



**THE DATASHEET OF  
ASEK770KCB-150U-T-DK**



## Thermally Enhanced, Fully Integrated, Hall-Effect-Based High-Precision Linear Current Sensor IC with 100 $\mu\Omega$ Current Conductor

### FEATURES AND BENEFITS

- Industry-leading total output accuracy achieved with new piecewise linear digital temperature compensation of offset and sensitivity
- Industry-leading noise performance through proprietary amplifier and filter design techniques
- 120 kHz typical bandwidth
- 4.1  $\mu\text{s}$  output rise time in response to step input current
- Integrated shield greatly reduces capacitive coupling from current conductor to die due to high  $dV/dt$  signals, and prevents offset drift in high-side, high-voltage applications
- Greatly improved total output error through digitally programmed and compensated gain and offset over the full operating temperature range
- Small package size, with easy mounting capability
- Monolithic Hall IC for high reliability
- Ultralow power loss: 100  $\mu\Omega$  internal conductor resistance
- Galvanic isolation allows use in economical, high-side current sensing in high-voltage systems
- 4.5 to 5.5 V, single supply operation
- Output voltage proportional to AC or DC currents
- Factory-trimmed for accuracy
- Extremely stable output offset voltage

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### PACKAGE: 5-pin package (suffix CB)



*Additional leadforms available for qualifying volumes*

### DESCRIPTION

The Allegro™ ACS770 family of current sensor ICs provides economical and precise solutions for AC or DC current sensing. Typical applications include motor control, load detection and management, power supply and DC-to-DC converter control, inverter control, and overcurrent fault detection.

The device consists of a precision, low-offset linear Hall circuit with a copper conduction path located near the die. Applied current flowing through this copper conduction path generates a magnetic field that is concentrated by a low magnetic hysteresis core, then converted by the Hall IC into a proportional voltage. Device accuracy is optimized through the close proximity of the magnetic signal to the Hall transducer. A precise, proportional output voltage is provided by the low-offset, chopper-stabilized BiCMOS Hall IC, which is programmed for accuracy at the factory. Proprietary digital temperature compensation technology greatly improves the IC accuracy and temperature stability without influencing the high-bandwidth operation of the analog output.

High-level immunity to current conductor  $dV/dt$  and stray electric fields is offered by Allegro proprietary integrated shield technology for low output voltage ripple and low offset drift in high-side, high-voltage applications.

The output of the device has a positive slope ( $>V_{CC}/2$  for bidirectional devices) when an increasing current flows through the primary copper conduction path (from terminal 4 to terminal 5), which is the path used for current sampling. The internal resistance of this conductive path is 100  $\mu\Omega$  typical, providing low power loss.

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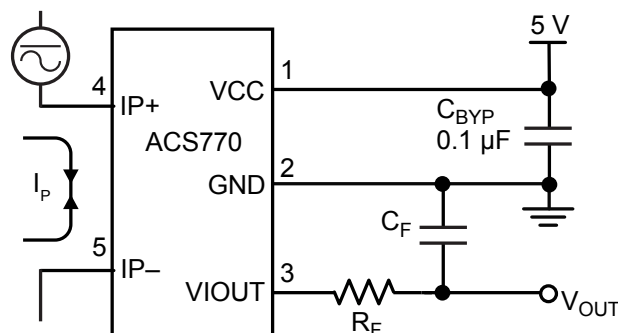


TÜV America  
Certificate Number:  
U8V 14 05 54214 037



CB Certificate Number:  
US-29755-UL

**Application 1:** the ACS770 outputs an analog signal,  $V_{OUT}$ , that varies linearly with the bidirectional AC or DC primary sampled current,  $I_P$ , within the range specified.  $R_F$  and  $C_F$  are for optimal noise management, with values that depend on the application.



**Typical Application**

### FEATURES AND BENEFITS (continued)

- Undervoltage lockout for  $V_{CC}$  below specification
- AEC-Q100 automotive qualified
- UL certified, File No. US-29755-UL

### DESCRIPTION (continued)

The thickness of the copper conductor allows survival of the device at high overcurrent conditions. The terminals of the conductive path are electrically isolated from the signal leads (pins 1 through 3). This allows the ACS770 family of sensor ICs to be used in applications requiring electrical isolation without the use of opto-isolators or other costly isolation techniques.

The device is fully calibrated prior to shipment from the factory. The ACS770 family is lead (Pb) free. All leads are plated with 100% matte tin, and there is no Pb inside the package. The heavy gauge leadframe is made of oxygen-free copper.

### SELECTION GUIDE

Part Number	Package		Primary Sampled Current, $I_p$ (A)	Sensitivity Sens (Typ.) (mV/A)	Current Directionality	$T_{OP}$ (°C)	Packing
	Terminals	Signal Pins					
ACS770LCB-050B-PFF-T	Formed	Formed	$\pm 50$	40	Bidirectional	-40 to 150	34 pieces per tube
ACS770LCB-050U-PFF-T	Formed	Formed	50	80	Unidirectional		
ACS770LCB-100B-PFF-T	Formed	Formed	$\pm 100$	20	Bidirectional		
ACS770LCB-100B-PSF-T	Straight	Formed					
ACS770LCB-100B-PSS-T	Straight	Straight					
ACS770LCB-100U-PFF-T	Formed	Formed	100	40	Unidirectional		
ACS770KCB-150B-PFF-T	Formed	Formed	$\pm 150$	13.3	Bidirectional	-40 to 125	
ACS770KCB-150B-PSS-T	Straight	Straight					
ACS770KCB-150U-PFF-T	Formed	Formed	150	26.7	Unidirectional		
ACS770KCB-150U-PSF-T	Straight	Formed					
ACS770ECB-200B-PFF-T	Formed	Formed	$\pm 200$	10	Bidirectional	-40 to 85	
ACS770ECB-200B-PSF-T	Straight	Formed					
ACS770ECB-200U-PFF-T	Formed	Formed	200	20	Unidirectional		
ACS770ECB-200U-PSF-T	Straight	Formed					

### ABSOLUTE MAXIMUM RATINGS

Characteristic	Symbol	Notes	Rating	Unit
Forward Supply Voltage	$V_{CC}$		6	V
Reverse Supply Voltage	$V_{RCC}$		-0.5	V
Forward Output Voltage	$V_{IOUT}$		25	V
Reverse Output Voltage	$V_{RIOUT}$		-0.5	V
Output Source Current	$I_{OUT(SOURCE)}$	V <sub>IOUT</sub> to GND	3	mA
Output Sink Current	$I_{OUT(SINK)}$	Minimum pull-up resistor of 500 $\Omega$ , from V <sub>CC</sub> to V <sub>IOUT</sub>	10	mA
Nominal Operating Ambient Temperature	$T_{OP}$	Range E	-40 to 85	$^{\circ}C$
		Range K	-40 to 125	$^{\circ}C$
		Range L	-40 to 150	$^{\circ}C$
Maximum Junction	$T_J(max)$		165	$^{\circ}C$
Storage Temperature	$T_{stg}$		-65 to 165	$^{\circ}C$

### ISOLATION CHARACTERISTICS

Characteristic	Symbol	Notes	Rating	Unit
Dielectric Surge Strength Test Voltage	$V_{SURGE}$	Tested $\pm 5$ pulses at 2/minute in compliance to IEC 61000-4-5 1.2 $\mu s$ (rise) / 50 $\mu s$ (width)	8000	V
Dielectric Strength Test Voltage [1]	$V_{ISO}$	Agency type-tested for 60 seconds per UL standard 62368-1, 2nd Edition. Tested at 4320 $V_{RMS}$ for 1 second in production.	4800	$V_{RMS}$
Working Voltage for Basic Isolation	$V_{WVBI}$	For basic (single) isolation per UL standard 62368-1, 2nd Edition	1358	$V_{PK}$ or $V_{DC}$
			960	$V_{RMS}$
Working Voltage for Reinforced Isolation	$V_{WFRI}$	For reinforced (double) isolation per UL standard 62368-1, 2nd Edition	672	$V_{PK}$ or $V_{DC}$
			475	$V_{RMS}$
Clearance	$D_{cl}$	Minimum distance through air from IP leads to signal leads	6.8	mm
Creepage	$D_{cr}$	Minimum distance along package body from IP leads to signal leads	6.8	mm
Distance Through Insulation	DTI	Minimum internal distance through insulation	0.47	mm
Comparative Tracking Index	CTI	Material Group II	400 to 599	V

[1] Allegro does not conduct 60-second testing. It is done only during the UL certification process.

### THERMAL CHARACTERISTICS: May require derating at maximum conditions

Characteristic	Symbol	Test Conditions [2]	Value	Unit
Package Thermal Resistance	$R_{\theta JA}$	Mounted on the Allegro evaluation board with 2800 mm <sup>2</sup> (1400 mm <sup>2</sup> on component side and 1400 mm <sup>2</sup> on opposite side) of 4 oz. copper connected to the primary leadframe and with thermal vias connecting the copper layers. Performance is based on current flowing through the primary leadframe and includes the power consumed by the PCB.	7	$^{\circ}C/W$

[2] Additional thermal information available on the Allegro website.

### TYPICAL OVERCURRENT CAPABILITIES [3][4]

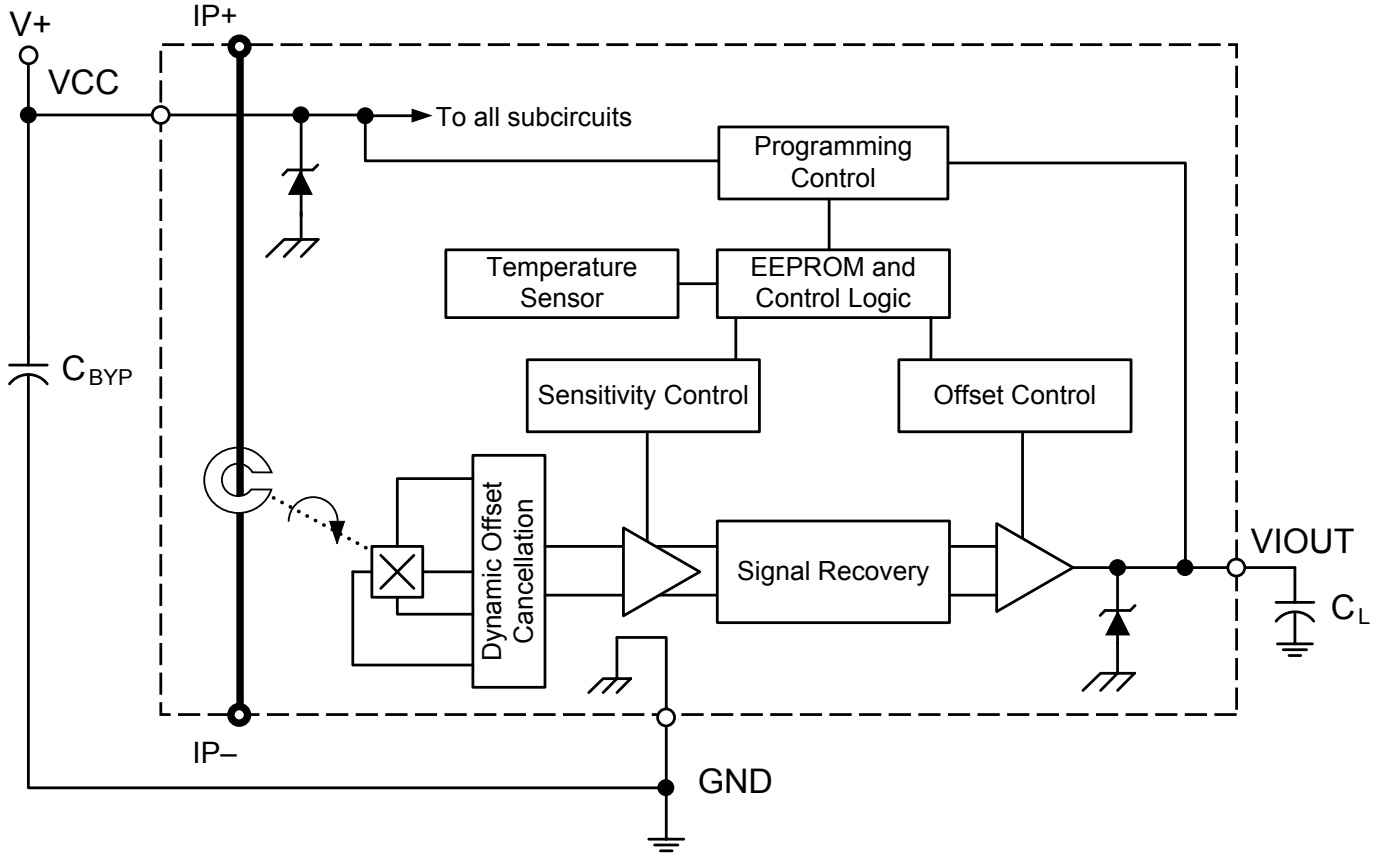
Characteristic	Symbol	Notes	Rating	Unit
Overcurrent	$I_{POC}$	$T_A = 25^{\circ}C$ , 1 second duration, 1% duty cycle	1200	A
		$T_A = 85^{\circ}C$ , 1 second duration, 1% duty cycle	900	A
		$T_A = 150^{\circ}C$ , 1 second duration, 1% duty cycle	600	A

[3] Test was done with Allegro evaluation board. The maximum allowed current is limited by  $T_J(max)$  only.

