



**THE DATASHEET OF  
HCPL-063L-060**



## HCPL-260L/060L/263L/063L

### High-Speed LVTTTL-Compatible 3.3V Optocouplers

#### Description

The Broadcom<sup>®</sup> HCPL-260L/060L/263L/063L are optically coupled gates that combine a GaAsP light-emitting diode and an integrated high gain photo detector. An enable input allows the detector to be strobed. The output of the detector IC is an open collector Schottky-clamped transistor. The internal shield provides a guaranteed common mode transient immunity specification of 15 kV/ $\mu$ s at 3.3V.

This unique design provides maximum AC and DC circuit isolation while achieving LVTTTL/LVCMOS compatibility. The optocoupler AC and DC operational parameters are guaranteed from  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  allowing trouble-free system performance.

These optocouplers are suitable for high-speed logic interfacing, input/output buffering, as line receivers in environments that conventional line receivers cannot tolerate and are recommended for use in extremely high ground or induced noise environments.

#### Features

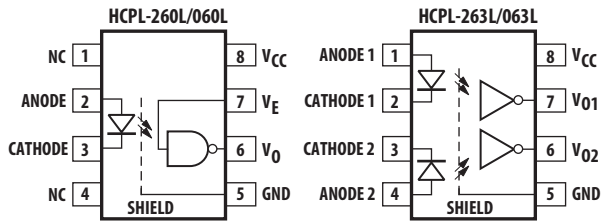
- 3.3V/5V dual-supply voltages
- Low power consumption
- 15 kV/ $\mu$ s minimum Common Mode Rejection (CMR) at  $V_{\text{CM}} = 1000\text{V}$
- High speed: 15 MBd typical
- LVTTTL/LVCMOS compatible
- Low input current capability: 5 mA
- Guaranteed AC and DC performance over temperature:  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$
- Available in 8-pin DIP, SOIC-8
- Strobeable output (single-channel products only)
- Safety approvals: UL, CSA, IEC/EN/DIN EN 60747-5-5

#### Applications

- Isolated line receiver
- Computer-peripheral interfaces
- Microprocessor system interfaces
- Digital isolation for A/D, D/A conversion
- Switching power supply
- Instrument input/output isolation
- Ground loop elimination
- Pulse transformer replacement
- Field buses

**CAUTION!** Take normal static precautions in handling and assembly of this component to prevent damage, degradation, or both that may be induced by ESD. The components featured in this data sheet are not to be used in military or aerospace applications or environments.

# Functional Diagram



**Truth Table  
(Positive Logic)**

LED	Enable	Output
ON	H	L
OFF	H	H
ON	L	H
OFF	L	H
ON	NC	L
OFF	NC	H

**Truth Table  
(Positive Logic)**

LED	Output
ON	L
OFF	H

# Ordering Information

HCPL-xxxx is UL Recognized with 3750 Vrms for 1 minute per UL1577.

Part Number	Option		Package	Surface Mount	Gull Wing	Tape and Reel	UL 5000 Vrms/1 Minute Rating	IEC/EN/DIN EN 60747-5-5	Quantity
	RoHS Compliant	Non-RoHS Compliant							
HCPL-260L	-000E	No option	300-mil DIP-8						50 per tube
	-300E	-300		X	X				50 per tube
	-500E	#500		X	X	X			1000 per reel
	-020E	-020					X		50 per tube
	-320E	-320		X	X		X		50 per tube
	-520E	-520		X	X	X	X		1000 per reel
	-060E	#060						X	50 per tube
	-560E	#560		X	X	X		X	1000 per reel
HCPL-263L	-000E	No option	300-mil DIP-8						50 per tube
	-300E	#300		X	X				50 per tube
	-500E	#500		X	X	X			1000 per reel
	-020E	#020					X		50 per tube
	-320E	-320		X	X		X		50 per tube
	-520E	#520		X	X	X	X		1000 per reel
	-060E	-060						X	50 per tube
	-560E	-560		X	X	X		X	1000 per reel
HCPL-060L	-000E	No option	SO-8	X					100 per tube
	-500E	#500		X		X			1500 per reel
	-060E	#060		X				X	100 per tube
	-560E	-560		X		X		X	1500 per reel
HCPL-063L	-000E	No option	SO-8	X					100 per tube
	-500E	#500		X		X			1500 per reel
	-060E	-060		X				X	100 per tube
	-560E	-560		X		X		X	1500 per reel

To order, choose a part number from the part number column and combine with the desired option from the option column to form an order entry. Combination of Option 020 and Option 060 is not available.

**Example 1:**

HCPL-260L-560E to order product of 300-mil DIP Gull Wing Surface Mount package in Tape and Reel packaging with IEC/EN/DIN EN 60747-5-5 Safety Approval in RoHS compliant.

