



THE DATASHEET OF TEA2209T/1J





TEA2209T

Active bridge rectifier controller

Rev. 1.1 — 14 April 2021

Product data sheet

1 General description

The TEA2209T is a product of a new generation of active bridge rectifier controllers replacing the traditional diode bridge.

Using the TEA2209T with low-ohmic high-voltage external MOSFETs significantly improves the efficiency of the power converter as the typical rectifier diode-forward conduction losses are eliminated. Efficiency can improve up to about 1.4 % at 90 V (AC) mains voltage.

The TEA2209T is designed in a silicon-on insulator (SOI) process.

2 Features and benefits

2.1 Efficiency features

- Forward conduction losses of the diode rectifier bridge are eliminated
- Very low IC power consumption (2 mW)

2.2 Application features

- Integrated high-voltage level shifters
- Directly drives all four rectifier MOSFETs
- Very low external part count
- Integrated X-capacitor discharge (2 mA)
- Self-supplying
- Full-wave drive improving total harmonic distortion (THD)
- S016 package

2.3 Control features

- Disable function for all external power FETs
- Undervoltage lockout (UVLO) for high-side and low-side drivers
- Drain-source overvoltage protection for all external power MOSFETs
- Gate pull-down currents at start-up for all external power MOSFETs



3 Applications

The TEA2209T is intended for power supplies with a boost-type power-factor controller as a first stage. The second stage can be a resonant controller, a flyback controller, or any other controller topology. It can be used in all power supplies requiring high efficiency:

- Adapters
- Power supplies for desktop PC and all-in-one PC
- Power supplies for television
- Power supplies for servers

4 Ordering information

Table 1. Ordering information

| Type number | Package | | Version |
|-------------|---------|--|----------|
| | Name | Description | |
| TEA2209T/1 | SO16 | plastic small outline package; 16 leads; body width 3.9 mm | SOT109-1 |

5 Marking

Table 2. Marking

| Type number | Marking code |
|-------------|--------------|
| TEA2209T/1 | TEA2209T |