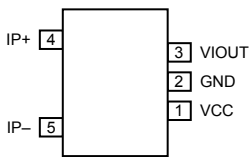
[4] For more overcurrent profiles, please see FAQ on the Allegro website, [www.allegromicro.com](http://www.allegromicro.com).

# ACS770xCB

Thermally Enhanced, Fully Integrated, Hall-Effect-Based  
High-Precision Linear Current Sensor IC with 100  $\mu\Omega$  Current Conductor



Functional Block Diagram



Pinout Diagram

Terminal List Table

Number	Name	Description
1	VCC	Device power supply terminal
2	GND	Signal ground terminal
3	VIOUT	Analog output signal
4	IP+	Terminal for current being sampled
5	IP-	Terminal for current being sampled

**COMMON OPERATING CHARACTERISTICS:** Valid at  $T_{OP} = -40^{\circ}\text{C}$  to  $150^{\circ}\text{C}$ ,  $C_{BYP} = 0.1 \mu\text{F}$ , and  $V_{CC} = 5 \text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Supply Voltage	$V_{CC}$		4.5	5.0	5.5	V
Supply Current	$I_{CC}$	Output open	–	10	15	mA
Supply Zener Voltage	$V_Z$	$T_A = 25^{\circ}\text{C}$ , $I_{CC} = 30 \text{ mA}$	6.5	7.5	–	V
Power-On Delay [1][2]	$t_{POD}$	$T_A = 25^{\circ}\text{C}$ , $C_{BYP} = \text{open}$	–	90	–	$\mu\text{s}$
Temperature Compensation Power-On Time [1]	$t_{TC}$	$T_A = 25^{\circ}\text{C}$ , $C_{BYP} = \text{open}$	–	90	–	$\mu\text{s}$
Undervoltage Lockout (UVLO) Threshold [1]	$V_{UVLOH}$	$T_A = 25^{\circ}\text{C}$ , $V_{CC}$ rising	–	3.8	–	V
	$V_{UVLOL}$	$T_A = 25^{\circ}\text{C}$ , $V_{CC}$ falling	–	3	–	V
UVLO Enable/Disable Delay Time [1][2]	$t_{UVLOE}$	$T_A = 25^{\circ}\text{C}$ , $C_{BYP} = \text{open}$ , $V_{CC}$ Fall Time (5 V to 3 V) = 1 $\mu\text{s}$	–	75	–	$\mu\text{s}$
	$t_{UVLOD}$	$T_A = 25^{\circ}\text{C}$ , $C_{BYP} = \text{Open}$ , $V_{CC}$ Recover Time (3 V to 5 V) = 1 $\mu\text{s}$	–	14	–	$\mu\text{s}$
Power-On Reset Voltage [1]	$V_{PORH}$	$T_A = 25^{\circ}\text{C}$ , $V_{CC}$ rising	–	2.9	–	V
	$V_{PORL}$	$T_A = 25^{\circ}\text{C}$ , $V_{CC}$ falling	–	2.7	–	V
Rise Time [1][2]	$t_r$	$I_P$ step = 60% of $I_{P+}$ , 10% to 90% rise time, $T_A = 25^{\circ}\text{C}$ , $C_L = 0.47 \text{ nF}$	–	4.1	–	$\mu\text{s}$
Propagation Delay Time [1][2]	$t_{PROP}$	$I_P$ step = 60% of $I_{P+}$ , 20% input to 20% output, $T_A = 25^{\circ}\text{C}$ , $C_L = 0.47 \text{ nF}$	–	2.4	–	$\mu\text{s}$
Response Time [1][2]	$t_{RESPONSE}$	$I_P$ step = 60% of $I_{P+}$ , 80% input to 80% output, $T_A = 25^{\circ}\text{C}$ , $C_{OUT} = 0.47 \text{ nF}$	–	4.6	–	$\mu\text{s}$
Internal Bandwidth	$BW_i$	–3 dB; $T_A = 25^{\circ}\text{C}$ , $C_L = 0.47 \text{ nF}$	–	120	–	kHz
Output Load Resistance	$R_L$	VIOUT to GND	4.7	–	–	k $\Omega$
Output Load Capacitance	$C_L$	VIOUT to GND	–	–	10	nF
Primary Conductor Resistance	$R_{PRIMARY}$	$T_A = 25^{\circ}\text{C}$	–	100	–	$\mu\Omega$
Quiescent Output Voltage [1]	$V_{IOUT(QBI)}$	Bidirectional variant, $I_P = 0 \text{ A}$ , $T_A = 25^{\circ}\text{C}$	–	$V_{CC}/2$	–	V
	$V_{IOUT(QUNI)}$	Unidirectional variant, $I_P = 0 \text{ A}$ , $T_A = 25^{\circ}\text{C}$	–	0.5	–	V
Ratiometry [1]	$V_{RAT}$	$V_{CC} = 4.5$ to $5.5 \text{ V}$	–	100	–	%

[1] See Characteristic Definitions section of this datasheet.

[2] See Timing Data Section of this datasheet.

### X050B PERFORMANCE CHARACTERISTICS [1]: $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , $C_{BYP} = 0.1 \mu\text{F}$ , $V_{CC} = 5 \text{ V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Primary Sampled Current	$I_P$		-50	-	50	A
Sensitivity [2]	$Sens_{TA}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	39.04	40	40.96	mV/A
	$Sens_{(TOP)HT}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	39.04	40	40.96	mV/A
	$Sens_{(TOP)LT}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	38.6	40	41.4	mV/A
Sensitivity Drift Over Lifetime [3]	$\Delta Sens_{LIFE}$	$T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-0.72	$\pm 0.24$	0.72	mV/A
Noise [4]	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	-	10	-	mV
Nonlinearity	$E_{LIN}$	Measured using full-scale and half-scale $I_P$	-1	-	1	%
Electrical Offset Voltage [5][6]	$V_{OE(TA)}$	$I_P = 0 \text{ A}$ , $T_A = 25^{\circ}\text{C}$	-10	$\pm 4$	10	mV
	$V_{OE(TOP)HT}$	$I_P = 0 \text{ A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-10	$\pm 6$	10	mV
	$V_{OE(TOP)LT}$	$I_P = 0 \text{ A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-20	$\pm 6$	20	mV
Electrical Offset Voltage Drift Over Lifetime [3]	$\Delta V_{OE(LIFE)}$	$I_P = 0 \text{ A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-5	$\pm 2$	5	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0 \text{ A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 50 A	-	120	300	mA
Total Output Error [7]	$E_{TOT(TA)}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	-2.4	$\pm 0.5$	2.4	%
	$E_{TOT(HT)}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-2.4	$\pm 1.5$	2.4	%
	$E_{TOT(LT)}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-3.5	$\pm 2$	3.5	%
Total Output Error Drift Over Lifetime [3]	$\Delta E_{TOT(LIFE)}$	$T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-1.9	$\pm 0.6$	1.9	%

[1] See Characteristic Performance Data page for parameter distributions over temperature range.

[2] This parameter may drift a maximum of  $\Delta Sens_{LIFE}$  over lifetime.

[3] Based on characterization data obtained during standardized stress test for Qualification of Integrated Circuits, including Package Hysteresis. Cannot be guaranteed. Drift is a function of customer application conditions. Contact Allegro MicroSystems for further information.

[4]  $\pm 3$  sigma noise voltage.

[5] Drift is referred to ideal  $V_{IOUT(QBI)} = 2.5 \text{ V}$ .

[6] This parameter may drift a maximum of  $\Delta V_{OE(LIFE)}$  over lifetime.

[7] This parameter may drift a maximum of  $\Delta E_{TOT(LIFE)}$  over lifetime.

### X050U PERFORMANCE CHARACTERISTICS [1]: $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , $C_{BYP} = 0.1 \mu\text{F}$ , $V_{CC} = 5 \text{V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Primary Sampled Current	$I_P$		0	–	50	A
Sensitivity [2]	$Sens_{TA}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	78.08	80	81.92	mV/A
	$Sens_{(TOP)HT}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	78.08	80	81.92	mV/A
	$Sens_{(TOP)LT}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	77.2	80	82.8	mV/A
Sensitivity Drift Over Lifetime [3]	$\Delta Sens_{LIFE}$	$T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-1.44	$\pm 0.48$	1.44	mV/A
Noise [4]	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	–	20	–	mV
Nonlinearity	$E_{LIN}$	Measured using full-scale and half-scale $I_P$	-1	–	1	%
Electrical Offset Voltage [5][6]	$V_{OE(TA)}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$	-10	$\pm 4$	10	mV
	$V_{OE(TOP)HT}$	$I_P = 0 \text{A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-10	$\pm 6$	10	mV
	$V_{OE(TOP)LT}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-20	$\pm 6$	20	mV
Electrical Offset Voltage Drift Over Lifetime [3]	$\Delta V_{OE(LIFE)}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-5	$\pm 2$	5	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 50 A	–	120	300	mA
Total Output Error [7]	$E_{TOT(TA)}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	-2.4	$\pm 0.5$	2.4	%
	$E_{TOT(HT)}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-2.4	$\pm 1.5$	2.4	%
	$E_{TOT(LT)}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-3.5	$\pm 2$	3.5	%
Total Output Error Drift Over Lifetime [3]	$\Delta E_{TOT(LIFE)}$	$T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-1.9	$\pm 0.6$	1.9	%

[1] See Characteristic Performance Data page for parameter distributions over temperature range.

[2] This parameter may drift a maximum of  $\Delta Sens_{LIFE}$  over lifetime.

[3] Based on characterization data obtained during standardized stress test for Qualification of Integrated Circuits, including Package Hysteresis. Cannot be guaranteed. Drift is a function of customer application conditions. Contact Allegro MicroSystems for further information.

[4]  $\pm 3$  sigma noise voltage.

[5] Drift is referred to ideal  $V_{IOUT(QBI)} = 0.5 \text{V}$ .

[6] This parameter may drift a maximum of  $\Delta V_{OE(LIFE)}$  over lifetime.

[7] This parameter may drift a maximum of  $\Delta E_{TOT(LIFE)}$  over lifetime.

### X100B PERFORMANCE CHARACTERISTICS [1]: $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , $C_{BYP} = 0.1 \mu\text{F}$ , $V_{CC} = 5 \text{V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Primary Sampled Current	$I_P$		-100	-	100	A
Sensitivity [2]	$Sens_{TA}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	19.52	20	20.48	mV/A
	$Sens_{(TOP)HT}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	19.52	20	20.48	mV/A
	$Sens_{(TOP)LT}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	19.3	20	20.7	mV/A
Sensitivity Drift Over Lifetime [3]	$\Delta Sens_{LIFE}$	$T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-0.36	$\pm 0.12$	0.36	mV/A
Noise [4]	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	-	6	-	mV
Nonlinearity	$E_{LIN}$	Measured using full-scale and half-scale $I_P$	-1	-	1	%
Electrical Offset Voltage [5][6]	$V_{OE(TA)}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$	-10	$\pm 4$	10	mV
	$V_{OE(TOP)HT}$	$I_P = 0 \text{A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-10	$\pm 6$	10	mV
	$V_{OE(TOP)LT}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-20	$\pm 6$	20	mV
Electrical Offset Voltage Drift Over Lifetime [3]	$\Delta V_{OE(LIFE)}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-5	$\pm 2$	5	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 100 A	-	170	400	mA
Total Output Error [7]	$E_{TOT(TA)}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	-2.4	$\pm 0.5$	2.4	%
	$E_{TOT(HT)}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-2.4	$\pm 1.5$	2.4	%
	$E_{TOT(LT)}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-3.5	$\pm 2$	3.5	%
Total Output Error Drift Over Lifetime [3]	$\Delta E_{TOT(LIFE)}$	$T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-1.9	$\pm 0.6$	1.9	%

[1] See Characteristic Performance Data page for parameter distributions over temperature range.

[2] This parameter may drift a maximum of  $\Delta Sens_{LIFE}$  over lifetime.

[3] Based on characterization data obtained during standardized stress test for Qualification of Integrated Circuits, including Package Hysteresis. Cannot be guaranteed. Drift is a function of customer application conditions. Contact Allegro MicroSystems for further information.

[4]  $\pm 3$  sigma noise voltage.

[5] Drift is referred to ideal  $V_{IOUT(QB)} = 2.5 \text{V}$ .

[6] This parameter may drift a maximum of  $\Delta V_{OE(LIFE)}$  over lifetime.

[7] This parameter may drift a maximum of  $\Delta E_{TOT(LIFE)}$  over lifetime.

