



**THE DATASHEET OF  
BQ25708RSNT**



# bq25708 SMBus Multi-Chemistry Battery Buck-Boost Charge Controller With System Power Monitor and Processor Hot Monitor

## 1 Features

- Charge 1- to 4-Cell Battery From Wide Range of Input Sources
  - 3.5-V to 24-V Input Operating Voltage
  - Supports USB2.0, USB 3.0, USB 3.1 (Type C), and USB\_PD Input Current Settings
  - Seamless Transition Between Buck and Boost Operation
  - Input current and Voltage Regulation (IDPM and VDPM) Against Source Overload
- Power/Current Monitor for CPU Throttling
  - Comprehensive PROCHOT Profile, IMVP8 Compliant
  - Input and Battery Current Monitor
  - System Power Monitor, IMVP8 Compliant
- Narrow-VDC (NVDC) Power Path Management
  - Instant-On With No Battery or Deeply Discharged Battery
  - Battery Supplements System When Adapter is Fully-Loaded
  - Ideal Diode Operation in Supplement Mode
- 800-kHz or 1.2-MHz Programmable Switching Frequency for 1- $\mu$ H to 3.3- $\mu$ H Inductor
- Host Control Interface for Flexible System Configuration
  - SMBus Port for Optimal System Performance and Status Reporting
  - Hardware Pin to Set Input Current Limit Without EC Control
- Integrated ADC to Monitor Voltage, Current and Power
- High Accuracy Regulation and Monitor
  - $\pm 0.5\%$  Charge Voltage Regulation
  - $\pm 2\%$  Input/Charge Current Regulation
  - $\pm 2\%$  Input/Charge Current Monitor
  - $\pm 5\%$  Power Monitor
- Safety
  - Thermal Shutdown
  - Input, System, Battery Overvoltage Protection
  - MOSFET Inductor Overcurrent Protection
- Low Battery Quiescent Current

- Input Current Optimizer (ICO) to Extract Max Input Power
- Charge any Battery Chemistry: Li+, LiFePO<sub>4</sub>, NiCd, NiMH, Lead Acid
- Package: 32-Pin 4 x 4 WQFN

## 2 Applications

- Ultra-Books, Notebooks, Detachable, and Tablet PCs and Power Bank
- Industrial and Medical Equipment
- Portable Equipment With Rechargeable Batteries

## 3 Description

The bq25708 is a synchronous NVDC battery buck-boost charge controller, offering low component count, high efficiency solution for space-constraint, multi-chemistry battery charging applications.

The NVDC-1 configuration allows the system to be regulated at battery voltage, but not drop below system minimum voltage. The system keeps operating even when the battery is completely discharged or removed. When load power exceeds input source rating, the battery goes into supplement mode and prevents the system from crashing.

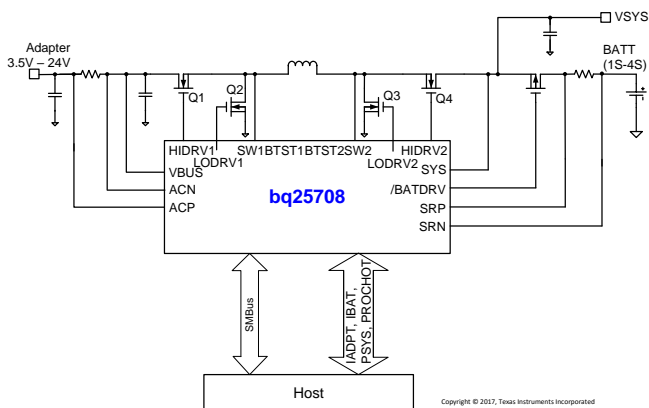
The bq25708 charges battery from a wide range of input sources including USB adapter, high voltage USB PD sources and traditional adapters.

### Device Information (1)

PART NUMBER	PACKAGE	BODY SIZE (NOM)
bq25708	WQFN (32)	4.00 mm x 4.00 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

### Application Diagram



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## 4 Revision History

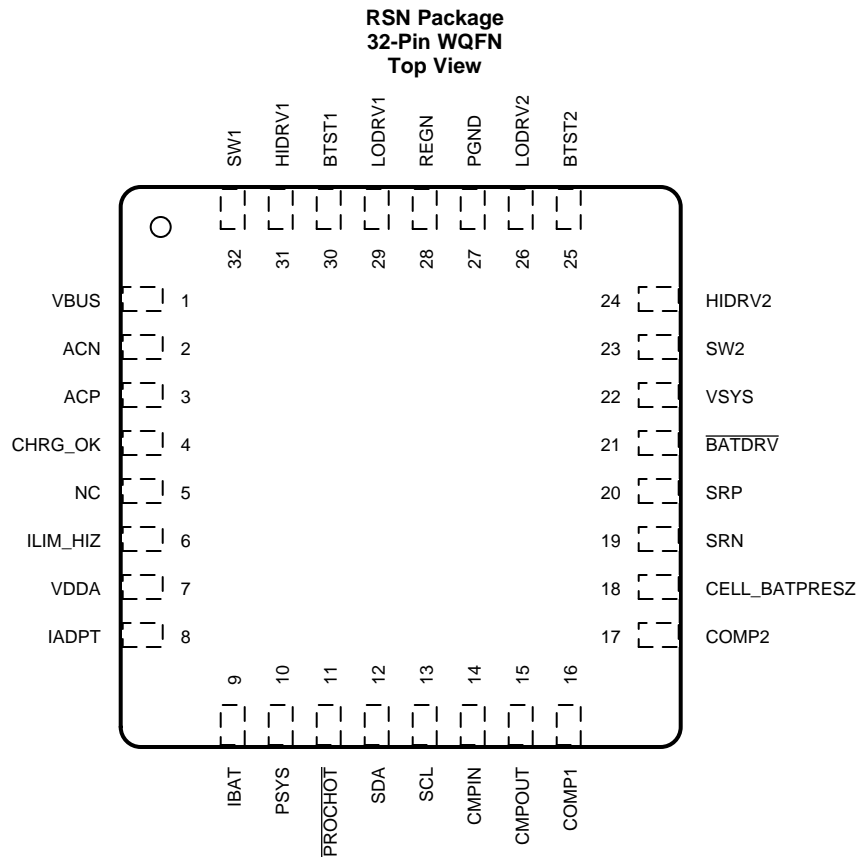
DATE	REVISION	NOTES
November 2017	*	Initial release.

## 5 Description (Continued)

During power up, the charger sets converter to buck, boost or buck-boost configuration based on input source and battery conditions. The charger automatically transits among buck, boost and buck-boost configuration without host control.

The bq25708 monitors adapter current, battery current and system power. The flexibly programmed  $\overline{\text{PROCHOT}}$  output goes directly to CPU for throttle back when needed.

## 6 Pin Configuration and Functions



### Pin Functions

PIN		I/O	DESCRIPTION
NAME	NUMBER		
ACN	2	PWR	Input current sense resistor negative input. The leakage on ACP and ACN are matched. The series resistors on the ACP and ACN pins are placed between sense resistor and filter cap. Refer to <a href="#">Application and Implementation</a> for ACP/ACN filter design.
ACP	3	PWR	Input current sense resistor positive input. The leakage on ACP and ACN are matched. The series resistors on the ACP and ACN pins are placed between sense resistor and filter cap. Refer to <a href="#">Application and Implementation</a> for ACP/ACN filter design.
$\overline{\text{BATDRV}}$	21	O	P-channel battery FET (BATFET) gate driver output. It is shorted to VSYS to turn off the BATFET. It goes 10 V below VSYS to fully turn on BATFET. BATFET is in linear mode to regulate VSYS at minimum system voltage when battery is depleted. BATFET is fully on during fast charge and supplement mode.
BTST1	30	PWR	Buck mode high side power MOSFET driver power supply. Connect a 0.047- $\mu\text{F}$ capacitor between SW1 and BTST1. The bootstrap diode between REGN and BTST1 is integrated.
BTST2	25	PWR	Boost mode high side power MOSFET driver power supply. Connect a 0.047- $\mu\text{F}$ capacitor between SW2 and BTST2. The bootstrap diode between REGN and BTST2 is integrated.
CELL_BATPRESZ	18	I	Battery cell selection pin for 1–4 cell battery setting. CELL_BATPRESZ pin is biased from VDDA. CELL_BATPRESZ pin also sets SYSOVP threshold to 5 V for 1-cell, 12 V for 2-cell, and 19.5 for 3-cell/4-cell. CELL_BATPRESZ pin is pulled below $V_{\text{CELL\_BATPRESZ\_FALL}}$ to indicate battery removal. The device exits LEARN mode, and disables charge. REG0x15() goes back to default.
CHRG_OK	4	O	Open drain active high indicator to inform the system good power source is connected to the charger input. Connect to the pullup rail via 10-k $\Omega$ resistor. When VBUS rises above 3.5V or falls below 24.5V, CHRG_OK is HIGH after 50ms deglitch time. When VBUS is falls below 3.2V or rises above 26V, CHRG_OK is LOW.

**Pin Functions (continued)**

PIN		I/O	DESCRIPTION
NAME	NUMBER		
CMPIN	14	I	Input of independent comparator. The independent comparator compares the voltage sensed on CMPIN pin to internal reference, and its output is on CMPOUT pin. Internal reference, output polarity and deglitch time is selectable by SMBus. With polarity HIGH (REG0x30[6]=1), place a resistor between CMPIN and CMPOUT to program hysteresis. With polarity LOW (REG0x30[6]=0), the internal hysteresis is 100 mV. If the independent comparator is not in use, tie CMPIN to ground.
CMPOUT	15	O	Open-drain output of independent comparator. Place pullup resistor from CMPOUT to pullup supply rail. Internal reference, output polarity and deglitch time are selectable by SMBus.
COMP2	17	I	Buck boost converter compensation pin 2. Refer to bq25700 EVM schematic for COMP2 pin RC network.
COMP1	16	I	Buck boost converter compensation pin 1. Refer to bq25700 EVM schematic for COMP1 pin RC network.
NC	5		Not Connected
HIDRV1	31	O	Buck mode high side power MOSFET (Q1) driver. Connect to high side n-channel MOSFET gate.
HIDRV2	24	O	Boost mode high side power MOSFET(Q4) driver. Connect to high side n-channel MOSFET gate.
IADPT	8	O	Buffered adapter current output. $V_{(IADPT)} = 20$ or $40 \times (V_{(ACP)} - V_{(ACN)})$ . With ratio selectable in REG0x12[4]. Place a resistor from the IADPT pin to ground corresponding to inductor in use. For 2.2 $\mu$ H, the resistor is 100 k $\Omega$ . Place 100-pF or less ceramic decoupling capacitor from IADPT pin to ground. IADPT output voltage is clamped below 3.3 V.
IBAT	9	O	Buffered battery current selected by SMBus. $V_{(IBAT)} = 8$ or $16 \times (V_{(SRP)} - V_{(SRN)})$ for charge current, or $V_{(IBAT)} = 8$ or $16 \times (V_{(SRN)} - V_{(SRP)})$ for discharge current, with ratio selectable in REG0x12[3]. Place 100-pF or less ceramic decoupling capacitor from IBAT pin to ground. This pin can be floating if not in use. Its output voltage is clamped below 3.3 V.
ILIM_HIZ	6	I	Input current limit input. Program ILIM_HIZ voltage by connecting a resistor divider from supply rail to ILIM_HIZ pin to ground. The pin voltage is calculated as: $V_{(ILIM\_HIZ)} = 1\text{ V} + 40 \times IDPM \times RAC$ , in which IDPM is the target input current. The input current limit used by the charger is the lower setting of ILIM_HIZ pin and REG0x3F(). When the pin voltage is below 0.4 V, the device enters Hi-Z mode with low quiescent current. When the pin voltage is above 0.8 V, the device is out of Hi-Z mode.
LODRV1	29	O	Buck mode low side power MOSFET (Q2) driver. Connect to low side n-channel MOSFET gate.
LODRV2	26	O	Boost mode low side power MOSFET (Q3) driver. Connect to low side n-channel MOSFET gate.
PGND	27	GND	Device power ground.
PROCHOT	11	O	Active low open drain output of processor hot indicator. It monitors adapter input current, battery discharge current, and system voltage. After any event in the PROCHOT profile is triggered, a pulse is asserted. The minimum pulse width is adjustable in REG0x33[5:2].
PSYS	10	O	Current mode system power monitor. The output current is proportional to the total power from the adapter and battery. The gain is selectable through SMBus. Place a resistor from PSYS to ground to generate output voltage. This pin can be floating if not in use. Its output voltage is clamped below 3.3 V. Place a capacitor in parallel with the resistor for filtering.
REGN	28	PWR	6-V linear regulator output supplied from VBUS or VSYS. The LDO is active when VBUS above UVLO. Connect a 2.2- or 3.3- $\mu$ F ceramic capacitor from REGN to power ground. REGN pin output is for power stage gate drive.
SCL	13	I	SMBus clock input. Connect to clock line from the host controller or smart battery. Connect a 10-k $\Omega$ pullup resistor according to SMBus specifications.
SDA	12	I/O	SMBus open-drain data I/O. Connect to data line from the host controller or smart battery. Connect a 10-k $\Omega$ pullup resistor according to SMBus specifications.
SRN	19	PWR	Charge current sense resistor negative input. SRN pin is for battery voltage sensing as well. Connect SRN pin with optional 0.1- $\mu$ F ceramic capacitor to GND for common-mode filtering. Connect a 0.1- $\mu$ F ceramic capacitor from SRP to SRN to provide differential mode filtering. The leakage current on SRP and SRN are matched. For reverse battery plug-in protection, 10- $\Omega$ series resistors are placed on SRP and SRN.

**Pin Functions (continued)**

PIN		I/O	DESCRIPTION
NAME	NUMBER		
SRP	20	PWR	Charge current sense resistor positive input. Connect 0.1- $\mu$ F ceramic capacitor from SRP to SRN to provide differential mode filtering. The leakage current on SRP and SRN are matched. For reverse battery plug-in protection, 10- $\Omega$ series resistors are placed on SRP and SRN. Connect SRP pin with optional 0.1-uF ceramic capacitor to GND for common-mode filtering.
SW1	32	PWR	Buck mode high side power MOSFET driver source. Connect to the source of the high side n-channel MOSFET.
SW2	23	PWR	Boost mode high side power MOSFET driver source. Connect to the source of the high side n-channel MOSFET.
VBUS	1	PWR	Charger input voltage. An input low pass filter of 1 $\Omega$ and 0.47 $\mu$ F (minimum) is recommended.
VDDA	7	PWR	Internal reference bias pin. Connect a 10- $\Omega$ resistor from REGN to VDDA and a 1- $\mu$ F ceramic capacitor from VDDA to power ground.
VSYS	22	PWR	Charger system voltage sensing. The system voltage regulation limit is programmed in REG0x15() and REG0x3E().
Thermal pad	–	–	Exposed pad beneath the IC. Analog ground and power ground star-connected near the IC's ground. Always solder thermal pad to the board, and have vias on the thermal pad plane connecting to power ground planes. It also serves as a thermal pad to dissipate the heat.

## 7 Specifications

### 7.1 Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted)<sup>(1)(2)</sup>

		MIN	MAX	UNIT
Voltage	SRN, SRP, ACN, ACP, VBUS, VSYS, $\overline{\text{BATDRV}}$	-0.3	30	V
	SW1, SW2	-2.0	30	
	BTST1, BTST2, HIDRV1, HIDRV2	-0.3	36	
	LODRV1, LODRV2 (25 ns)	-4.0	7	
	HIDRV1, HIDRV2 (25 ns)	-4.0	36	
	SW1, SW2 (25 ns)	-4.0	30	
	SDA, SCL, REGN, CHRГ_OK, CELL_BATPRESZ, ILIM_HIZ, LODRV1, LODRV2, VDDA, COMP1, COMP2, CMPIN, CMPOUT	-0.3	7	
	$\overline{\text{PROCHOT}}$	-0.3	5.5	
	IADPT, IBAT, PSYS	-0.3	3.6	
Differential voltage	BTST1-SW1, BTST2-SW2, HIDRV1-SW1, HIDRV2-SW2	-0.3	7	V
	SRP-SRN, ACP-ACN	-0.5	0.5	
Junction temperature range, T <sub>J</sub>		-40	155	°C
Storage temperature, T <sub>stg</sub>		-40	155	°C

- Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- All voltages are with respect to GND if not specified. Currents are positive into, negative out of the specified terminal. Consult Packaging Section of the data book for thermal limitations and considerations of packages.

### 7.2 ESD Ratings

		VALUE	UNIT
V <sub>(ESD)</sub> Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±2000	V
	Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±500	

- JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

### 7.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
Voltage	ACN, ACP, VBUS	0	24	V
	SRN, SRP, VSYS, $\overline{\text{BATDRV}}$	0	19.2	
	SW1, SW2	-2	24	
	BTST1, BTST2, HIDRV1, HIDRV2	0	30	
	SDA, SCL, REGN, CHRГ_OK, CELL_BATPRESZ, ILIM_HIZ, LODRV1, LODRV2, VDDA, COMP1, COMP2, CMPIN, CMPOUT	0	6.5	
	$\overline{\text{PROCHOT}}$	0	5.3	
	IADPT, IBAT, PSYS	0	3.3	
Differential voltage	BTST1-SW1, BTST2-SW2, HIDRV1-SW1, HIDRV2-SW2	0	6.5	V
	SRP-SRN, ACP-ACN	-0.35	0.35	
Junction temperature, T <sub>J</sub>		-40	125	°C
Operating free-air temperature, T <sub>A</sub>		-40	85	°C

## 7.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		bq25708	
		RSN (WQFN)	
		32 PINS	
Symbol	Description	Value	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	37.2	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	26.1	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	7.8	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	0.3	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	7.8	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	2.3	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) application report.

## 7.5 Electrical Characteristics

over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
$V_{INPUT\_OP}$	Input voltage operating range	3.5		26	V
<b>REGULATION ACCURACY</b>					
<b>MAX SYSTEM VOLTAGE REGULATION</b>					
$V_{SYSMAX\_RNG}$	System voltage regulation, measured on $V_{SYS}$	1.024		19.2	V
$V_{SYSMAX\_ACC}$	System voltage regulation accuracy (charge disable)	REG0x15() = 0x41A0H (16.800 V)	$V_{SRN} + 160$ mV		V
			-2%	2%	
		REG0x15() = 0x3130H (12.592 V)	$V_{SRN} + 160$ mV		V
			-2%	2%	
	REG0x15() = 0x20D0H (8.400 V)	$V_{SRN} + 160$ mV		V	
		-3%	3%		
	REG0x15() = 0x1060H (4.192 V)	$V_{SRN} + 160$ mV		V	
		-3%	3%		
<b>MINIMUM SYSTEM VOLTAGE REGULATION</b>					
$V_{SYSMIN\_RNG}$	System voltage regulation, measured on $V_{SYS}$	1.024		19.2	V
$V_{SYSMIN\_REG\_ACC}$	Minimum system voltage regulation accuracy (charge enable, VBAT below REG0x3E() setting)	REG0x3E() = 0x3000H	12.288		V
			-2%	2%	
		REG0x3E() = 0x2400H	9.216		V
			-2%	2%	
	REG0x3E() = 0x1800H	6.144		V	
		-3%	3%		
	REG0x3E() = 0x0E00H	3.584		V	
		-3%	4%		
<b>CHARGE VOLTAGE REGULATION</b>					
$V_{BAT\_RNG}$	Battery voltage regulation	1.024		19.2	V
$V_{BAT\_REG\_ACC}$	Battery voltage regulation accuracy (charge enable) (0°C to 85°C)	REG0x15() = 0x41A0H	16.8		V
			-0.5%	0.5%	
		REG0x15() = 0x3130H	12.592		V
			-0.5%	0.5%	
	REG0x15() = 0x20D0H	8.4		V	
		-0.6%	0.6%		
	REG0x15() = 0x1060H	4.192		V	
		-1.1%	1.2%		

## Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT	
<b>CHARGE CURRENT REGULATION IN FAST CHARGE</b>							
$V_{\text{IREG\_CHG\_RNG}}$	Charge current regulation differential voltage range	$V_{\text{IREG\_CHG}} = V_{\text{SRP}} - V_{\text{SRN}}$	0		81.28	mV	
$I_{\text{CHRG\_REG\_ACC}}$	Charge current regulation accuracy 10-m $\Omega$ current sensing resistor, $V_{\text{BAT}}$ above 0x3E() setting (0°C to 85°C)	REG0x14() = 0x1000H		4096		mA	
			-3%		2%		
		REG0x14() = 0x0800H		2048		3%	mA
			-4%				
		REG0x14() = 0x0400H		1024		5%	mA
			-5%				
		REG0x14() = 0x0200H		512		mA	
			-12%		12%		
<b>CHARGE CURRENT REGULATION IN LDO MODE</b>							
$I_{\text{CLAMP}}$	Pre-charge current clamp	CELL 2s-4s		384		mA	
		CELL 1 s, $V_{\text{SRN}} < 3 \text{ V}$		384		mA	
		CELL 1 s, $3 \text{ V} < V_{\text{SRN}} < V_{\text{SYS\_MIN}}$		2		A	
$I_{\text{PRECHRG\_REG\_ACC}}$	Pre-charge current regulation accuracy with 10-m $\Omega$ SRP/SRN series resistor, $V_{\text{BAT}}$ below REG0x3E() setting (0°C to 85°C)	REG0x14() = 0x0180H		384		mA	
		2S-4S	-15%		15%		
		1S	-25%		25%		
		REG0x14() = 0x0100H		256		mA	
		2S-4S	-20%		20%		
		1S	-35%		35%		
		REG0x14() = 0x00C0H		192		mA	
		2S-4S	-25%		25%		
		1S	-50%		50%		
		REG0x14() = 0x0080H		128		mA	
		2S-4S	-30%		30%		
$I_{\text{LEAK\_SRP\_SRN}}$	SRP, SRN leakage current mismatch (0°C to 85°C)		-12		-10	$\mu\text{A}$	
<b>INPUT CURRENT REGULATION</b>							
$V_{\text{IREG\_DPM\_RNG}}$	Input current regulation differential voltage range	$V_{\text{IREG\_DPM}} = V_{\text{ACP}} - V_{\text{ACN}}$	0.5		64	mV	
$I_{\text{DPM\_REG\_ACC}}$	Input current regulation accuracy with 10-m $\Omega$ ACP/ACN series resistor (-40°C to 105°C)	REG0x3F() = 0x4FFFH		3820		4000 mA	
		REG0x3F() = 0x3BFFH		2830		3000 mA	
		REG0x3F() = 0x1DFFH		1350		1500 mA	
		REG0x3F() = 0x09FFH		340		500 mA	
$I_{\text{LEAK\_ACP\_ACN}}$	ACP, ACN leakage current mismatch		-16		10	$\mu\text{A}$	
$V_{\text{IREG\_DPM\_RNG\_ILIM}}$	Voltage Range for input current regulation		1		4	V	
$I_{\text{DPM\_REG\_ACC\_ILIM}}$	Input Current Regulation Accuracy on ILIM_HIZ pin $V_{\text{ILIM\_HIZ}} = 1 \text{ V} + 40 \times I_{\text{DPM}} \times R_{\text{AC}}$ , with 10-m $\Omega$ ACP/ACN series resistor	$V_{\text{ILIM\_HIZ}} = 2.6 \text{ V}$		3800	4000	4200 mA	
		$V_{\text{ILIM\_HIZ}} = 2.2 \text{ V}$		2800	3000	3200 mA	
		$V_{\text{ILIM\_HIZ}} = 1.6 \text{ V}$		1300	1500	1700 mA	
		$V_{\text{ILIM\_HIZ}} = 1.2 \text{ V}$		300	500	700 mA	
$I_{\text{LEAK\_ILIM}}$	$I_{\text{LIM\_HIZ}}$ pin leakage		-1		1	$\mu\text{A}$	

## Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>INPUT VOLTAGE REGULATION</b>						
V <sub>I</sub> REG_DPM_RNG	Input voltage regulation range	Voltage on VBUS	3.2		19.52	V
V <sub>DPM</sub> _REG_ACC	Input voltage regulation accuracy	REG0x3D()=0x3C80H		18688		mV
			-2%		2%	
		REG0x3D()=0x1E00H		10880		mV
			-2.5%		2.5%	
		REG0x3D()=0x0500H		4480		mV
			-3%		5%	
<b>REFERENCE AND BUFFER</b>						
<b>REGN REGULATOR</b>						
V <sub>REGN</sub> _REG	REGN regulator voltage (0 mA–60 mA)	V <sub>VBUS</sub> = 10 V	5.7	6	6.3	V
V <sub>DROPOUT</sub>	REGN voltage in drop out mode	V <sub>VBUS</sub> = 5 V, I <sub>LOAD</sub> = 20 mA	3.8	4.3	4.6	V
I <sub>REGN</sub> _LIM_Charging	REGN current limit when converter is enabled	V <sub>VBUS</sub> = 10 V, force V <sub>REGN</sub> = 4 V	50	65		mA
C <sub>REGN</sub>	REGN output capacitor required for stability	I <sub>LOAD</sub> = 100 μA to 50 mA	2.2			μF
C <sub>VDDA</sub>	REGN output capacitor required for stability	I <sub>LOAD</sub> = 100 μA to 50 mA	1			μF
<b>QUIESCENT CURRENT</b>						
I <sub>BAT</sub> _BATFET_ON	System powered by battery. BATFET on. I <sub>SRN</sub> + I <sub>SRP</sub> + I <sub>SW2</sub> + I <sub>BTST2</sub> + I <sub>SW1</sub> + I <sub>BTST1</sub> + I <sub>ACP</sub> + I <sub>ACN</sub> + I <sub>VBUS</sub> + I <sub>VSYS</sub>	V <sub>BAT</sub> = 18 V, REG0x12[15] = 1, in low power mode		22	45	μA
		V <sub>BAT</sub> = 18 V, REG0x12[15] = 1, in low power mode, REG0x30[14:13] = 01, REGN off		105	175	μA
		V <sub>BAT</sub> =18 V, REG0x12[15] = 1, in low power mode, REG0x30[14:13]= 10, REGN off		60	90	μA
		V <sub>BAT</sub> = 18 V, REG0x12[15] = 0, REG0x30[12] = 0, REGN on, EN_PSYS		860	1150	μA
		V <sub>BAT</sub> = 18 V, REG0x12[15] = 0, REG0x30[12] = 1, REGN on		960	1250	
I <sub>AC</sub> _SW_LIGHT_buck	Input current during PFM in buck mode, no load, I <sub>VBUS</sub> + I <sub>ACP</sub> + I <sub>ACN</sub> + I <sub>VSYS</sub> + I <sub>SRP</sub> + I <sub>SRN</sub> + I <sub>SW1</sub> + I <sub>BTST</sub> + I <sub>SW2</sub> + I <sub>BTST2</sub>	V <sub>IN</sub> = 20 V, V <sub>BAT</sub> = 12.6 V, 3 s, REG0x12[10] = 0; MOSFET Q <sub>g</sub> = 4 nC		2.2		mA
I <sub>AC</sub> _SW_LIGHT_boost	Input current during PFM in boost mode, no load, I <sub>VBUS</sub> + I <sub>ACP</sub> + I <sub>ACN</sub> + I <sub>VSYS</sub> + I <sub>SRP</sub> + I <sub>SRN</sub> + I <sub>SW1</sub> + I <sub>BTST2</sub> + I <sub>SW2</sub> + I <sub>BTST2</sub>	V <sub>IN</sub> = 5 V, V <sub>BAT</sub> = 8.4 V, 2 s, REG0x12[10] = 0; MOSFET Q <sub>g</sub> = 4 nC		2.7		mA
I <sub>AC</sub> _SW_LIGHT_buckboost	Input current during PFM in buck boost mode, no load, I <sub>VBUS</sub> + I <sub>ACP</sub> + I <sub>ACN</sub> + I <sub>VSYS</sub> + I <sub>SRP</sub> + I <sub>SRN</sub> + I <sub>SW1</sub> + I <sub>BTST1</sub> + I <sub>SW2</sub> + I <sub>BTST2</sub>	V <sub>IN</sub> = 12 V, V <sub>BAT</sub> = 12 V, REG0x12[10] = 0; MOSFET Q <sub>g</sub> = 4 nC		2.4		mA
V <sub>ACP/N</sub> _OP	Input common mode range	Voltage on ACP/ACN	3.8		26	V
V <sub>IADPT</sub> _CLAMP	I <sub>ADPT</sub> output clamp voltage		3.1	3.2	3.3	V
I <sub>IADPT</sub>	I <sub>ADPT</sub> output current				1	mA

## Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
A <sub>IADPT</sub>	Input current sensing gain	$V_{(IADPT)} / V_{(ACP-ACN)}$ , REG0x12[4] = 0	20		V/V	
		$V_{(IADPT)} / V_{(ACP-ACN)}$ , REG0x12[4] = 1	40		V/V	
V <sub>IADPT_ACC</sub>	Input current monitor accuracy	$V_{(ACP-ACN)} = 40.96$ mV	-2%	2%		
		$V_{(ACP-ACN)} = 20.48$ mV	-3%	3%		
		$V_{(ACP-ACN)} = 10.24$ mV	-6%	6%		
		$V_{(ACP-ACN)} = 5.12$ mV	-10%	10%		
C <sub>IADPT_MAX</sub>	Maximum output load capacitance			100	pF	
V <sub>SRP/N_OP</sub>	Battery common mode range	Voltage on SRP/SRN		2.5	18	V
V <sub>IBAT_CLAMP</sub>	IBAT output clamp voltage	3.1	3.2	3.3	V	
I <sub>IBAT</sub>	IBAT output current			1	mA	
A <sub>IBAT</sub>	Charge and discharge current sensing gain on IBAT pin	$V_{(IBAT)} / V_{(SRN-SRP)}$ , REG0x12[3] = 0,	8		V/V	
		$V_{(IBAT)} / V_{(SRN-SRP)}$ , REG0x12[3] = 1,	16		V/V	
I <sub>IBAT_CHG_ACC</sub>	Charge and discharge current monitor accuracy on IBAT pin	$V_{(SRN-SRP)} = 40.96$ mV	-2%	2%		
		$V_{(SRN-SRP)} = 20.48$ mV	-3%	4%		
		$V_{(SRN-SRP)} = 10.24$ mV	-6%	6%		
		$V_{(SRN-SRP)} = 5.12$ mV	-12%	12%		
C <sub>IBAT_MAX</sub>	Maximum output load capacitance			100	pF	
<b>SYSTEM POWER SENSE AMPLIFIER</b>						
V <sub>PSYS</sub>	PSYS output voltage range		0	3.3	V	
I <sub>PSYS</sub>	PSYS output current		0	160	μA	
A <sub>PSYS</sub>	PSYS system gain	$V_{(PSYS)} / (P_{(IN)} + P_{(BAT)})$ , REG0x30[9] = 1	1		μA/W	
V <sub>PSYS_ACC</sub>	PSYS gain accuracy (REG0x3B[9] = 1), T <sub>A</sub> = 25°C	Adapter only with system power = 19.5 V / 45 W	-5%	5%		
		Battery only with system power 11 V / 44 W (25°C)	-5%	5%		
V <sub>PSYS_CLAMP</sub>	PSYS clamp voltage	3		3.3	V	
<b>COMPARATOR</b>						
<b>VBUS UNDER VOLTAGE LOCKOUT COMPARATOR</b>						
V <sub>VBUS_UVLOZ</sub>	VBUS undervoltage rising threshold	VBUS rising	2.34	2.55	2.77	V
V <sub>VBUS_UVLO</sub>	VBUS undervoltage falling threshold	VBUS falling	2.2	2.4	2.6	V
V <sub>VBUS_UVLO_HYST</sub>	VBUS undervoltage hysteresis			150	mV	
V <sub>VBUS_CONVEN</sub>	VBUS converter enable rising threshold	VBUS rising	3.2	3.5	3.9	V
V <sub>VBUS_CONVENZ</sub>	VBUS converter enable falling threshold	VBUS falling	2.9	3.2	3.5	V
V <sub>VBUS_CONVEN_HYST</sub>	VBUS converter enable hysteresis			400	mV	

## Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>BATTERY UNDER VOLTAGE LOCKOUT COMPARATOR</b>						
V <sub>VBAT_UVLOZ</sub>	VBAT undervoltage rising threshold	VSRN rising	2.35	2.55	2.75	V
V <sub>VBAT_UVLO</sub>	VBAT undervoltage falling threshold	VSRN falling	2.2	2.4	2.6	V
V <sub>VBAT_UVLO_HYST</sub>	VBAT undervoltage hysteresis			150		mV
<b>BATTERY LOWV COMPARATOR (Pre-charge to Fast Charge Thresold for 1S)</b>						
V <sub>BATLV_FALL</sub>	BATLOWV falling threshold	1 s		2.80		V
V <sub>BATLV_RISE</sub>	BATLOWV rising threshold			3.00		V
V <sub>BATLV_RHYST</sub>	BATLOWV hysteresis			200		mV
<b>INPUT OVER-VOLTAGE COMPARATOR (ACOV)</b>						
V <sub>ACOV_RISE</sub>	VBUS overvoltage rising threshold	VBUS rising	25	26	27	V
V <sub>ACOV_FALL</sub>	VBUS overvoltage falling threshold	VBUS falling	24	24.5	25	V
V <sub>ACOV_HYST</sub>	VBUS overvoltage hysteresis			1.5		V
t <sub>ACOV_RISE_DEG</sub>	VBUS overvoltage rising deglitch	VBUS rising to stop converter		100		μs
t <sub>ACOV_FALL_DEG</sub>	VBUS overvoltage falling deglitch	VBUS falling to start converter		1		ms
<b>INPUT OVER CURRENT COMPARATOR (ACOC)</b>						
V <sub>ACOC</sub>	ACP to ACN rising threshold, w.r.t. ILIM2 in REG0x33[15:11]	Voltage across input sense resistor rising, Reg0x31[2] = 1	195%	210%	225%	
V <sub>ACOC_FLOOR</sub>	Measure between ACP and ACN	Set IDPM to minimum	44	50	56	mV
V <sub>ACOC_CEILING</sub>	Measure between ACP and ACN	Set IDPM to maximum	172	180	188	mV
t <sub>ACOC_DEG_RISE</sub>	Rising deglitch time	Deglitch time to trigger ACOC		250		μs
t <sub>ACOC_RELAX</sub>	Relax time	Relax time before converter starts again		250		ms
<b>SYSTEM OVER-VOLTAGE COMPARATOR (SYSOVP)</b>						
V <sub>SYSOVP_RISE</sub>	System overvoltage rising threshold to turn off converter	1 s	4.85	5	5.1	V
		2 s	11.7	12	12.2	V
		3s	18	18.5	19	V
V <sub>SYSOVP_FALL</sub>	System overvoltage falling threshold	1 s		4.8		V
		2 s		11.5		V
		3s		18		V
I <sub>SYSOVP</sub>	Discharge current when SYSOVP stop switching was triggered	on SYS		20		mA
<b>BAT OVER-VOLTAGE COMPARATOR (BATOVP)</b>						
V <sub>BATOVP_RISE</sub>	Overvoltage rising threshold as percentage of V <sub>BAT_REG</sub> in REG0x15()	1 s, 4.2 V	102.5%	104%	105.7%	
		2 s - 4 s	102.5%	104%	105%	
V <sub>BATOVP_FALL</sub>	Overvoltage falling threshold as percentage of V <sub>BAT_REG</sub> in REG0x15()	1 s	100%	102%	104%	
		2 s - 4 s	100%	102%	103%	
V <sub>BATOVP_HYST</sub>	Overvoltage hysteresis as percentage of V <sub>BAT_REG</sub> in REG0x15()	1 s		2%		
		2 s - 4 s		2%		

## Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
$I_{BATOVP}$	Discharge current during BATOVP	on SRP and SRN		20		mA
$t_{BATOVP\_RISE}$	Ovoltage rising deglitch to turn off BATDRV to disable charge			20		ms
<b>CONVERTER OVER-CURRENT COMPARATOR (Q2)</b>						
VOCP_limit_Q2	Converter Over-Current Limit	Reg0x31[5]=1		150		mV
		Reg0x31[5]=0		210		
VOCP_limit_SYSSH ORT_Q2	System Short or SRN<2.5 V	Reg0x31[5]=1		45		mV
		Reg0x31[5]=0		60		
<b>CONVERTER OVER-CURRENT COMPARATOR (ACX)</b>						
VOCP_limit_ACX	Converter Over-Current Limit	Reg0x31[4]=1		150		mV
		Reg0x31[4]=0		280		
VOCP_limit_SYSSH ORT_ACX	System Short or SRN<2.5 V	Reg0x31[4]=1		90		mV
		Reg0x31[4]=0		150		
<b>THERMAL SHUTDOWN COMPARATOR</b>						
$T_{SHUT\_RISE}$	Thermal shutdown rising temperature	Temperature increasing		155		°C
$T_{SHUT\_FALL}$	Thermal shutdown falling temperature	Temperature reducing		135		°C
$T_{SHUT\_HYS}$	Thermal shutdown hysteresis			20		°C
$t_{SHUT\_RDEG}$	Thermal shutdown rising deglitch			100		µs
$t_{SHUT\_FHYS}$	Thermal shutdown falling deglitch			12		ms
<b>VSYS PROCHOT COMPARATOR</b>						
$V_{SYS\_PROCHOT}$	$V_{SYS}$ threshold falling threshold	Reg0x36[7:6] = 00, 1 s		2.85		V
		Reg0x36[7:6] = 00, 2–4 s		5.75		V
		Reg0x36[7:6] = 01, 1 s	2.95	3.1	3.25	V
		Reg0x36[7:6] = 01, 2–4 s	5.8	5.95	6.1	V
		Reg0x36[7:6] = 10, 1 s		3.3		V
		Reg0x36[7:6] = 10, 2–4 s		6.25		V
		Reg0x36[7:6] = 11, 1 s		3.5		V
		Reg0x36[7:6] = 11, 2–4 s		6.5		V
$t_{SYS\_PRO\_RISE\_DEG}$	$V_{SYS}$ rising deglitch for throttling			8		µs
<b>ICRIT PROCHOT COMPARATOR</b>						
$V_{ICRIT\_PRO}$	Input current rising threshold for throttling as percentage above ILIM2 (REG0x33[15:11])	Reg0x36[15:11] = 00000	105%	110%	116%	
		Reg0x36[15:11] = 01001	142%	150%	155%	
		Reg0x36[15:11] = 11110	410%	430%	455%	
<b>INOM PROCHOT COMPARATOR</b>						
$V_{INOM\_PRO}$	INOM rising threshold as percentage of IIN (REG0x3F())		105%	110%	116%	
<b>IDCHG PROCHOT COMPARATOR</b>						
$V_{IDCHG\_PRO}$	IDCHG threshold for throttling for IDSCHG of 6 A	Reg0x38 [15:10]=001100		6144		mA
			95%		102%	

## Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT	
<b>INDEPENDENT COMPARATOR</b>						
V <sub>INDEP_CMP</sub>	Independent comparator threshold	Reg0x30[7] = 1, CMPIN falling	1.17	1.2	1.23	V
		Reg0x30[7] = 0, CMPIN falling	2.27	2.3	2.33	V
V <sub>INDEP_CMP_HYS</sub>	Independent comparator hysteresis	Reg0x06[6] = 0, CMPIN falling		100		mV
<b>POWER MOSFET DRIVER</b>						
<b>PWM OSCILLATOR AND RAMP</b>						
F <sub>SW</sub>	PWM switching frequency	Reg0x00[9] = 0	1020	1200	1380	kHz
		Reg0x00[9] = 1	680	800	920	kHz
<b>BATFET GATE DRIVER (BATDRV)</b>						
V <sub>BATDRV_ON</sub>	Gate drive voltage on BATFET		8.5	10	11.5	V
V <sub>BATDRV_DIODE</sub>	Drain-source voltage on BATFET during ideal diode operation			30		mV
R <sub>BATDRV_ON</sub>	Measured by sourcing 10-μA current to BATDRV		3	4	6	kΩ
R <sub>BATDRV_OFF</sub>	Measured by sinking 10-μA current from BATDRV			1.2	2.1	kΩ
<b>PWM HIGH SIDE DRIVER (HIDRV Q1)</b>						
R <sub>DS_HI_ON_Q1</sub>	High side driver (HSD) turnon resistance	V <sub>BTST1</sub> – V <sub>SW1</sub> = 5 V		6		Ω
R <sub>DS_HI_OFF_Q1</sub>	High side driver turnoff resistance	V <sub>BTST1</sub> – V <sub>SW1</sub> = 5 V		1.3	2.2	Ω
V <sub>BTST1_REFRESH</sub>	Bootstrap refresh comparator falling threshold voltage	V <sub>BTST1</sub> – V <sub>SW1</sub> when low side refresh pulse is requested	3.2	3.7	4.6	V
<b>PWM HIGH SIDE DRIVER (HIDRV Q4)</b>						
R <sub>DS_HI_ON_Q4</sub>	High side driver (HSD) turnon resistance	V <sub>BTST2</sub> – V <sub>SW2</sub> = 5 V		6		Ω
R <sub>DS_HI_OFF_Q4</sub>	High side driver turnoff resistance	V <sub>BTST2</sub> – V <sub>SW2</sub> = 5 V		1.5	2.4	Ω
V <sub>BTST2_REFRESH</sub>	Bootstrap refresh comparator falling threshold voltage	V <sub>BTST2</sub> – V <sub>SW2</sub> when low side refresh pulse is requested	3.3	3.7	4.6	V
<b>PWM LOW SIDE DRIVER (LODRV Q2)</b>						
R <sub>DS_LO_ON_Q2</sub>	Low side driver (LSD) turnon resistance	V <sub>BTST1</sub> – V <sub>SW1</sub> = 5.5 V		6		Ω
R <sub>DS_LO_OFF_Q2</sub>	Low side driver turnoff resistance	V <sub>BTST1</sub> – V <sub>SW1</sub> = 5.5 V		1.7	2.6	Ω
<b>PWM LOW SIDE DRIVER (LODRV Q3)</b>						
R <sub>DS_LO_ON_Q3</sub>	Low side driver (LSD) turnon resistance	V <sub>BTST2</sub> – V <sub>SW2</sub> = 5.5 V		7.6		Ω
R <sub>DS_LO_OFF_Q3</sub>	Low side driver turnoff resistance	V <sub>BTST2</sub> – V <sub>SW2</sub> = 5.5 V		2.9	4.6	Ω
<b>INTERNAL SOFT START During Charge Enable</b>						
SSSTEP_DAC	Soft Start Step Size			64		mA
SSSTEP_DAC	Soft Start Step Time			8		μs

## Electrical Characteristics (continued)

over operating free-air temperature range (unless otherwise noted)