6 Block diagram

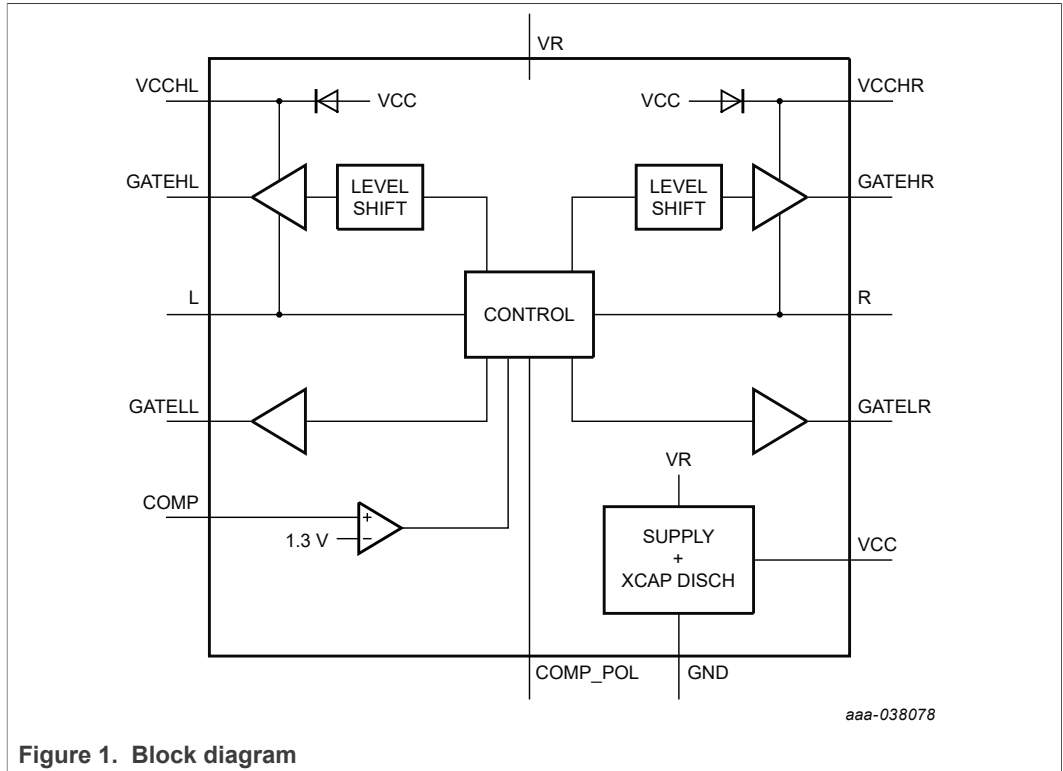


Figure 1. Block diagram

7 Pinning information

7.1 Pinning

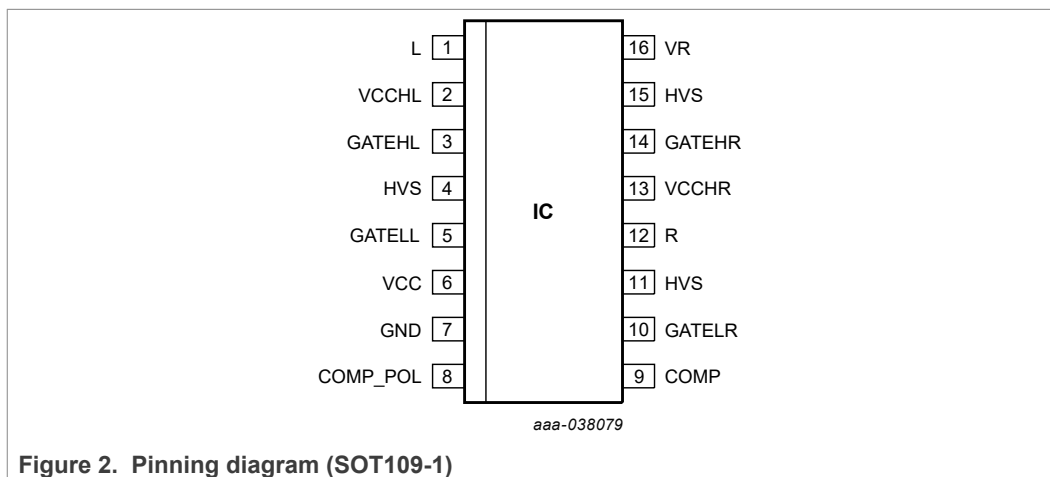


Figure 2. Pinning diagram (SOT109-1)

7.2 Pin description

Table 3. Pin description

| Symbol | Pin | Description |
|----------|-----|---|
| L | 1 | left input, source of upper left MOSFET |
| VCCHL | 2 | left high-side floating supply |
| GATEHL | 3 | gate driver left high side |
| HVS | 4 | high-voltage spacer; not to be connected |
| GATELL | 5 | gate driver left low side |
| VCC | 6 | supply voltage |
| GND | 7 | ground |
| COMP_POL | 8 | comparator polarity setting |
| COMP | 9 | comparator input |
| GATELR | 10 | gate driver right low side |
| HVS | 11 | high-voltage spacer; not to be connected |
| R | 12 | right input, source of upper right MOSFET |
| VCCHR | 13 | right high-side floating supply |
| GATEHR | 14 | gate driver right high side |
| HVS | 15 | high-voltage spacer; not to be connected |
| VR | 16 | rectified mains voltage |

8 Functional description

8.1 Introduction

The TEA2209T is a controller IC for an active bridge rectifier. It can directly drive the four MOSFETs in an active bridge. Figure 3 shows a typical configuration. Since the output is a rectified sine wave, a boost-type power-factor circuit must follow the application.

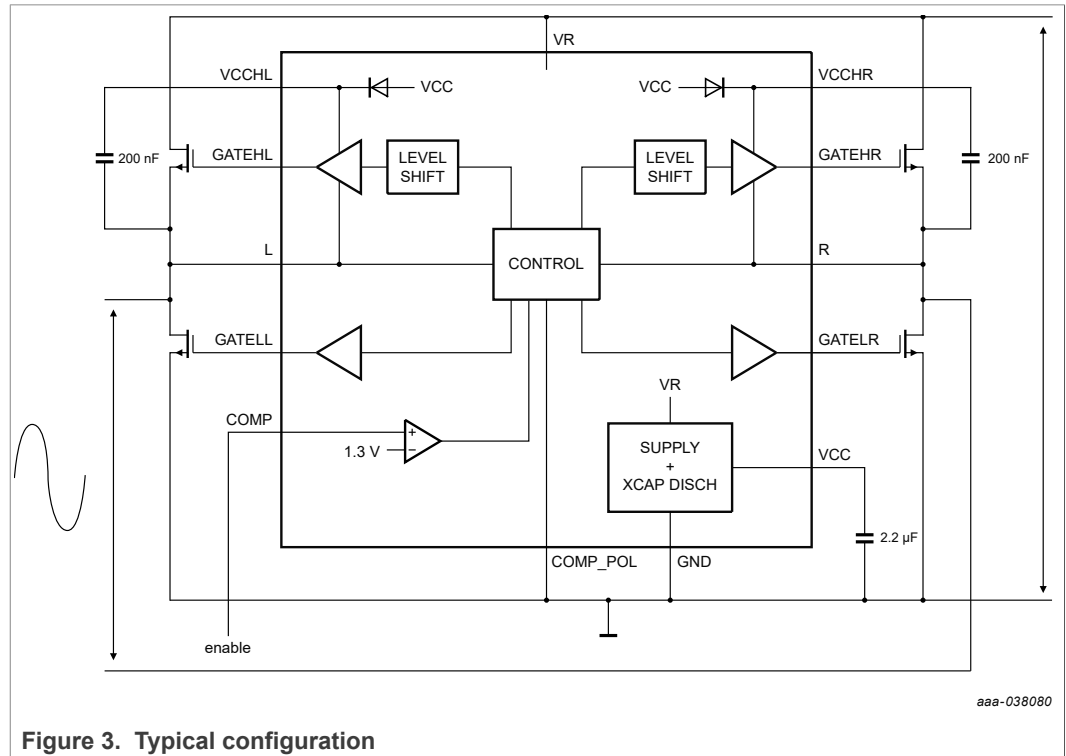


Figure 3. Typical configuration

8.2 Operation

The control circuit of the TEA2209T senses the polarity of the mains voltage between pins L and R. Depending on the polarity, diagonal pairs of power MOSFETs are switched on or off. Depending on the slope polarity, the comparator in the control circuit, which compares the L and R voltages, has thresholds of +250 mV and -250 mV.