### X100U PERFORMANCE CHARACTERISTICS [1]: $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , $C_{BYP} = 0.1 \mu\text{F}$ , $V_{CC} = 5 \text{V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Primary Sampled Current	$I_P$		0	–	100	A
Sensitivity [2]	$Sens_{TA}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	39.04	40	40.96	mV/A
	$Sens_{(TOP)HT}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	39.04	40	40.96	mV/A
	$Sens_{(TOP)LT}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	38.6	40	41.4	mV/A
Sensitivity Drift Over Lifetime [3]	$\Delta Sens_{LIFE}$	$T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-0.72	$\pm 0.24$	0.72	mV/A
Noise [4]	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	–	12	–	mV
Nonlinearity	$E_{LIN}$	Measured using full-scale and half-scale $I_P$	-1	–	1	%
Electrical Offset Voltage [5][6]	$V_{OE(TA)}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$	-10	$\pm 4$	10	mV
	$V_{OE(TOP)HT}$	$I_P = 0 \text{A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-10	$\pm 6$	10	mV
	$V_{OE(TOP)LT}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-20	$\pm 6$	20	mV
Electrical Offset Voltage Drift Over Lifetime [3]	$\Delta V_{OE(LIFE)}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-5	$\pm 2$	5	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 100 A	–	170	400	mA
Total Output Error [7]	$E_{TOT(TA)}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	-2.4	$\pm 0.5$	2.4	%
	$E_{TOT(HT)}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $150^{\circ}\text{C}$	-2.4	$\pm 1.5$	2.4	%
	$E_{TOT(LT)}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-3.5	$\pm 2$	3.5	%
Total Output Error Drift Over Lifetime [3]	$\Delta E_{TOT(LIFE)}$	$T_{OP} = -40^{\circ}\text{C}$ to $150^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-1.9	$\pm 0.6$	1.9	%

[1] See Characteristic Performance Data page for parameter distributions over temperature range.

[2] This parameter may drift a maximum of  $\Delta Sens_{LIFE}$  over lifetime.

[3] Based on characterization data obtained during standardized stress test for Qualification of Integrated Circuits, including Package Hysteresis. Cannot be guaranteed. Drift is a function of customer application conditions. Contact Allegro MicroSystems for further information.

[4]  $\pm 3$  sigma noise voltage.

[5] Drift is referred to ideal  $V_{IOUT(QB)} = 0.5 \text{V}$ .

[6] This parameter may drift a maximum of  $\Delta V_{OE(LIFE)}$  over lifetime.

[7] This parameter may drift a maximum of  $\Delta E_{TOT(LIFE)}$  over lifetime.

### X150B PERFORMANCE CHARACTERISTICS [1]: $T_{OP} = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ , $C_{BYP} = 0.1 \mu\text{F}$ , $V_{CC} = 5 \text{V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Primary Sampled Current	$I_P$		-150	-	150	A
Sensitivity [2]	$Sens_{TA}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	13.01	13.33	13.65	mV/A
	$Sens_{(TOP)HT}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $125^{\circ}\text{C}$	13.01	13.33	13.65	mV/A
	$Sens_{(TOP)LT}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	12.86	13.33	13.8	mV/A
Sensitivity Drift Over Lifetime [3]	$\Delta Sens_{LIFE}$	$T_{OP} = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-0.24	$\pm 0.08$	0.24	mV/A
Noise [4]	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	-	4	-	mV
Nonlinearity	$E_{LIN}$	Measured using full-scale and half-scale $I_P$	-1	-	1	%
Electrical Offset Voltage [5][6]	$V_{OE(TA)}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$	-10	$\pm 4$	10	mV
	$V_{OE(TOP)HT}$	$I_P = 0 \text{A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $125^{\circ}\text{C}$	-10	$\pm 6$	10	mV
	$V_{OE(TOP)LT}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-20	$\pm 6$	20	mV
Electrical Offset Voltage Drift Over Lifetime [3]	$\Delta V_{OE(LIFE)}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-5	$\pm 2$	5	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 150 A	-	225	400	mA
Total Output Error [7]	$E_{TOT(TA)}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	-2.4	$\pm 0.5$	2.4	%
	$E_{TOT(HT)}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $125^{\circ}\text{C}$	-2.4	$\pm 1.5$	2.4	%
	$E_{TOT(LT)}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-3.5	$\pm 2$	3.5	%
Total Output Error Drift Over Lifetime [3]	$\Delta E_{TOT(LIFE)}$	$T_{OP} = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-1.9	$\pm 0.6$	1.9	%
Symmetry	$E_{SYM}$	Over half-scale of $I_P$	99	100	101	%

[1] See Characteristic Performance Data page for parameter distributions over temperature range.

[2] This parameter may drift a maximum of  $\Delta Sens_{LIFE}$  over lifetime.

[3] Based on characterization data obtained during standardized stress test for Qualification of Integrated Circuits, including Package Hysteresis. Cannot be guaranteed. Drift is a function of customer application conditions. Contact Allegro MicroSystems for further information.

[4]  $\pm 3$  sigma noise voltage.

[5] Drift is referred to ideal  $V_{IOUT(QB)} = 2.5 \text{V}$ .

[6] This parameter may drift a maximum of  $\Delta V_{OE(LIFE)}$  over lifetime.

[7] This parameter may drift a maximum of  $\Delta E_{TOT(LIFE)}$  over lifetime.

### X150U PERFORMANCE CHARACTERISTICS [1]: $T_{OP} = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ , $C_{BYP} = 0.1 \mu\text{F}$ , $V_{CC} = 5 \text{V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Primary Sampled Current	$I_P$		0	–	150	A
Sensitivity [2]	$Sens_{TA}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	26.02	26.66	27.30	mV/A
	$Sens_{(TOP)HT}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $125^{\circ}\text{C}$	26.02	26.66	27.30	mV/A
	$Sens_{(TOP)LT}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	25.73	26.66	27.59	mV/A
Sensitivity Drift Over Lifetime [3]	$\Delta Sens_{LIFE}$	$T_{OP} = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-0.48	$\pm 0.16$	0.48	mV/A
Noise [4]	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	–	6	–	mV
Nonlinearity	$E_{LIN}$	Measured using full-scale and half-scale $I_P$	-1	–	1	%
Electrical Offset Voltage [5][6]	$V_{OE(TA)}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$	-10	$\pm 4$	10	mV
	$V_{OE(TOP)HT}$	$I_P = 0 \text{A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $125^{\circ}\text{C}$	-10	$\pm 6$	10	mV
	$V_{OE(TOP)LT}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-20	$\pm 6$	20	mV
Electrical Offset Voltage Drift Over Lifetime [3]	$\Delta V_{OE(LIFE)}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-5	$\pm 2$	5	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 150 A	–	225	400	mA
Total Output Error [7]	$E_{TOT(TA)}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	-2.4	$\pm 0.5$	2.4	%
	$E_{TOT(HT)}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $125^{\circ}\text{C}$	-2.4	$\pm 1.5$	2.4	%
	$E_{TOT(LT)}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-3.5	$\pm 2$	3.5	%
Total Output Error Drift Over Lifetime [3]	$\Delta E_{TOT(LIFE)}$	$T_{OP} = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-1.9	$\pm 0.6$	1.9	%

[1] See Characteristic Performance Data page for parameter distributions over temperature range.

[2] This parameter may drift a maximum of  $\Delta Sens_{LIFE}$  over lifetime.

[3] Based on characterization data obtained during standardized stress test for Qualification of Integrated Circuits, including Package Hysteresis. Cannot be guaranteed. Drift is a function of customer application conditions. Contact Allegro MicroSystems for further information.

[4]  $\pm 3$  sigma noise voltage.

[5] Drift is referred to ideal  $V_{IOUT(QB)} = 0.5 \text{V}$ .

[6] This parameter may drift a maximum of  $\Delta V_{OE(LIFE)}$  over lifetime.

[7] This parameter may drift a maximum of  $\Delta E_{TOT(LIFE)}$  over lifetime.

### X200B PERFORMANCE CHARACTERISTICS [1]: $T_{OP} = -40^{\circ}\text{C}$ to $85^{\circ}\text{C}$ , $C_{BYP} = 0.1 \mu\text{F}$ , $V_{CC} = 5 \text{V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Primary Sampled Current	$I_P$		-200	-	200	A
Sensitivity [2]	$Sens_{TA}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	9.76	10	10.24	mV/A
	$Sens_{(TOP)HT}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	9.76	10	10.24	mV/A
	$Sens_{(TOP)LT}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	9.65	10	10.35	mV/A
Sensitivity Drift Over Lifetime [3]	$\Delta Sens_{LIFE}$	$T_{OP} = -40^{\circ}\text{C}$ to $85^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-0.18	$\pm 0.06$	0.18	mV/A
Noise [4]	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	-	3	-	mV
Nonlinearity	$E_{LIN}$	Measured using full-scale and half-scale $I_P$	-1	-	1	%
Electrical Offset Voltage [5][6]	$V_{OE(TA)}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$	-10	$\pm 4$	10	mV
	$V_{OE(TOP)HT}$	$I_P = 0 \text{A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	-10	$\pm 6$	10	mV
	$V_{OE(TOP)LT}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-20	$\pm 6$	20	mV
Electrical Offset Voltage Drift Over Lifetime [3]	$\Delta V_{OE(LIFE)}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $85^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-5	$\pm 2$	5	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 200 A	-	250	575	mA
Total Output Error [7]	$E_{TOT(TA)}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	-2.4	$\pm 0.5$	2.4	%
	$E_{TOT(HT)}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	-2.4	$\pm 1.5$	2.4	%
	$E_{TOT(LT)}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-3.5	$\pm 2$	3.5	%
Total Output Error Drift Over Lifetime [3]	$\Delta E_{TOT(LIFE)}$	$T_{OP} = -40^{\circ}\text{C}$ to $85^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-1.9	$\pm 0.6$	1.9	%

[1] See Characteristic Performance Data page for parameter distributions over temperature range.

[2] This parameter may drift a maximum of  $\Delta Sens_{LIFE}$  over lifetime.

[3] Based on characterization data obtained during standardized stress test for Qualification of Integrated Circuits, including Package Hysteresis. Cannot be guaranteed. Drift is a function of customer application conditions. Contact Allegro MicroSystems for further information.

[4]  $\pm 3$  sigma noise voltage.

[5] Drift is referred to ideal  $V_{IOUT(QB)} = 2.5 \text{V}$ .

[6] This parameter may drift a maximum of  $\Delta V_{OE(LIFE)}$  over lifetime.

[7] This parameter may drift a maximum of  $\Delta E_{TOT(LIFE)}$  over lifetime.

### X200U PERFORMANCE CHARACTERISTICS [1]: $T_{OP} = -40^{\circ}\text{C}$ to $85^{\circ}\text{C}$ , $C_{BYP} = 0.1 \mu\text{F}$ , $V_{CC} = 5 \text{V}$ , unless otherwise specified

Characteristic	Symbol	Test Conditions	Min.	Typ.	Max.	Unit
Primary Sampled Current	$I_P$		0	–	200	A
Sensitivity [2]	$Sens_{TA}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	19.52	20	20.48	mV/A
	$Sens_{(TOP)HT}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	19.52	20	20.48	mV/A
	$Sens_{(TOP)LT}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	19.3	20	20.7	mV/A
Sensitivity Drift Over Lifetime [3]	$\Delta Sens_{LIFE}$	$T_{OP} = -40^{\circ}\text{C}$ to $85^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-0.36	$\pm 0.12$	0.36	mV/A
Noise [4]	$V_{NOISE}$	$T_A = 25^{\circ}\text{C}$ , 10 nF on VIOUT pin to GND	–	6	–	mV
Nonlinearity	$E_{LIN}$	Measured using full-scale and half-scale $I_P$	-1	–	1	%
Electrical Offset Voltage [5][6]	$V_{OE(TA)}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$	-10	$\pm 4$	10	mV
	$V_{OE(TOP)HT}$	$I_P = 0 \text{A}$ , $T_{OP} = 25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	-10	$\pm 6$	10	mV
	$V_{OE(TOP)LT}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-20	$\pm 6$	20	mV
Electrical Offset Voltage Drift Over Lifetime [3]	$\Delta V_{OE(LIFE)}$	$I_P = 0 \text{A}$ , $T_{OP} = -40^{\circ}\text{C}$ to $85^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-5	$\pm 2$	5	mV
Magnetic Offset Error	$I_{ERROM}$	$I_P = 0 \text{A}$ , $T_A = 25^{\circ}\text{C}$ , after excursion of 200 A	–	250	575	mA
Total Output Error [7]	$E_{TOT(TA)}$	Measured using full-scale $I_P$ , $T_A = 25^{\circ}\text{C}$	-2.4	$\pm 0.5$	2.4	%
	$E_{TOT(HT)}$	Measured using full-scale $I_P$ , $T_{OP} = 25^{\circ}\text{C}$ to $85^{\circ}\text{C}$	-2.4	$\pm 1.5$	2.4	%
	$E_{TOT(LT)}$	Measured using full-scale $I_P$ , $T_{OP} = -40^{\circ}\text{C}$ to $25^{\circ}\text{C}$	-3.5	$\pm 2$	3.5	%
Total Output Error Drift Over Lifetime [3]	$\Delta E_{TOT(LIFE)}$	$T_{OP} = -40^{\circ}\text{C}$ to $85^{\circ}\text{C}$ , shift after AEC-Q100 grade 0 qualification testing	-1.9	$\pm 0.6$	1.9	%

[1] See Characteristic Performance Data page for parameter distributions over temperature range.

[2] This parameter may drift a maximum of  $\Delta Sens_{LIFE}$  over lifetime.