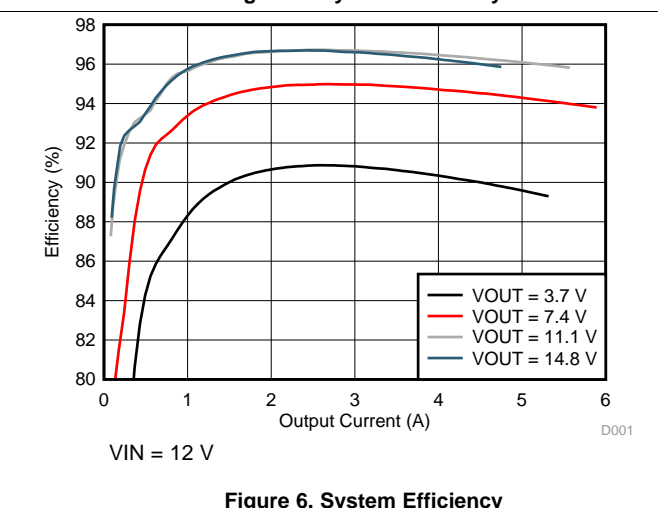
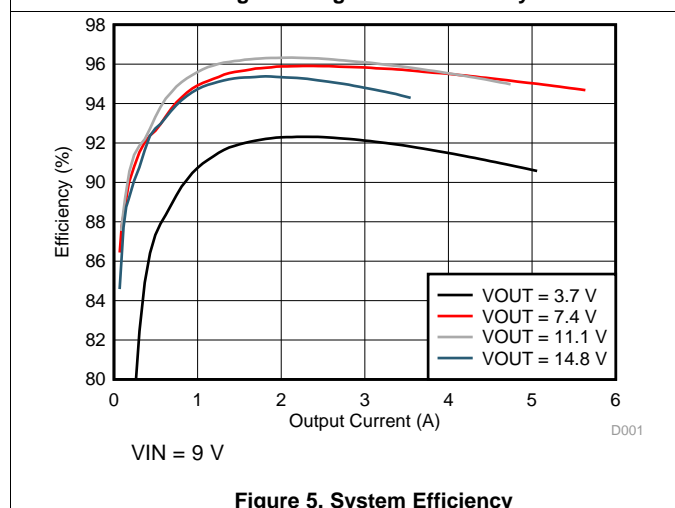
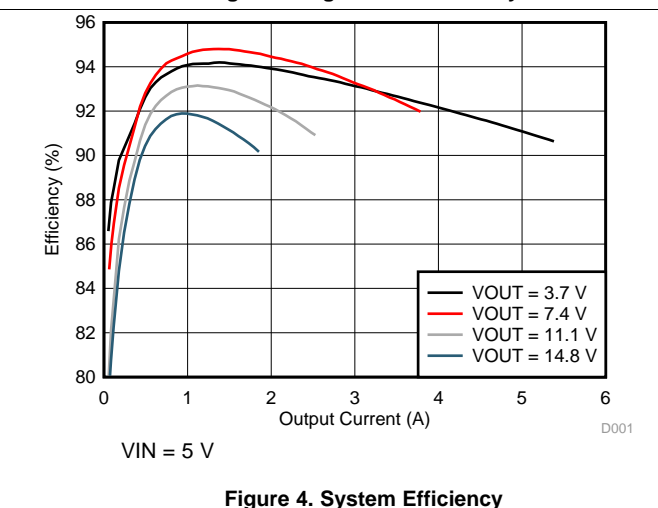
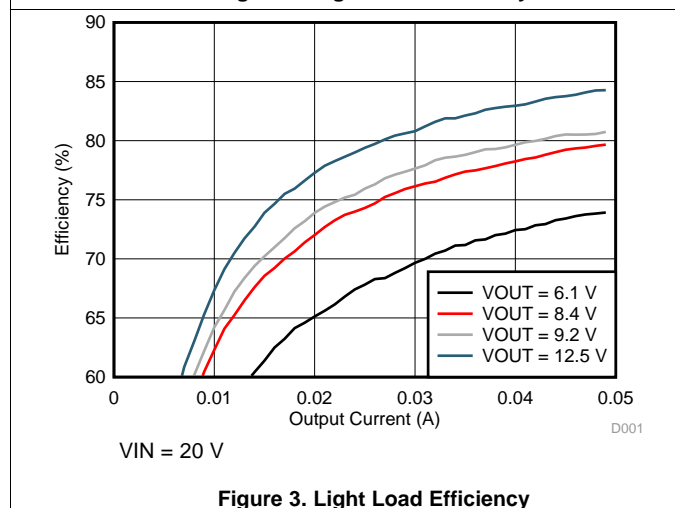
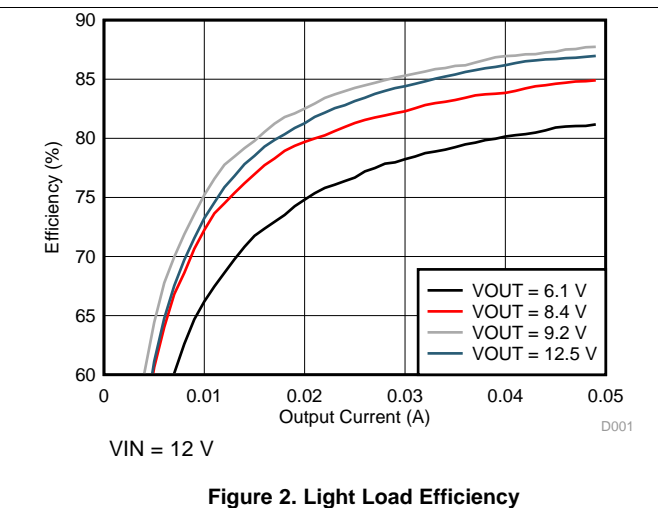
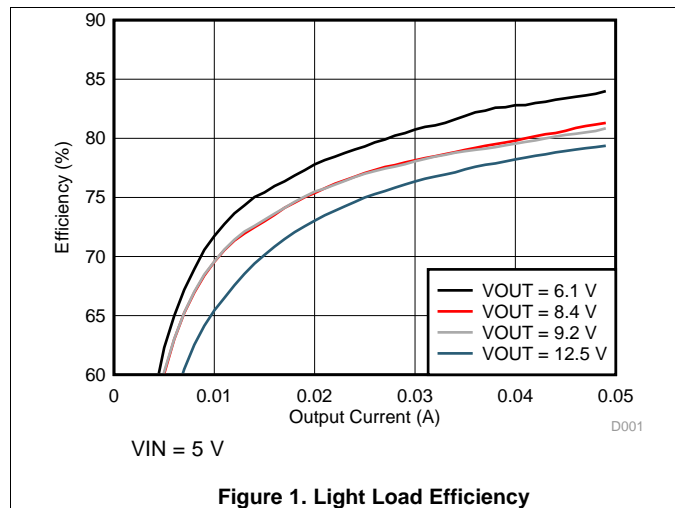
PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>INTEGRATED BTST DIODE (D1)</b>						
V <sub>F_D1</sub>	Forward bias voltage	I <sub>F</sub> = 20 mA at 25°C		0.8		V
V <sub>R_D1</sub>	Reverse breakdown voltage	I <sub>R</sub> = 2 μA at 25°C			20	V
<b>INTEGRATED BTST DIODE (D2)</b>						
V <sub>F_D2</sub>	Forward bias voltage	I <sub>F</sub> = 20 mA at 25°C		0.8		V
V <sub>R_D2</sub>	Reverse breakdown voltage	I <sub>R</sub> = 2 μA at 25°C			20	V
<b>PWM DRIVERS TIMING</b>						
<b>INTERFACE</b>						
<b>LOGIC INPUT (SDA, SCL)</b>						
V <sub>IN_LO</sub>	Input low threshold	SMBus			0.8	V
V <sub>IN_HI</sub>	Input high threshold	SMBus (bq25708)	2.1			V
<b>LOGIC OUTPUT OPEN DRAIN (SDA, CHRG_OK, CMPOUT)</b>						
V <sub>OUT_LO</sub>	Output saturation voltage	5-mA drain current			0.4	V
V <sub>OUT_LEAK</sub>	Leakage current	V = 7 V	–1		1	mA
<b>LOGIC OUTPUT OPEN DRAIN SDA</b>						
V <sub>OUT_LO_SDA</sub>	Output Saturation Voltage	5 mA drain current			0.4	V
V <sub>OUT_LEAK_SDA</sub>	Leakage Current	V = 7V	–1		1	mA
<b>LOGIC OUTPUT OPEN DRAIN CHRG_OK</b>						
V <sub>OUT_LO_CHRG_OK</sub>	Output Saturation Voltage	5 mA drain current			0.4	V
V <sub>OUT_LEAK_CHRG_OK</sub>	Leakage Current	V = 7V	–1		1	mA
<b>LOGIC OUTPUT OPEN DRAIN CMPOUT</b>						
V <sub>OUT_LO_CMPOUT</sub>	Output Saturation Voltage	5 mA drain current			0.4	V
V <sub>OUT_LEAK_CMPOUT</sub>	Leakage Current	V = 7V	–1		1	mA
<b>LOGIC OUTPUT OPEN DRAIN (PROCHOT)</b>						
V <sub>OUT_LO_PROCHOT</sub>	Output saturation voltage	50-Ω pullup to 1.05 V / 5-mA load			300	mV
V <sub>OUT_LEAK_PROCHOT</sub>	Leakage current	V = 5.5 V	–1		1	mA
<b>ANALOG INPUT (ILIM_HIZ)</b>						
V <sub>HIZ_LO</sub>	Voltage to get out of HIZ mode	ILIM_HIZ pin rising	0.8			V
V <sub>HIZ_HIGH</sub>	Voltage to enable HIZ mode	ILIM_HIZ pin falling			0.4	V
<b>ANALOG INPUT (CELL_BATPRESZ)</b>						
V <sub>CELL_4S</sub>	4S	REGN = 6 V, as percentage of REGN	68.4%	75%		
V <sub>CELL_3S</sub>	3s	REGN = 6 V, as percentage of REGN	51.7%	55%	65%	
V <sub>CELL_2S</sub>	2S	REGN = 6 V, as percentage of REGN	35%	40%	49.1%	
V <sub>CELL_1S</sub>	1S	REGN = 6 V, as percentage of REGN	18.4%	25%	31.6%	
V <sub>CELL_BATPRESZ_RISE</sub>	Battery is present		18%			
V <sub>CELL_BATPRESZ_FALL</sub>	Battery is removed	CELL_BATPRESZ falling			15%	
<b>ANALOG INPUT (COMP1, COMP2)</b>						
I <sub>LEAK_COMP1</sub>	COMP1 Leakage		–120		120	nA
I <sub>LEAK_COMP2</sub>	COMP2 Leakage		–120		120	nA

## 7.6 Timing Requirements

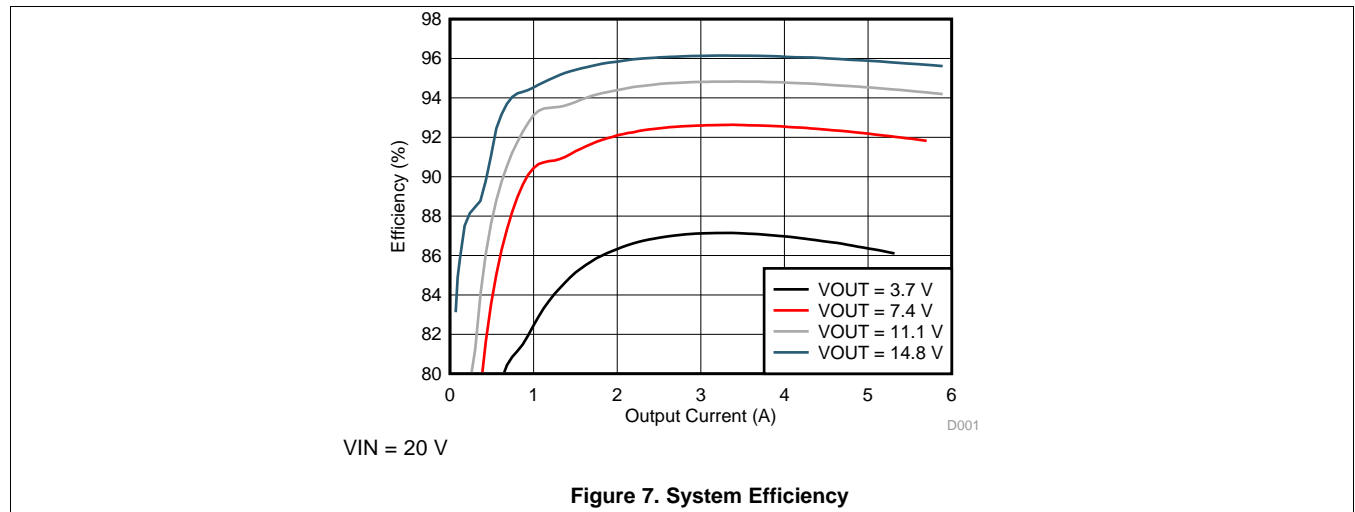
		MIN	TYP	MAX	UNIT
<b>SMBus TIMING CHARACTERISTICS</b>					
$t_r$	SCLK/SDATA rise time			1	$\mu$ s
$t_f$	SCLK/SDATA fall time			300	ns
$t_{W(H)}$	SCLK pulse width high	4		50	$\mu$ s
$t_{W(L)}$	SCLK Pulse Width Low	4.7			$\mu$ s
$t_{SU(STA)}$	Setup time for START condition	4.7			$\mu$ s
$t_{H(STA)}$	START condition hold time after which first clock pulse is generated	4			$\mu$ s
$t_{SU(DAT)}$	Data setup time	250			$\mu$ s
$t_{H(DTA)}$	Data hold time	300			$\mu$ s
$t_{SU(STOP)}$	Setup time for STOP condition	4			$\mu$ s
$t_{(BUF)}$	Bus free time between START and STOP condition	4.7			$\mu$ s
$F_{S(CL)}$	Clock Frequency	10		100	KHz
<b>HOST COMMUNICATION FAILURE</b>					
$t_{timeout}$	SMBus bus release timeout <sup>(1)</sup>	25		35	ms
$t_{BOOT}$	Deglitch for watchdog reset signal	10			ms
$t_{WDI}$	Watchdog timeout period, ChargeOption() bit [14:13] = 01 <sup>(2)</sup>	35	44	53	s
	Watchdog timeout period, ChargeOption() bit [14:13] = 10 <sup>(2)</sup>	70	88	105	s
	Watchdog timeout period, ChargeOption() bit [14:13] = 11 <sup>(2)</sup> (default)	140	175	210	s

- (1) Devices participating in a transfer will timeout when any clock low exceeds the 25ms minimum timeout period. Devices that have detected a timeout condition must reset the communication no later than the 35 ms maximum timeout period. Both a master and a slave must adhere to the maximum value specified as it incorporates the cumulative stretch limit for both a master (10 ms) and a slave (25 ms).
- (2) User can adjust threshold via SMBus ChargeOption() REG0x12.

## 7.7 Typical Characteristics



**Typical Characteristics (continued)**



## 8 Detailed Description

### 8.1 Overview

The bq25708 is a buck boost NVDC (narrow voltage DC) charge controller for multi-chemistry portable applications such as notebook, detachable, ultrabook, tablet and other mobile devices with rechargeable batteries. It provides seamless transition between converter operation modes (buck, boost, or buck boost), fast transient response, and high light load efficiency.

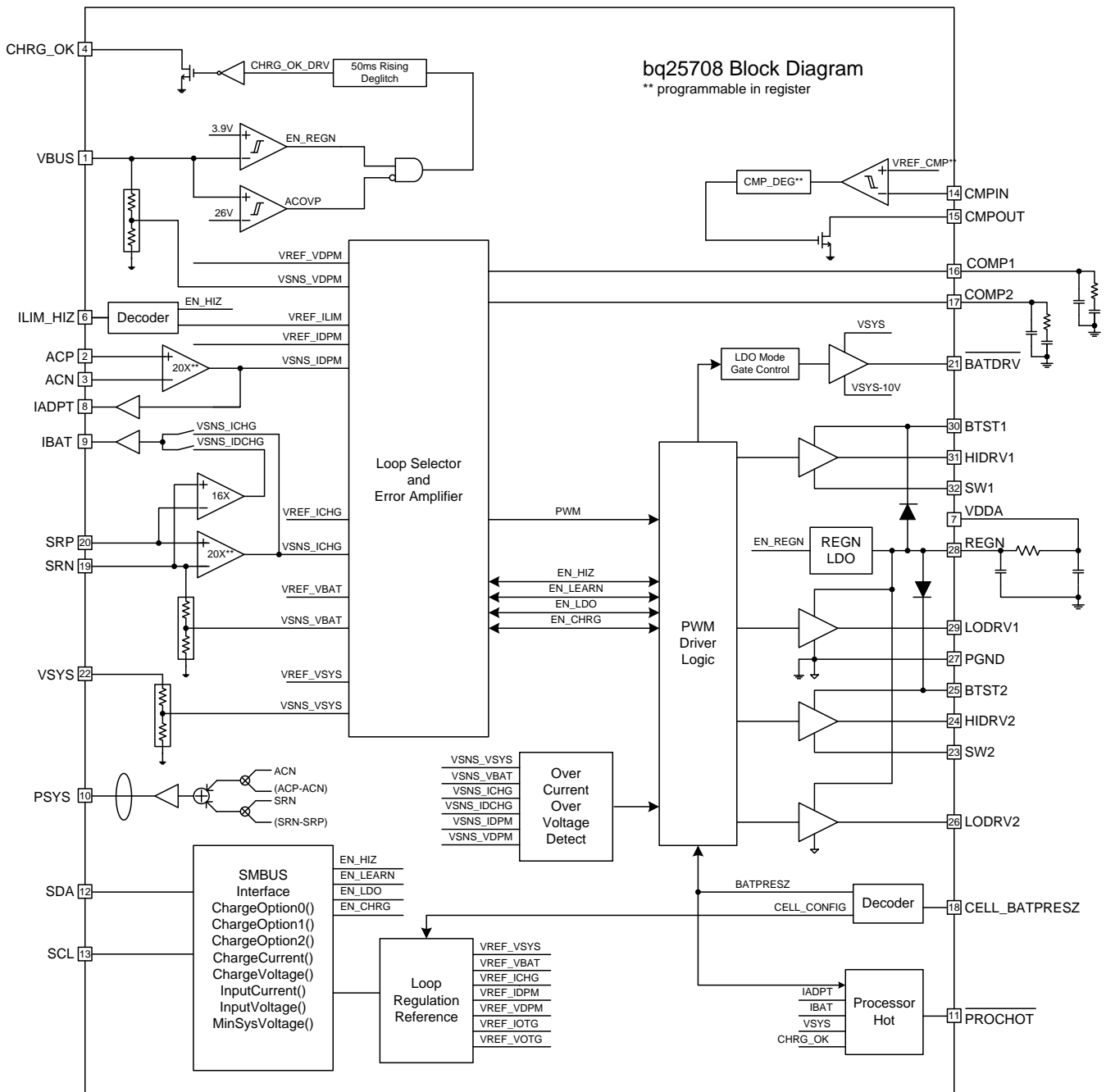
The bq25708 supports wide range of power sources, including USB PD ports, legacy USB ports, traditional ACDC adapters, etc. It takes input voltage from 3.5 V to 24 V, and charges battery of 1-4 series.

The bq25708 features Dynamic Power Management (DPM) to limit the input power and avoid AC adapter overloading. During battery charging, as the system power increases, the charging current will reduce to maintain total input current below adapter rating. If system power demand temporarily exceeds adapter rating, the bq25708 supports NVDC architecture to allow battery discharge energy to supplement system power. For details, refer to [System Voltage Regulation](#) section.

In order to be compliant with an Intel IMVP8 compliant system, the bq25708 includes PSYS function to monitor the total platform power from adapter and battery. Besides PSYS, it provides both an independent input current buffer (IADPT) and a battery current buffer (IBAT) with highly accurate current sense amplifiers. If the platform power exceeds the available power from adapter and battery, a  $\overline{\text{PROCHOT}}$  signal is asserted to CPU so that the CPU optimizes its performance to the power available to the system.

The bq25708 controls input current, charge current and charge voltage registers with high resolution, high accuracy regulation limits. It also sets the  $\overline{\text{PROCHOT}}$  timing and threshold profile to meet system requirements.

## 8.2 Functional Block Diagram



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## 8.3 Feature Description

### 8.3.1 Power-Up from Battery Without DC Source

If only battery is present and the voltage is above  $V_{VBAT\_UVLOZ}$ , the BATFET turns on and connects battery to system. By default, the charger is in low power mode ( $REG0x12[15] = 1$ ) with lowest quiescent current. The LDO stays off. When device moves to performance mode ( $REG0x12[15] = 0$ ), The host enables IBAT buffer through SMBus to monitor discharge current. For PSYS,  $\overline{PROCHOT}$  or independent comparator, REGN LDO is enabled for an accurate reference.

### 8.3.2 Power-Up From DC Source

When an input source plugs in, the charger checks the input source voltage to turn on LDO and all the bias circuits. It sets the input current limit before the converter starts.

The power-up sequence from DC source is as follows:

1. 50 ms after VBUS above  $V_{VBUS\_CONVEN}$ , enable 6V LDO and CHRГ\_OK goes HIGH
2. Input voltage and current limit setup
3. Battery CELL configuration
4. 150 ms after VBUS above  $V_{VBUS\_CONVEN}$ , converter powers up.

#### 8.3.2.1 CHRГ\_OK Indicator

CHRГ\_OK is an active HIGH open drain indicator. It indicates the charger is in normal operation when the following conditions are valid:

- VBUS is above  $V_{VBUS\_CONVEN}$
- VBUS is below  $V_{ACOV}$
- No MOSFET/inductor fault

#### 8.3.2.2 Input Voltage and Current Limit Setup

When CHRГ\_OK goes LOW, the charger sets default input current limit in  $REG0x3F()$  to 3.30A. The actual input current limit is the lower setting of  $REG0x3F()$  and ILIM\_HIZ pin.

Charger initiates a VBUS voltage measurement without any load (VBUS at no load condition). The default VINDPM threshold is  $VBUS@noLoad - 1.28$  V.

After input current and voltage limits are set, the charger device is ready to power up. The host can always update input current and voltage limit based on input source type.

#### 8.3.2.3 Battery Cell Configuration

CELL\_BATPRESZ pin is biased with resistors from REGN to CELL\_BATPRESZ to GND. After VDDA LDO is activated, the device detects the battery configuration through CELL\_BATPRESZ pin bias voltage. Refer to [Electrical Characteristics](#) for CELL setting thresholds.

**Table 1. Battery Cell Configuration**

CELL COUNT	PIN VOLTAGE w.r.t. VDDA	BATTERY VOLTAGE (REG0x15)	SYSOVP
4S	75%	16.800V	19.5V
3S	55%	12.592V	19.5V
2S	40%	8.400V	12V
1S	25%	4.192V	5V

#### 8.3.2.4 Device Hi-Z State

The charger enters Hi-Z mode when ILIM\_HIZ pin voltage is below 0.4 V or  $REG0x32[15]$  is set to 1. During Hi-Z mode, the input source is present, and the charger is in the low quiescent current mode with REGN LDO enabled.

### 8.3.3 Converter Operation

The charger employs a synchronous buck-boost converter that allows charging from a standard 5-V or a high-voltage power source. The charger operates in buck, buck-boost and boost mode. The buck-boost can operate uninterruptedly and continuously across the three operation modes.

**Table 2. MOSFET Operation**

MODE	BUCK	BUCK-BOOST	BOOST
Q1	Switching	Switching	ON
Q2	Switching	Switching	OFF
Q3	OFF	Switching	Switching
Q4	ON	Switching	Switching

#### 8.3.3.1 Inductor Setting through IADPT Pin

The charger reads the inductor value through the IADPT pin.

**Table 3. Inductor Setting on IADPT Pin**

INDUCTOR IN USE	RESISTOR ON IADPT PIN
1 $\mu$ H	93 k $\Omega$
2.2 $\mu$ H	137 k $\Omega$
3.3 $\mu$ H	169 k $\Omega$

#### 8.3.3.2 Continuous Conduction Mode (CCM)

With sufficient charge current, the inductor current does not cross 0 A, which is defined as CCM. The controller starts a new cycle with ramp coming up from 200 mV. As long as error amplifier output voltage is above the ramp voltage, the high-side MOSFET (HSFET) stays on. When the ramp voltage exceeds error amplifier output voltage, HSFET turns off and low-side MOSFET (LSFET) turns on. At the end of the cycle, ramp gets reset and LSFET turns off, ready for the next cycle. There is always break-before-make logic during transition to prevent cross-conduction and shoot-through. During the dead time when both MOSFETs are off, the body-diode of the low-side power MOSFET conducts the inductor current.

During CCM, the inductor current always flows and creates a fixed two-pole system. Having the LSFET turn-on when the HSFET is off, keeps the power dissipation low and allows safe charging at high currents.

#### 8.3.3.3 Pulse Frequency Modulation (PFM)

In order to improve converter light-load efficiency, the bq25708 switches to PFM control at light load condition. The effective switching frequency will decrease accordingly when system load decreases. The minimum frequency can be limit to 25kHz (ChargeOption0() bit[10]=1).

### 8.3.4 Current and Power Monitor

#### 8.3.4.1 High-Accuracy Current Sense Amplifier (IADPT and IBAT)

As an industry standard, a high-accuracy current sense amplifier (CSA) is used to monitor the input current during forward charging, and the charge/discharge current (IBAT). IADPT voltage is 20 $\times$  or 40 $\times$  the differential voltage across ACP and ACN. IBAT voltage is 8 $\times$ /16 $\times$  (during charging), or 8 $\times$ /16 $\times$  (during discharging) of the differential across SRP and SRN. After input voltage or battery voltage is above UVLO, IADPT output becomes valid. To lower the voltage on current monitoring, a resistor divider from CSA output to GND can be used and accuracy over temperature can still be achieved.

- $V_{(IADPT)} = 20 \text{ or } 40 \times (V_{(ACP)} - V_{(ACN)})$  during forward mode.
- $V_{(IBAT)} = 8 \text{ or } 16 \times (V_{(SRP)} - V_{(SRN)})$  during forward charging mode.
- $V_{(IBAT)} = 8 \text{ or } 16 \times (V_{(SRN)} - V_{(SRP)})$  during forward supplement mode.

A maximum 100-pF capacitor is recommended to connect on the output for decoupling high-frequency noise. An additional RC filter is optional, if additional filtering is desired. Note that adding filtering also adds additional response delay. The CSA output voltage is clamped at 3.3 V.

### 8.3.4.2 High-Accuracy Power Sense Amplifier (PSYS)

The charger monitors total system power. During forward mode, the input adapter powers system. The ratio of PSYS current and total power  $K_{PSYS}$  can be programmed in REG0x30[9] with default 1  $\mu\text{A}/\text{W}$ . The input and charge sense resistors (RAC and RSR) are programmed in REG0x30[11:10]. PSYS voltage can be calculated with Equation 1 where  $I_{IN} > 0$  when adapter is in forward charging, and  $I_{BAT} > 0$  when the battery is in discharge when the battery is in discharge.

$$V_{PSYS} = R_{PSYS} \times K_{PSYS} (V_{ACP} \times I_{IN} + V_{BAT} \times I_{BAT}) \quad (1)$$

For proper PSYS functionality, RAC and RSR values are limited to 10m $\Omega$  and 20m $\Omega$ .

### 8.3.5 Input Source Dynamic Power Manage

Refer to [Input Current and Input Voltage Registers for Dynamic Power Management](#).

### 8.3.6 Two-Level Adapter Current Limit (Peak Power Mode)

Usually adapter can supply current higher than DC rating for a few milliseconds to tens of milliseconds. The charger employs two-level input current limit, or peak power mode, to fully utilize the overloading capability and minimize battery discharge during CPU turbo mode. Peak power mode is enabled in REG0x31[13] (EN\_PKPWR\_IDPM) or REG0x31[12] (EN\_PKPWR\_VSYS). The DC current limit, or  $I_{LIM1}$ , is the same as adapter DC current, set in REG0x3F(). The overloading current, or  $I_{LIM2}$ , is set in REG0x33[15:11], as a percentage of  $I_{LIM1}$ .