The gate drivers are high-current rail-to-rail MOS output drivers. An on-chip supply circuit which draws current from the rectified sine-wave pin VR generates the gate driver voltage. After a zero-crossing of the mains voltage, the supply capacitor C_{VCC} is charged to the regulation level V_{reg} . Then the discharge state is entered. The resulting power dissipation from the mains voltage is about 1 mW, excluding gate charge losses of the external power MOSFETs. These gate charge losses typically add a 1 mW dissipation.

At start-up, the body diodes of the power MOSFETs act as a traditional diode bridge. They cause a peak rectified voltage at pin VR. From this high voltage, the supply capacitor is first charged to the V_{start} voltage and then enters the start-up state. After a next zero-crossing of the mains voltage, the supply capacitor is charged to V_{reg} in the charging state. When the voltage at the supply capacitor exceeds V_{dis} , the gate driver outputs are enabled. The high-side drivers start up later than the low-side drivers. The floating supplies must be charged first and the drain-source voltage of the high-side

power MOSFETs must be less than the drain-source protection voltage. When all drivers are active, the MOSFETs take over the role of the diodes. The result is a much lower power loss than with a passive diode rectifier bridge.

In the discharge state, when the mains voltage is disconnected, the internal bias current discharges the supply capacitor. When the voltage at pin VCC drops to below V_{dis} the X-capacitor discharge state is entered, which draws a 2 mA current from pin VR to discharge the X-capacitor. The waiting time, t_d until the X-capacitor discharge starts is:

$$t_d = C_{VCC} * (V_{reg} - V_{dis}) / 23 \mu A = 0.11E6 * C_{VCC} \tag{1}$$

Using a typical value of 2.2 μF for C_{VCC} yields about 0.24 s. While the VR pin discharges the X-capacitor, the mains can be reconnected. In that case, the charge mode is entered again.

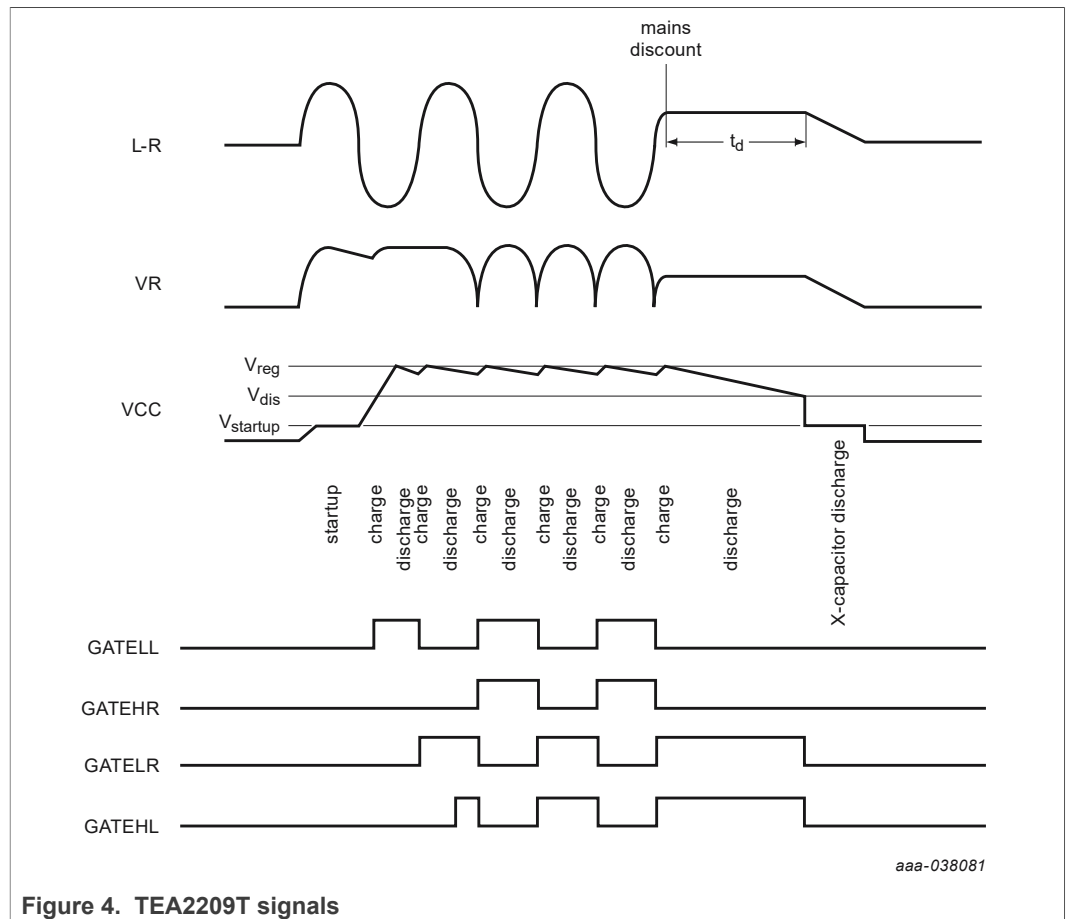


Figure 4. TEA2209T signals

Table 4. TEA2209T states

| State | Description | I _{VR} | I _{VCC} |
|-----------------------|--|-----------------|------------------|
| start-up | supply capacitor kept stable at 4.8 V | 2 mA | 0 |
| charge | supply capacitor is charged from pin VR with 2 mA | +2 mA | -2 mA |
| discharge | internal bias currents and gate charge losses discharge the supply capacitor | 1 µA | 20 µA |
| X-capacitor discharge | supply capacitor and X-capacitor at pin VR are discharged by 2 mA | +2 mA | -2 mA |