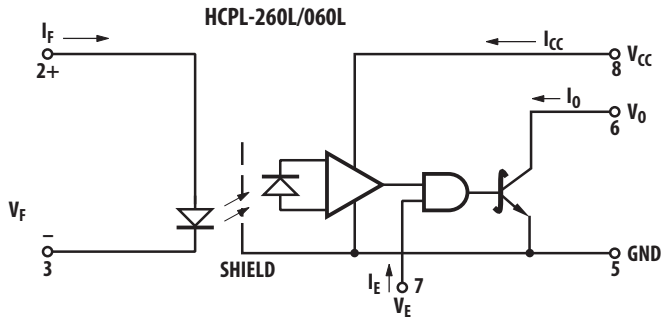
**Example 2:**

HCPL-263L to order product of 300-mil DIP package in tube packaging and non-RoHS compliant.

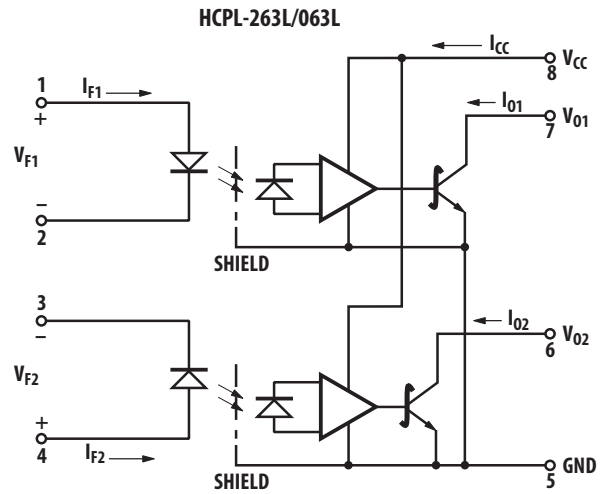
Option data sheets are available. Contact your Broadcom sales representative or authorized distributor for information.

Remarks: The notation '#XXX' is used for existing products, while (new) products launched since July 15, 2001 and RoHS compliant option will use '-XXxE'.

# Schematic

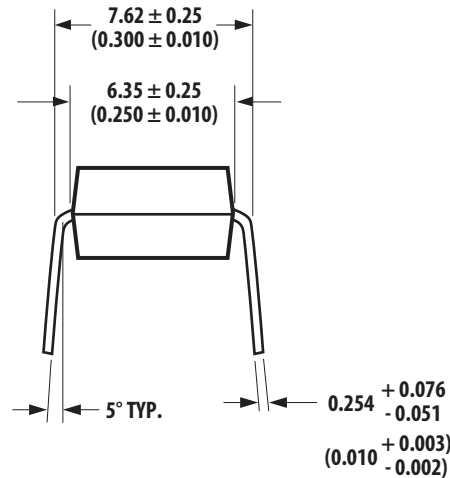
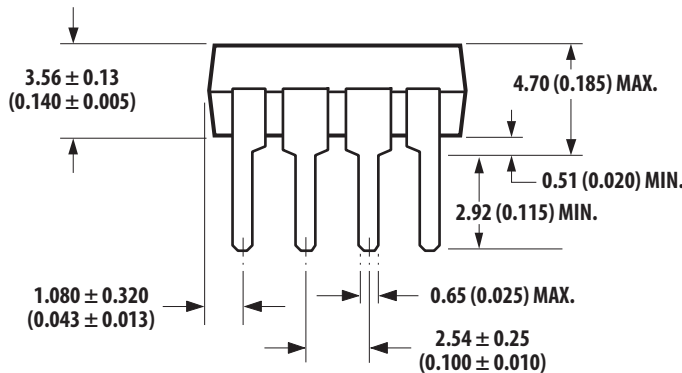
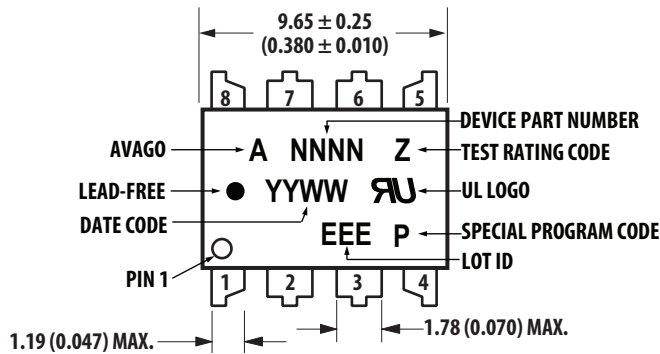


USE OF A 0.1  $\mu$ F BYPASS CAPACITOR CONNECTED BETWEEN PINS 5 AND 8 IS RECOMMENDED (SEE NOTE 5).