When the charger detects input current surge and battery discharge due to load transient, it applies  $I_{LIM2}$  for  $T_{OVL D}$  in REG0x31[15:14], first, and then  $I_{LIM1}$  for up to  $T_{MAX} - T_{OVL D}$  time.  $T_{MAX}$  is programmed in REG0x31[9:8]. After  $T_{MAX}$ , if the load is still high, another peak power cycle starts. Charging is disabled during  $T_{MAX}$ ; once  $T_{MAX}$  expires, charging continues. If  $T_{OVL D}$  is programmed higher than  $T_{MAX}$ , then peak power mode is always on.

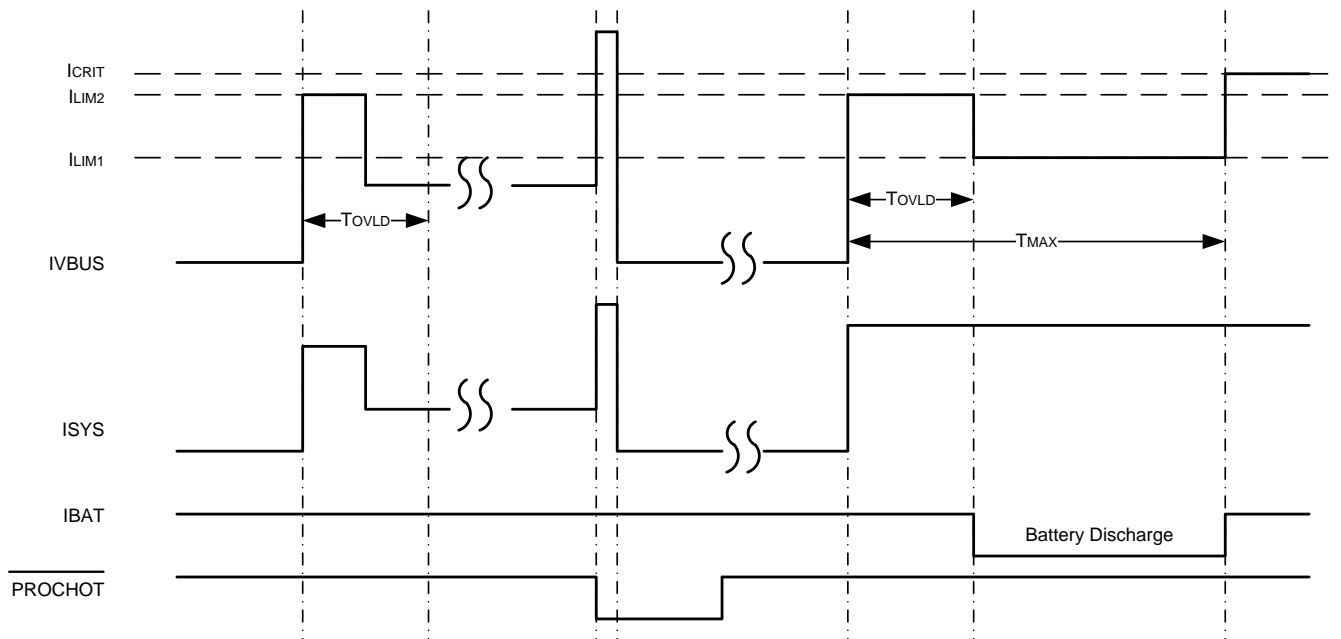


Figure 8. Two-Level Adapter Current Limit Timing Diagram

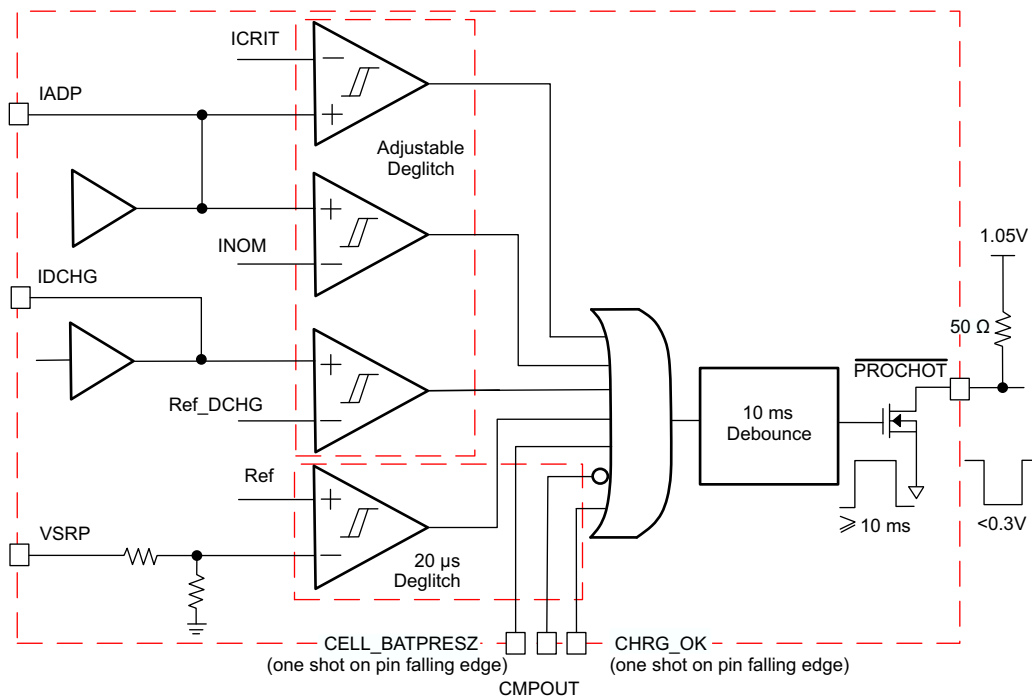
### 8.3.7 Processor Hot Indication

When CPU is running turbo mode, the system peak power may exceed available power from adapter and battery together. The adapter current and battery discharge peak current, or system voltage drop is an indication that system power is too high. The charger processor hot function monitors these events, and  $\overline{\text{PROCHOT}}$  pulse is asserted. Once CPU receives  $\overline{\text{PROCHOT}}$  pulse from charger, it slows down to reduce the system power. The processor hot function monitors these events, and  $\overline{\text{PROCHOT}}$  pulse is asserted.

The  $\overline{\text{PROCHOT}}$  triggering events include:

- ICRIT: adapter peak current, as 110% of  $I_{LIM2}$
- INOM: adapter average current (110% of input current limit)
- IDCHG: battery discharge current
- VSYS: system voltage on VSYS
- Adapter Removal: upon adapter removal (CHRG\_OK pin HIGH to LOW)
- Battery Removal: upon battery removal (CELL\_BATPRESZ pin goes LOW)
- CMPOUT: Independent comparator output (CMPOUT pin HIGH to LOW)

The threshold of ICRIT, IDCHG or VSYS, and the deglitch time of ICRIT, INOM, IDCHG or CMPOUT are programmable through SMBus. Each triggering event can be individually enabled in REG0x34[6:0]. When any event in PROCHOT profile is triggered,  $\overline{\text{PROCHOT}}$  is asserted low for minimum 10 ms programmable in 0x33[4:3]. At the end of the 10 ms, if the  $\overline{\text{PROCHOT}}$  event is still active, the pulse gets extended.



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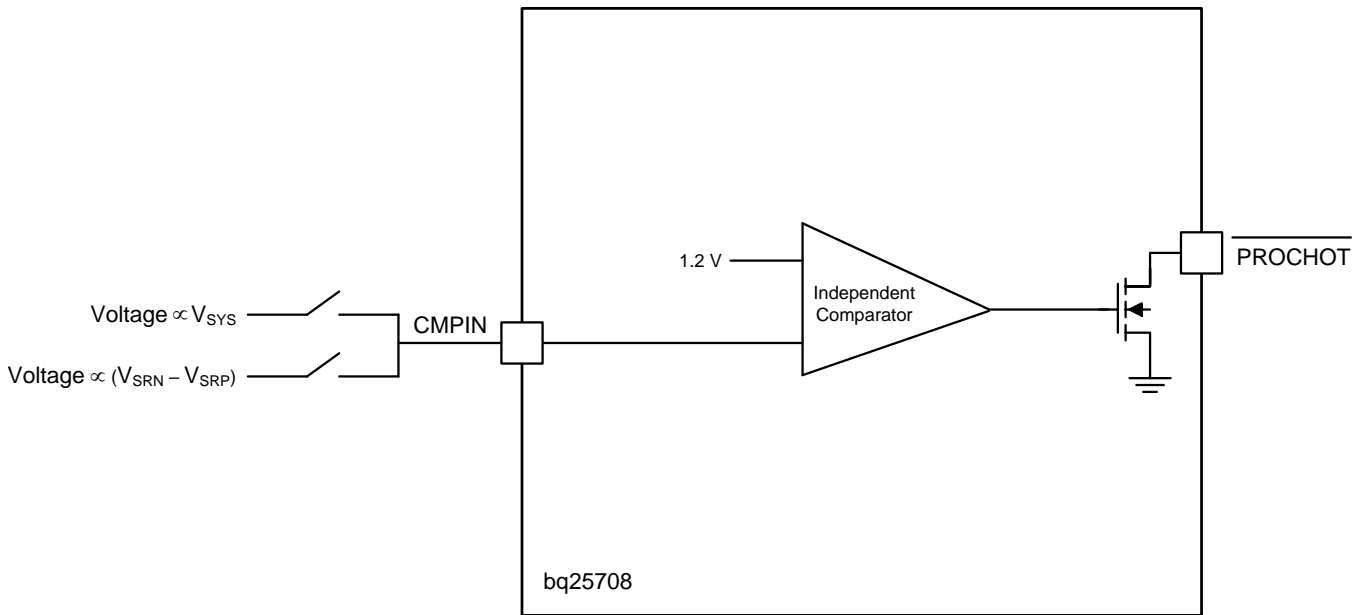
**Figure 9. PROCHOT Profile**

### 8.3.7.1 $\overline{\text{PROCHOT}}$ During Low Power Mode

During low power mode (REG0x12[15]=1), the charger offers a low quiescent current (~150 uA). Low power  $\overline{\text{PROCHOT}}$  function uses the independent comparator to monitor battery discharge current and system voltage, and assert  $\overline{\text{PROCHOT}}$  to CPU.

Below lists the register setting to enable  $\overline{\text{PROCHOT}}$  during low power mode.

- REG0x12[15]=1
- REG0x34[5:0]=000000
- REG0x30[6:4]=100
- Independent comparator threshold is always 1.2V
- When REG0x30[14]=1, charger monitors discharge current. Connect CMPIN to voltage proportional to IBAT pin.  $\overline{\text{PROCHOT}}$  triggers from HIGH to LOW when CMPIN voltage falls below 1.2V.
- When REG0x30[13]=1, charger monitors system voltage. Connect CMPIN to voltage proportional to system.  $\overline{\text{PROCHOT}}$  triggers from HIGH to LOW when CMPIN voltage rises above 1.2V.



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Figure 10.  $\overline{\text{PROCHOT}}$  Low Power Mode Implementation

### 8.3.7.2 $\overline{\text{PROCHOT}}$ Status

REG0x21[6:0] reports which event in the profile triggers  $\overline{\text{PROCHOT}}$  by setting the corresponding bit to 1. The status bit can be reset back to 0 after it is read by host, and current  $\overline{\text{PROCHOT}}$  event is no longer active.

Assume there are two  $\overline{\text{PROCHOT}}$  events, event A and event B. Event A triggers  $\overline{\text{PROCHOT}}$  first, but event B is also active. Both status bits will be HIGH. At the end of the 10ms  $\overline{\text{PROCHOT}}$  pulse, if  $\overline{\text{PROCHOT}}$  is still active (either by A or B), the  $\overline{\text{PROCHOT}}$  pulse is extended.

### 8.3.8 Device Protection

#### 8.3.8.1 Watchdog Timer

The charger includes watchdog timer to terminate charging if the charger does not receive a write MaxChargeVoltage() or write ChargeCurrent() command within 175 s (adjustable via REG0x12[14:13]). When watchdog timeout occurs, all register values are kept unchanged except ChargeCurrent() resets to zero. Battery charging is suspended. Write MaxChargeVoltage() or write ChargeCurrent() commands must be re-sent to reset watchdog timer and resume charging. Writing REG0x12[14:13] = 00 to disable watchdog timer also resumes charging.

#### 8.3.8.2 Input Overvoltage Protection (ACOV)

The charger has fixed ACOV voltage. When VBUS pin voltage is higher than ACOV, it is considered as adapter over voltage. CHRG\_OK will be pulled low, and converter will be disabled. As system falls below battery voltage, BATFET will be turned on. When VBUS pin voltage falls below ACOV, it is considered as adapter voltage returns back to normal voltage. CHRG\_OK will be pulled high by external pull up resistor. The converter will resume if enable conditions are valid.

#### 8.3.8.3 Input Overcurrent Protection (ACOC)

If the input current exceeds the 1.25x or 2x (REG0x31[2]) of  $I_{\text{LIM2\_VTH}}$  (REG0x33[15:11]) set point, converter stops switching. After 300 ms, converter will start switching again.

#### 8.3.8.4 System Overvoltage Protection (YSOVP)

When the converter starts up, the bq25708 reads CELL pin configuration and sets MaxChargeVoltage() and YSOVP threshold (1s – 5 V, 2s – 12 V, 3s/4s – 19.5 V). Before REGx15() is written by host, the battery configuration will change with CELL pin voltage. When YSOVP happens, the device latches off the converter. REG20[4] is set as 1. The user can clear the latch-off by either writing 0 to YSOVP bit or removing and plugging in adapter again. After the latch-off is cleared, converter starts again.

#### 8.3.8.5 Battery Overvoltage Protection (BATOVP)

Battery over-voltage may happen when battery is removed during charging or the user plugs in a wrong battery. The BATOVP threshold is 104% (1 s) or 102% (2 s to 4 s) of regulation voltage set in REG0x15().

#### 8.3.8.6 Battery Short

If BAT voltage falls below SYSMIN during charging, the maximum current is limited to 384 mA.

#### 8.3.8.7 Thermal Shutdown (TSHUT)

The WQFN package has low thermal impedance, which provides good thermal conduction from the silicon to the ambient, to keep junction temperatures low. As added level of protection, the charger converter turns off for self-protection whenever the junction temperature exceeds the 155°C. The charger stays off until the junction temperature falls below 135°C. During thermal shut down, the LDO current limit is reduced to 16 mA and REGN LDO stays off. When the temperature falls below 135°C, charge can be resumed with soft start.

### 8.4 Device Functional Modes

#### 8.4.1 Forward Mode

When input source is connected to VBUS, bq25708 is in forward mode to regulate system and charge battery.

##### 8.4.1.1 System Voltage Regulation with Narrow VDC Architecture

The bq25708 employs Narrow VDC architecture (NVDC) with BATFET separating system from battery. The minimum system voltage is set by MinSystemVoltage(). Even with a deeply depleted battery, the system is regulated above the minimum system voltage.

When the battery is below minimum system voltage setting, the BATFET operates in linear mode (LDO mode).

As the battery voltage rises above the minimum system voltage, BATFET is fully on when charging or in supplement mode and the voltage difference between the system and battery is the VDS of BATFET. System voltage is regulated 160 mV above battery voltage when BATFET is off (no charging or no supplement current).

See [System Voltage Regulation](#) for details on system voltage regulation and register programming.

##### 8.4.1.2 Battery Charging

The bq25708 charges 1-4 cell battery in constant current (CC), and constant voltage (CV) mode. Based on CELL\_BATPREZ pin setting, the charger sets default battery voltage 4.2V/cell to ChargeVoltage(), or REG0x15(). According to battery capacity, the host programs appropriate charge current to ChargeCurrent(), or REG0x14(). When battery is full or battery is not in good condition to charge, host terminates charge by setting REG0x12[0] to 1, or setting ChargeCurrent() to zero.

See [Feature Description](#) for details on register programming.

## 8.5 Programming

The charger supports battery-charger commands that use either Write-Word or Read-Word protocols, as summarized in [SMBus Write-Word and Read-Word Protocols](#). The SMBUS address is 12h (0001001\_X), where X is the read/write bit. The ManufacturerID and DeviceID registers are assigned identify the charger device. The ManufacturerID register command always returns 40h.

### 8.5.1 SMBus Interface

The bq25700 device operates as a slave, receiving control inputs from the embedded controller host through the SMBus interface. The bq25700 device uses a simplified subset of the commands documented in *System Management Bus Specification V1.1*, which can be downloaded from [www.smbus.org](http://www.smbus.org). The bq25700 device uses the SMBus read-word and write-word protocols (shown in [Table 4](#) and [Table 5](#)) to communicate with the smart battery. The bq25700 device performs only as a SMBus slave device with address 0b00010010 (0x12H) and does not initiate communication on the bus. In addition, the bq25700 device has two identification registers, a 16-bit device ID register (0xFFH) and a 16-bit manufacturer ID register (0xFEH).

SMBus communication starts when VCC is above  $V_{(UVLO)}$ .

The data (SDA) and clock (SCL) pins have Schmitt-trigger inputs that can accommodate slow edges. Choose pullup resistors (10 k $\Omega$ ) for SDA and SCL to achieve rise times according to the SMBus specifications. Communication starts when the master signals a start condition, which is a high-to-low transition on SDA, while SCL is high. When the master has finished communicating, the master issues a stop condition, which is a low-to-high transition on SDA, while SCL is high. The bus is then free for another transmission. [Figure 11](#) and [Figure 12](#) show the timing diagram for signals on the SMBus interface. The address byte, command byte, and data bytes are transmitted between the start and stop conditions. The SDA state changes only while SCL is low, except for the start and stop conditions. Data is transmitted in 8-bit bytes and is sampled on the rising edge of SCL. Nine clock cycles are required to transfer each byte in or out of the bq25708 device because either the master or the slave acknowledges the receipt of the correct byte during the ninth clock cycle. The bq25708 supports the charger commands listed in [Table 4](#).

## Programming (continued)

### 8.5.1.1 SMBus Write-Word and Read-Word Protocols

**Table 4. Write-Word Format**

S <sup>(1)(2)</sup>	SLAVE ADDRESS <sup>(1)</sup>	W <sup>(1)(3)</sup>	ACK <sup>(4)(5)</sup>	COMMAND BYTE <sup>(1)</sup>	ACK <sup>(4)(5)</sup>	LOW DATA BYTE <sup>(1)</sup>	ACK <sup>(4)(5)</sup>	HIGH DATA BYTE <sup>(1)</sup>	ACK <sup>(4)(5)</sup>	P <sup>(1)(6)</sup>
	7 bits	1b	1b	8 bits	1b	8 bits	1b	8 bits	1b	
	MSB LSB	0	0	MSB LSB	0	MSB LSB	0	MSB LSB	0	

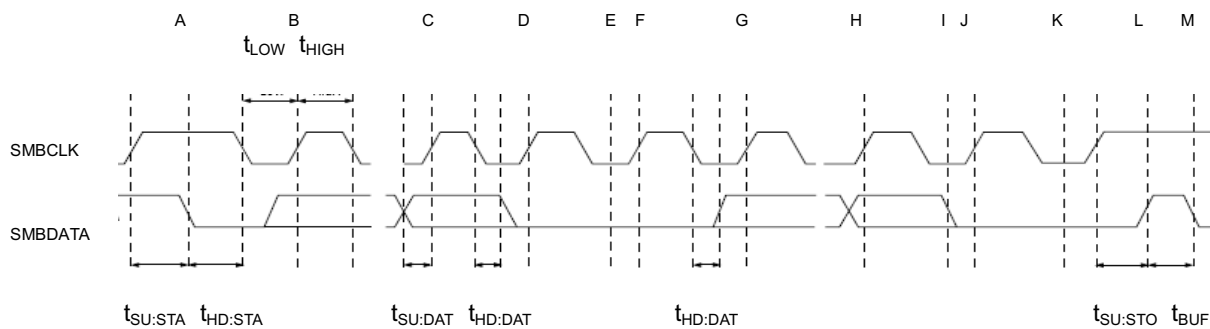
- (1) Master to slave
- (2) S = Start condition or repeated start condition
- (3) W = Write bit (logic-low)
- (4) Slave to master (shaded gray)
- (5) ACK = Acknowledge (logic-low)
- (6) P = Stop condition

**Table 5. Read-Word Format**

S <sup>(1)(2)</sup>	SLAVE ADDRESS <sup>(1)</sup>	W <sup>(1)(3)</sup>	ACK <sup>(4)(5)</sup>	COMMAND BYTE <sup>(1)</sup>	ACK <sup>(4)(5)</sup>	S <sup>(1)(2)</sup>	SLAVE ADDRESS <sup>(1)</sup>	R <sup>(1)(6)</sup>	ACK <sup>(4)(5)</sup>	LOW DATA BYTE <sup>(4)</sup>	ACK <sup>(1)(5)</sup>	HIGH DATA BYTE <sup>(4)</sup>	NACK <sup>(1)(7)</sup>	P <sup>(1)(8)</sup>
	7 bits	1b	1b	8 bits	1b		7 bits	1b	1b	8 bits	1b	8 bits	1b	
	MSB LSB	0	0	MSB LSB	0		MSB LSB	1	0	MSB LSB	0	MSB LSB	1	

- (1) Master to slave
- (2) S = Start condition or repeated start condition
- (3) W = Write bit (logic-low)
- (4) Slave to master (shaded gray)
- (5) ACK = Acknowledge (logic-low)
- (6) R = Read bit (logic-high)
- (7) NACK = Not acknowledge (logic-high)
- (8) P = Stop condition

### 8.5.1.2 Timing Diagrams



- A = Start condition
- B = MSB of address clocked into slave
- C = LSB of address clocked into slave
- D = R/W bit clocked into slave
- E = Slave pulls SMBDATA line low
- F = ACKNOWLEDGE bit clocked into master
- G = MSB of data clocked into slave
- H = LSB of data clocked into slave
- I = Slave pulls SMBDATA line low
- J = Acknowledge clocked into master
- K = Acknowledge clock pulse
- L = Stop condition, data executed by slave
- M = New start condition