When there is hardly any load current or no load current at all on pin VR, the dissipation in the capacitor connected between pin VR and GND, although very low by itself, can contribute relatively much to the total low-load power consumption when the TEA2209T is enabled. So, an external control signal at pin COMP can disable the gate drivers. A comparator with 1.3 V input threshold and 350 mV hysteresis is used at pin COMP. Pin COMP_POL can select the polarity of the comparator. Pin COMP has an internal pull-up and pull-down current which pin COMP_POL selects. The selection is such that with an open pin at COMP, the TEA2209T is enabled. Pin COMP_POL has an internal 0.5 µA pull-down current. Connect pin COMP_POL to either GND or VCC. Do not drive the COMP_POL pin with an external signal.

Table 5. COMP functionality

| COMP_POL = GND | COMP_POL = VCC |
|---|---|
| COMP = low: all gate drivers disabled; internal pull-up current = 0.25 µA | COMP = low; all gate drivers enabled; internal pull-down current = 0.5 µA |

8.3 Protections

8.3.1 Gate pull-down

All gate driver outputs have a pull-down circuit. It ensures that, if a driver supply voltage is lower than the undervoltage lockout level, the discharge of the gate driver output discharges to less than 2 V.

8.3.2 Power MOSFET drain-source protection

If the drain-source voltage of the external power MOSFET exceeds $V_{VCC} - 2\text{ V}$ (low side), $V_{VCC_{HL}} - 3.5\text{ V}$ (high side left), or $V_{VCC_{HR}} - 3.5\text{ V}$ (high side right), all gate driver outputs are disabled. Disabling the gate driver outputs avoids high dissipation and high current peaks in the power MOSFETs during start-up.

8.3.3 Minimum mains voltage

Only when the voltage at either node L or R exceeds 22 V, the charge state is entered.

9 Limiting values

Table 6. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134). All voltages are measured with respect to ground (pin 7). Positive currents flow into the chip. Voltage ratings are valid provided other ratings are not violated. Current ratings are valid provided the other ratings are not violated. The internal IC clearances comply with all NXP design standards and regulations. Moreover, at final testing every chip is checked against the maximum voltage rating in the data sheet.

| Symbol | Parameter | Conditions | Min | Max | Unit |
|----------------------|---|---|------|------|------|
| Voltages | | | | | |
| V _{VR} | voltage on pin VR | operating | -0.4 | 440 | V |
| | | mains transient: maximum 10 minutes over lifetime | -0.4 | 700 | V |
| V _{VCCHL} | voltage on pin VCCHL | operating | -0.4 | 440 | V |
| | | mains transient: maximum 10 minutes over lifetime | -0.4 | 700 | V |
| V _{VCCHR} | voltage on pin VCCHR | operating | -0.4 | 440 | V |
| | | mains transient: maximum 10 minutes over lifetime | -0.4 | 700 | V |
| V _L | voltage on pin L | operating | -5 | +440 | V |
| | | mains transient: maximum 10 minutes over lifetime | -5 | +700 | V |
| V _R | voltage on pin R | operating | -5 | +440 | V |
| | | mains transient: maximum 10 minutes over lifetime | -5 | +700 | V |
| ΔV _(VR-L) | voltage difference between pins VR and L | operating | -10 | +440 | V |
| | | mains transient: maximum 10 minutes over lifetime | -10 | +700 | V |
| ΔV _(VR-R) | voltage difference between pins VR and R | operating | -10 | +440 | V |
| | | mains transient: maximum 10 minutes over lifetime | -10 | +700 | V |
| V _{GATEHR} | voltage on pin GATEHR | operating | -5 | +440 | V |
| | | mains transient: maximum 10 minutes over lifetime | -5 | +700 | V |
| V _{GATEHL} | voltage on pin GATEHL | operating | -5 | +440 | V |
| | | mains transient: maximum 10 minutes over lifetime | -5 | +700 | V |

Table 6. Limiting values...continued

In accordance with the Absolute Maximum Rating System (IEC 60134). All voltages are measured with respect to ground (pin 7). Positive currents flow into the chip. Voltage ratings are valid provided other ratings are not violated. Current ratings are valid provided the other ratings are not violated. The internal IC clearances comply with all NXP design standards and regulations. Moreover, at final testing every chip is checked against the maximum voltage rating in the data sheet.