# Package Outline Drawings

## 8-Pin DIP Package

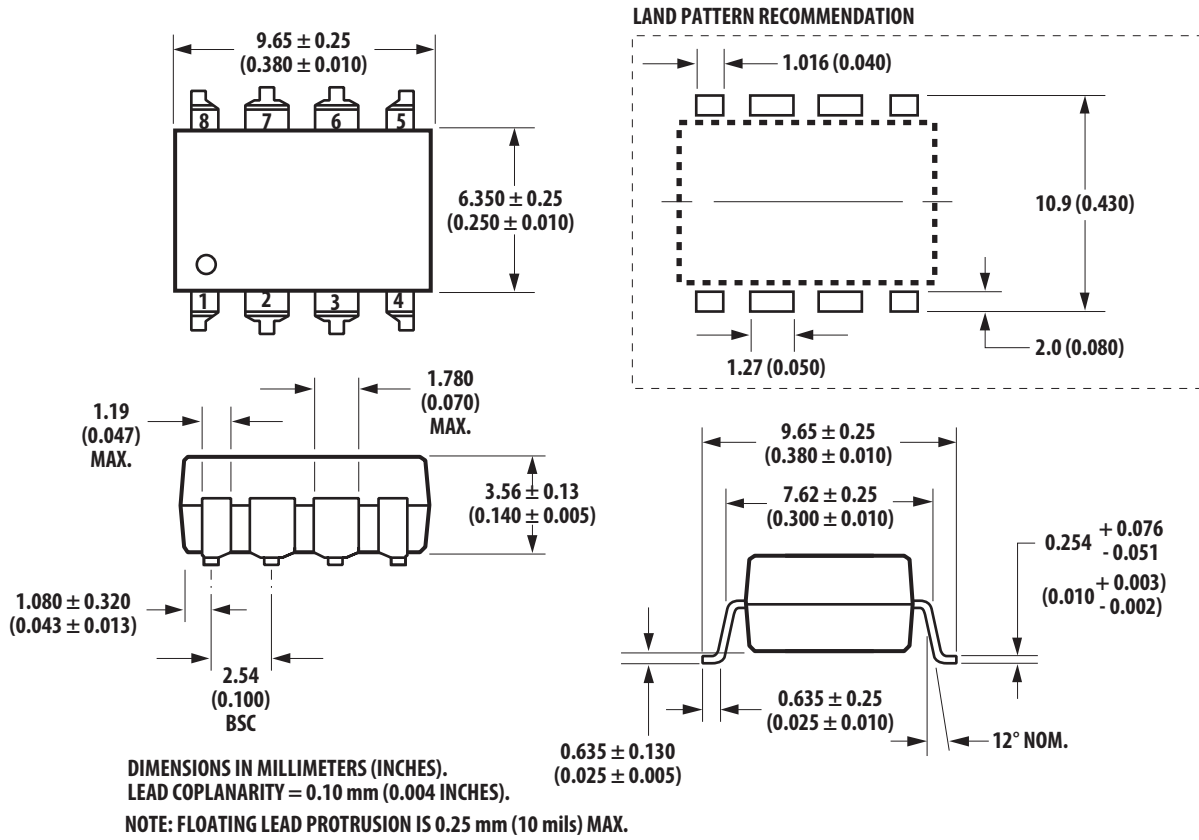


DIMENSIONS IN MILLIMETERS (INCHES).

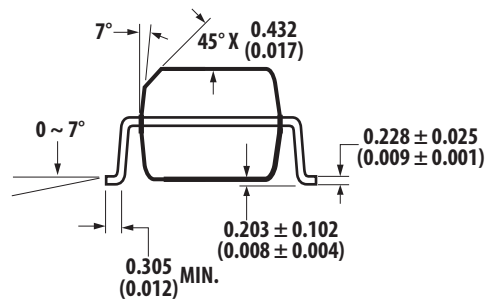
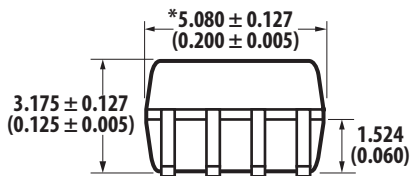
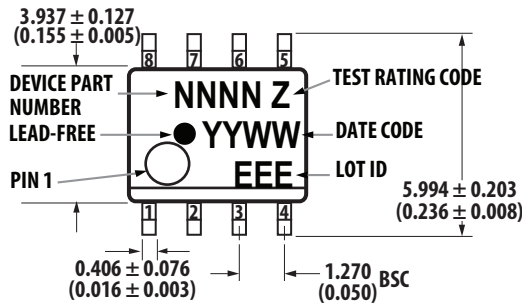
\*MARKING CODE LETTER FOR OPTION NUMBERS  
 "V" = OPTION 060  
 OPTION NUMBERS 300 AND 500 NOT MARKED.

NOTE: FLOATING LEAD PROTRUSION IS 0.25 mm (10 mils) MAX.