**Figure 11. SMBus Write Timing**

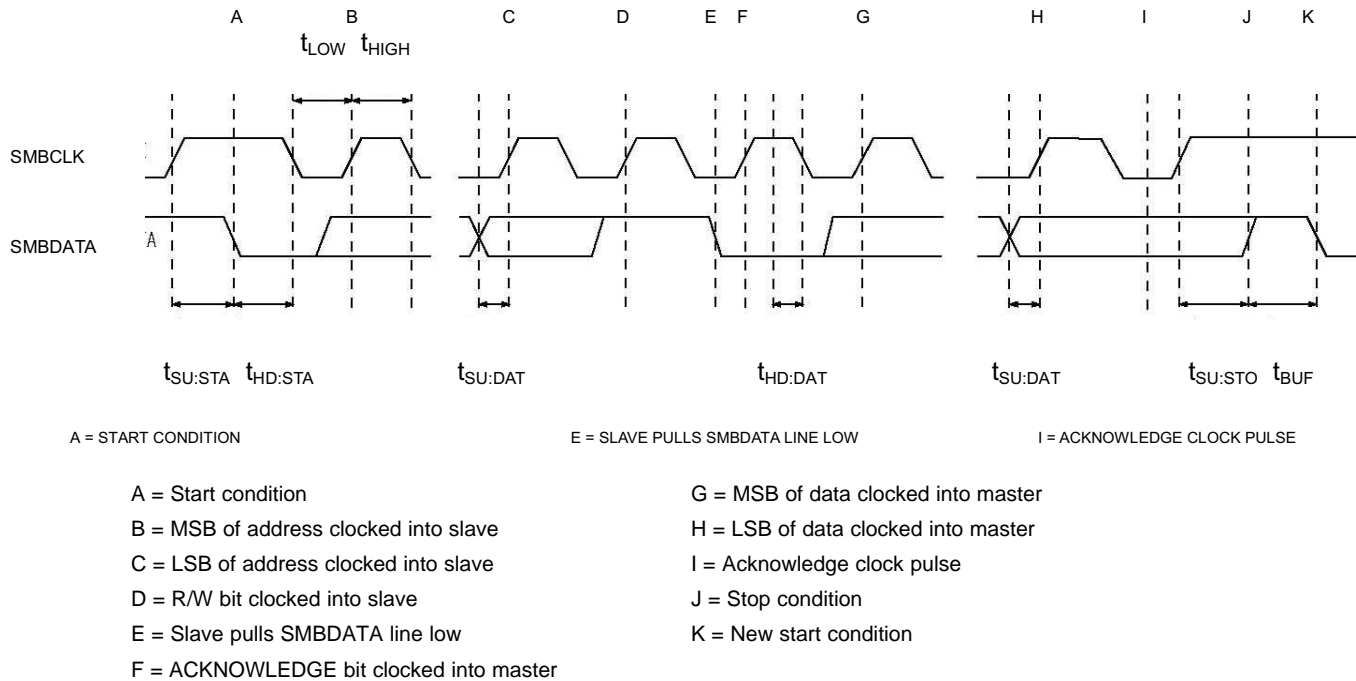


Figure 12. SMBus Read Timing

## 8.6 Register Map

Table 6. Charger Command Summary

SMBus ADDR	REGISTER NAME	TYPE	DESCRIPTION	LINKS
12h	ChargeOption0()	R/W	Charge Option 0	<a href="#">Go</a>
14h	ChargeCurrent()	R/W	7-bit charge current setting LSB 64 mA, Range 8128 mA	<a href="#">Go</a>
15h	MaxChargeVoltage()	R/W	11-bit charge voltage setting LSB 16 mV, Default: 1S-4192mV, 2S-8400mV, 3S-12592mV, 4S-16800mV	<a href="#">Go</a>
30h	ChargeOption1()	R/W	Charge Option 1	<a href="#">Go</a>
31h	ChargeOption2()	R/W	Charge Option 2	<a href="#">Go</a>
32h	ChargeOption3()	R/W	Charge Option 3	<a href="#">Go</a>
33h	ProchotOption0()	R/W	PROCHOT Option 0	<a href="#">Go</a>
34h	ProchotOption1()	R/W	PROCHOT Option 1	<a href="#">Go</a>
35h	ADCOption()	R/W	ADC Option	<a href="#">Go</a>
20h	ChargerStatus()	R	Charger Status	<a href="#">Go</a>
21h	ProchotStatus()	R	Prochot Status	<a href="#">Go</a>
22h	IIN_DPM()	R	7-bit input current limit in use LSB: 50mA, Range: 50mA-6400mA	<a href="#">Go</a>
23h	ADCVBUS/PSYS()	R	8-bit digital output of input voltage, 8-bit digital output of system power PSYS: Full range: 3.06V, LSB: 12mV VBUS: Full range: 3.2V-19.52V, LSB 64mV	<a href="#">Go</a>
24h	ADCIBAT()	R	8-bit digital output of battery charge current, 8-bit digital output of battery discharge current ICHG: Full range 8.128A, LSB 64mA IDCHG: Full range: 32.512A, LSB: 256mA	<a href="#">Go</a>

**Register Map (continued)**
**Table 6. Charger Command Summary (continued)**

SMBus ADDR	REGISTER NAME	TYPE	DESCRIPTION	LINKS
25h	ADCIIINCMPIN()	R	8-bit digital output of input current, 8-bit digital output of CMPIN voltage POR State - IIN: Full range: 12.75A, LSB 50mA CMPIN: Full range 3.06V, LSB: 12mV	<a href="#">Go</a>
26h	ADCVSYSVBAT()	R	8-bit digital output of system voltage, 8-bit digital output of battery voltage VSYS: Full range: 2.88V-19.2V, LSB: 64mV VBAT: Full range : 2.88V-19.2V, LSB 64mV	<a href="#">Go</a>
3Bh	Reserved	R/W		
3Ch	Reserved	R/W		
3Dh	InputVoltage()	R/W	8-bit input voltage setting LSB 64mV, Range: 3200 mV – 19520 mV	<a href="#">Go</a>
3Eh	MinSystemVoltage()	R/W	6-Bit minimum system voltage setting LSB: 256mV, Range: 1024mV-16182mV Default: 1S-3.584V, 2S-6.144V, 3S-9.216V, 4S-12.288V	<a href="#">Go</a>
3Fh	IIN_HOST()	R/W	6-bit Input current limit set by host LSB: 50mA, Range: 50mA-6400mA	<a href="#">Go</a>
FEh	ManufacturerID()	R	Manufacturer ID - 0x0040H	<a href="#">Go</a>
FFh	DeviceID()	R	Device ID	<a href="#">Go</a>

## 8.6.1 Setting Charge and PROCHOT Options

### 8.6.1.1 ChargeOption0 Register (SMBus address = 12h) [reset = E20Eh]

**Figure 13. ChargeOption0 Register (SMBus address = 12h) [reset = E20Eh]**

15	14	13	12	11	10	9	8
EN_LWPWR	WDTMR_ADJ		IDPM_AUTO_DISABLE	Reserved	EN_OOA	PWM_FREQ	Reserved
R/W	R/W		R/W	R/W	R/W	R/W	R/W
7	6	5	4	3	2	1	0
Reserved	EN_LEARN	IADPT_GAIN	IBAT_GAIN	EN_LDO	EN_IDPM	CHRG_INHIBIT	
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 7. ChargeOption0 Register (SMBus address = 12h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15	EN_LWPWR	R/W	1b	<p>Low Power Mode Enable</p> <p>0b: Disable Low Power Mode. Device in performance mode with battery only. The <u>PROCHOT</u>, current/power monitor buffer and comparator follow register setting.</p> <p>1b: Enable Low Power Mode. Device in low power mode with battery only for lowest quiescent current. /PROCHOT, discharge current monitor buffer, power monitor buffer and independent comparator are disabled. Independent comparator can be enabled by setting either REG0x30()[14] or [13] to 1. &lt;default at POR&gt;</p>
14-13	WDTMR_ADJ	R/W	11b	<p>WATCHDOG Timer Adjust</p> <p>Set maximum delay between consecutive SMBus write of charge voltage or charge current command.</p> <p>If device does not receive a write on the REG0x15() or the REG0x14() within the watchdog time period, the charger will be suspended by setting the REG0x14() to 0 mA.</p> <p>After expiration, the timer will resume upon the write of REG0x14(), REG0x15() or REG0x12[14:13]. The charger will resume if the values are valid.</p> <p>00b: Disable Watchdog Timer            01b: Enabled, 5 sec            10b: Enabled, 88 sec            11b: Enable Watchdog Timer, 175 sec &lt;default at POR&gt;</p>
12	IDPM_AUTO_DISABLE	R/W	0b	<p>IDPM Auto Disable</p> <p>When CELL_BATPRESZ pin is LOW, the charger automatically disables the IDPM function by setting EN_IDPM (REG0x12[1]) to 0. The host can enable IDPM function later by writing EN_IDPM bit (REG0x12[1]) to 1.</p> <p>0b: Disable this function. IDPM is not disabled when CELL_BATPRESZ goes LOW. &lt;default at POR&gt;</p> <p>1b: Enable this function. IDPM is disabled when CELL_BATPRESZ goes LOW.</p>
11	Reserved	R/W		Reserved
10	EN_OOA	R/W	0b	<p>Out-of-Audio Enable</p> <p>0b: No limit of PFM burst frequency &lt;default at POR&gt;</p> <p>1b: Set minimum PFM burst frequency to above 25 kHz to avoid audio noise</p>

**Table 7. ChargeOption0 Register (SMBus address = 12h) Field Descriptions (continued)**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
9	PWM_FREQ	R/W	1b	Switching Frequency Two converter switching frequencies. One for small inductor and the other for big inductor. Currently, customer uses 800 kHz with 2.2 $\mu$ H or 3.3 $\mu$ H, and 1.2 MHz with 1 $\mu$ H or 1.5 $\mu$ H. 0b: 1.2 MHz 1b: 800 kHz <default at POR>
8	Reserved	R/W	0b	Reserved

**Table 8. ChargeOption0 Register (SMBus address = 12h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7-6	Reserved	R/W	00b	Reserved
5	EN_LEARN	R/W	0b	LEARN function allows the battery to discharge while the adapter is present. It calibrates the battery gas gauge over a complete discharge/charge cycle. When the battery voltage is below battery depletion threshold, the system switches back to adapter input by the host. When CELL_BATPRESZ pin is LOW, the device exits LEARN mode and this bit is set back to 0. 0b: Disable LEARN Mode <default at POR> 1b: Enable LEARN Mode
4	IADPT_GAIN	R/W	0b	IADPT Amplifier Ratio The ratio of voltage on IADPT and voltage across ACP and ACN. 0b: 20x <default at POR> 1b: 40x
3	IBAT_GAIN	R/W	1b	IBAT Amplifier Ratio The ratio of voltage on IBAT and voltage across SRP and SRN 0b: 8x 1b: 16x <default at POR>
2	EN_LDO	R/W	1b	LDO Mode Enable When battery voltage is below minimum system voltage (REG0x3E()), the charger is in pre-charge with LDO mode enabled. 0b: Disable LDO mode, BATFET fully ON. Precharge current is set by battery pack internal resistor. The system is regulated by the MaxChargeVoltage register. 1b: Enable LDO mode, Precharge current is set by the ChargeCurrent register and clamped below 384 mA (2 cell – 4 cell) or 2A (1 cell). The system is regulated by the MinSystemVoltage register. <default at POR>
1	EN_IDPM	R/W	1b	IDPM Enable Host writes this bit to enable IDPM regulation loop. When the IDPM is disabled by the charger (refer to IDPM_AUTO_DISABLE), this bit goes LOW. 0b: IDPM disabled 1b: IDPM enabled <default at POR>
0	CHRG_INHIBIT	R/W	0b	Charge Inhibit When this bit is 0, battery charging will start with valid values in the MaxChargeVoltage register and the ChargeCurrent register. 0b: Enable Charge <default at POR> 1b: Inhibit Charge

**8.6.1.2 ChargeOption1 Register (SMBus address = 30h) [reset = 211h]**
**Figure 14. ChargeOption1 Register (SMBus address = 30h) [reset = 211h]**

15		14		13		12		11		10		9		8	
EN_IBAT		EN_PROCHOT_LPWR		EN_PSYS		RSNS_RAC		RSNS_RSR		PSYS_RATIO		Reserved			
R/W		R/W		R/W		R/W		R/W		R/W		R/W		R/W	
7		6		5		4		3		2		1		0	
CMP_REF		CMP_POL		CMP_DEG		FORCE_LATCHOFF		Reserved		EN_SHIP_DCHG		AUTO_WAKEUP_EN			
R/W		R/W		R/W		R/W		R/W		R/W		R/W		R/W	

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 9. ChargeOption1 Register (SMBus address = 30h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15	EN_IBAT	R/W	0b	IBAT Enable Enable the IBAT output buffer. In low power mode (REG0x12[15] = 1), IBAT buffer is always disabled regardless of this bit value. 0b Turn off IBAT buffer to minimize Iq <default at POR> 1b: Turn on IBAT buffer
14-13	EN_PROCHOT_LPWR	R/W	00b	Enable $\overline{\text{PROCHOT}}$ during battery only low power mode With battery only, enable IDCHG or VSYS in $\overline{\text{PROCHOT}}$ with low power consumption. Do not enable this function with adapter present. Refer to <a href="#">PROCHOT During Low Power Mode</a> for more details. 00b: Disable low power $\overline{\text{PROCHOT}}$ <default at POR> 01b: Enable IDCHG low power $\overline{\text{PROCHOT}}$ 10b: Enable VSYS low power $\overline{\text{PROCHOT}}$ 11b: Reserved
12	EN_PSYS	R/W	0b	PSYS Enable Enable PSYS sensing circuit and output buffer (whole PSYS circuit). In low power mode (REG0x12[15] = 1), PSYS sensing and buffer are always disabled regardless of this bit value. 0b: Turn off PSYS buffer to minimize Iq <default at POR> 1b: Turn on PSYS buffer
11	RSNS_RAC	R/W	0b	Input sense resistor RAC 0b: 10 m $\Omega$ <default at POR> 1b: 20 m $\Omega$
10	RSNS_RSR	R/W	0b	Charge sense resistor RSR 0b: 10 m $\Omega$ <default at POR> 1b: 20 m $\Omega$
9	PSYS_RATIO	R/W	1b	PSYS Gain Ratio of PSYS output current vs total input and battery power with 10-m $\Omega$ sense resistor. 0b: 0.25 $\mu\text{A/W}$ 1b: 1 $\mu\text{A/W}$ <default at POR>
8	Reserved	R/W	0b	Reserved

**Table 10. ChargeOption1 Register (SMBus address = 30h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7	CMP_REF	R/W	0b	Independent Comparator Internal Reference 0b: 2.3 V <default at POR> 1b: 1.2 V

**Table 10. ChargeOption1 Register (SMBus address = 30h) Field Descriptions (continued)**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
6	CMP_POL	R/W	0b	Independent Comparator Output Polarity 0b: When CMPIN is above internal threshold, CMPOUT is LOW (internal hysteresis) <default at POR> 1b: When CMPIN is below internal threshold, CMPOUT is LOW (external hysteresis)
5-4	CMP_DEG	R/W	01b	Independent Comparator Deglitch Time, only applied to the falling edge of CMPOUT (HIGH → LOW). 00b: Independent comparator is disabled 01b: Independent comparator is enabled with output deglitch time 1 μs <default at POR> 10b: Independent comparator is enabled with output deglitch time of 2 ms 11b: Independent comparator is enabled with output deglitch time of 5 sec
3	FORCE_LATCNOFF	R/W	0b	Force Power Path Off When the independent comparator triggers, charger turns off Q1 and Q4 (same as disable converter) so that the system is disconnected from the input source. At the same time, CHRG_OK signal goes to LOW to notify the system. 0b: Disable this function <default at POR> 1b: Enable this function
2	Reserved	R/W	0b	Reserved
1	EN_SHIP_DCHG	R/W	0b	Discharge SRN for Shipping Mode When this bit is 1, discharge SRN pin down below 3.8 V in 140 ms. When 140 ms is over, this bit is reset to 0. 0b: Disable shipping mode <default at POR> 1b: Enable shipping mode
0	AUTO_WAKEUP_EN	R/W	1b	Auto Wakeup Enable When this bit is HIGH, if the battery is below minimum system voltage (REG0x3E()), the device will automatically enable 128 mA charging current for 30 mins. When the battery is charged up above minimum system voltage, charge will terminate and the bit is reset to LOW. 0b: Disable 1b: Enable <default at POR>

**8.6.1.3 ChargeOption2 Register (SMBus address = 31h) [reset = 2B7]**
**Figure 15. ChargeOption2 Register (SMBus address = 31h) [reset = 2B7]**

15	14	13	12	11	10	9	8
PKPWR_TOVLD_DEG		EN_PKPWR_IDPM	EN_PKPWR_VSYS	PKPWR_OVLD_STAT	PKPWR_RELAX_STAT	PKPWR_TMAX[1:0]	
R/W		R/W	R/W	R/W	R/W	R/W	
7	6	5	4	3	2	1	0
EN_EXTILIM	EN_ICHG_IDCHG	Q2_OCP	ACX_OCP	EN_ACOC	ACOC_VTH	EN_BATOC	BATOC_VTH
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 11. ChargeOption2 Register (SMBus address = 31h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15-14	PKPWR_TOVLD_DEG	R/W	00b	Input Overload time in Peak Power Mode 00b: 1 ms <default at POR> 01b: 2 ms 10b: 10 ms 11b: 20 ms
13	EN_PKPWR_IDPM	R/W	0b	Enable Peak Power Mode triggered by input current overshoot If REG0x31[13:12] are 00b, peak power mode is disabled. Upon adapter removal, the bits are reset to 00b. 0b: Disable peak power mode triggered by input current overshoot <default at POR> 1b: Enable peak power mode triggered by input current overshoot.
12	EN_PKPWR_VSYS	R/W	0b	Enable Peak Power Mode triggered by system voltage under-shoot If REG0x31[13:12] are 00b, peak power mode is disabled. Upon adapter removal, the bits are reset to 00b. 0b: Disable peak power mode triggered by system voltage under-shoot <default at POR> 1b: Enable peak power mode triggered by system voltage under-shoot.
11	PKPWR_OVLD_STAT	R/W	0b	Indicator that the device is in overloading cycle. Write 0 to get out of overloading cycle. 0b: Not in peak power mode. <default at POR> 1b: In peak power mode.
10	PKPWR_RELAX_STAT	R/W	0b	Indicator that the device is in relaxation cycle. Write 0 to get out of relaxation cycle. 0b: Not in relaxation cycle. <default at POR> 1b: In relaxation mode.
9-8	PKPWR_TMAX[1:0]	R/W	10b	Peak power mode overload and relax cycle time. When REG0x31[15:14] is programmed longer than REG0x31[9:8], there is no relax time. 00b: 5 ms 01b: 10 ms 10b: 20 ms <default at POR> 11b: 40 ms

**Table 12. ChargeOption2 Register (SMBus address = 31h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7	EN_EXTILIM	R/W	1b	Enable ILIM_HIZ pin to set input current limit 0b: Input current limit is set by REG0x3F. 1b: Input current limit is set by the lower value of ILIM_HIZ pin and REG0x3F. <default at POR>
6	EN_ICHG_IDCHG	R/W	0b	0b: IBAT pin as discharge current. <default at POR> 1b: IBAT pin as charge current.
5	Q2_OCP	R/W	1b	Q2 OCP threshold by sensing Q2 VDS 0b: 300 mV 1b: 500 mV <default at POR>
4	ACX_OCP	R/W	1b	Input current OCP threshold by sensing ACP-ACN. 0b: 150 mV 1b: 280 mV <default at POR>
3	EN_ACOC	R/W	0b	ACOC Enable Input overcurrent (ACOC) protection by sensing the voltage across ACP and ACN. Upon ACOC (after 100- $\mu$ s blank-out time), converter is disabled. 0b: Disable ACOC <default at POR> 1b: ACOC threshold 125% or 200% ICRIT
2	ACOC_VTH	R/W	1b	ACOC Limit Set MOSFET OCP threshold as percentage of IDPM with current sensed from R <sub>AC</sub> . 0b: 125% of ICRIT 1b: 200% of ICRIT <default at POR>
1	EN_BATOC	R/W	1b	BATOC Enable Battery discharge overcurrent (BATOC) protection by sensing the voltage across SRN and SRP. Upon BATOC, converter is disabled. 0b: Disable BATOC 1b: BATOC threshold 125% or 200% $\overline{\text{PROCHOT}}$ IDCHG <default at POR>
0	BATOC_VTH	R/W	1b	Set battery discharge overcurrent threshold as percentage of $\overline{\text{PROCHOT}}$ battery discharge current limit. 0b: 125% of $\overline{\text{PROCHOT}}$ IDCHG 1b: 200% of $\overline{\text{PROCHOT}}$ IDCHG <default at POR>

**8.6.1.4 ChargeOption3 Register (SMBus address = 32h) [reset = 0h]**
**Figure 16. ChargeOption3 Register (SMBus address = 32h) [reset = 0h]**

15	14	13	12	11	10	9	8
EN_HIZ	RESET_REG	RESET_VINDPM	Reserved	EN_ICO_MODE	Reserved		
R/W	R/W	R/W	R/W	R/W	R/W		
7	6	5	4	3	2	1	0
Reserved						BATFETOFF_HIZ	Reserved
R/W						R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 13. ChargeOption3 Register (SMBus address = 32h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15	EN_HIZ	R/W	0b	Device Hi-Z Mode Enable When the charger is in Hi-Z mode, the device draws minimal quiescent current. With VBUS above UVLO. REGN LDO stays on, and system powers from battery. 0b: Device not in Hi-Z mode <default at POR> 1b: Device in Hi-Z mode
14	RESET_REG	R/W	0b	Reset Registers All the registers go back to the default setting except the VINDPM register. 0b: Idle <default at POR> 1b: Reset all the registers to default values. After reset, this bit goes back to 0.
13	RESET_VINDPM	R/W	0b	Reset VINDPM Threshold 0b: Idle 1b: Converter is disabled to measure VINDPM threshold. After VINDPM measurement is done, this bit goes back to 0 and converter starts.
12	Reserved	R/W		Reserved
11	EN_ICO_MODE	R/W	0b	Enable ICO Algorithm 0b: Disable ICO algorithm. <default at POR> 1b: Enable ICO algorithm.
10-8	Reserved	R/W	000b	Reserved

**Table 14. ChargeOption3 Register (SMBus address = 32h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7-2	Reserved	R/W	000000b	Reserved
1	BATFETOFF_HIZ	R/W	0b	Control BATFET during HIZ mode. 0b: BATFET on during Hi-Z <default at POR> 1b: BATFET off during Hi-Z
0	Reserved	R/W		

**8.6.1.5 ProchotOption0 Register (SMBus address = 33h) [reset = 04A54h]**
**Figure 17. ProchotOption0 Register (SMBus address = 33h) [reset = 04A54h]**

15-11			10-9		8
ILIM2_VTH			ICRIT_DEG		Reserved
R/W			R/W		R/W
7-6		5	4-3		2
VSYS_VTH		EN_PROCHOT_EXT	PROCHOT_WIDTH		PROCHOT_CLEAR
R/W		R/W	R/W		R/W
			1		0
			INOM_DEG		Reserved
			R/W		R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 15. ProchotOption0 Register (SMBus address = 33h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15-11	ILIM2_VTH	R/W	01001b	$I_{LIM2}$ Threshold 5 bits, percentage of IDPM in 0x3FH. Measure current between ACP and ACN. Trigger when the current is above this threshold: 00001b - 11001b: 110% - 230%, step 5% 11010b - 11110b: 250% - 450%, step 50% 11111b: Out of Range (Ignored) Default 150%, or 01001
10-9	ICRIT_DEG	R/W	01b	ICRIT Deglitch time ICRIT threshold is set to be 110% of $I_{LIM2}$ . Typical ICRIT deglitch time to trigger $\overline{PROCHOT}$ . 00b: 15 $\mu$ s 01b: 100 $\mu$ s <default at POR> 10b: 400 $\mu$ s (max 500 us) 11b: 800 $\mu$ s (max 1 ms)
8	Reserved	R/W	0b	Reserved

**Table 16. ProchotOption0 Register (SMBus address = 33h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7-6	VSYS_VTH	R/W	01b	VSYS Threshold Measure on VSYS with fixed 20- $\mu$ s deglitch time. Trigger when SYS pin voltage is below the threshold. 00b: 5.75 V (2-4 s) or 2.85 V (1 s) 01b: 6 V (2-4 s) or 3.1 V (1 s) <default at POR> 10b: 6.25 V (2-4 s) or 3.35 V (1 s) 11b: 6.5 V (2-4 s) or 3.6 V (1 s)
5	EN_PROCHOT_EXT	R/W	0b	$\overline{PROCHOT}$ Pulse Extension Enable When pulse extension is enabled, keep the $\overline{PROCHOT}$ pin voltage LOW until host writes 0x33[2] = 0. 0b: Disable pulse extension <default at POR> 1b: Enable pulse extension
4-3	PROCHOT_WIDTH	R/W	10b	$\overline{PROCHOT}$ Pulse Width Minimum $\overline{PROCHOT}$ pulse width when REG0x33[5] = 0 00b: 100 $\mu$ s 01b: 1 ms 10b: 10 ms <default at POR> 11b: 5 ms

**Table 16. ProchotOption0 Register (SMBus address = 33h) Field Descriptions (continued)**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
2	PROCHOT_CLEAR	R/W	1b	$\overline{\text{PROCHOT}}$ Pulse Clear Clear $\overline{\text{PROCHOT}}$ pulse when 0x3C[5] = 1. 0b: Clear $\overline{\text{PROCHOT}}$ pulse and drive $\overline{\text{PROCHOT}}$ pin HIGH. 1b: Idle <default at POR>
1	INOM_DEG	R/W	0b	INOM Deglitch Time INOM is always 10% above IDPM in 0x3FH. Measure current between ACP and ACN. Trigger when the current is above this threshold. 0b: 1 ms (must be max) <default at POR> 1b: 50 ms (max 60 ms)
0	Reserved	R/W	0b	Reserved

**8.6.1.6 ProchotOption1 Register (SMBus address = 34h) [reset = 8120h]**
**Figure 18. ProchotOption1 Register (SMBus address = 34h) [reset = 8120h]**

15-10				9-8			
IDCHG_VTH				IDCHG_DEG			
R/W				R/W			
7	6	5	4	3	2	1	0
Reserved	PROCHOT_PROFILE_IC	PP_ICRIT	PP_INOM	PP_IDCHG	PP_VSYS	PP_BATPRES	PP_ACOK
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 17. ProchotOption1 Register (SMBus address = 34h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15-10	IDCHG_VTH	R/W	000000b	IDCHG Threshold 6 bit, range, range 0 A to 32256 mA, step 512 mA. There is a 128 mA offset measured current between SRN and SRP. Trigger when the discharge current is above the threshold. If the value is programmed to 000000b, $\overline{\text{PROCHOT}}$ is always triggered. Default: 16384 mA or 100000
9-8	IDCHG_DEG	R/W	01b	IDCHG Deglitch Time 00b: 1.6 ms 01b: 100 $\mu\text{s}$ <default at POR> 10b: 6 ms 11b: 12 ms

**Table 18. ProchotOption1 Register (SMBus address = 34h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7	Reserved	R/W	0b	Reserved
6	PROCHOT_PROFILE_COMP	R/W	0b	$\overline{\text{PROCHOT}}$ Profile When all the REG0x34[6:0] bits are 0, $\overline{\text{PROCHOT}}$ function is disabled. Bit6 Independent comparator 0b: disable <default at POR> 1b: enable
5	PROCHOT_PROFILE_ICRIT	R/W	1b	0b: disable 1b: enable <default at POR>
4	PROCHOT_PROFILE_INOM	R/W	0b	0b: disable <default at POR> 1b: enable
3	PROCHOT_PROFILE_IDCHG	R/W	0b	0b: disable <default at POR> 1b: enable
2	PROCHOT_PROFILE_VSYS	R/W	0b	0b: disable <default at POR> 1b: enable
1	PROCHOT_PROFILE_BATPRES	R/W	0b	0b: disable <default at POR> 1b: enable (one-shot falling edge triggered) If BATPRES is enabled in $\overline{\text{PROCHOT}}$ after the battery is removed, it will immediately send out one-shot $\overline{\text{PROCHOT}}$ pulse.

