3000 | 180.0 | 8.4 | 3.2 | 3.2 | 1.4 | 4.0 | 8.0 | Q3 |
| TLV1851QDBVRQ1 | SOT-23 | DBV | 5 | 3000 | 180.0 | 8.4 | 3.2 | 3.2 | 1.4 | 4.0 | 8.0 | Q3 |
| TLV1861QDBVRQ1 | SOT-23 | DBV | 5 | 3000 | 180.0 | 8.4 | 3.2 | 3.2 | 1.4 | 4.0 | 8.0 | Q3 |
| TLV1861QDBVRQ1 | SOT-23 | DBV | 5 | 3000 | 180.0 | 8.4 | 3.2 | 3.2 | 1.4 | 4.0 | 8.0 | Q3 |

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|----------------|--------------|-----------------|------|------|-------------|------------|-------------|
| TLV1851QDBVRQ1 | SOT-23 | DBV | 5 | 3000 | 210.0 | 185.0 | 35.0 |
| TLV1851QDBVRQ1 | SOT-23 | DBV | 5 | 3000 | 210.0 | 185.0 | 35.0 |
| TLV1861QDBVRQ1 | SOT-23 | DBV | 5 | 3000 | 210.0 | 185.0 | 35.0 |
| TLV1861QDBVRQ1 | SOT-23 | DBV | 5 | 3000 | 210.0 | 185.0 | 35.0 |

EXAMPLE BOARD LAYOUT

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE
EXPOSED METAL SHOWN
SCALE:15X



SOLDER MASK DETAILS

4214839/K 08/2024

NOTES: (continued)

- 6. Publication IPC-7351 may have alternate designs.
- 7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE
BASED ON 0.125 mm THICK STENCIL
SCALE:15X

4214839/K 08/2024

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you fully indemnify TI and its representatives against any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to [TI's Terms of Sale](#), [TI's General Quality Guidelines](#), or other applicable terms available either on [ti.com](#) or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products. Unless TI explicitly designates a product as custom or customer-specified, TI products are standard, catalog, general purpose devices.

TI objects to and rejects any additional or different terms you may propose.

Copyright © 2025, Texas Instruments Incorporated

Last updated 10/2025

Looking for pricing, stock, or lifecycle information?

Click below to explore more details on WIN SOURCE:

 [View TLV1851QDBVRQ1 on WIN SOURCE](#)

 [Texas Instruments](#) Information

Optimize Your Supply Chain with WIN SOURCE Solutions

-  Global Sourcing Solution
-  Obsolete Management
-  Cost Control Management
-  Shortage Management
-  Alternative Solution
-  Excess Inventory Management