# 8-Pin DIP Package with Gull Wing Surface Mount in Option 500 (HCPL-260L, HCPL-263L)



## Small Outline SO-8 Package



\* TOTAL PACKAGE LENGTH (INCLUSIVE OF MOLD FLASH)  
 $5.207 \pm 0.254$  ( $0.205 \pm 0.010$ )

DIMENSIONS IN MILLIMETERS (INCHES).  
 LEAD COPLANARITY =  $0.10$  mm ( $0.004$  INCHES) MAX.

OPTION NUMBER 500 NOT MARKED.

NOTE: FLOATING LEAD PROTRUSION IS  $0.15$  mm (6 mils) MAX.

## Reflow Soldering Profile

Recommended reflow condition as per JEDEC Standard, J-STD-020 (latest revision). Non-halide flux should be used.

## Regulatory Information

The HCPL-260L/060L/263L/063L have been approved by the following organizations.

UL	Approval under UL 1577, Component Recognition Program, File E55361.
CSA	Approval under CSA Component Acceptance Notice #5, File CA 88324.
IEC/EN/DIN EN 60747-5-5	

## Insulation and Safety Related Specifications

Parameter	Symbol	8-Pin DIP (300 Mil) Value	SO-8 Value	Units	Conditions
Minimum External Air Gap (External Clearance)	L (101)	7.1	4.9	mm	Measured from input terminals to output terminals, shortest distance through air.
Minimum External Tracking (External Creepage)	L (102)	7.4	4.8	mm	Measured from input terminals to output terminals, shortest distance path along body.
Minimum Internal Plastic Gap (Internal Clearance)		0.08	0.08	mm	Through insulation distance, conductor to conductor, usually the direct distance between the photoemitter and photodetector inside the optocoupler cavity.
Tracking Resistance (Comparative Tracking Index)	CTI	200	200	V	DIN IEC 112/VDE 0303 Part 1
Isolation Group		IIIa	IIIa		Material Group (DIN VDE 0110, 1/89, Table 1)

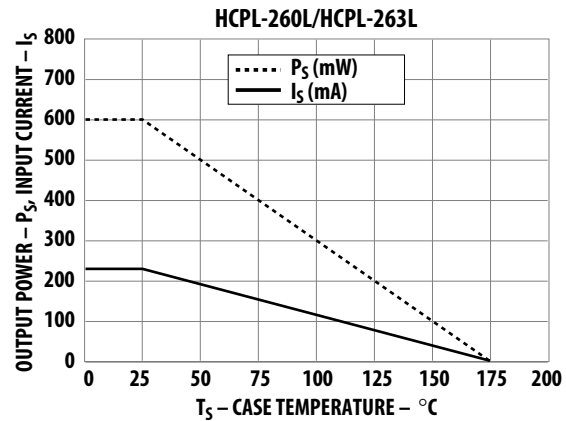
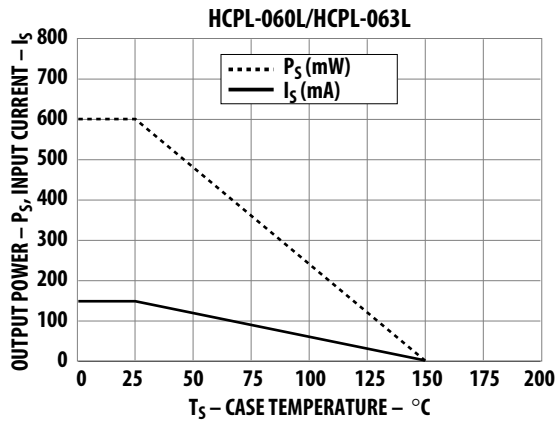
## IEC/EN/DIN EN 60747-5-5 Insulation Characteristics<sup>a</sup>

Description	Symbol	PDIP Option 060	SO-8 Option 060	Units
Installation classification per DIN VDE 0110, Table 1 For rated mains voltage ≤ 150 Vrms For rated mains voltage ≤ 300 Vrms For rated mains voltage ≤ 600 Vrms		I-IV I-IV I-III	I-IV I-IV I-III	
Climatic Classification		40/85/21	40/85/21	
Pollution Degree (DIN VDE 0110/39)		2	2	
Maximum Working Insulation Voltage	$V_{IORM}$	630	567	Vpeak
Input to Output Test Voltage, Method b <sup>a</sup> $V_{IORM} \times 1.875 = V_{PR}$ , 100% Production Test with $t_m=1s$ , Partial discharge < 5 pC	VPR	1181	1063	Vpeak
Input to Output Test Voltage, Method a <sup>a</sup> $V_{IORM} \times 1.6 = V_{PR}$ , Type and Sample Test, $t_m=10s$ , Partial discharge < 5 pC	VPR	1008	907	Vpeak
Highest Allowable Overvoltage (Transient Overvoltage $t_{ini} = 60s$ )	VIOTM	6000	6000	Vpeak
Safety-Limiting Values – maximum values allowed in the event of a failure.				
Case Temperature	$T_S$	175	150	°C
Input Current	$I_{S, INPUT}$	230	150	mA
Output Power	$P_{S, OUTPUT}$	600	600	mW
Insulation Resistance at $T_S$ , $V_{IO} = 500V$	$R_S$	$\geq 10^9$	$\geq 10^9$	$\Omega$

a. Refer to the front of the optocoupler section of the current catalog, under Product Safety Regulations section IEC/EN/DIN EN 60747-5-5, for a detailed description.