**Table 18. ProchotOption1 Register (SMBus address = 34h) Field Descriptions (continued)**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
0	PROCHOT_PROFILE_ACOK	R/W	0b	0b: disable <default at POR> 1b: enable (one-shot falling edge triggered) ChargeOption0[15] = 0 to pull $\overline{\text{PROCHOT}}$ low after adapter removal. If BATPRES is enabled in $\overline{\text{PROCHOT}}$ after the battery is removed, it will immediately send out one-shot $\overline{\text{PROCHOT}}$ pulse.

**8.6.1.7 ADCOption Register (SMBus address = 35h) [reset = 2000h]**
**Figure 19. ADCOption Register (SMBus address = 35h) [reset = 2000h]**

15		14		13		12-8									
ADC_CONV		ADC_START		ADC_FULLSCALE		Reserved									
R/W		R/W		R/W		R/W									
7		6		5		4		3		2		1		0	
EN_ADC_CMPIN		EN_ADC_VBUS		EN_ADC_PSYS		EN_ADC_IIN		EN_ADC_IDCHG		EN_ADC_ICHG		EN_ADC_VSYS		EN_ADC_VBAT	
R/W		R/W		R/W		R/W		R/W		R/W		R/W		R/W	

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 19. ADCOption Register (SMBus address = 35h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15	ADC_CONV	R/W	0b	Typical ADC conversion time is 10 ms. 0b: One-shot update. Do one set of conversion updates to registers REG0x23(), REG0x24(), REG0x25(), and REG0x26() after ADC_START = 1. 1b: Continuous update. Do a set of conversion updates to registers REG0x23(), REG0x24(), REG0x25(), and REG0x26() every 1 sec.
14	ADC_START	R/W	0b	0b: No ADC conversion 1b: Start ADC conversion. After the one-shot update is complete, this bit automatically resets to zero
13	ADC_FULLSCALE	R/W	1b	ADC input voltage range. When input voltage is below 5V, or battery is 1S, full scale 2.04V is recommended. 0b: 2.04 V 1b: 3.06 V <default at POR>
12-8	Reserved	R/W	00000b	Reserved

**Table 20. ADCOption Register (SMBus address = 35h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7	EN_ADC_CMPIN	R/W	0b	0b: Disable <default at POR> 1b: Enable
6	EN_ADC_VBUS	R/W	0b	0b: Disable <default at POR> 1b: Enable
5	EN_ADC_PSYS	R/W	0b	0b: Disable <default at POR> 1b: Enable
4	EN_ADC_IIN	R/W	0b	0b: Disable <default at POR> 1b: Enable
3	EN_ADC_IDCHG	R/W	0b	0b: Disable <default at POR> 1b: Enable
2	EN_ADC_ICHG	R/W	0b	0b: Disable <default at POR> 1b: Enable
1	EN_ADC_VSYS	R/W	0b	0b: Disable <default at POR> 1b: Enable
0	EN_ADC_VBAT	R/W	0b	0b: Disable <default at POR> 1b: Enable

## 8.6.2 Charge and PROCHOT Status

### 8.6.2.1 ChargerStatus Register (SMBus address = 20h) [reset = 0000h]

**Figure 20. ChargerStatus Register (SMBus address = 20h) [reset = 0000h]**

15	14	13	12	11	10	9	8
AC_STAT	ICO_DONE	Reserved	IN_VINDPM	IN_IINDPM	IN_FCHRG	IN_PCHRG	Reserved
R	R	R	R	R	R	R	R
7	6	5	4	3	2	1	0
Fault ACOV	Fault BATOC	Fault ACOC	YSOVP_STAT	Reserved	Fault Latchoff	Reserved	Reserved
R	R	R	R	R	R	R	R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 21. ChargerStatus Register (SMBus address = 20h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15	AC_STAT	R	0b	Input source status, same as CHRG_OK pin 0b: Input not present 1b: Input is present
14	ICO_DONE	R	0b	After the ICO routine is successfully executed, the bit goes 1. 0b: ICO is not complete 1b: ICO is complete
13	Reserved	R	0b	Reserved
12	IN_VINDPM	R	0b	0b: Charger is not in VINDPM during forward mode 1b: Charger is in VINDPM during forward mode
11	IN_IINDPM	R	0b	0b: Charger is not in IINDPM 1b: Charger is in IINDPM
10	IN_FCHRG	R	0b	0b: Charger is not in fast charge 1b: Charger is in fast charger
9	IN_PCHRG	R	0b	0b: Charger is not in pre-charge 1b: Charger is in pre-charge
8	Reserved	R		Reserved

**Table 22. ChargerStatus Register (SMBus address = 20h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7	Fault ACOV	R	0b	The faults are latched until a read from host. 0b: No fault 1b: ACOV
6	Fault BATOC	R	0b	The faults are latched until a read from host. 0b: No fault 1b: BATOC
5	Fault ACOC	R	0b	The faults are latched until a read from host. 0b: No fault 1b: ACOC

**Table 22. ChargerStatus Register (SMBus address = 20h) Field Descriptions (continued)**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
4	SYSOVP_STAT	R	0b	SYSOVP Status and Clear When the SYSOVP occurs, this bit is HIGH. During the SYSOVP, the converter is disabled. After the SYSOVP is removed, the user must write a 0 to this bit or unplug the adapter to clear the SYSOVP condition to enable the converter again. 0b: Not in SYSOVP <default at POR> 1b: In SYSOVP. When SYSOVP is removed, write 0 to clear the SYSOVP latch.
3	Reserved	R	0b	Reserved
2	Fault Latchoff	R	0b	The faults are latched until a read from host. 0b: No fault 1b: Latch off (REG0x30[3])
1	Reserved	R		Reserved
0	Reserved	R		Reserved

**8.6.2.2 ProchotStatus Register (SMBus address = 21h) [reset = 0h]**
**Figure 21. ProchotStatus Register (SMBus address = 21h) [reset = 0h]**

15-8							
Reserved							
R							
7	6	5	4	3	2	1	0
Reserved	STAT_COMP	STAT_ICRIT	STAT_INOM	STAT_IDCHG	STAT_VSYS	STAT_Battery_ Removal	STAT_Adapter_ Removal
R	R	R	R	R	R	R	R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 23. ProchotStatus Register (SMBus address = 21h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15-8	Reserved	R	0000000 0b	Reserved

**Table 24. ProchotStatus Register (SMBus address = 21h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7	Reserved	R	0b	Reserved
6	STAT_COMP	R	0b	0b: Not triggered 1b: Triggered
5	STAT_ICRIT	R	0b	0b: Not triggered 1b: Triggered
4	STAT_INOM	R	0b	0b: Not triggered 1b: Triggered
3	STAT_IDCHG	R	0b	0b: Not triggered 1b: Triggered
2	STAT_VSYS	R	0b	0b: Not triggered 1b: Triggered
1	STAT_Battery_Removal	R	0b	0b: Not triggered 1b: Triggered
0	STAT_Adapter_Removal	R	0b	0b: Not triggered 1b: Triggered

### 8.6.3 ChargeCurrent Register (SMBus address = 14h) [reset = 0h]

To set the charge current, write a 16-bit ChargeCurrent() command (REG0x14()) using the data format listed in Figure 22

With 10-mΩ sense resistor, the charger provides charge current range of 64 mA to 8.128 A, with a 64-mA step resolution. Upon POR, ChargeCurrent() is 0 A when auto wakeup is not active. Any conditions for CHRГ\_OK low except ACOV will reset ChargeCurrent() to zero. CELL\_BATPRESZ going LOW (battery removal) will reset the ChargeCurrent() register to 0 A.

Charge current is not reset in ACOC, TSHUT, power path latch off (REG0x30[1]), and SYSOVP.

A 0.1-μF capacitor between SRP and SRN for differential mode filtering is recommended; an optional 0.1-μF capacitor between SRN and ground, and an optional 0.1-μF capacitor between SRP and ground for common mode filtering. Meanwhile, the capacitance on SRP should not be higher than 0.1 μF in order to properly sense the voltage across SRP and SRN for cycle-by-cycle current detection.

The SRP and SRN pins are used to sense voltage drop across RSR with default value of 10 mΩ. However, resistors of other values can also be used. For a larger sense resistor, a larger sense voltage is given, and a higher regulation accuracy; but, at the expense of higher conduction loss. A current sensing resistor value no more than 20 mΩ is suggested.

**Figure 22. ChargeCurrent Register With 10-mΩ Sense Resistor (SMBus address = 14h) [reset = 0h]**

15	14	13	12	11	10	9	8
Reserved			Charge Current, bit 6	Charge Current, bit 5	Charge Current, bit 4	Charge Current, bit 3	Charge Current, bit 2
R/W			R/W	R/W	R/W	R/W	R/W
7	6	5	4	3	2	1	0
Charge Current, bit 1	Charge Current, bit 0	Reserved	Reserved				
R/W	R/W	R/W	R/W				

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 25. Charge Current Register (14h) With 10-mΩ Sense Resistor (SMBus address = 14h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15-13	Reserved	R/W	000b	Not used. 1 = invalid write.
12	Charge Current, bit 6	R/W	0b	0 = Adds 0 mA of charger current. 1 = Adds 4096 mA of charger current.
11	Charge Current, bit 5	R/W	0b	0 = Adds 0 mA of charger current. 1 = Adds 2048 mA of charger current.
10	Charge Current, bit 4	R/W	0b	0 = Adds 0 mA of charger current. 1 = Adds 1024 mA of charger current.
9	Charge Current, bit 3	R/W	0b	0 = Adds 0 mA of charger current. 1 = Adds 512 mA of charger current.
8	Charge Current, bit 2	R/W	0b	0 = Adds 0 mA of charger current. 1 = Adds 256 mA of charger current.

**Table 26. Charge Current Register (14h) With 10-mΩ Sense Resistor (SMBus address = 14h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7	Charge Current, bit 1	R/W	0b	0 = Adds 0 mA of charger current. 1 = Adds 128 mA of charger current.
6	Charge Current, bit 0	R/W	0b	0 = Adds 0 mA of charger current. 1 = Adds 64 mA of charger current.
5-0	Reserved	R/W	000000b	Not used. Value Ignored.

### 8.6.3.1 Battery Pre-Charge Current Clamp

During pre-charge, BATFET works in linear mode or LDO mode (default REG0x12[2] = 1). For 2-4 cell battery, the system is regulated at minimum system voltage in REG0x3E() and the pre-charge current is clamped at 384 mA. For 1 cell battery, the pre-charge to fast charge threshold is 3V, and the pre-charge current is clamped at 384mA. However, the BATFET stays in LDO mode operation till battery voltage is above minimum system voltage (~3.6V). During battery voltage from 3V to 3.6V, the fast charge current is clamped at 2A.

### 8.6.4 MaxChargeVoltage Register (SMBus address = 15h) [reset value based on CELL\_BATPRESZ pin setting]

To set the output charge voltage, write a 16-bit ChargeVoltage register command (REG0x15()) using the data format listed in [Figure 23](#). The charger provides charge voltage range from 1.024 V to 19.200 V, with 16-mV step resolution. Any write below 1.024 V or above 19.200 V is ignored.

Upon POR, REG0x15() is by default set as 4192 mV for 1 s, 8400 mV for 2 s, 12592 mV for 3 s or 16800 mV for 4 s. After CHRG\_OK, if host writes REG0x14() before REG0x15(), the charge will start after the write to REG0x14(). If the battery is different from 4.2 V/cell, the host has to write to REG0x15() before REG0x14() for correct battery voltage setting. Writing REG0x15() to 0 will set REG0x15() to default value on CELL\_BATPRESZ pin, and force REG0x14() to zero to disable charge.

The SRN pin is used to sense the battery voltage for voltage regulation and should be connected as close to the battery as possible, and directly place a decoupling capacitor (0.1- $\mu$ F recommended) as close to the device as possible to decouple high frequency noise.

**Figure 23. MaxChargeVoltage Register (SMBus address = 15h) [reset value based on CELL\_BATPRESZ pin setting]**

15	14	13	12	11	10	9	8
Reserved	Max Charge Voltage, bit 10	Max Charge Voltage, bit 9	Max Charge Voltage, bit 8	Max Charge Voltage, bit 7	Max Charge Voltage, bit 6	Max Charge Voltage, bit 5	Max Charge Voltage, bit 4
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
7	6	5	4	3	2	1	0
Max Charge Voltage, bit 3	Max Charge Voltage, bit 2	Max Charge Voltage, bit 1	Max Charge Voltage, bit 0	Reserved			
R/W	R/W	R/W	R/W	R/W			

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 27. MaxChargeVoltage Register (SMBus address = 15h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15	Reserved	R/W	0b	Not used. 1 = invalid write.
14	Max Charge Voltage, bit 10	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 16384 mV of charger voltage.
13	Max Charge Voltage, bit 9	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 8192 mV of charger voltage
12	Max Charge Voltage, bit 8	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 4096 mV of charger voltage.
11	Max Charge Voltage, bit 7	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 2048 mV of charger voltage.
10	Max Charge Voltage, bit 6	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 1024 mV of charger voltage.
9	Max Charge Voltage, bit 5	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 512 mV of charger voltage.
8	Max Charge Voltage, bit 4	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 256 mV of charger voltage.

**Table 28. MaxChargeVoltage Register (SMBus address = 15h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7	Max Charge Voltage, bit 3	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 128 mV of charger voltage.
6	Max Charge Voltage, bit 2	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 64 mV of charger voltage.
5	Max Charge Voltage, bit 1	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 32 mV of charger voltage.
4	Max Charge Voltage, bit 0	R/W	0b	0 = Adds 0 mV of charger voltage. 1 = Adds 16 mV of charger voltage.
3-0	Reserved	R/W	0000b	Not used. Value Ignored.

**8.6.5 MinSystemVoltage Register (SMBus address = 3Eh) [reset value based on CELL\_BATPRESZ pin setting]**

To set the minimum system voltage, write a 16-bit MinSystemVoltage register command (REG0x3E()) using the data format listed in Figure 25. The charger provides minimum system voltage range from 1.024 V to 16.128 V, with 256-mV step resolution. Any write below 1.024 V or above 16.128 V is ignored. Upon POR, the MinSystemVoltage register is 3.584 V for 1 S, 6.144 V for 2 S and 9.216 V for 3 S, and 12.288 V for 4 S.

**Figure 24. MinSystemVoltage Register (SMBus address = 3Eh) [reset value based on CELL\_BATPRESZ pin setting]**

15	14	13	12	11	10	9	8
Reserved		Min System Voltage, bit 5	Min System Voltage, bit 4	Min System Voltage, bit 3	Min System Voltage, bit 2	Min System Voltage, bit 1	Min System Voltage, bit 0
R/W		R/W	R/W	R/W	R/W	R/W	R/W
7	6	5	4	3	2	1	0
Reserved							
R/W							

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 29. MinSystemVoltage Register (SMBus address = 3Eh) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15-14	Reserved	R/W	00b	Not used. 1 = invalid write.
13	Min System Voltage, bit 5	R/W	0b	0 = Adds 0 mV of system voltage. 1 = Adds 8192 mV of system voltage.
12	Min System Voltage, bit 4	R/W	0b	0 = Adds 0 mV of system voltage. 1 = Adds 4096mV of system voltage.
11	Min System Voltage, bit 3	R/W	0b	0 = Adds 0 mV of system voltage. 1 = Adds 2048 mV of system voltage.
10	Min System Voltage, bit 2	R/W	0b	0 = Adds 0 mV of system voltage. 1 = Adds 1024 mV of system voltage.
9	Min System Voltage, bit 1	R/W	0b	0 = Adds 0 mV of system voltage. 1 = Adds 512 mV of system voltage.
8	Min System Voltage, bit 0	R/W	0b	0 = Adds 0 mV of system voltage. 1 = Adds 256 mV of system voltage.

**Table 30. MinSystemVoltage Register (SMBus address = 3Eh) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7-0	Reserved	R/W	0000000 0b	Not used. Value Ignored.

### 8.6.5.1 System Voltage Regulation

The device employs Narrow VDC architecture (NVDC) with BATFET separating system from battery. The minimum system voltage is set by REG0x3E(). Even with a deeply depleted battery, the system is regulated above the minimum system voltage with BATFET.

When the battery is below minimum system voltage setting, the BATFET operates in linear mode (LDO mode), and the system is regulated above the minimum system voltage setting. As the battery voltage rises above the minimum system voltage, BATFET is fully on when charging or in supplement mode and the voltage difference between the system and battery is the VDS of BATFET. System voltage is regulated 160mV above battery voltage when BATFET is off (no charging or no supplement current).

When BATFET is removed, the system node VSYS is shorted to SRP. Before the converter starts operation, LDO mode needs to be disabled. The following sequence is required to configure charger without BATFET.

1. Before adapter plugs in, put the charger into HIZ mode. (either pull pin 6 ILIM\_HIZ to ground, or set REG0x32[15] to 1)
2. Set 0x12[2] to 0 to disable LDO mode.
3. Set 0x30[0] to 0 to disable auto-wakeup mode.
4. Check if battery voltage is properly programmed (REG0x15)
5. Set pre-charge/charge current (REG0x14)
6. Put the device out of HIZ mode. (Release ILIM\_HIZ from ground and set REG0x32[15]=0).

In order to prevent any accidental SW mistakes, the host sets low input current limit (a few hundred milliamps) when device is out of HIZ.

### 8.6.6 Input Current and Input Voltage Registers for Dynamic Power Management

The charger supports Dynamic Power Management (DPM). Normally, the input power source provides power for the system load or to charge the battery. When the input current exceeds the input current setting, or the input voltage falls below the input voltage setting, the charger decreases the charge current to provide priority to the system load. As the system current rises, the available charge current drops accordingly towards zero. If the system load keeps increasing after the charge current drops down to zero, the system voltage starts to drop. As the system voltage drops below the battery voltage, the battery will discharge to supply the heavy system load.

#### 8.6.6.1 Input Current Registers

To set the maximum input current limit, write a 16-bit IIN\_HOST register command (REG0x3F()) using the data format listed in [Figure 25](#). When using a 10-mΩ sense resistor, the charger provides an input-current limit range of 50 mA to 6400 mA, with 50-mA resolution. The default current limit is 3.3 A. Due to the USB current setting requirement, the register setting specifies the maximum current instead of the typical current. Upon adapter removal, the input current limit is reset to the default value of 3.3 A. The register setting is clamped at 50mA. With code 00h and 01h, the input current limit settings are both 50mA.

The ACP and ACN pins are used to sense R<sub>AC</sub> with the default value of 10 mΩ. For a 20 mΩ sense resistor, a larger sense voltage is given and a higher regulation accuracy, but at the expense of higher conduction loss.

Instead of using the internal DPM loop, the user can build up an external input current regulation loop and have the feedback signal on the ILIM\_HIZ pin.

$$V_{ILIM\_HIZ} = 1V + 40 \times (V_{ACP} - V_{ACN}) = 1 + 40 \times I_{DPM} \times R_{AC} \quad (2)$$

In order to disable ILIM\_HIZ pin, the host can write to 0x31[7] to disable ILIM\_HIZ pin, or pull ILIM\_HIZ pin above 4 V.

**8.6.6.1.1 IIN\_HOST Register With 10-mΩ Sense Resistor (SMBus address = 3Fh) [reset = 4000h]**
**Figure 25. IIN\_HOST Register With 10-mΩ Sense Resistor (SMBus address = 3Fh) [reset = 4100h]**

15	14	13	12	11	10	9	8
Reserved	Input Current set by host, bit 6	Input Current set by host, bit 5	Input Current set by host, bit 4	Input Current set by host, bit 3	Input Current set by host, bit 2	Input Current set by host, bit 1	Input Current set by host, bit 0
R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
7	6	5	4	3	2	1	0
Reserved							
R							

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 31. IIN\_HOST Register With 10-mΩ Sense Resistor (SMBus address = 3Fh) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15	Reserved	R	0b	Not used. 1 = invalid write.
14	Input Current set by host, bit 6	R	1b	0 = Adds 0 mA of input current. 1 = Adds 3200 mA of input current.
13	Input Current set by host, bit 5	R	0b	0 = Adds 0 mA of input current. 1 = Adds 1600 mA of input current.
12	Input Current set by host, bit 4	R	0b	0 = Adds 0 mA of input current. 1 = Adds 800 mA of input current.
11	Input Current set by host, bit 3	R	0b	0 = Adds 0 mA of input current. 1 = Adds 400 mA of input current.
10	Input Current set by host, bit 2	R	0b	0 = Adds 0 mA of input current. 1 = Adds 200 mA of input current.
9	Input Current set by host, bit 1	R	0b	0 = Adds 0 mA of input current. 1 = Adds 100 mA of input current.
8	Input Current set by host, bit 0	R	0b	0 = Adds 0 mA of input current. 1 = Adds 50 mA of input current.

**Table 32. IIN\_HOST Register With 10-mΩ Sense Resistor (SMBus address = 3Fh) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7-0	Reserved	R	0000000 0b	Not used. Value Ignored.