[3] Based on characterization data obtained during standardized stress test for Qualification of Integrated Circuits, including Package Hysteresis. Cannot be guaranteed. Drift is a function of customer application conditions. Contact Allegro MicroSystems for further information.

[4]  $\pm 3$  sigma noise voltage.

[5] Drift is referred to ideal  $V_{IOUT(QBI)} = 0.5 \text{V}$ .

[6] This parameter may drift a maximum of  $\Delta V_{OE(LIFE)}$  over lifetime.

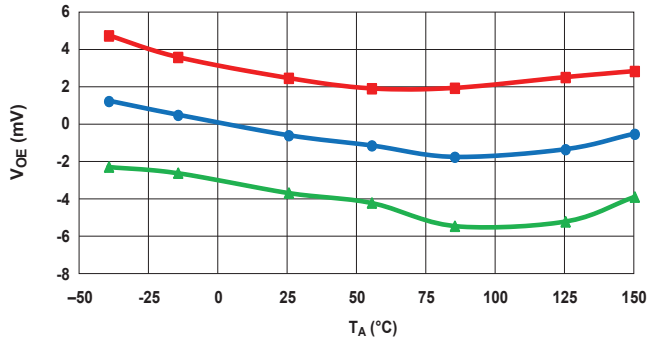
[7] This parameter may drift a maximum of  $\Delta E_{TOT(LIFE)}$  over lifetime.

## CHARACTERISTIC PERFORMANCE DATA

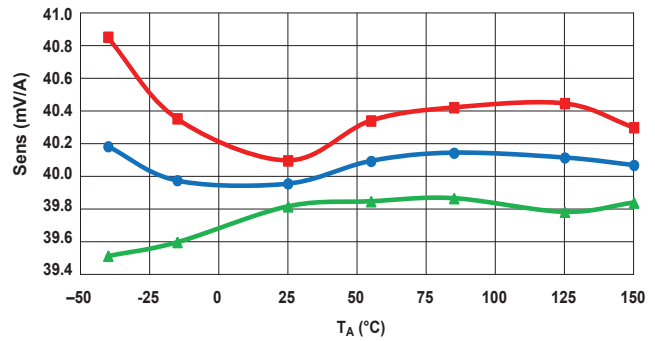
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### Accuracy Data

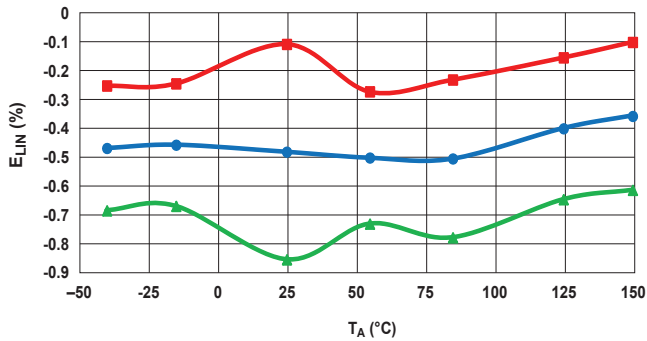
Electrical Offset Voltage versus Ambient Temperature



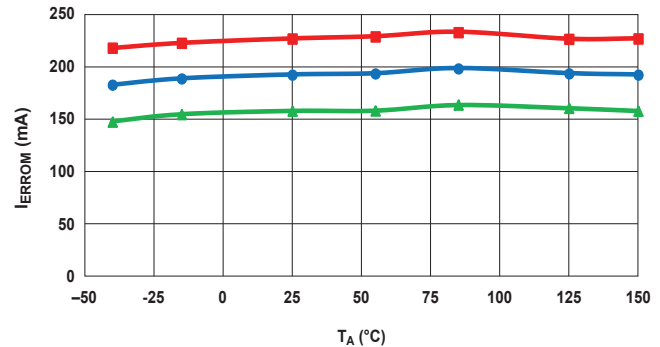
Sensitivity versus Ambient Temperature



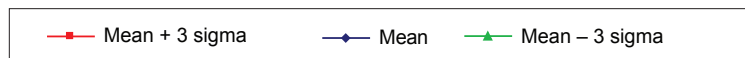
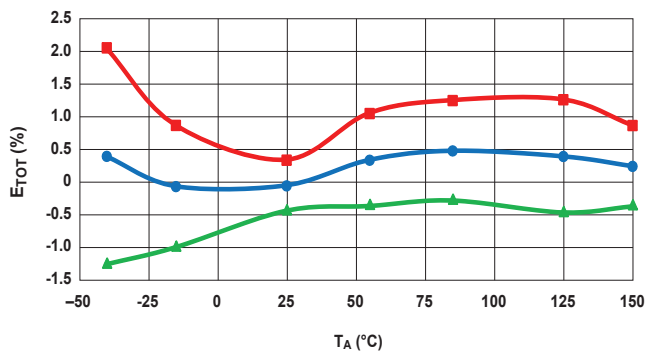
Nonlinearity versus Ambient Temperature



Magnetic Offset Error versus Ambient Temperature



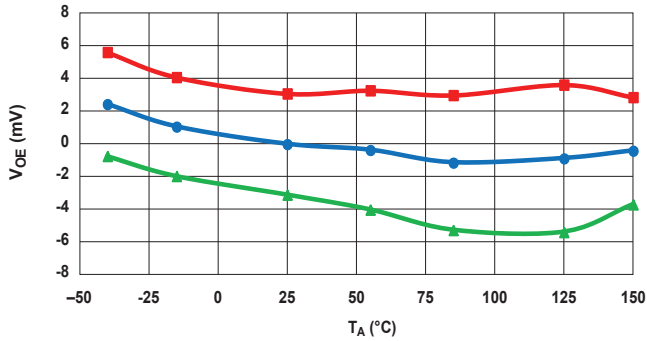
Total Output Error versus Ambient Temperature



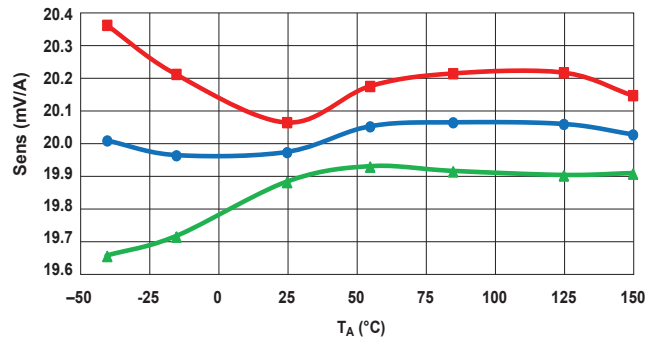
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### Accuracy Data

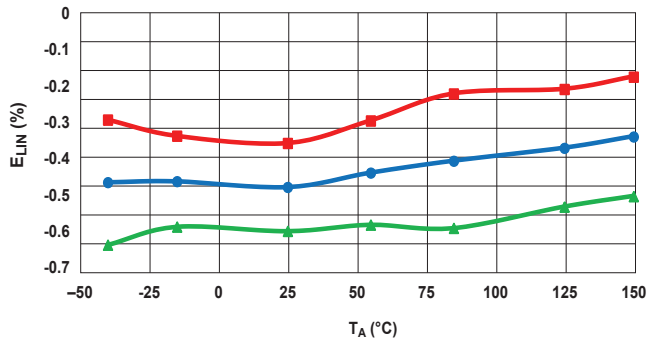
Electrical Offset Voltage versus Ambient Temperature



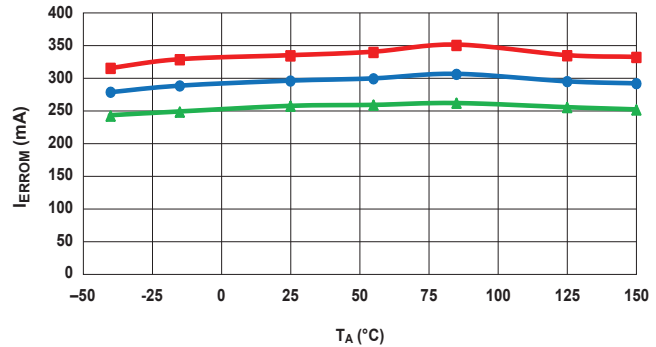
Sensitivity versus Ambient Temperature



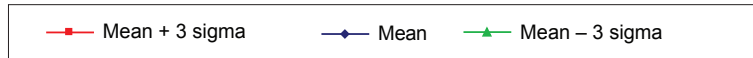
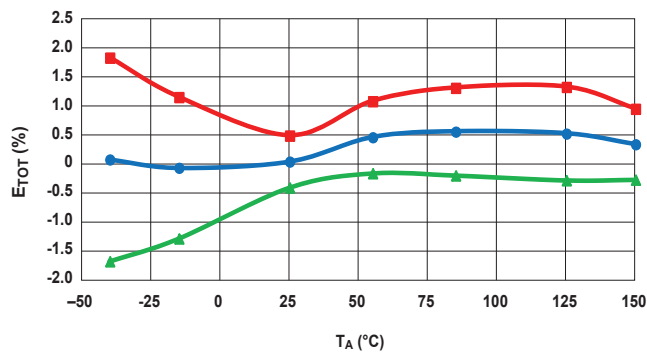
Nonlinearity versus Ambient Temperature



Magnetic Offset Error versus Ambient Temperature



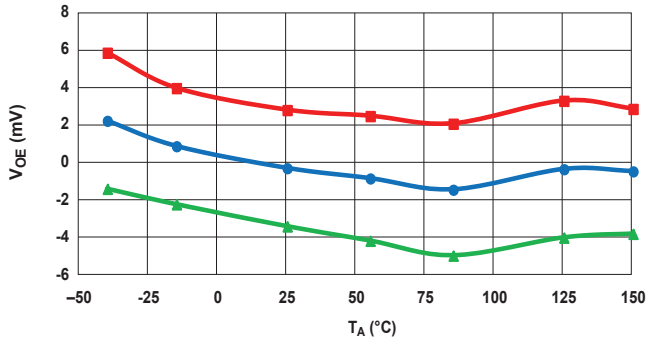
Total Output Error versus Ambient Temperature



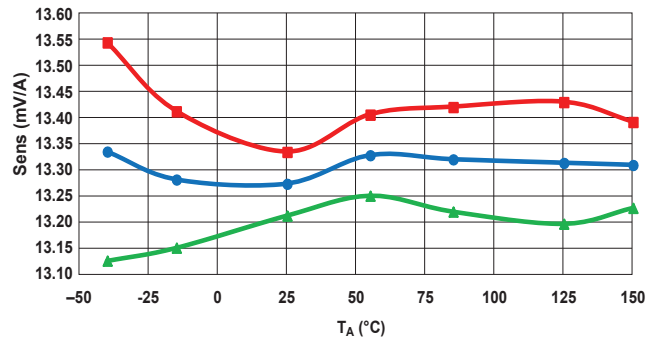
Data Taken using the ACS770KCB-150B

### Accuracy Data

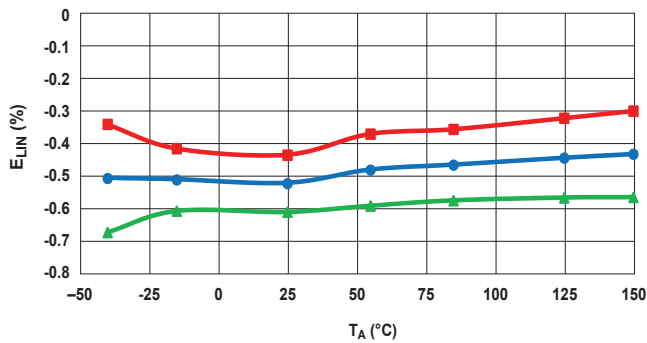
Electrical Offset Voltage versus Ambient Temperature



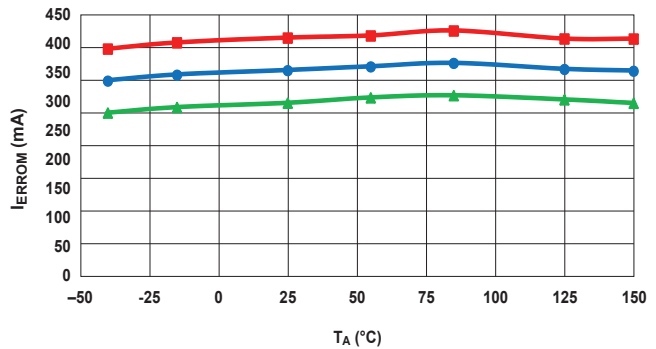
Sensitivity versus Ambient Temperature



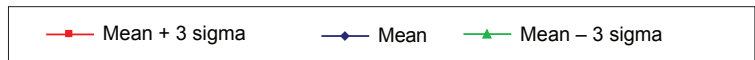
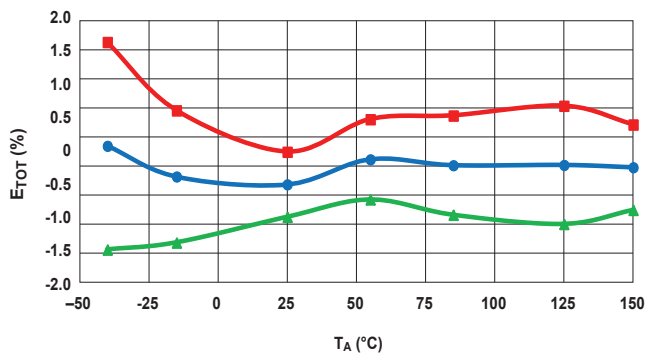
Nonlinearity versus Ambient Temperature