**NOTE:** Isolation characteristics are guaranteed only within the safety maximum ratings which must be ensured by protective circuits in application.

## Thermal Derating Curve Figures



## Absolute Maximum Ratings (No Derating Required up to 85°C)

Parameter	Symbol	Package <sup>a</sup>	Min.	Max.	Units	Note
Storage Temperature	$T_S$		-55	125	°C	
Operating Temperature	$T_A$		-40	85	°C	
Average Forward Input Current	$I_F$	Single 8-Pin DIP Single SO-8	—	20	mA	b
		Dual 8-Pin DIP Dual SO-8	—	15		
Reverse Input Voltage	$V_R$	8-Pin DIP, SO-8	—	5	V	c
Input Power Dissipation	$P_I$		—	40	mW	
Supply Voltage (1 Minute Maximum)	$V_{CC}$		—	7	V	
Enable Input Voltage (Not to Exceed $V_{CC}$ by more than 500 mV)	$V_E$	Single 8-Pin DIP Single SO-8	—	$V_{CC} + 0.5$	V	
			—	5	mA	
Output Collector Current	$I_O$		—	50	mA	c
Output Collector Voltage	$V_O$		—	7	V	c
Output Collector Power Dissipation	$P_O$	Single 8-Pin DIP Single SO-8	—	85	mW	
		Dual 8-Pin DIP Dual SO-8	—	60		

- a. Ratings apply to all devices except otherwise noted in the Package column.
- b. Peaking circuits may produce transient input currents up to 50 mA, 50-ns maximum pulse width, provided average current does not exceed 20 mA.
- c. Each channel.
- d. Peaking circuits may produce transient input currents up to 50 mA, 50-ns maximum pulse width, provided average current does not exceed 15 mA.
- e. Derate linearly above +80°C free-air temperature at a rate of 2.7 mW/°C for the SOIC-8 package.

## Recommended Operating Conditions

Parameter	Symbol	Min.	Max.	Units
Input Current, Low Level	$I_{FL}^a$	0	250	$\mu\text{A}$
Input Current, High Level <sup>b</sup>	$I_{FH}^c$	5	15	mA
Power Supply Voltage	$V_{CC}$	2.7	3.6	V
		4.5	5.5	
Low Level Enable Voltage	$V_{EL}$	0	0.8	V
High Level Enable Voltage	$V_{EH}$	2.0	$V_{CC}$	V
Operating Temperature	$T_A$	-40	85	$^{\circ}\text{C}$
Fan Out (at $R_L = 1\text{ k}\Omega$ ) <sup>b</sup>	N	—	5	TTL Loads
Output Pull-up Resistor	$R_L$	330	4 k	$\Omega$

a. The off condition can also be guaranteed by ensuring that  $V_{FL} \leq 0.8\text{V}$ .

b. Each channel.

c. The initial switching threshold is 5 mA or less. It is recommended that 6.3 mA to 10 mA be used for best performance and to permit at least a 20% LED degradation guardband.

## Electrical Specifications

Over recommended operating conditions ( $T_A = -40^\circ\text{C}$  to  $+85^\circ\text{C}$ ,  $2.7\text{V} \leq V_{CC} \leq 3.6\text{V}$ ), unless otherwise specified. All Typicals at  $V_{CC} = 3.3\text{V}$ ,  $T_A = 25^\circ\text{C}$ . All enable test conditions apply to single-channel products only.

**NOTE:** Bypassing of the power supply line is required, with a 0.1- $\mu\text{F}$  ceramic disc capacitor adjacent to each optocoupler as illustrated in [Figure 11](#). Total lead length between both ends of the capacitor and the isolator pins should not exceed 20 mm.