**8.6.6.1.2 IIN\_DPM Register With 10-mΩ Sense Resistor (SMBus address = 022h) [reset = 0h]**

IIN\_DPM register reflects the actual input current limit programmed in the register, either from host or from ICO.

After ICO, the current limit used by DPM regulation may differ from the IIN\_HOST register settings. The actual DPM limit is reported in REG0x22().

**Figure 26. IIN\_DPM Register With 10-mΩ Sense Resistor (SMBus address = 022h) [reset = 0h]**

15	14	13	12	11	10	9	8
Reserved	Input Current in DPM, bit 6	Input Current in DPM, bit 5	Input Current in DPM, bit 4	Input Current in DPM, bit 3	Input Current in DPM, bit 2	Input Current in DPM, bit 1	Input Current in DPM, bit 0
R	R	R	R	R	R	R	R
7	6	5	4	3	2	1	0
Reserved							
R							

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 33. IIN\_DPM Register With 10-mΩ Sense Resistor (SMBus address = 022h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15	Reserved	R	0b	Not used. 1 = invalid write.
14	Input Current in DPM, bit 6	R	0b	0 = Adds 0 mA of input current. 1 = Adds 3200 mA of input current.
13	Input Current in DPM, bit 5	R	0b	0 = Adds 0 mA of input current. 1 = Adds 1600 mA of input current.
12	Input Current in DPM, bit 4	R	0b	0 = Adds 0 mA of input current. 1 = Adds 800mA of input current
11	Input Current in DPM, bit 3	R	0b	0 = Adds 0 mA of input current. 1 = Adds 400 mA of input current.
10	Input Current in DPM, bit 2	R	0b	0 = Adds 0 mA of input current. 1 = Adds 200 mA of input current.
9	Input Current in DPM, bit 1	R	0b	0 = Adds 0 mA of input current. 1 = Adds 100 mA of input current.
8	Input Current in DPM, bit 0	R	0b	0 = Adds 0 mA of input current. 1 = Adds 50 mA of input current.

**Table 34. IIN\_DPM Register With 10-mΩ Sense Resistor (SMBus address = 022h) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7-0	Reserved	R	0000000b	Not used. Value Ignored.

**8.6.6.1.3 InputVoltage Register (SMBus address = 3Dh) [reset = VBUS-1.28V]**

To set the input voltage limit, write a 16-bit InputVoltage register command (REG0x3D()) using the data format listed in [Figure 27](#).

If the input voltage drops more than the InputVoltage register allows, the device enters DPM and reduces the charge current. The default offset voltage is 1.28 V below the no-load VBUS voltage. The DC offset is 3.2 V (0000000).

**Figure 27. InputVoltage Register (SMBus address = 3Dh) [reset = VBUS-1.28V]**

15	14	13	12	11	10	9	8
Reserved		Input Voltage, bit 7	Input Voltage, bit 6	Input Voltage, bit 5	Input Voltage, bit 4	Input Voltage, bit 3	Input Voltage, bit 2
R/W		R/W	R/W	R/W	R/W	R/W	R/W
7	6	5	4	3	2	1	0
Input Voltage, bit 1	Input Voltage, bit 0	Reserved					
R/W	R/W	R/W					

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 35. InputVoltage Register (SMBus address = 3Dh) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
15-14	Reserved	R/W	00b	Not used. 1 = invalid write.
13	Input Voltage, bit 7	R/W	0b	0 = Adds 0 mV of input voltage. 1 = Adds 8192 mV of input voltage.
12	Input Voltage, bit 6	R/W	0b	0 = Adds 0 mV of input voltage. 1 = Adds 4096mV of input voltage.
11	Input Voltage, bit 5	R/W	0b	0 = Adds 0 mV of input voltage. 1 = Adds 2048 mV of input voltage.
10	Input Voltage, bit 4	R/W	0b	0 = Adds 0 mV of input voltage. 1 = Adds 1024 mV of input voltage.
9	Input Voltage, bit 3	R/W	0b	0 = Adds 0 mV of input voltage. 1 = Adds 512 mV of input voltage.
8	Input Voltage, bit 2	R/W	0b	0 = Adds 0 mV of input voltage. 1 = Adds 256 mV of input voltage.

**Table 36. InputVoltage Register (SMBus address = 3Dh) Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION
7	Input Voltage, bit 1	R/W	0b	0 = Adds 0 mV of input voltage. 1 = Adds 128 mV of input voltage.
6	Input Voltage, bit 0	R/W	0b	0 = Adds 0 mV of input voltage. 1 = Adds 64 mV of input voltage
5-0	Reserved	R/W	000000b	Not used. Value Ignored.

**8.6.7 ADCVBUS/PSYS Register (SMBus address = 23h)**

- PSYS: Full range: 3.06 V, LSB: 12 mV
- VBUS: Full range: 3200 mV to 19520 mV, LSB: 64 mV

**Figure 28. ADCVBUS/PSYS Register (SMBus address = 23h)**

15	14	13	12	11	10	9	8
R	R	R	R	R	R	R	R
7	6	5	4	3	2	1	0
R	R	R	R	R	R	R	R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 37. ADCVBUS/PSYS Register Field Descriptions**

BIT	FIELD	TYPE	RESET	DESCRIPTION
15-8		R		8-bit Digital Output of Input Voltage
7-0		R		8-bit Digital Output of System Power

**8.6.8 ADCIBAT Register (SMBus address = 24h)**

- ICHG: Full range: 8.128 A, LSB 64: mA
- IDCHG: Full range: 32.512 A, LSB: 256 mA

**Figure 29. ADCIBAT Register (SMBus address = 24h)**

15	14	13	12	11	10	9	8
Reserved	R	R	R	R	R	R	R
7	6	5	4	3	2	1	0
Reserved	R	R	R	R	R	R	R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 38. ADCIBAT Register Field Descriptions**

BIT	FIELD	TYPE	RESET	DESCRIPTION
15	Reserved	R		Not used. Value ignored.
14-8		R		7-bit Digital Output of Battery Charge Current
7	Reserved	R		Not used. Value ignored.
6-0		R		7-bit Digital Output of Battery Discharge Current

### 8.6.9 ADCIINCMPIN Register (SMBus address = 25h)

- IIN: Full range: 12.75 A, LSB: 50 mA
- CMPIN: Full range: 3.06 V, LSB: 12 mV

**Figure 30. ADCIINCMPIN Register (SMBus address = 25h)**

15	14	13	12	11	10	9	8
R	R	R	R	R	R	R	R
7	6	5	4	3	2	1	0
R	R	R	R	R	R	R	R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 39. ADCIINCMPIN Register Field Descriptions**

BIT	FIELD	TYPE	RESET	DESCRIPTION
15-8		R		8-bit Digital Output of Input Current
7-0		R		8-bit Digital Output of CMPIN voltage

**8.6.10 ADCVSYSVBAT Register (SMBus address = 26h) (reset = )**

- VSYS: Full range: 2.88 V to 19.2 V, LSB: 64 mV
- VBAT: Full range: 2.88 V to 19.2 V, LSB: 64 mV

**Figure 31. ADCVSYSVBAT Register (SMBus address = 26h) (reset = )**

15	14	13	12	11	10	9	8
R	R	R	R	R	R	R	R
7	6	5	4	3	2	1	0
R	R	R	R	R	R	R	R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 40. ADCVSYSVBAT Register Field Descriptions**

BIT	FIELD	TYPE	RESET	DESCRIPTION
15-8		R		8-bit Digital Output of System Voltage
7-0		R		8-bit Digital Output of Battery Voltage

## 8.6.11 ID Registers

### 8.6.11.1 ManufactureID Register (SMBus address = FEh) [reset = 0040h]

**Figure 32. ManufactureID Register (SMBus address = FEh) [reset = 0040h]**

15-0
MANUFACTURE_ID
R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 41. ManufactureID Register Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION (READ ONLY)
15-0	MANUFACTURE_ID	R		40h

### 8.6.11.2 Device ID (DeviceAddress) Register (SMBus address = FFh) [reset = 0h]

**Figure 33. Device ID (DeviceAddress) Register (SMBus address = FFh) [reset = 0h]**

15-8
Reserved
R
7-0
DEVICE_ID
R

LEGEND: R/W = Read/Write; R = Read only; -n = value after reset

**Table 42. Device ID (DeviceAddress) Register Field Descriptions**

SMBus BIT	FIELD	TYPE	RESET	DESCRIPTION (READ ONLY)
15-8	Reserved	R	0b	Reserved
7-0	DEVICE_ID	R	0b	SMBus: 7Ch



## Typical Application (continued)

DESIGN PARAMETER	EXAMPLE VALUE
Minimum System Voltage <sup>(2)</sup>	6144mV for 2s battery

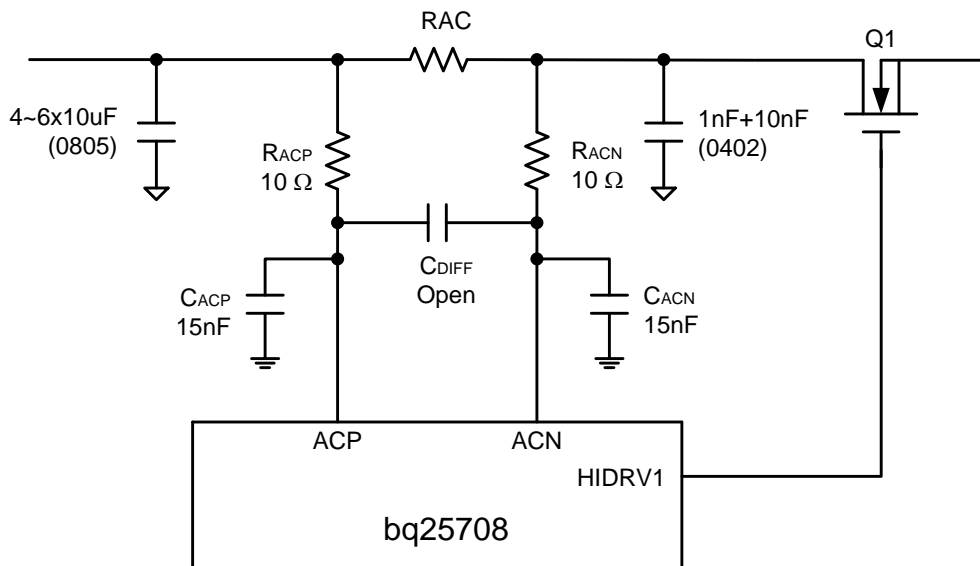
### 9.2.2 Detailed Design Procedure

The parameters are configurable using the evaluation software. The simplified application circuit (see Figure 39, as the application diagram) shows the minimum component requirements. Inductor, capacitor, and MOSFET selection are explained in the rest of this section. Refer to the EVM user's guide (SLUUBG6) for the complete application schematic.

#### 9.2.2.1 ACP-ACN Input Filter

The bq25708 has average current mode control. The input current sensing through ACP/ACN is critical to recover inductor current ripple. Parasitic inductance on board will generate high frequency ringing on ACP-ACN which overwhelms converter sensed inductor current information, so it is difficult to manage parasitic inductance created based on different PCB layout. Bigger parasitic inductance will generate bigger sense current ringing which will cause the average current control loop to go into oscillation.

For real system board condition, we suggest to use below circuit design to get best result and filter noise induced from different PCB parasitic factor. With time constant of filter from 47nsec to 200nsec, the filtering on ringing is effective and in the meantime, the delay of on the sensed signal is small and therefore poses no concern for average current mode control.



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Figure 35. ACN-ACP Input Filter

#### 9.2.2.2 Inductor Selection

The bq25708 has two selectable fixed switching frequency. Higher switching frequency allows the use of smaller inductor and capacitor values. Inductor saturation current should be higher than the charging current ( $I_{CHG}$ ) plus half the ripple current ( $I_{RIPPLE}$ ):

$$I_{SAT} \geq I_{CHG} + (1/2) I_{RIPPLE} \quad (3)$$

The inductor ripple current in buck operation depends on input voltage ( $V_{IN}$ ), duty cycle ( $D_{BUCK} = V_{OUT}/V_{IN}$ ), switching frequency ( $f_S$ ) and inductance ( $L$ ):

$$I_{RIPPLE\_BUCK} = \frac{V_{IN} \times D \times (1 - D)}{f_S \times L} \quad (4)$$

During boost operation, the duty cycle is:

$$D_{\text{BOOST}} = 1 - (V_{\text{BAT}}/V_{\text{IN}})$$

and the ripple current is:

$$I_{\text{RIPPLE\_BOOST}} = D_{\text{BOOST}} \times f_{\text{S}} \times V_{\text{BAT}}/L_{\text{BOOST}}$$

The maximum inductor ripple current happens with  $D = 0.5$  or close to 0.5. For example, the battery charging voltage range is from 9 V to 12.6 V for 3-cell battery pack. For 20 V adapter voltage, 10 V battery voltage gives the maximum inductor ripple current. Another example is 4-cell battery, the battery voltage range is from 12 V to 16.8 V, and 12 V battery voltage gives the maximum inductor ripple current.

Usually inductor ripple is designed in the range of (20-40%) maximum charging current as a trade-off between inductor size and efficiency for a practical design.

### 9.2.2.3 Input Capacitor

Bulk input capacitors should be located in front of input current sensing resistor. Do not recommend to put bulk input capacitors between input sensing resistor and switching MOSFET. Input capacitor should have enough ripple current rating to absorb input switching ripple current. The worst case RMS ripple current is half of the charging current when duty cycle is 0.5 in buck mode. If the converter does not operate at 50% duty cycle, then the worst case capacitor RMS current occurs where the duty cycle is closest to 50% and can be estimated by Equation 5:

$$I_{\text{CIN}} = I_{\text{CHG}} \times \sqrt{D \times (1 - D)} \quad (5)$$

Low ESR ceramic capacitor such as X7R or X5R is preferred for input decoupling capacitor and should be placed to the drain of the high side MOSFET and source of the low side MOSFET as close as possible. Voltage rating of the capacitor must be higher than normal input voltage level. 25 V rating or higher capacitor is preferred for 19-20 V input voltage. Minimum 10uF effective capacitance (7 pcs of 10uF 0805 size capacitor) is suggested for 45 W-65 W adapter.

Ceramic capacitors show a dc-bias effect. This effect reduces the effective capacitance when a dc-bias voltage is applied across a ceramic capacitor, as on the input capacitor of a charger. The effect may lead to a significant capacitance drop, especially for high input voltages and small capacitor packages. See the manufacturer's datasheet about the performance with a dc bias voltage applied. It may be necessary to choose a higher voltage rating or nominal capacitance value in order to get the required value at the operating point.

### 9.2.2.4 Output Capacitor

Output capacitor also should have enough ripple current rating to absorb output switching ripple current. In buck mode the output capacitor RMS current is given:

To get good loop stability, the resonant frequency of the output inductor and output capacitor should be designed between 10 kHz and 20 kHz. The preferred ceramic capacitor is 25V X7R or X5R for output capacitor. Minimum 10uF effective capacitance (7 pcs of 10uF 0805 size capacitor) is suggested to be placed after the Q4 drain, and 50uF effective distributed capacitance on Vsys output. Place minimum 10uF after the charge current sense resistor for best stability.

Ceramic capacitors show a dc-bias effect. This effect reduces the effective capacitance when a dc-bias voltage is applied across a ceramic capacitor, as on the output capacitor of a charger. The effect may lead to a significant capacitance drop, especially for high output voltages and small capacitor packages. See the manufacturer's data sheet about the performance with a dc bias voltage applied. It may be necessary to choose a higher voltage rating or nominal capacitance value in order to get the required value at the operating point.

### 9.2.2.5 Power MOSFETs Selection

Four external N-channel MOSFETs are used for a synchronous switching battery charger. The gate drivers are internally integrated into the IC with 6V of gate drive voltage. 30 V or higher voltage rating MOSFETs are preferred for 19-20 V input voltage.

Figure-of-merit (FOM) is usually used for selecting proper MOSFET based on a tradeoff between the conduction loss and switching loss. For the top side MOSFET, FOM is defined as the product of a MOSFET's on-resistance,  $R_{\text{DS(ON)}}$ , and the gate-to-drain charge,  $Q_{\text{GD}}$ . For the bottom side MOSFET, FOM is defined as the product of the MOSFET's on-resistance,  $R_{\text{DS(ON)}}$ , and the total gate charge,  $Q_{\text{G}}$ .

$$\text{FOM}_{\text{top}} = R_{\text{DS(on)}} \times Q_{\text{GD}}; \text{FOM}_{\text{bottom}} = R_{\text{DS(on)}} \times Q_{\text{G}} \quad (6)$$

The lower the FOM value, the lower the total power loss. Usually lower  $R_{DS(ON)}$  has higher cost with the same package size.

The top-side MOSFET loss includes conduction loss and switching loss. It is a function of duty cycle ( $D=V_{OUT}/V_{IN}$ ), charging current ( $I_{CHG}$ ), MOSFET's on-resistance ( $R_{DS(ON)}$ ), input voltage ( $V_{IN}$ ), switching frequency ( $f_s$ ), turn on time ( $t_{on}$ ) and turn off time ( $t_{off}$ ):

$$P_{top} = D \times I_{CHG}^2 \times R_{DS(on)} + \frac{1}{2} \times V_{IN} \times I_{CHG} \times (t_{on} + t_{off}) \times f_s \quad (7)$$

The first item represents the conduction loss. Usually MOSFET  $R_{DS(ON)}$  increases by 50% with 100°C junction temperature rise. The second term represents the switching loss. The MOSFET turn-on and turn-off times are given by:

$$t_{on} = \frac{Q_{SW}}{I_{on}}, \quad t_{off} = \frac{Q_{SW}}{I_{off}} \quad (8)$$

where  $Q_{SW}$  is the switching charge,  $I_{on}$  is the turn-on gate driving current and  $I_{off}$  is the turn-off gate driving current. If the switching charge is not given in MOSFET datasheet, it can be estimated by gate-to-drain charge ( $Q_{GD}$ ) and gate-to-source charge ( $Q_{GS}$ ):

$$Q_{SW} = Q_{GD} + \frac{1}{2} \times Q_{GS} \quad (9)$$

Gate driving current can be estimated by REGN voltage ( $V_{REGN}$ ), MOSFET plateau voltage ( $V_{plt}$ ), total turn-on gate resistance ( $R_{on}$ ) and turn-off gate resistance ( $R_{off}$ ) of the gate driver:

$$I_{on} = \frac{V_{REGN} - V_{plt}}{R_{on}}, \quad I_{off} = \frac{V_{plt}}{R_{off}} \quad (10)$$

The conduction loss of the bottom-side MOSFET is calculated with the following equation when it operates in synchronous continuous conduction mode:

$$P_{bottom} = (1 - D) \times I_{CHG}^2 \times R_{DS(on)} \quad (11)$$

When charger operates in non-synchronous mode, the bottom-side MOSFET is off. As a result all the freewheeling current goes through the body-diode of the bottom-side MOSFET. The body diode power loss depends on its forward voltage drop ( $V_F$ ), non-synchronous mode charging current ( $I_{NONSYN}$ ), and duty cycle ( $D$ ).

$$P_D = V_F \times I_{NONSYN} \times (1 - D) \quad (12)$$

The maximum charging current in non-synchronous mode can be up to 0.25 A for a 10 mΩ charging current sensing resistor or 0.5 A if battery voltage is below 2.5 V. The minimum duty cycle happens at lowest battery voltage. Choose the bottom-side MOSFET with either an internal Schottky or body diode capable of carrying the maximum non-synchronous mode charging current.

9.2.3 Application Curves

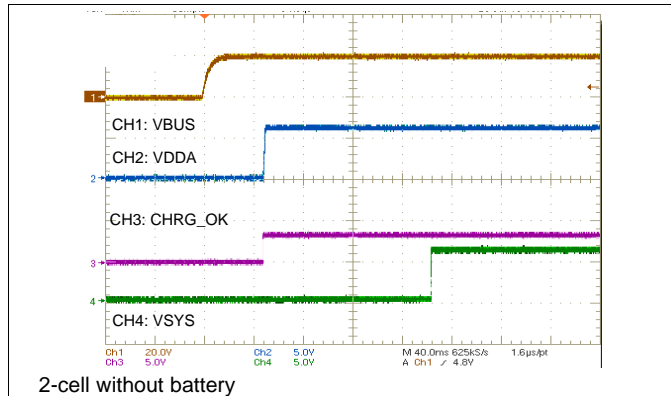


Figure 36. Power Up from 20 V

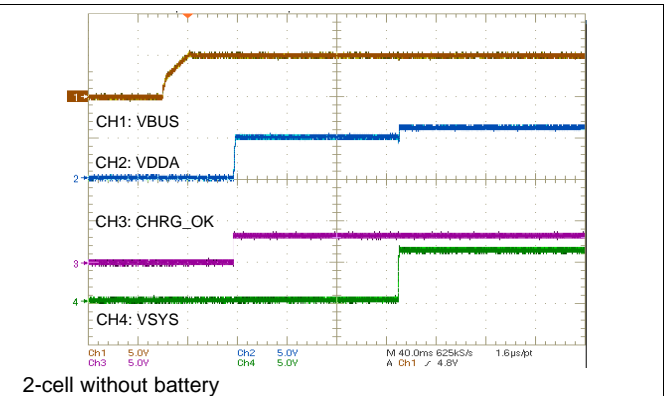


Figure 37. Power Up from 5 V

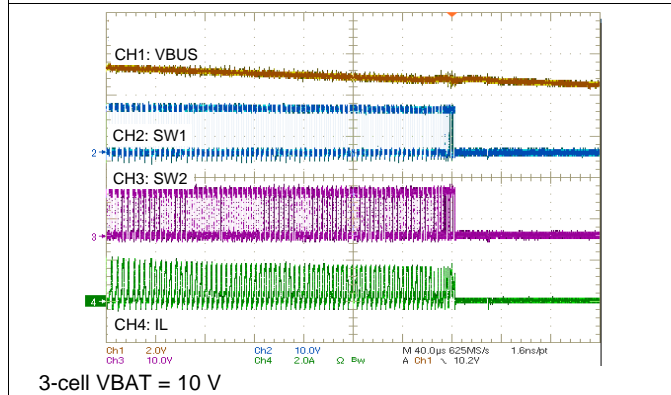


Figure 38. Power Off from 12 V

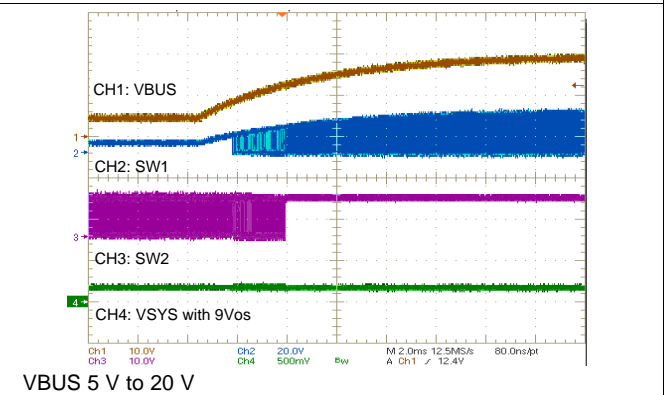


Figure 39. System Regulation

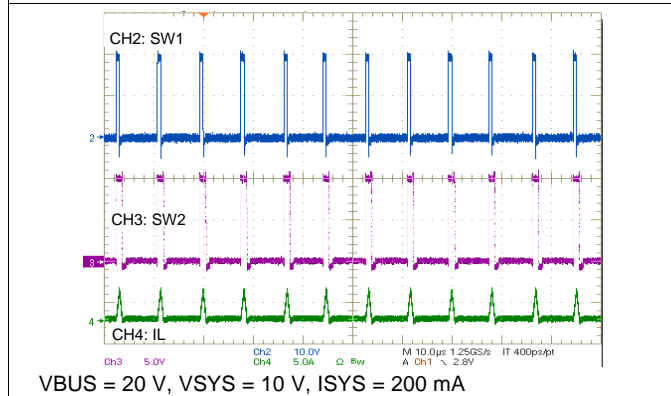


Figure 40. PFM Operation

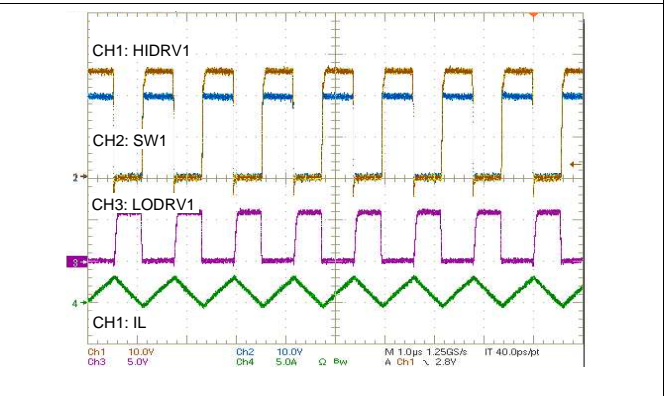
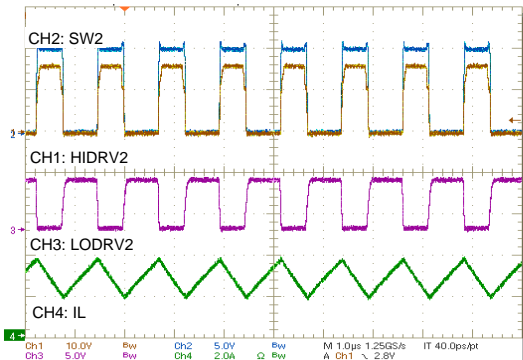
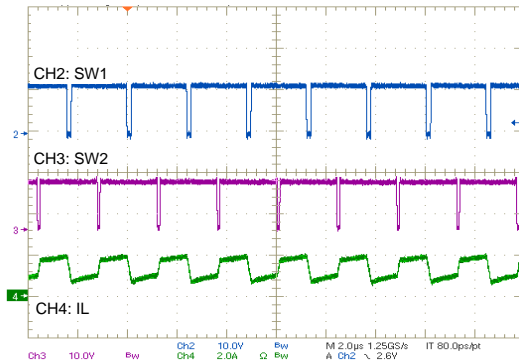