Magnetic Offset Error versus Ambient Temperature



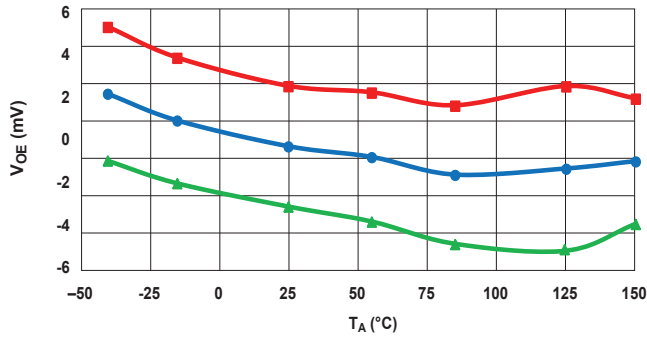
Total Output Error versus Ambient Temperature



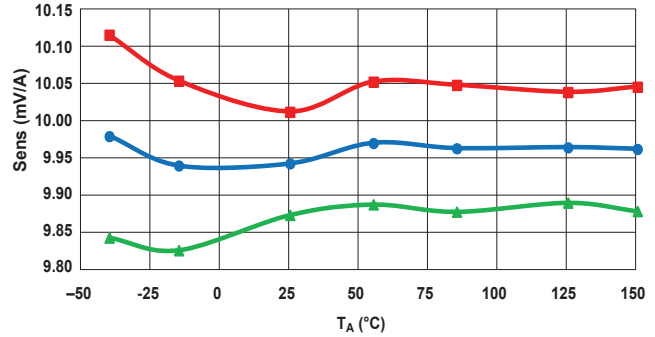
Data Taken using the ACS770ECB-200B

### Accuracy Data

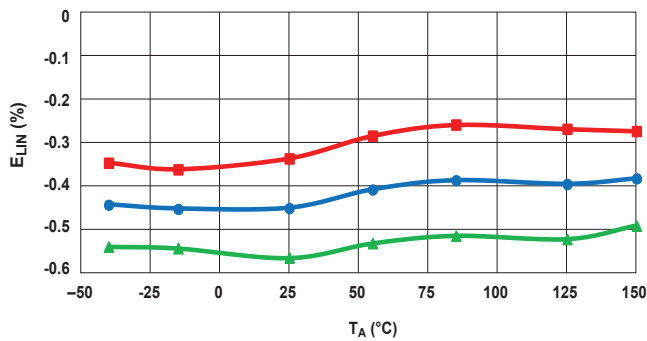
Electrical Offset Voltage versus Ambient Temperature



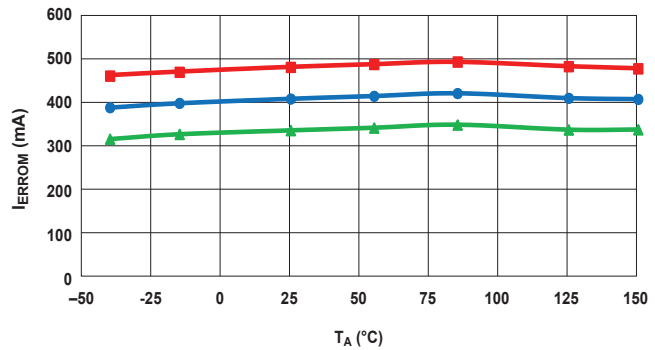
Sensitivity versus Ambient Temperature



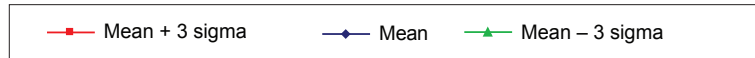
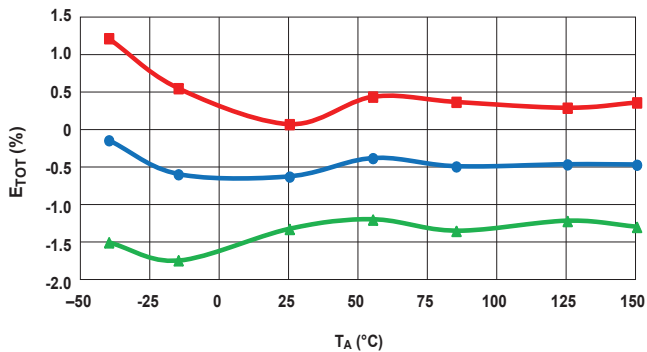
Nonlinearity versus Ambient Temperature



Magnetic Offset Error versus Ambient Temperature

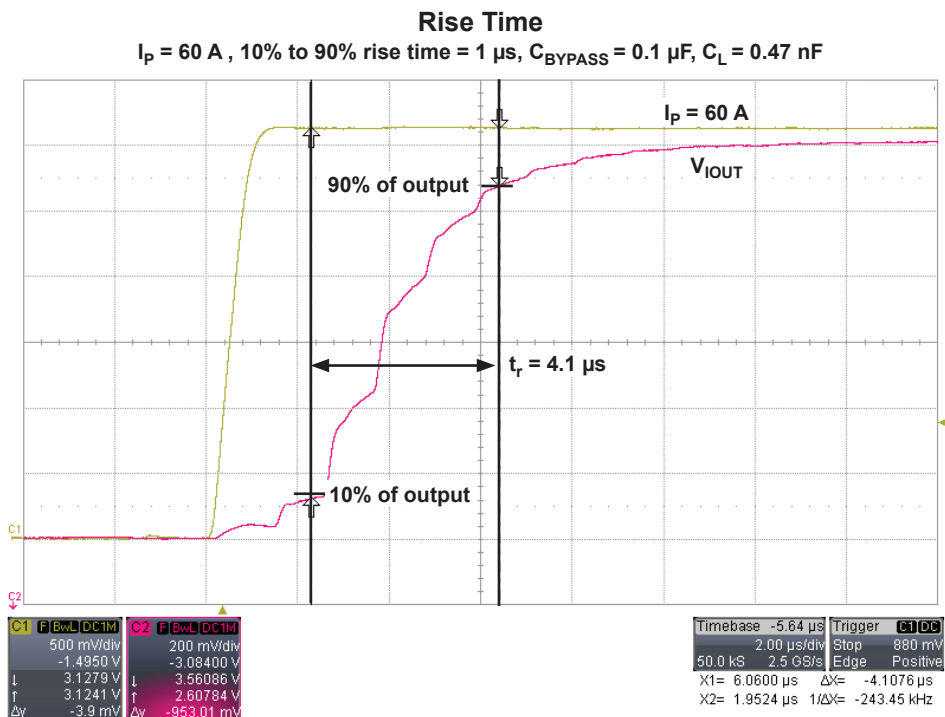
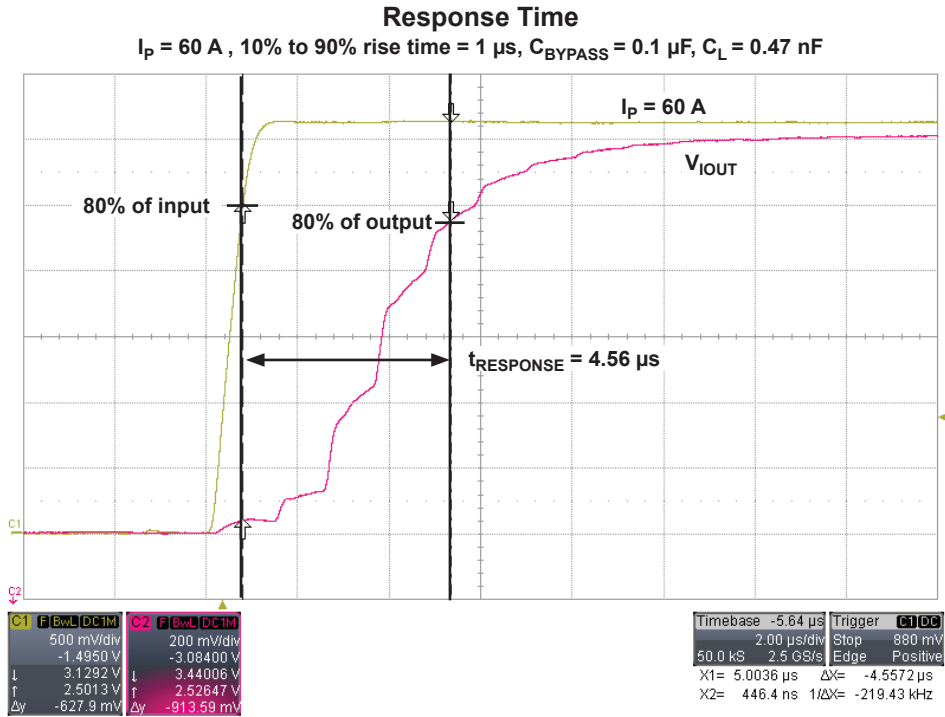


Total Output Error versus Ambient Temperature



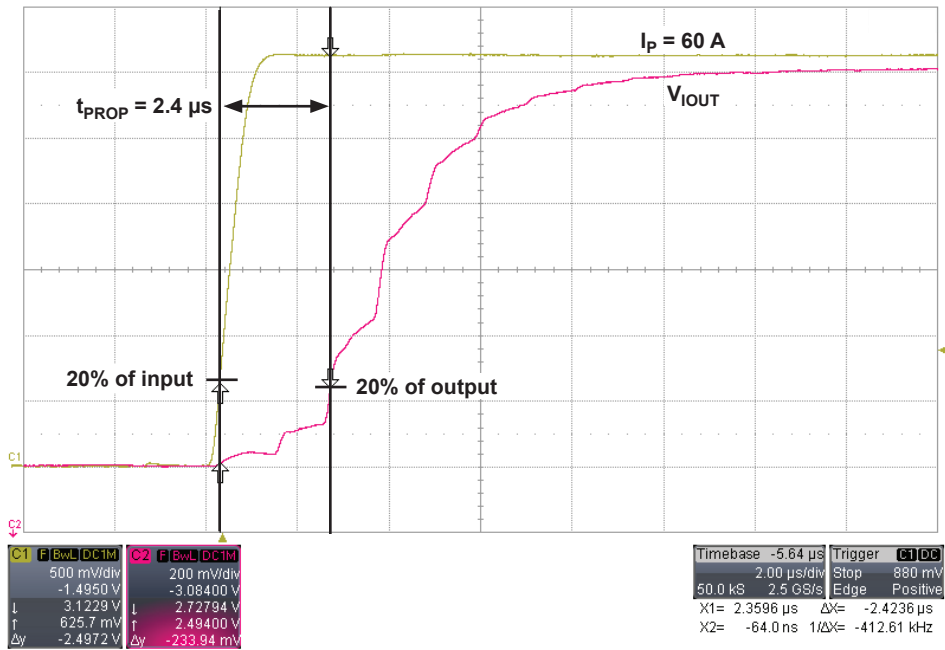
### Data Taken using the ACS770LCB-100B

#### Timing Data



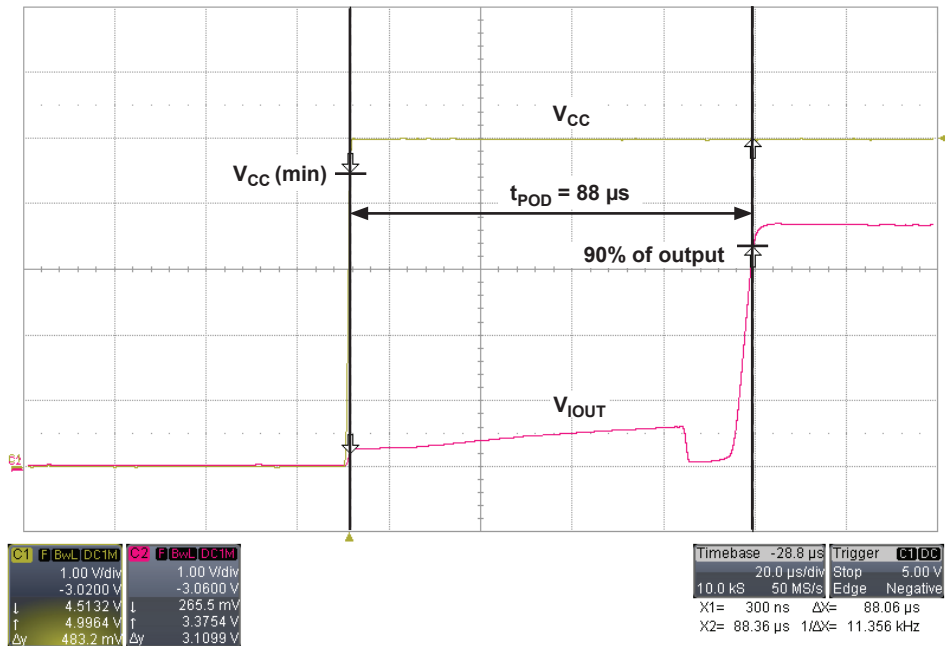
### Propagation Time

$I_p = 60 \text{ A}$ , 10% to 90% rise time = 1  $\mu\text{s}$ ,  $C_{\text{BYPASS}} = 0.1 \mu\text{F}$ ,  $C_L = 0.47 \text{ nF}$