Parameter	Symbol	Device	Min.	Typ.	Max.	Units	Test Conditions	Figure	Note
High Level Output Current	$I_{OH}^a$		—	4.5	50	$\mu\text{A}$	$V_{CC} = 3.3\text{V}$ , $V_E = 2.0\text{V}$ , $V_O = 3.3\text{V}$ , $I_F = 250\ \mu\text{A}$	1	b, c
Input Threshold Current	$I_{TH}$		—	3.0	5.0	mA	$V_{CC} = 3.3\text{V}$ , $V_E = 2.0\text{V}$ , $V_O = 0.6\text{V}$ , $I_{OL}$ (Sinking) = 13 mA	2	c
Low Level Output Voltage	$V_{OL}^a$		—	0.35	0.6	V	$V_{CC} = 3.3\text{V}$ , $V_E = 2.0\text{V}$ , $I_F = 5\text{ mA}$ , $I_{OL}$ (Sinking) = 13 mA	3	c
High Level Supply Current	$I_{CCH}$	Single	—	4.7	7.0	mA	$V_E = 0.5\text{V}$ , $I_F = 0\text{ mA}$		
		Dual	—	6.9	10.0		$V_{CC} = 3.3\text{V}$		
Low Level Supply Current	$I_{CCL}$	Single	—	7.0	10.0	mA	$V_E = 0.5\text{V}$ , $I_F = 10\text{ mA}$		
		Dual	—	8.7	15.0		$V_{CC} = 3.3\text{V}$		
High Level Enable Current	$I_{EH}$	Single	—	-0.5	-1.2	mA	$V_{CC} = 3.3\text{V}$ , $V_E = 2.0\text{V}$		
Low Level Enable Current	$I_{EL}^a$	Single	—	-0.5	-1.2	mA	$V_{CC} = 3.3\text{V}$ , $V_E = 0.5\text{V}$		
High Level Enable Voltage	$V_{EH}$	Single	2.0	—	—	V			c
Low Level Enable Voltage	$V_{EL}$	Single	—	—	0.8	V			
Input Forward Voltage	$V_F$		1.4	1.5	1.75 <sup>a</sup>	V	$T_A = 25^\circ\text{C}$ , $I_F = 10\text{ mA}$	5	b
Input Reverse Breakdown Voltage	$BV_R^a$		5	—	—	V	$I_R = 10\ \mu\text{A}$		b
Input Diode Temperature Coefficient	$\Delta V_F/\Delta T_A$		—	-1.6	—	mV $^\circ\text{C}$	$I_F = 10\text{ mA}$		b
Input Capacitance	$C_{IN}$		—	60	—	pF	$f = 1\text{ MHz}$ , $V_F = 0\text{V}$		b

a. The JEDEC Registration specifies  $0^\circ\text{C}$  to  $+70^\circ\text{C}$ . Broadcom specifies  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$ .

b. Each channel.

c. No external pull-up is required for a high logic state on the enable input. If the  $V_E$  pin is not used, tying  $V_E$  to  $V_{CC}$  will result in improved CMR performance. For single channel products only. See [Application Information](#).

## Electrical Specifications (DC)

Over recommended operating conditions ( $T_A = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$ ), unless otherwise specified. All typicals at  $V_{CC} = 5\text{V}$ ,  $T_A = 25^{\circ}\text{C}$ .

Parameter	Symbol	Channel	Min.	Typ. <sup>a</sup>	Max.	Units	Test Conditions	Figure	Note
High Level Output Current	$I_{OH}$		—	5.5	100	$\mu\text{A}$	$V_{CC} = 5.5\text{V}$ , $V_O = 5.5\text{V}$ , $I_{FL} = 250\ \mu\text{A}$	1	b, c
Input Threshold Current	$I_{TH}$	Single	—	2.0	5.0	mA	$V_{CC} = 5.5\text{V}$ , $V_O = 0.6\text{V}$ , $I_{OL} > 13\ \text{mA}$	2	c
		Dual	—	2.5	—				
Low Level Output Voltage	$V_{OL}$		—	0.35	0.6	V	$V_{CC} = 5.5\text{V}$ , $I_F = 5\ \text{mA}$ , $I_{OL} (\text{Sinking}) = 13\ \text{mA}$	3	c
High Level Supply Current	$I_{CCH}$	Single	—	7.0	10.0	mA	$V_E = 0.5\text{V}$ , $V_{CC} = 5.5\text{V}$ , $I_F = 0\ \text{mA}$		
				6.5	—		$V_E = V_{CC}$ , $V_{CC} = 5.5\text{V}$ , $I_F = 0\ \text{mA}$		
		Dual	—	10.0	15.0	$V_{CC} = 5.5\text{V}$ , $I_F = 0\ \text{mA}$			
Low Level Supply Current	$I_{CCL}$	Single	—	9.0	13.0	mA	$V_E = 0.5\text{V}$ , $V_{CC} = 5.5\text{V}$ , $I_F = 10\ \text{mA}$		
				8.5	—		$V_E = V_{CC}$ , $V_{CC} = 5.5\text{V}$ , $I_F = 10\ \text{mA}$		
		Dual	—	13.0	21.0	$V_{CC} = 5.5\text{V}$ , $I_F = 10\ \text{mA}$			
High Level Enable Current	$I_{EH}$	Single	—	-0.7	-1.6	mA	$V_{CC} = 5.5\text{V}$ , $V_E = 2.0\text{V}$		
Low Level Enable Current	$I_{EL}$	Single	—	-0.9	-1.6	mA	$V_{CC} = 5.5\text{V}$ , $V_E = 0.5\text{V}$		
High Level Enable Voltage	$V_{EH}$	Single	2.0	—	—	V			c
Low Level Enable Voltage	$V_{EL}$	Single	—	—	0.8	V			
Input Forward Voltage	$V_F$		1.4	1.5	1.75	V	$T_A = 25^{\circ}\text{C}$ , $I_F = 10\ \text{mA}$		d
			1.3	—	1.8	V	$I_F = 10\ \text{mA}$		
Input Reverse Breakdown Voltage	$BV_R$		5	—	—	V	$I_R = 10\ \mu\text{A}$		b
Input Diode Temperature Coefficient	$\Delta V_F / \Delta T_A$		—	-1.6	—	mV/ $^{\circ}\text{C}$	$I_F = 10\ \text{mA}$		b
Input Capacitance	$C_{IN}$		—	60	—	pF	$f = 1\ \text{MHz}$ , $V_F = 0\text{V}$		b