| Symbol | Parameter | Conditions | Min | Max | Unit |
|--------------------------------------|---------------------------------|---|-------|-------|------|
| SR _{max} | maximum slew rate | pins VR, L, R, VCCHL, VCCHR, GATEHL, GATEHR | - | 50 | V/ns |
| V _{VCC} | voltage on pin VCC | | -0.4 | 14 | V |
| V _{GATELR} | voltage on pin GATELR | | -0.4 | 14 | V |
| V _{GATELL} | voltage on pin GATELL | | -0.4 | 14 | V |
| V _{COMP} | voltage on pin COMP | | -0.4 | 14 | V |
| V _{COMP_POL} | voltage on pin COMP_POL | | -0.4 | 14 | V |
| V _{DD(float)} | float supply voltage | pins GATEHL-L, GATEHR-R, VCCHR-R, VCCHL-L | -0.4 | 14 | V |
| General | | | | | |
| T _j | junction temperature | | -40 | +125 | °C |
| T _{stg} | storage temperature | | -55 | +150 | °C |
| Electrostatic discharge (ESD) | | | | | |
| V _{ESD} | electrostatic discharge voltage | human body model (HBM) | | | |
| | | pins VR, L, R, VCCHL, VCCHR, GATEHL, and GATEHR | -1000 | +1000 | V |
| | | other pins | -2000 | +2000 | V |
| | | charge device model (CDM) | -500 | +500 | V |

10 Thermal characteristics

Table 7. Thermal characteristics

| Symbol | Parameter | Conditions | Typ | Unit |
|----------------------|---|--|---------|------|
| R _{th(j-c)} | thermal resistance from junction to case | in free air | [1] 46 | K/W |
| R _{th(j-a)} | thermal resistance from junction to ambient | in free air; 1-layer PCB | [1] 148 | K/W |
| | | in free air; 4-layer PCB; JEDEC test board | [1] 106 | K/W |

[1] Given thermal resistance values are based on simulation results.

11 Characteristics

Table 8. Characteristics

$T_{amb} = 25\text{ }^{\circ}\text{C}$; all voltages are measured with respect to GND; currents are positive when flowing into the IC; unless otherwise specified.

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|---|---|---|---------|------|------|---------------|
| VR pin | | | | | | |
| I_{on} | on-state current | charging state; X-capacitor discharge state; start-up state | 1.5 | 2 | 2.75 | mA |
| I_{off} | off-state current | discharge state | 0.5 | 0.8 | 1.2 | μA |
| V_{start} | start voltage | high-voltage start-up | 9 | - | - | V |
| VCC pin | | | | | | |
| I_{dch} | discharge current | X-capacitor discharge | 3 | 4 | 5.5 | mA |
| I_{bias} | bias current | discharge state | 15 | 23 | 33 | μA |
| I_{ch} | charge current | charge state | 1.5 | 2 | 2.75 | mA |
| V_{UVLO} | undervoltage lockout voltage | | 3.6 | 4.2 | 4.9 | V |
| $V_{startup}$ | start-up voltage | start-up state | 4.3 | 4.8 | 5.3 | V |
| V_{dis} | disable voltage | high level | 9.2 | 9.7 | 10.2 | V |
| | | hysteresis | 1.1 | 1.5 | 1.8 | V |
| V_{regd} | regulated output voltage | | 10.2 | 10.7 | 11.2 | V |
| Floating supply pins (VCCHL, VCCHR) | | | | | | |
| $I_{I(VCCHL)}$ | input current on pin VCCHL | $V_L = 0\text{ V}$ | 1.4 | 1.8 | 2.5 | μA |
| | | $V_L = 200\text{ V}$ | 4 | 7 | 12 | μA |
| $I_{I(VCCHR)}$ | input current on pin VCCHR | $V_L = 0\text{ V}$ | 1.4 | 1.8 | 2.5 | μA |
| | | $V_L = 200\text{ V}$ | 4 | 7 | 12 | μA |
| $V_{DD(float)UVLO}$ | undervoltage lockout float supply voltage | | 3.6 | 4.2 | 5.0 | V |
| $V_{d(bs)}$ | bootstrap diode voltage | current on diode = 1 mA | 0.8 | 1 | 1.3 | V |
| Gate driver output pins (GATELL, GATELR, GATEHL, GATEHR) | | | | | | |
| I_{source} | source current | $V_{VCC} = 12\text{ V}$; $V_{GATELL} = V_{GATEHL} = 6\text{ V}$; $V_{GATELR} = V_{GATEHR} = 6\text{ V}$ | [1] 125 | 200 | 400 | mA |
| I_{sink} | sink current | $V_{VCC} = 12\text{ V}$; $V_{GATELL} = V_{GATEHL} = 6\text{ V}$; $V_{GATELR} = V_{GATEHR} = 6\text{ V}$ | [1] 150 | 200 | 500 | mA |
| I_{pd} | pull-down current | off-state current; $V_{VCC} = 2\text{ V}$; $V_{GATELL} = V_{GATEHL} = 2\text{ V}$; $V_{GATELR} = V_{GATEHR} = 2\text{ V}$ | 100 | 200 | 250 | μA |
| R_{on} | on-state resistance | | 11 | 15 | 20 | Ω |
| R_{off} | off-state resistance | | 7 | 10 | 14 | Ω |
| $V_{prot(G)}$ | gate driver protection voltage | VR-VCCHR; VR-VCCHL | -5 | -3.5 | -2 | V |
| | | L-VCC; R-VCC | -3 | -2.3 | -1 | V |