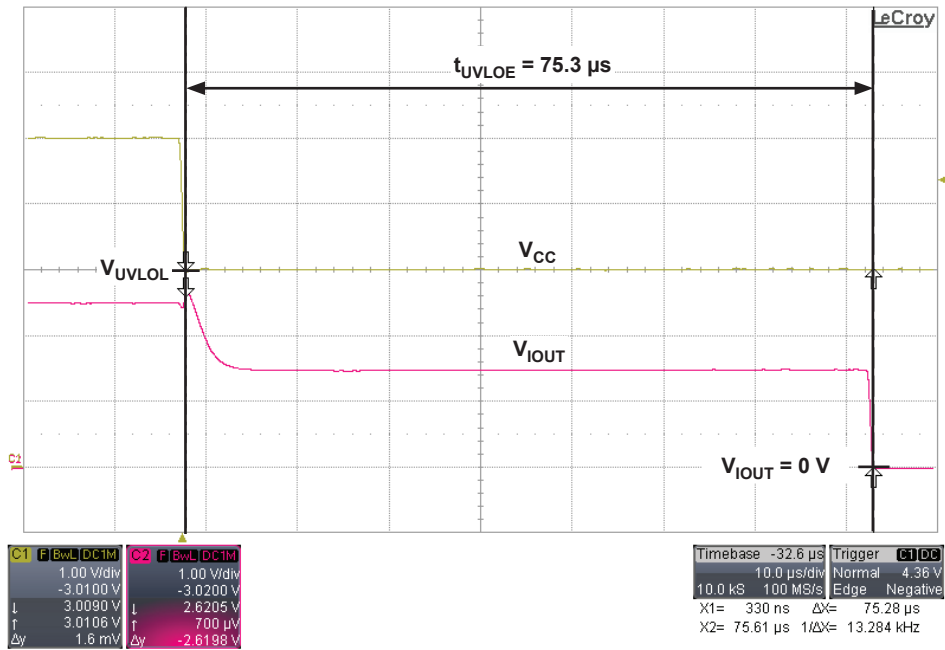


### Power-On Delay

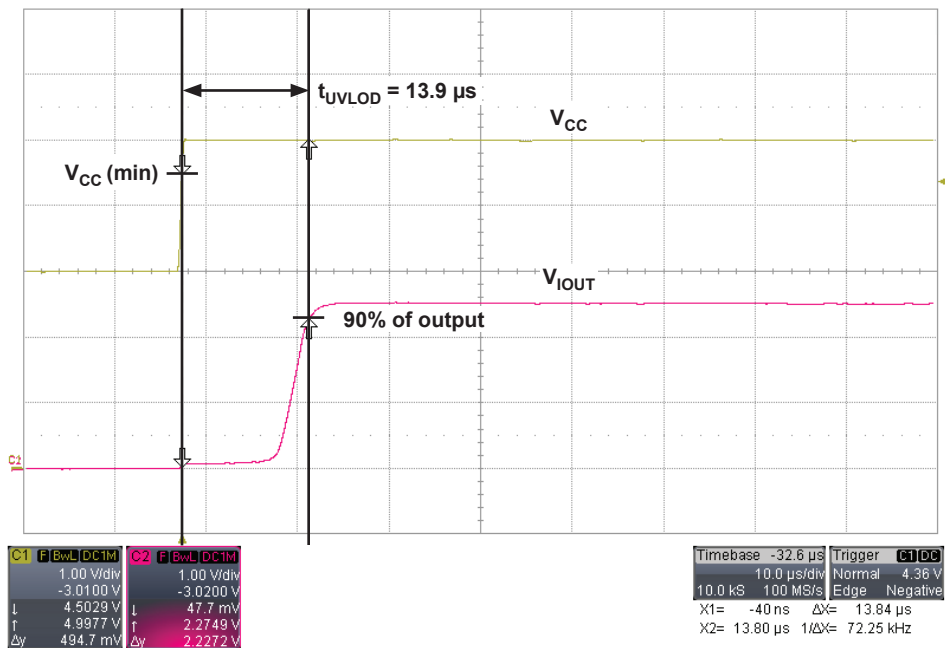
$I_p = 60 \text{ A DC}$ ,  $C_{\text{BYPASS}} = \text{Open}$ ,  $C_L = 0.47 \text{ nF}$



**UVLO Enable Time ( $t_{UVLOE}$ )**  
 $I_P = 0\text{ A}$ ,  $C_{BYPASS} = \text{Open}$ ,  $C_L = \text{Open}$ ,  $V_{CC}$  5 V to 3 V fall time = 1  $\mu\text{s}$



**UVLO Disable Time ( $t_{UVLOD}$ )**  
 $I_P = 0\text{ A}$ ,  $C_{BYPASS} = \text{Open}$ ,  $C_L = \text{Open}$ ,  $V_{CC}$  3 V to 5 V recovery time = 1  $\mu\text{s}$



### CHARACTERISTIC DEFINITIONS

#### Definitions of Accuracy Characteristics

##### SENSITIVITY (Sens)

The change in device output in response to a 1 A change through the primary conductor. The sensitivity is the product of the magnetic circuit sensitivity (G/A) and the linear IC amplifier gain (mV/G). The linear IC amplifier gain is programmed at the factory to optimize the sensitivity (mV/A) for the half-scale current of the device.

$$\Delta\text{Sens}_{(\Delta V)} = \frac{\text{Sens}_{(V_{CC})} / \text{Sens}_{(5V)}}{V_{CC} / 5 V} \times 100 (\%)$$

##### NOISE ( $V_{\text{NOISE}}$ )

The noise floor is derived from the thermal and shot noise observed in Hall elements. Dividing the noise (mV) by the sensitivity (mV/A) provides the smallest current that the device is able to resolve.

##### NONLINEARITY ( $E_{\text{LIN}}$ )

The ACS770 is designed to provide a linear output in response to a ramping current. Consider two current levels: I1 and I2. Ideally, the sensitivity of a device is the same for both currents, for a given supply voltage and temperature. Nonlinearity is present when there is a difference between the sensitivities measured at I1 and I2. Nonlinearity is calculated separately for the positive ( $E_{\text{LINpos}}$ ) and negative ( $E_{\text{LINneg}}$ ) applied currents as follows:

$$E_{\text{LINpos}} = 100 (\%) \times \{1 - (\text{Sens}_{\text{IPOS2}} / \text{Sens}_{\text{IPOS1}})\}$$

$$E_{\text{LINneg}} = 100 (\%) \times \{1 - (\text{Sens}_{\text{INEG2}} / \text{Sens}_{\text{INEG1}})\}$$

where:

$$\text{Sens}_{\text{Ix}} = (V_{\text{IOUT(Ix)}} - V_{\text{IOUT(Q)}}) / \text{Ix}$$

and  $\text{I}_{\text{POSx}}$  and  $\text{I}_{\text{NEGx}}$  are positive and negative currents.

Then:

$$E_{\text{LIN}} = \max(E_{\text{LINpos}}, E_{\text{LINneg}})$$

##### RATIOMETRY

The device features a ratiometric output. This means that the quiescent voltage output,  $V_{\text{IOUTQ}}$ , and the magnetic sensitivity, Sens, are proportional to the supply voltage,  $V_{\text{CC}}$ . The ratiometric change (%) in the quiescent voltage output is defined as:

$$\Delta V_{\text{IOUTQ}(\Delta V)} = \frac{V_{\text{IOUTQ}(V_{CC})} / V_{\text{IOUTQ}(5V)}}{V_{CC} / 5 V} \times 100 (\%)$$

and the ratiometric change (%) in sensitivity is defined as:

##### QUIESCENT OUTPUT VOLTAGE ( $V_{\text{IOUT(Q)}}$ )

The output of the device when the primary current is zero. For bidirectional current flow, it nominally remains at  $V_{\text{CC}}/2$ . Thus,  $V_{\text{CC}} = 5 \text{ V}$  translates into  $V_{\text{IOUT(QBI)}} = 2.5 \text{ V}$ . For unidirectional devices, when  $V_{\text{CC}} = 5 \text{ V}$ ,  $V_{\text{IOUT(QUNI)}} = 0.5 \text{ V}$ . Variation in  $V_{\text{IOUT(Q)}}$  can be attributed to the resolution of the Allegro linear IC quiescent voltage trim, magnetic hysteresis, and thermal drift.

##### ELECTRICAL OFFSET VOLTAGE ( $V_{\text{OE}}$ )

The deviation of the device output from its ideal quiescent value of  $V_{\text{CC}}/2$  for bidirectional sensor ICs and 0.5 V for unidirectional sensor ICs, due to nonmagnetic causes.

##### MAGNETIC OFFSET ERROR ( $I_{\text{ERROM}}$ )

The magnetic offset is due to the residual magnetism (remnant field) of the core material. The magnetic offset error is highest when the magnetic circuit has been saturated, usually when the device has been subjected to a full-scale or high-current overload condition. The magnetic offset is largely dependent on the material used as a flux concentrator.

##### TOTAL OUTPUT ERROR ( $E_{\text{TOT}}$ )

The maximum deviation of the actual output from its ideal value, also referred to as *accuracy*, illustrated graphically in the output voltage versus current chart on the following page.

$E_{\text{TOT}}$  is divided into four areas:

- **0 A at 25°C.** Accuracy at the zero current flow at 25°C, without the effects of temperature.
- **0 A over  $\Delta$  temperature.** Accuracy at the zero current flow including temperature effects.
- **Full-scale current at 25°C.** Accuracy at the full-scale current at 25°C, without the effects of temperature.
- **Full-scale current over  $\Delta$  temperature.** Accuracy at the full-scale current flow including temperature effects.

$$E_{\text{TOT(IP)}} = \frac{V_{\text{IOUT(IP)}} - V_{\text{IOUT\_IDEAL(IP)}}}{\text{Sens}_{\text{IDEAL}} \times I_p} \times 100 (\%)$$

where

$$V_{\text{IOUT\_IDEAL(IP)}} = V_{\text{IOUT(Q)}} + (\text{Sens}_{\text{IDEAL}} \times I_p)$$

### Definitions of Dynamic Response Characteristics

#### POWER-ON DELAY ( $t_{POD}$ )

When the supply is ramped to its operating voltage, the device requires a finite time to power its internal components before responding to an input magnetic field.

Power-On Delay,  $t_{POD}$ , is defined as the time it takes for the output voltage to settle within  $\pm 10\%$  of its steady-state value under an applied magnetic field, after the power supply has reached its minimum specified operating voltage,  $V_{CC(min)}$ , as shown in the chart at right.

#### TEMPERATURE COMPENSATION POWER-ON TIME ( $t_{TC}$ )

After Power-On Delay,  $t_{POD}$ , elapses,  $t_{TC}$  also is required before a valid temperature compensated output.

#### RISE TIME ( $t_r$ )

The time interval between a) when the device reaches 10% of its full-scale value, and b) when it reaches 90% of its full-scale value. Both  $t_r$  and  $t_{RESPONSE}$  are detrimentally affected by eddy current losses observed in the conductive IC ground plane.

#### RESPONSE TIME ( $t_{RESPONSE}$ )

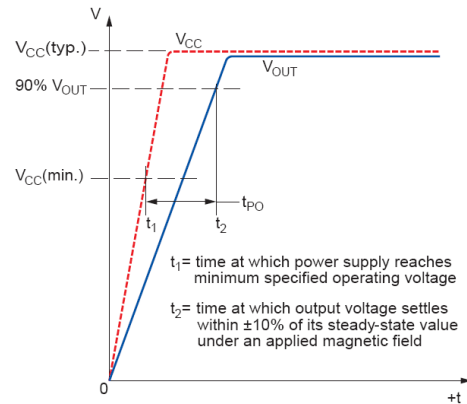
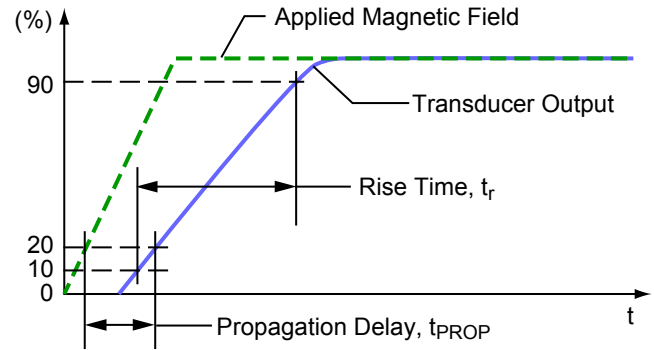
The time interval between a) when the applied current reaches 80% of its final value, and b) when the sensor reaches 80% of its output corresponding to the applied current.