- a. The JEDEC Registration specifies  $0^{\circ}\text{C}$  to  $+70^{\circ}\text{C}$ . Broadcom specifies  $-40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$ .
- b. Each channel.
- c. No external pull-up is required for a high logic state on the enable input. If the  $V_E$  pin is not used, tying  $V_E$  to  $V_{CC}$  will result in improved  $CM_R$  performance. For single-channel products only. See [Application Information](#).
- d. Bypassing of the power supply line is required, with a 0.1- $\mu\text{F}$  ceramic disc capacitor adjacent to each optocoupler as illustrated in [Figure 11](#). Total lead length between both ends of the capacitor and the isolator pins should not exceed 20 mm.

## Switching Specifications

Over recommended operating conditions ( $T_A = -40^\circ\text{C}$  to  $+85^\circ\text{C}$ ,  $2.7\text{V} \leq V_{CC} \leq 3.6\text{V}$ ),  $I_F = 7.5\text{ mA}$ , unless otherwise specified. All Typical at  $T_A = 25^\circ\text{C}$ ,  $V_{CC} = 3.3\text{V}$ .

Parameter	Symbol	Min.	Typ.	Max.	Units	Test Conditions	Figure	Note
Propagation Delay Time to High Output Level	$t_{PLH}$	—	—	90	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$	6, 7	a, b, c
Propagation Delay Time to Low Output Level	$t_{PHL}$	—	—	75	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$		a, c, d
Pulse Width Distortion	$ \mathit{t}_{PHL} - \mathit{t}_{PLH} $	—	—	25	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$	8	c, e
Propagation Delay Skew	$t_{PSK}$	—	—	40	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$		c, e, f
Output Rise Time (10% to 90%)	$t_r$	—	45	—	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$		a, c
Output Fall Time (90% to 10%)	$t_f$	—	20	—	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$		a, c
Propagation Delay Time of Enable from $V_{EH}$ to $V_{EL}$	$t_{ELH}$	—	45	—	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$ , $V_{EL} = 0\text{V}$ , $V_{EH} = 3\text{V}$	9	g
Propagation Delay Time of Enable from $V_{EL}$ to $V_{EH}$	$t_{EHL}$	—	30	—	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$ , $V_{EL} = 0\text{V}$ , $V_{EH} = 3\text{V}$	9	h

- a. Each channel.
- b. The  $t_{PLH}$  propagation delay is measured from the 3.75 mA point on the falling edge of the input pulse to the 1.5V point on the rising edge of the output pulse.
- c. No external pull up is required for a high logic state on the enable input. If the VE pin is not used, tying  $V_E$  to  $V_{CC}$  will result in improved  $CM_R$  performance. For single channel products only. See [Application Information](#).
- d. The  $t_{PHL}$  propagation delay is measured from the 3.75 mA point on the rising edge of the input pulse to the 1.5V point on the falling edge of the output pulse.
- e. See [Figure 9](#) for measurement details.
- f.  $t_{PSK}$  is equal to the worst case difference in  $t_{PHL}$ ,  $t_{PLH}$ , or both that will be seen between units at any given temperature and specified test conditions.
- g. The  $t_{ELH}$  enable propagation delay is measured from the 1.5V point on the falling edge of the enable input pulse to the 1.5V point on the rising edge of the output pulse.
- h. The  $t_{EHL}$  enable propagation delay is measured from the 1.5V point on the rising edge of the enable input pulse to the 1.5V point on the falling edge of the output pulse.

## Switching Specifications (AC)

Over recommended operating conditions  $T_A = -40^\circ\text{C}$  to  $85^\circ\text{C}$ ,  $4.5\text{V} \leq V_{CC} \leq 5.5\text{V}$ ,  $I_F = 7.5\text{ mA}$ , unless otherwise specified. All Typical at  $V_{CC} = 5\text{V}$ ,  $T_A = 25^\circ\text{C}$ .