Table 8. Characteristics...continued

$T_{amb} = 25\text{ }^{\circ}\text{C}$; all voltages are measured with respect to GND; currents are positive when flowing into the IC; unless otherwise specified.

| Symbol | Parameter | Conditions | Min | Typ | Max | Unit |
|--|-----------------------------------|---|------|------|------|---------------|
| Control circuit (pins L and R) | | | | | | |
| V_{th} | threshold voltage | peak detector threshold voltage | 15 | 22 | 32 | V |
| I_{det} | detection current | peak detector current | 0.4 | 0.5 | 0.6 | μA |
| V_{offset} | offset voltage | zero-crossing comparator offset voltage | 150 | 250 | 350 | mV |
| t_d | delay time | zero-crossing comparator delay time | | | | |
| | | $dV/dt = 0.1\text{ V}/\mu\text{s}$ ^[2] | 1200 | 1500 | 2500 | ns |
| | | $dV/dt = 10\text{ V}/\mu\text{s}$ ^[2] | 550 | 700 | 1200 | ns |
| Disable circuit (pin COMP and COMP_POL) | | | | | | |
| $V_{th(COMP)}$ | threshold voltage on pin COMP | high level | 1.2 | 1.3 | 1.4 | V |
| | | hysteresis | 0.28 | 0.35 | 0.42 | V |
| $I_{i(COMP)}$ | input current on pin COMP | pull-up current | 0.18 | 0.25 | 0.32 | μA |
| | | pull-down current | 0.2 | 0.44 | 0.7 | μA |
| $V_{th(COMP_POL)}$ | threshold voltage on pin COMP_POL | high level | 3.5 | 4.2 | 5.0 | V |
| | | hysteresis | 0.2 | 0.27 | 0.4 | V |
| $I_{i(COMP_POL)}$ | input current on pin COMP_POL | pull-down current | 0.33 | 0.5 | 0.65 | μA |

[1] Covered by correlating measurement.

[2] Guaranteed by design and validation.

12 Application information

A switched-mode power supply (SMPS) with the TEA2209T typically consists of a mains filter in front of the TEA2209T followed by a boost-type power factor controller. A resonant controller, flyback controller, or any other topology can follow this boost-type PFC.

Special attention must be given to the connection of the VR, L, and R pins of the TEA2209T. Mains transients or surges must be limited to voltages below 700 V.

If a 2 kV ESD rating is required on all pins, a 100 pF capacitor from pins L, R, and VR to ground can be used to achieve the 2 kV ESD.

Typical values for the three external capacitors are 1 μF to 2.2 μF (supply capacitor) and 100 nF to 220 nF (bootstrap capacitors). Supply capacitors with higher values increase the delay time (t_d) for the X-capacitor discharge. They may also increase the dissipation because the supply capacitor C_{VCC} may not be charged every half-mains cycle. Bootstrap capacitors with lower value may cause a voltage drop that is too high because of the gate charge losses.

When there is hardly any load current or no load current at all on pin VR, the dissipation in the capacitor connected between pins VR and GND, although very low by itself, can contribute relatively much to the total low-load power consumption when the TEA2209T is enabled. So, to minimize power consumption, the TEA2209T can be switched off at low power. Switching off at low power can be done in several ways. One option is a filter connected to the PFC gate signal. The pin COMP_POL is grounded such that, at a low duty cycle of the PFC signal, the voltage at pin COMP is low. It disables the TEA2209T.