#### PROPAGATION DELAY ( $t_{PROP}$ )

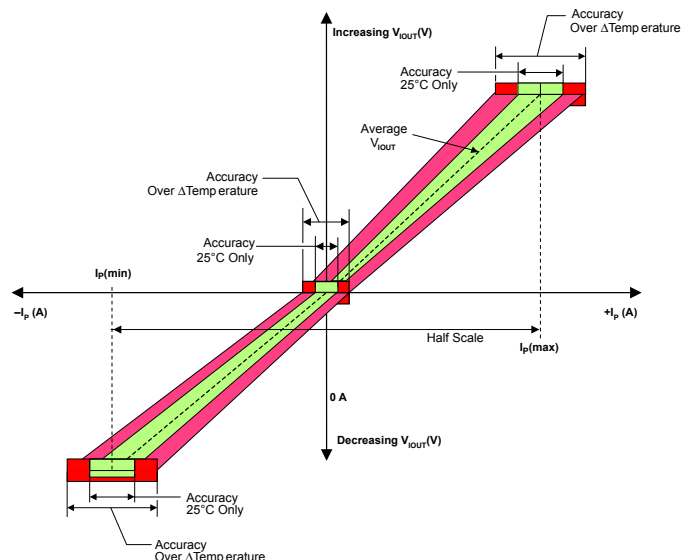
The time interval between a) when the input current reaches 20% of its final value, and b) when the output reaches 20% of its final value.

#### POWER-ON RESET VOLTAGE ( $V_{POR}$ )

At power-up, to initialize to a known state and avoid current spikes, the ACS770 is held in Reset state. The Reset signal is disabled when  $V_{CC}$  reaches  $V_{UVLOH}$  and time  $t_{PORR}$  has elapsed, allowing output voltage to go from a high-impedance state into normal operation. During power-down, the Reset signal is enabled when  $V_{CC}$  reaches  $V_{PORL}$ , causing output voltage to go into a high-impedance state. (Note that a detailed description of POR and UVLO operation can be found in the Functional Description section.)



**Output Voltage versus Sampled Current**  
**Total Output Error at 0 A and at Full-Scale Current**



### POWER-ON RESET RELEASE TIME ( $t_{PORR}$ )

When  $V_{CC}$  rises to  $V_{PORH}$ , the Power-On Reset Counter starts. The ACS770 output voltage will transition from a high-impedance state to normal operation only when the Power-On Reset Counter has reached  $t_{PORR}$  and  $V_{CC}$  has exceeded  $V_{UVLOH}$ .

### UNDERVOLTAGE LOCKOUT THRESHOLD ( $V_{UVLO}$ )

If  $V_{CC}$  drops below  $V_{UVLOL}$ , output voltage will be locked to GND. If  $V_{CC}$  starts rising, the ACS770 will come out of the locked state when  $V_{CC}$  reaches  $V_{UVLOH}$ .

### SYMMETRY ( $E_{SYM}$ )

The degree to which the absolute voltage output from the IC varies in proportion to either a positive or negative half-scale primary current. The following equation is used to derive symmetry:

$$100 \left( \frac{V_{IOUT_{+half-scale\ amperes}} - V_{IOUT(Q)}}{V_{IOUT(Q)} - V_{IOUT_{-half-scale\ amperes}}} \right)$$

### UVLO ENABLE/DISABLE RELEASE TIME ( $t_{UVLO}$ )

When a falling  $V_{CC}$  reaches  $V_{UVLOL}$ , time  $t_{UVLOE}$  is required to engage Undervoltage Lockout state. When  $V_{CC}$  rises above  $V_{UVLOH}$ , time  $t_{UVLOD}$  is required to disable UVLO and have a valid output voltage.

## FUNCTIONAL DESCRIPTION

### Power-On Reset (POR) and Undervoltage Lock-Out (UVLO) Operation

The descriptions in this section assume:

$$\text{Temperature} = 25^{\circ}\text{C},$$

$$V_{CC} = 5\text{ V},$$

no output load, and  
no significant current flow through the sensor IC.

Voltage levels shown are specific to a bidirectional ACS770; however, the POR and UVLO functionality described also applies to unidirectional sensors.

The reference numbers section refer to figures 1 and 2.

#### Power-Up

At power-up, as  $V_{CC}$  ramps up, the output is in a high-impedance state. When  $V_{CC}$  crosses  $V_{PORH}$  (location [1] in figure 1 and [1'] in figure 2), the POR Release counter starts counting for  $t_{PORR}$ . At this point, if  $V_{CC}$  exceeds  $V_{UVLOH}$  [2'], the output will go to  $V_{CC}/2$  after  $t_{UVLOD}$  [3']. If  $V_{CC}$  does not exceed  $V_{UVLOH}$  [2], the output will stay in the high-impedance state until  $V_{CC}$  reaches  $V_{UVLOH}$  [3] and then will go to  $V_{CC}/2$  after  $t_{UVLOD}$  [4].

### EEPROM Error Checking And Correction

Hamming code methodology is implemented for EEPROM checking and correction. The device has ECC enabled after power-up. If an uncorrectable error has occurred, the VOUT pin will go to high impedance and the device will not respond to applied magnetic field.

#### $V_{CC}$ drops below $V_{CC}(\text{min}) = 4.5\text{ V}$

If  $V_{CC}$  drops below  $V_{UVLOL}$  [4', 5], the UVLO Enable Counter starts counting. If  $V_{CC}$  is still below  $V_{UVLOL}$  when the counter reaches  $t_{UVLOE}$ , the UVLO function will be enabled and the output will be pulled near GND [6]. If  $V_{CC}$  exceeds  $V_{UVLOL}$  before the UVLO Enable Counter reaches  $t_{UVLOE}$  [5'], the output will continue to be  $V_{CC}/2$ .

#### Coming Out of UVLO

While UVLO is enabled [6], if  $V_{CC}$  exceeds  $V_{UVLOH}$  [7], UVLO will be disabled after  $t_{UVLOD}$ , and the output will be  $V_{CC}/2$  [8].

#### Power-Down

As  $V_{CC}$  ramps down below  $V_{UVLOL}$  [6', 9], the UVLO Enable Counter will start counting. If  $V_{CC}$  is higher than  $V_{PORL}$  when the counter reaches  $t_{UVLOE}$ , the UVLO function will be enabled and the output will be pulled near GND [10]. The output will enter a high-impedance state as  $V_{CC}$  goes below  $V_{PORL}$  [11]. If  $V_{CC}$  falls below  $V_{PORL}$  before the UVLO Enable Counter reaches  $t_{UVLOE}$ , the output will transition directly into a high-impedance state [7'].

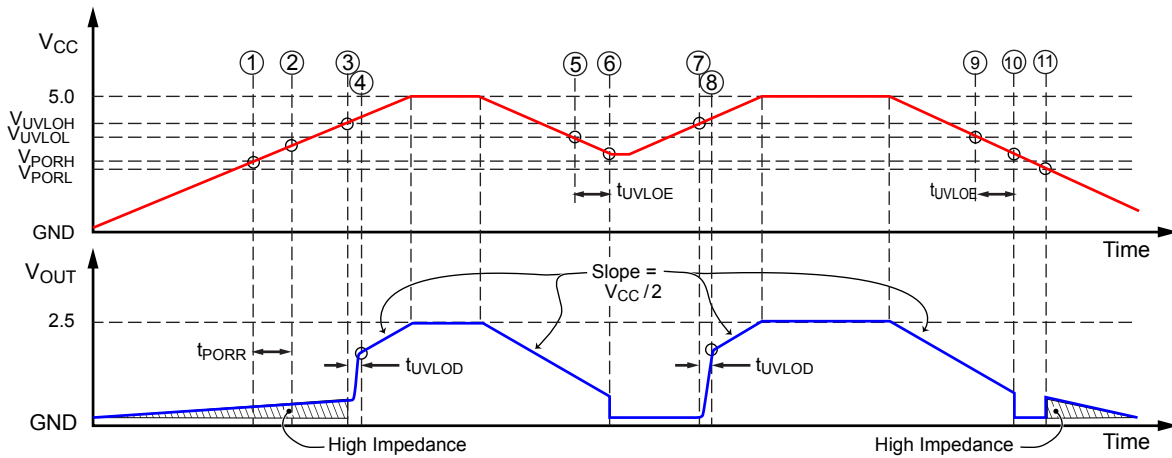


Figure 1: POR and UVLO Operation: Slow Rise Time Case

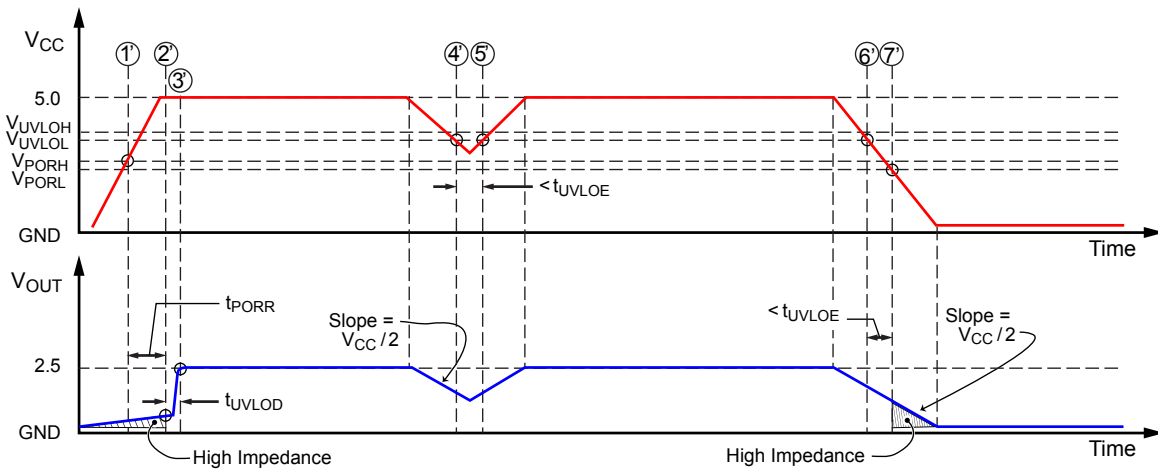


Figure 2: POR and UVLO Operation: Fast Rise Time Case

### Chopper Stabilization Technique

When using Hall-effect technology, a limiting factor for switchpoint accuracy is the small signal voltage developed across the Hall element. This voltage is disproportionately small relative to the offset that can be produced at the output of the Hall sensor IC. This makes it difficult to process the signal while maintaining an accurate, reliable output over the specified operating temperature and voltage ranges.

Chopper stabilization is a unique approach used to minimize Hall offset on the chip. Allegro employs a technique to remove key sources of the output drift induced by thermal and mechanical stresses. This offset reduction technique is based on a signal modulation-demodulation process. The undesired offset signal is separated from the magnetic field-induced signal in the frequency domain, through modulation. The subsequent demodulation acts as a modulation process for the offset, causing the magnetic field-induced signal to recover its original spectrum at baseband, while the DC offset becomes a high-frequency signal. The magnetic-

sourced signal then can pass through a low-pass filter, while the modulated DC offset is suppressed.

In addition to the removal of the thermal and stress related offset, this novel technique also reduces the amount of thermal noise in the Hall sensor IC while completely removing the modulated residue resulting from the chopper operation. The chopper stabilization technique uses a high-frequency sampling clock. For demodulation process, a sample-and-hold technique is used. This high-frequency operation allows a greater sampling rate, which results in higher accuracy and faster signal-processing capability. This approach desensitizes the chip to the effects of thermal and mechanical stresses, and produces devices that have extremely stable quiescent Hall output voltages and precise recoverability after temperature cycling. This technique is made possible through the use of a BiCMOS process, which allows the use of low-offset, low-noise amplifiers in combination with high-density logic integration and sample-and-hold circuits.