Figure 41. PWM Operation



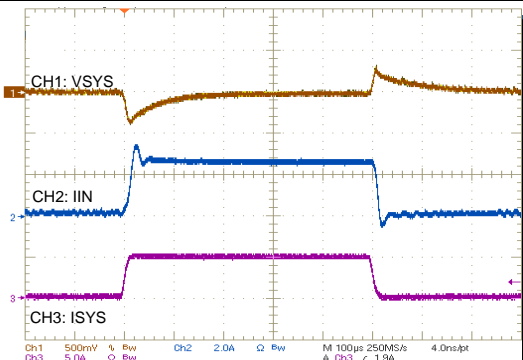
VBUS = 5 V, VBAT = 10 V

Figure 42. Switching During Boost Mode



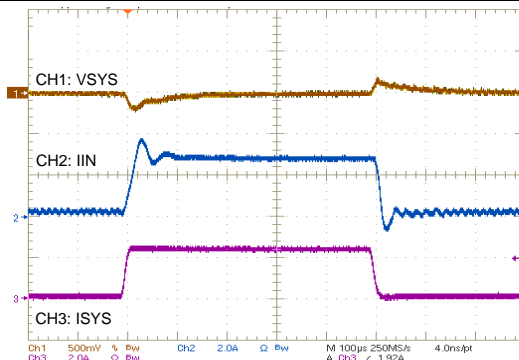
VBUS = 12 V, VBAT = 12 V

Figure 43. Switching During Buck Boost Mode



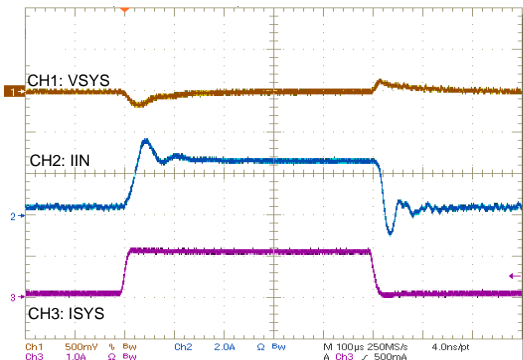
VBUS = 12 V/3.3 A, 3-cell, VSYS = 9 V, Without battery

Figure 44. System Regulation in Buck Mode



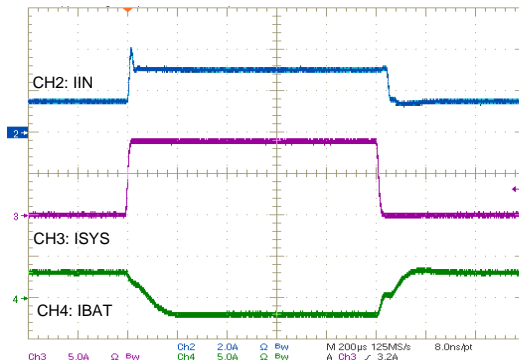
VBUS = 9 V/3.3 A, 3-cell, VSYS = 9 V, Without battery

Figure 45. System Regulation in Buck Boost Mode



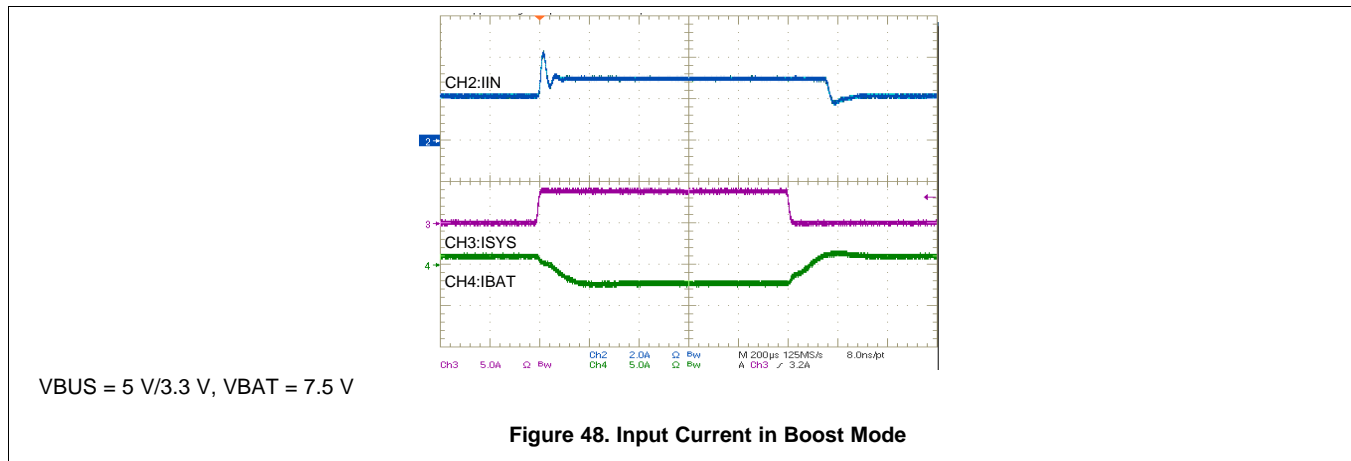
VBUS = 5 V/3.3 A, 3-cell, VSYS = 9 V, Without battery

Figure 46. System Regulation in Boost Mode



VBUS = 20 V/3.3 V, VBAT = 7.5 V

Figure 47. Input Current Regulation in Buck Mode



## 10 Power Supply Recommendations

The valid adapter range is from 3.5V (VBUS\_UVLOZ) to 24V (ACOV) with at least 500mA current rating. When CHRG\_OK goes HIGH, the system is powered from adapter through the charger. When adapter is removed, the system is connected to battery through BATFET. Typically the battery depletion threshold should be greater than the minimum system voltage so that the battery capacity can be fully utilized for maximum battery life.

## 11 Layout

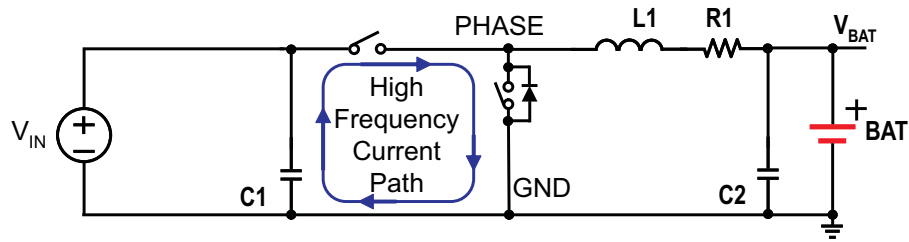
### 11.1 Layout Guidelines

The switching node rise and fall times should be minimized for minimum switching loss. Proper layout of the components to minimize high frequency current path loop (see [Layout Example](#) section) is important to prevent electrical and magnetic field radiation and high frequency resonant problems. Here is a PCB layout priority list for proper layout. Layout PCB according to this specific order is essential.

1. Place the input capacitor as close as possible to the supply of the switching MOSFET and ground connections. Use a short copper trace connection. These parts must be placed on the same layer of PCB using vias to make this connection.
2. The device must be placed close to the gate pins of the switching MOSFET. Keep the gate drive signal traces short for a clean MOSFET drive. The device can be placed on the other side of the PCB of switching MOSFETs.
3. Place an inductor input pin as close as possible to the output pin of the switching MOSFET. Minimize the copper area of this trace to lower electrical and magnetic field radiation but make the trace wide enough to carry the charging current. Do not use multiple layers in parallel for this connection. Minimize parasitic capacitance from this area to any other trace or plane.
4. The charging current sensing resistor should be placed right next to the inductor output. Route the sense leads connected across the sensing resistor back to the device in same layer, close to each other (minimize loop area) and do not route the sense leads through a high-current path (see [Figure 50](#) for Kelvin connection for best current accuracy). Place a decoupling capacitor on these traces next to the device.
5. Place an output capacitor next to the sensing resistor output and ground.
6. Output capacitor ground connections must be tied to the same copper that connects to the input capacitor ground before connecting to system ground.
7. Use a single ground connection to tie the charger power ground to the charger analog ground. Just beneath the device, use analog ground copper pour but avoid power pins to reduce inductive and capacitive noise coupling.
8. Route analog ground separately from power ground. Connect analog ground and connect power ground separately. Connect analog ground and power ground together using power pad as the single ground connection point. Or using a 0- $\Omega$  resistor to tie analog ground to power ground (power pad should tie to analog ground in this case if possible).
9. Decoupling capacitors must be placed next to the device pins. Make trace connection as short as possible.
10. It is critical that the exposed power pad on the backside of the device package be soldered to the PCB ground.
11. The via size and number should be enough for a given current path. See the EVM design ([SLUUBG6](#)) for the recommended component placement with trace and via locations. For the WQFN information, see [SLUA271](#).

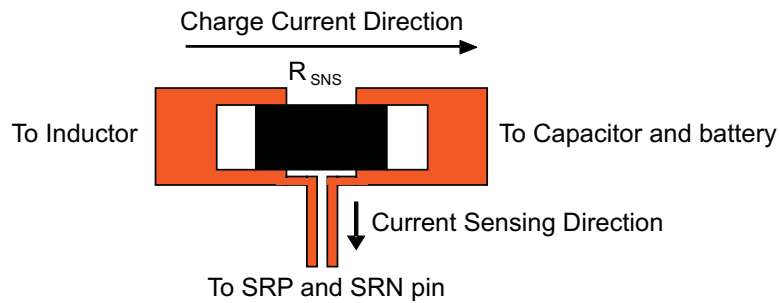
## 11.2 Layout Example

### 11.2.1 Layout Consideration of Current Path



**Figure 49. High Frequency Current Path**

### 11.2.2 Layout Consideration of Short Circuit Protection



**Figure 50. Sensing Resistor PCB Layout**

## 12 Device and Documentation Support

### 12.1 Device Support

#### 12.1.1 Third-Party Products Disclaimer

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### 12.2 Documentation Support

#### 12.2.1 Related Documentation

For related documentation see the following:

- *Semiconductor and IC Package Thermal Metrics* Application Report ([SPRA953](#))
- *bq25700 Evaluation Module User's Guide* ([SLUUBG6](#))
- *QFN/SOP PCB Attachment* Application Report ([SLUA271](#))

### 12.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

**TI E2E™ Online Community** *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At [e2e.ti.com](#), you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

### 12.4 Trademarks

E2E is a trademark of Texas Instruments.  
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### 12.5 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### 12.6 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
BQ25708RSNR	ACTIVE	QFN	RSN	32	3000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ25708	<a href="#">Samples</a>
BQ25708RSNT	ACTIVE	QFN	RSN	32	250	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 85	BQ25708	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

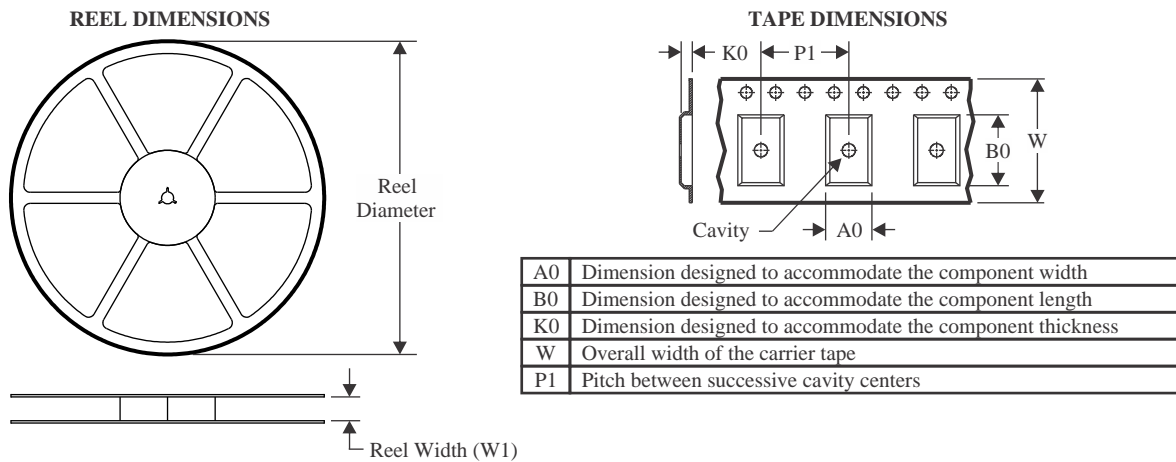
(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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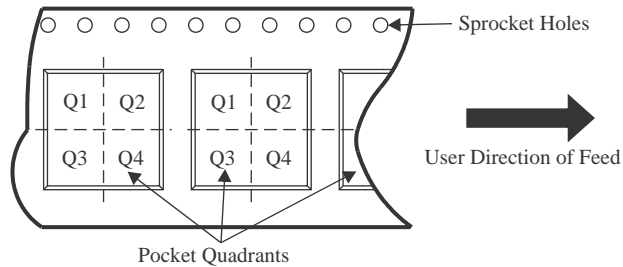
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## TAPE AND REEL INFORMATION

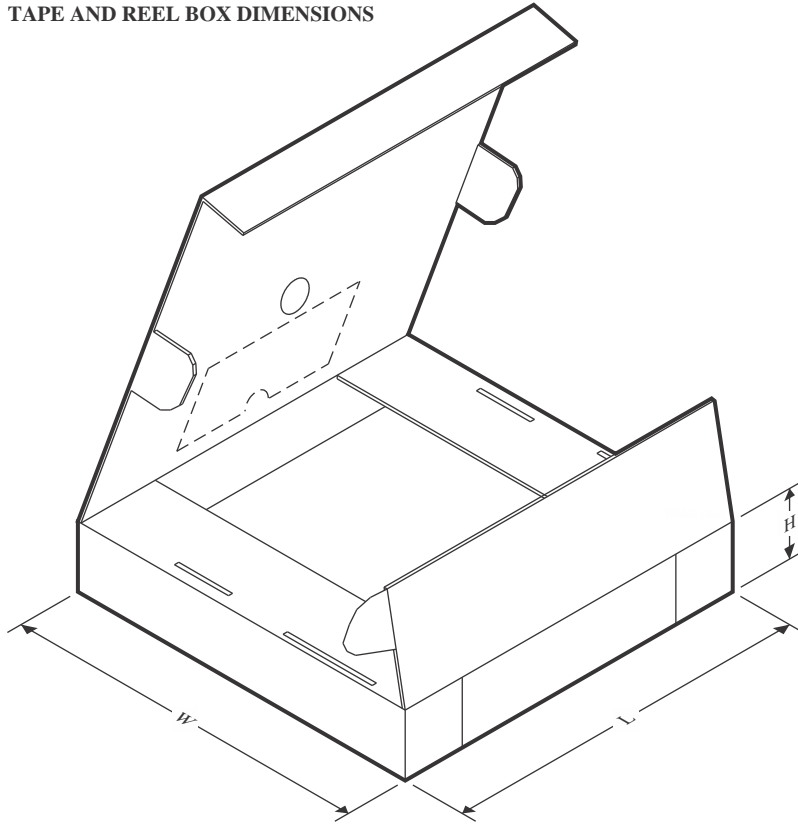


### QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
BQ25708RSNR	QFN	RSN	32	3000	330.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2
BQ25708RSNT	QFN	RSN	32	250	180.0	12.4	4.25	4.25	1.15	8.0	12.0	Q2

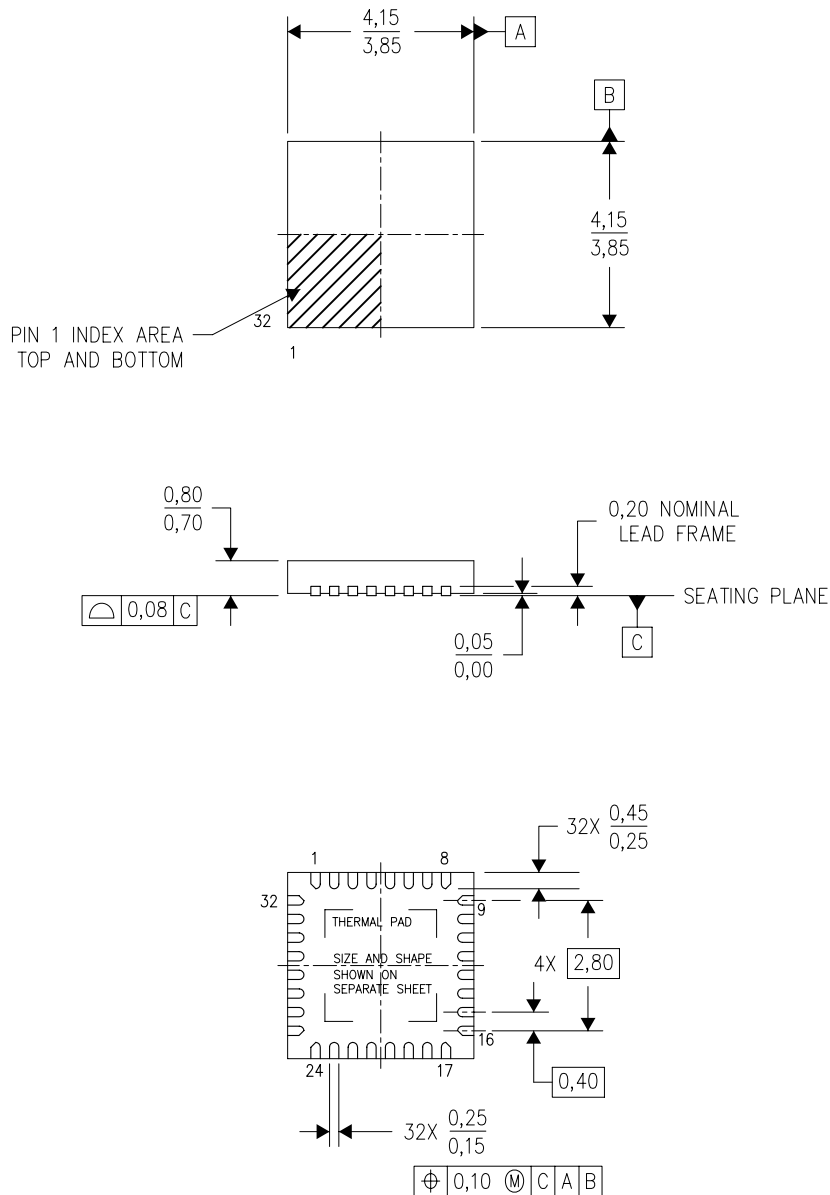
**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
BQ25708RSNR	QFN	RSN	32	3000	367.0	367.0	35.0
BQ25708RSNT	QFN	RSN	32	250	210.0	185.0	35.0

RSN (S-PWQFN-N32)

PLASTIC QUAD FLATPACK NO-LEAD



4207561/C 08/11

- NOTES:
- All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - This drawing is subject to change without notice.
  - QFN (Quad Flatpack No-Lead) Package configuration.
  - The package thermal pad must be soldered to the board for thermal and mechanical performance.
  - See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.

## THERMAL PAD MECHANICAL DATA

RSN (S-PWQFN-N32)

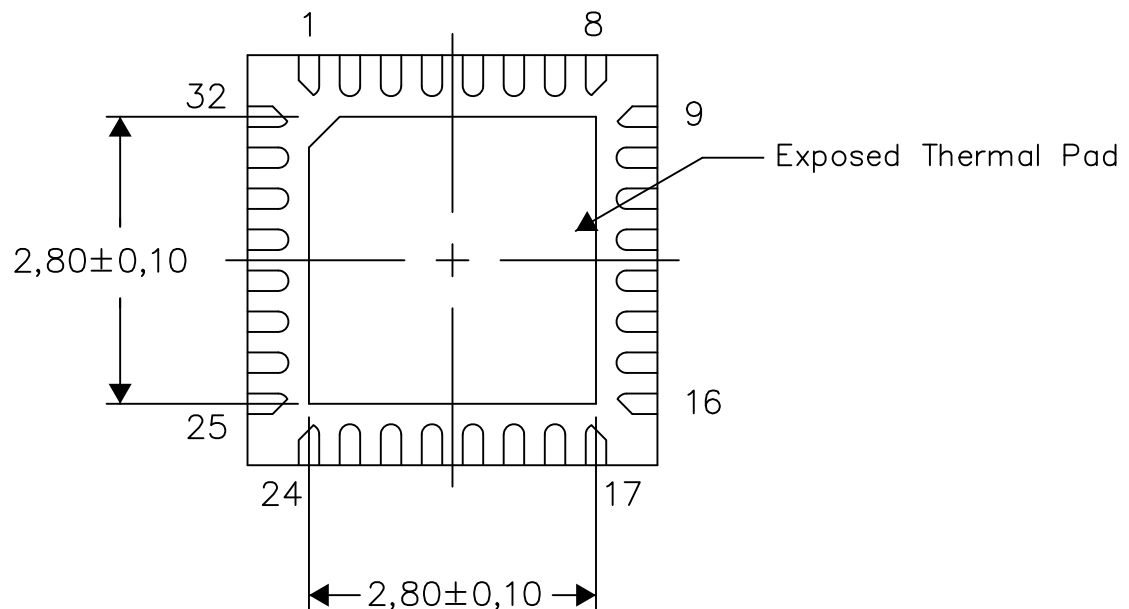
PLASTIC QUAD FLATPACK NO-LEAD

### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at [www.ti.com](http://www.ti.com).

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

Exposed Thermal Pad Dimensions

4209775-2/F 03/14

NOTE: All linear dimensions are in millimeters



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