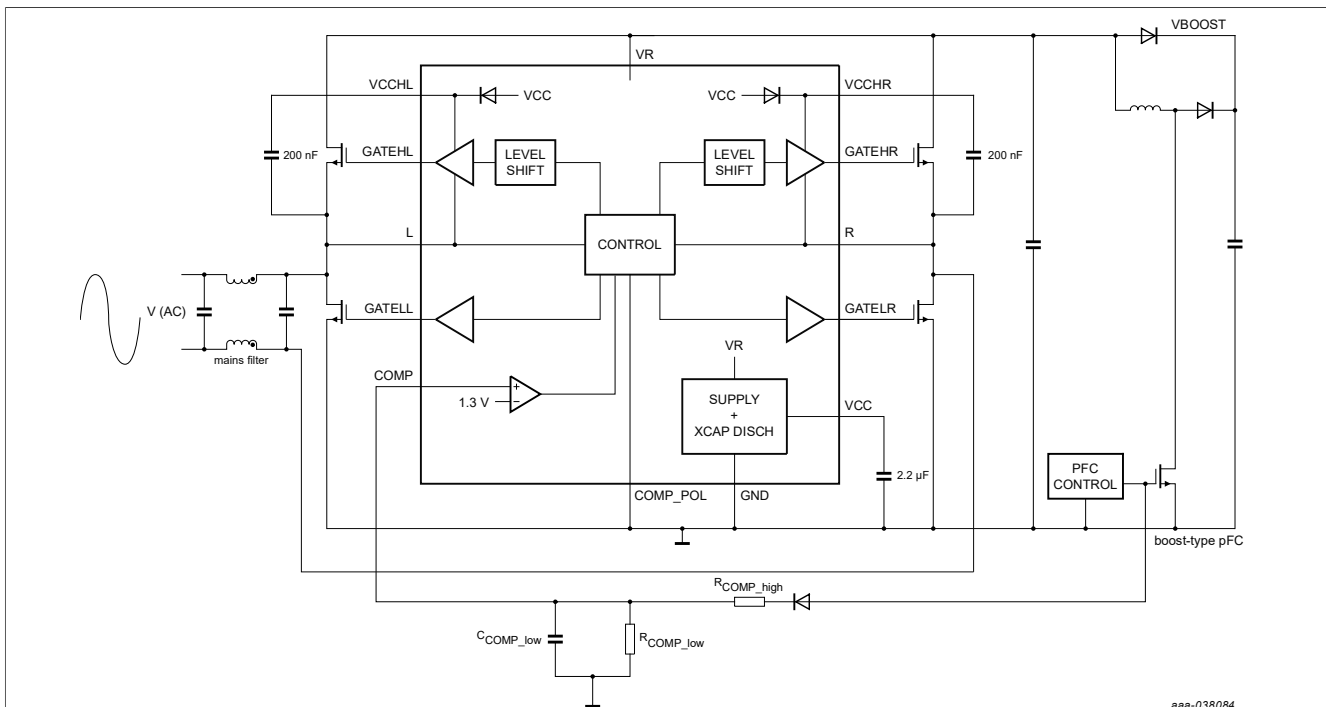
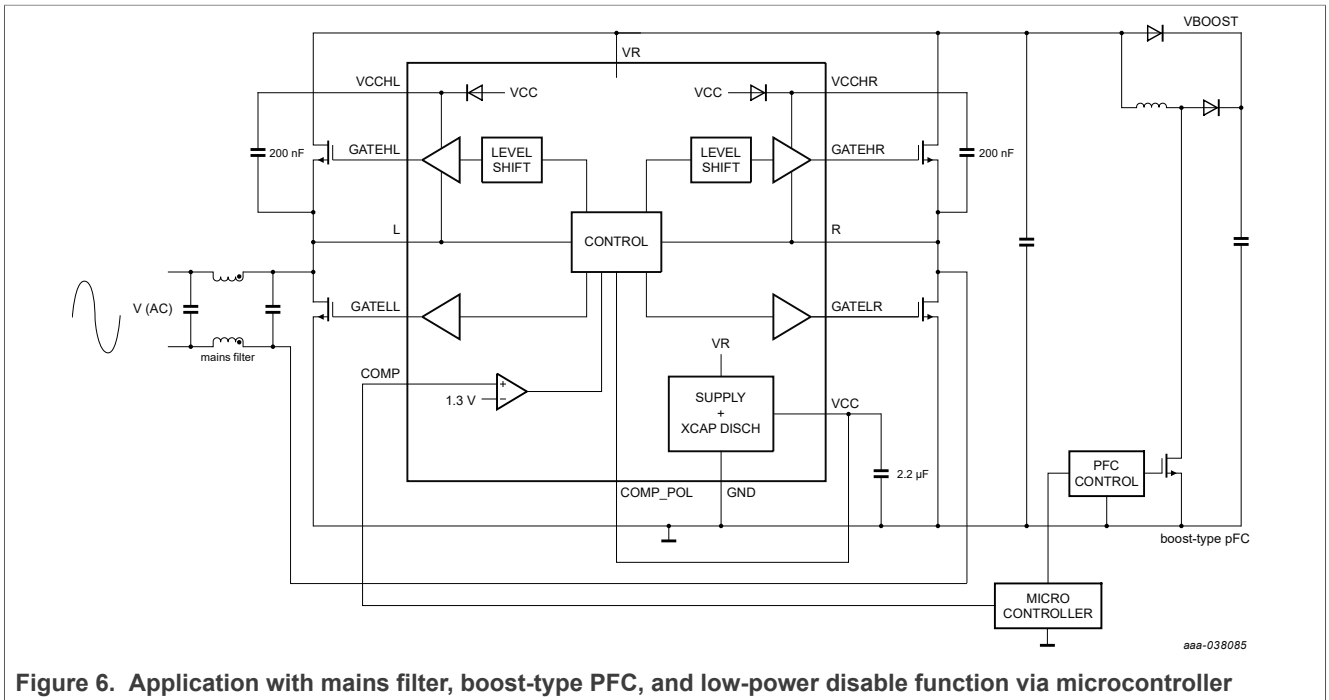


Figure 5. Application with mains filter, boost-type PFC, and low-power disable function via PFC gate signal

A microcontroller can also disable the TEA2209T. An application with a microcontroller is shown in [Figure 6](#). Pin COMP_POL is connected to VCC. If pin COMP is high, the TEA2209T is disabled.



13 Package outline

Table 9.