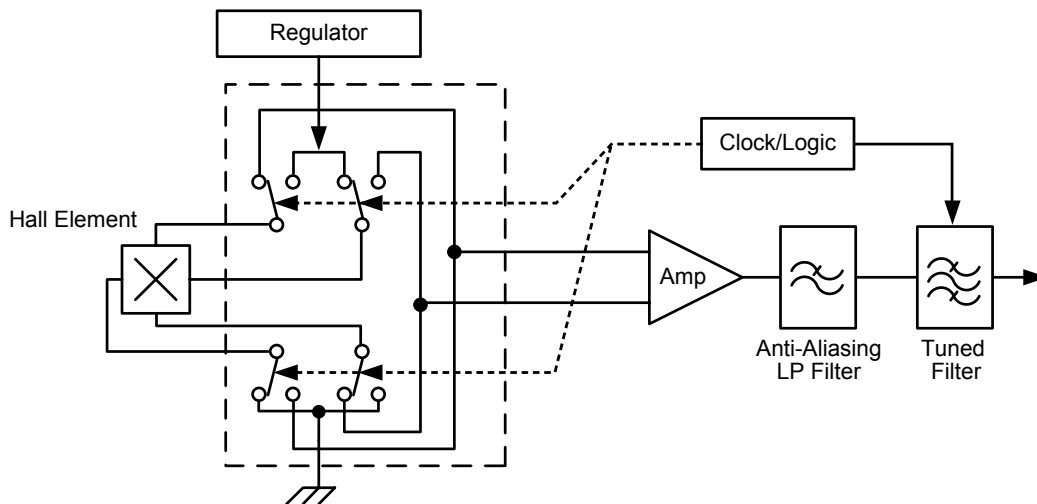


Figure 3: Concept of Chopper Stabilization Technique

### PACKAGE OUTLINE DRAWINGS

#### For Reference Only – Not for Tooling Use

(Reference DWG-9111 & DWG-9110)  
 Dimensions in millimeters – NOT TO SCALE  
 Dimensions exclusive of mold flash, gate burrs, and dambar protrusions  
 Exact case and lead configuration at supplier discretion within limits shown

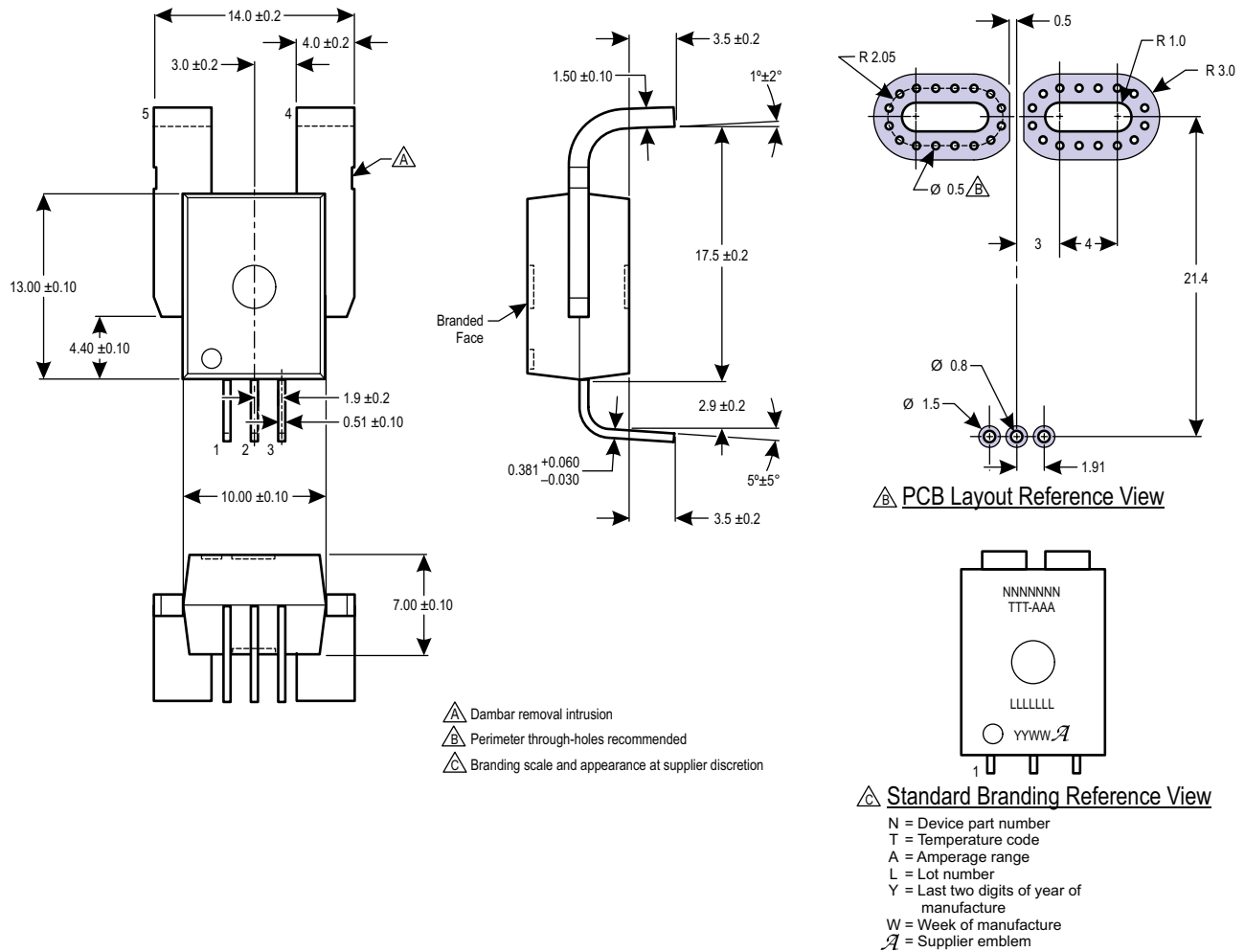
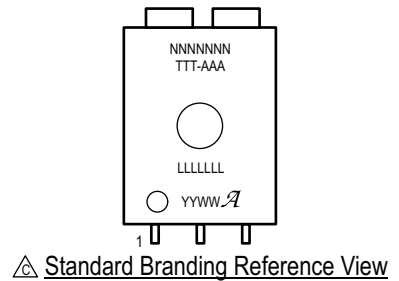
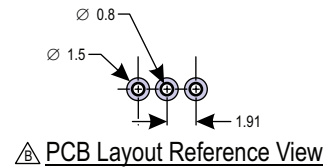
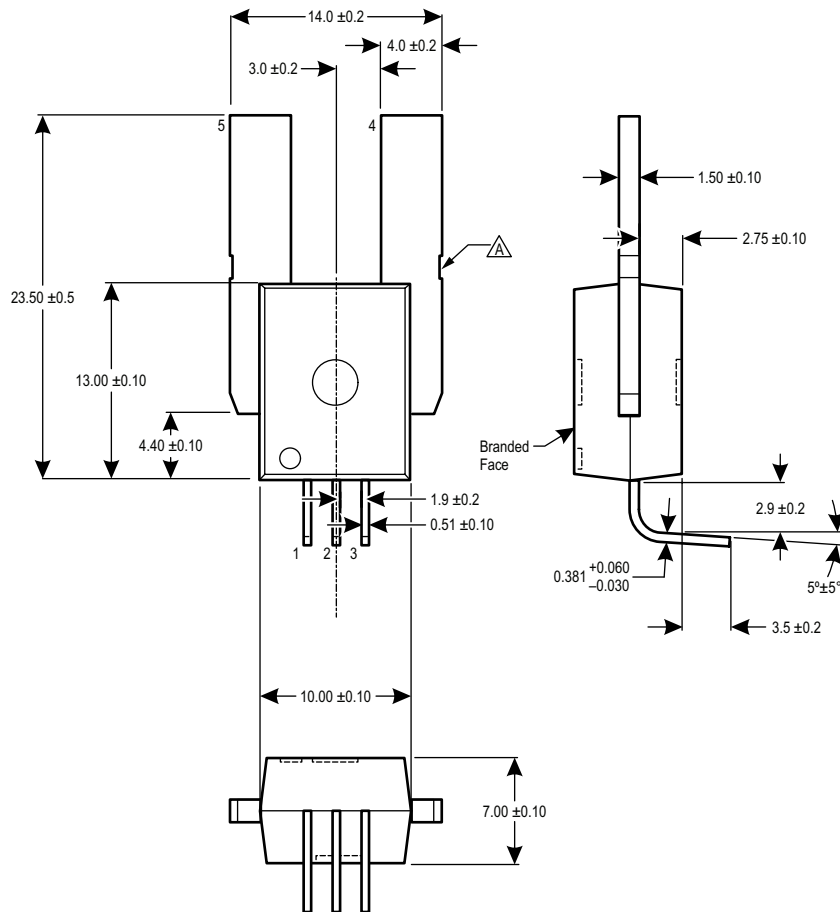


Figure 4: Package CB, 5-Pin, Leadform PFF

### For Reference Only – Not for Tooling Use

(Reference DWG-9111, DWG-9110)  
 Dimensions in millimeters – NOT TO SCALE  
 Dimensions exclusive of mold flash, gate burs, and dambar protrusions  
 Exact case and lead configuration at supplier discretion within limits shown



N = Device part number  
 T = Temperature code  
 A = Amperage range  
 L = Lot number  
 Y = Last two digits of year of manufacture  
 W = Week of manufacture  
 A = Supplier emblem

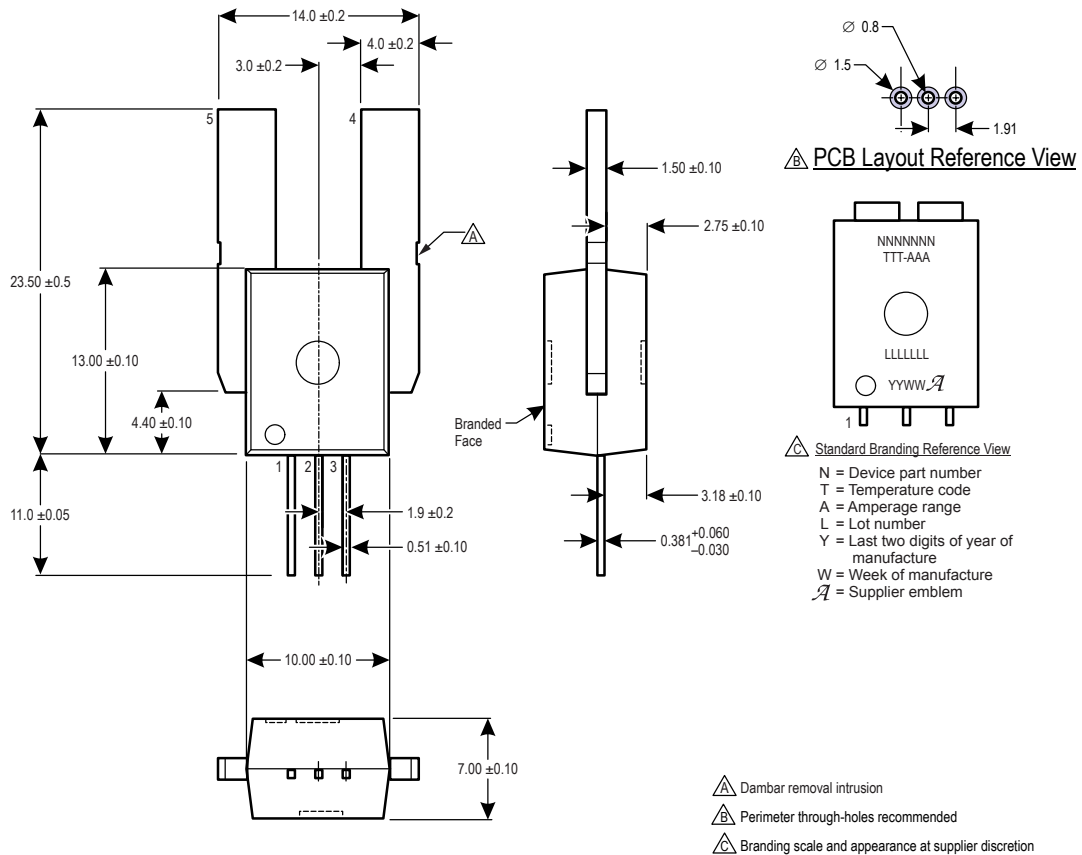
- Dambar removal intrusion
- Perimeter through-holes recommended
- Branding scale and appearance at supplier discretion

Creepage distance, current terminals to signal pins: 7.25 mm  
 Clearance distance, current terminals to signal pins: 7.25 mm  
 Package mass: 4.63 g typical

**Figure 5: Package CB, 5-Pin, Leadform PSF**

### For Reference Only – Not for Tooling Use

(Reference DWG-9111, DWG-9110)  
 Dimensions in millimeters – NOT TO SCALE  
 Dimensions exclusive of mold flash, gate burrs, and dambar protrusions  
 Exact case and lead configuration at supplier discretion within limits shown



Creepage distance, current terminals to signal pins: 7.25 mm  
 Clearance distance, current terminals to signal pins: 7.25 mm  
 Package mass: 4.63 g typical

Figure 6: Package CB, 5-Pin, Leadform PSS

### Revision History

Number	Date	Description
1	December 8, 2014	Revised Selection Guide
2	January 20, 2015	Revised $V_{PORH}$ Typical Value
3	March 11, 2015	Revised $V_{RCC}$ , $V_{RIOUT}$ , $I_{OUT(SOURCE)}$ , $I_{ERROM}$ (100 A and 150 A) values, and added Symmetry to X150B PERFORMANCE CHARACTERISTICS table
4	April 8, 2015	Updated TUV certification
5	November 2, 2016	Updated PCB Layout Reference View in Package Outline Drawing on page 27
6	June 5, 2017	Updated status of ACS770LCB-100U-PSF-T part variant to Last-Time Buy
7	November 16, 2017	Added PSS leadform
8	January 30, 2018	Added Dielectric Surge Strength Test Voltage characteristic (page 3) and EEPROM Error Checking and Correction section (page 24)
9	April 18, 2018	Corrected TUV certification
10	November 6, 2018	Added UL certificate number and minor editorial updates
11	May 28, 2019	Updated TUV certificate mark
12	October 12, 2021	Added ACS770LCB-100B-PSF-T part variant (page 2), removed Last-Time Buy ACS770LCB-100U-PSF-T variant from selection guide (page 2), removed footnote [3] from selection guide (page 2); removed Pb-free symbol (page 2)
13	July 10, 2023	Updated Isolation Characteristics table (page 3); removed footnote [1] and [2] from selection guide (page 2); removed ACS770KCB-150B-PSF-T part variant (all pages)

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