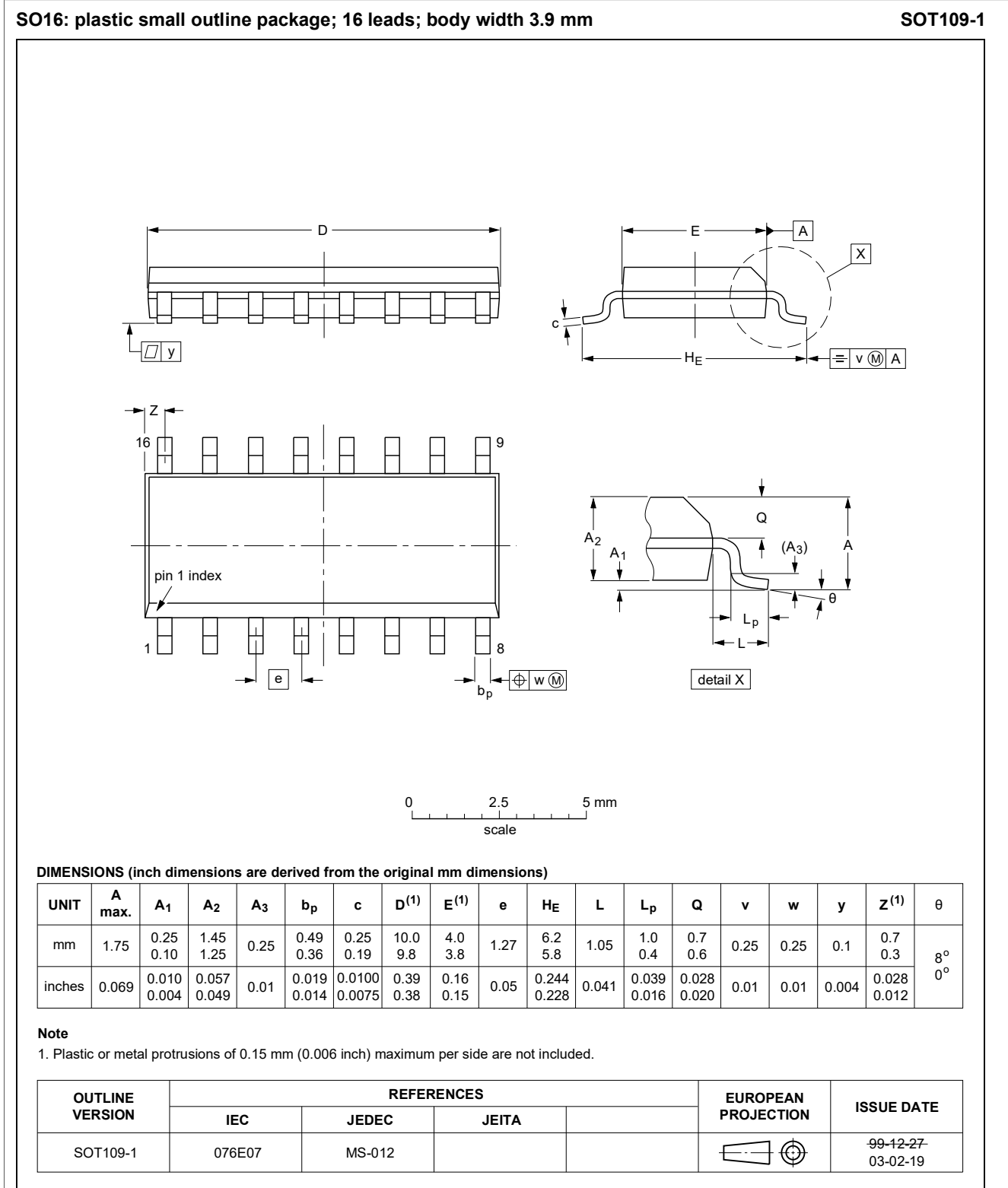


Figure 7. Package outline SOT109-1 (SO16)

14 Abbreviations

Table 10. Abbreviations

| Acronym | Description |
|---------|---|
| CDM | change device model |
| ESD | electrostatic discharge |
| HBM | human body model |
| MOSFET | metal–oxide–semiconductor field-effect transistor |
| MOV | metal-oxide varistor |
| PFC | power-factor controller |
| SMPS | switched-mode power supply |
| SOI | silicon-on insulator |
| THD | total harmonic distortion |
| UVLO | undervoltage lockout |

15 Revision history

Table 11. Revision history

| Document ID | Release date | Data sheet status | Change notice | Supersedes |
|----------------|--|--------------------|---------------|--------------|
| TEA2209T v.1.1 | 20210414 | Product data sheet | - | TEA2209T v.1 |
| Modifications: | • Section 11 "Characteristics" has been updated. | | | |
| TEA2209T v.1 | 20210324 | Product data sheet | - | - |

16 Legal information

16.1 Data sheet status

| Document status ^{[1][2]} | Product status ^[3] | Definition |
|-----------------------------------|-------------------------------|---|
| Objective [short] data sheet | Development | This document contains data from the objective specification for product development. |
| Preliminary [short] data sheet | Qualification | This document contains data from the preliminary specification. |
| Product [short] data sheet | Production | This document contains the product specification. |

[1] Please consult the most recently issued document before initiating or completing a design.

[2] The term 'short data sheet' is explained in section "Definitions".

[3] The product status of device(s) described in this document may have changed since this document was published and may differ in case of multiple devices. The latest product status information is available on the Internet at URL <http://www.nxp.com>.

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