Parameter	Symbol	Min.	Typ.	Max.	Units	Test Conditions	Figure	Note
Propagation Delay Time to High Output Level	$t_{PLH}$	20	48	75	ns	$T_A = 25^\circ\text{C}$ , $R_L = 350\Omega$ , $C_L = 15\text{ pF}$	6, 7	a, b, c
		—	—	100				
Propagation Delay Time to Low Output Level	$t_{PHL}$	25	50	75	ns	$T_A = 25^\circ\text{C}$ , $R_L = 350\Omega$ , $C_L = 15\text{ pF}$	6, 7	a, c, d
		—	—	100				
Pulse Width Distortion	$ t_{PHL} - t_{PLH} $	—	3.5	35	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$	8	c, e
Propagation Delay Skew	$T_{PSK}$	—	—	40	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$		c, e, f
Output Rise Time (10% to 90%)	$t_r$	—	24	—	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$		a, c
Output Fall Time (10% to 90%)	$t_f$	—	10	—	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$		a, c
Propagation Delay Time of Enable from $V_{EH}$ to $V_{EL}$	$t_{ELH}$	—	30	—	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$ , $V_{EL} = 0\text{V}$ , $V_{EH} = 3\text{ V}$	9	g
Propagation Delay Time of Enable from $V_{EL}$ to $V_{EH}$	$t_{EHL}$	—	20	—	ns	$R_L = 350\Omega$ , $C_L = 15\text{ pF}$ , $V_{EL} = 0\text{V}$ , $V_{EH} = 3\text{V}$	9	h

- Each channel.
- The  $t_{PLH}$  propagation delay is measured from the 3.75 mA point on the falling edge of the input pulse to the 1.5V point on the rising edge of the output pulse.
- No external pull-up is required for a high logic state on the enable input. If the  $V_E$  pin is not used, tying  $V_E$  to  $V_{CC}$  will result in improved  $CM_R$  performance. For single-channel products only. See [Application Information](#).
- The  $t_{PHL}$  propagation delay is measured from the 3.75 mA point on the rising edge of the input pulse to the 1.5V point on the falling edge of the output pulse.
- See test circuit for measurement details.
- $t_{PSK}$  is equal to the worst case difference in  $t_{PHL}$ ,  $t_{PLH}$ , or both that will be seen between units at any given temperature and specified test conditions.
- The  $t_{ELH}$  enable propagation delay is measured from the 1.5V point on the falling edge of the enable input pulse to the 1.5V point on the rising edge of the output pulse.
- The  $t_{EHL}$  enable propagation delay is measured from the 1.5V point on the rising edge of the enable input pulse to the 1.5V point on the falling edge of the output pulse.

Parameter	Symbol	Device	Min.	Typ.	Units	Test Conditions	Figure	Note
Output High Level Common Mode Transient Immunity	CMH	HCPL-263L HCPL-063L HCPL-260L HCPL-060L	15	25	kV/μs	V <sub>CC</sub> = 3.3V, I <sub>F</sub> = 0 mA, V <sub>O(MIN)</sub> = 2V, R <sub>L</sub> = 350Ω, T <sub>A</sub> = 25°C, V <sub>CM</sub> = 1000V and V <sub>CM</sub> = 10V	10	a, b, c
Output Low Level Common Mode Transient Immunity	CML	HCPL-263L HCPL-063L HCPL-260L HCPL-060L	15	25	kV/μs	V <sub>CC</sub> = 3.3V, I <sub>F</sub> = 7.5 mA, V <sub>O(MAX)</sub> = 0.8V, R <sub>L</sub> = 350Ω, T <sub>A</sub> = 25°C, V <sub>CM</sub> = 1000V and V <sub>CM</sub> = 10V	10	b, c, d
Output High Level Common Mode Transient Immunity	CMH	HCPL-263L HCPL-063L HCPL-260L HCPL-060L	10	15	kV/μs	V <sub>CC</sub> = 5V, I <sub>F</sub> = 0 mA, V <sub>O(MIN)</sub> = 2V, R <sub>L</sub> = 350 Ω, T <sub>A</sub> = 25°C, V <sub>CM</sub> = 1000V	10	a, b, c
Output Low Level Common Mode Transient Immunity	CML	HCPL-263L HCPL-063L HCPL-260L HCPL-060L	10	15	kV/μs	V <sub>CC</sub> = 5V, I <sub>F</sub> = 7.5 mA, V <sub>O(MAX)</sub> = 0.8V, R <sub>L</sub> = 350Ω, T <sub>A</sub> = 25°C, V <sub>CM</sub> = 1000V	10	b, c, d

- a. CM<sub>H</sub> is the maximum tolerable rate of rise on the common mode voltage to assure that the output will remain in a high logic state (that is, V<sub>O</sub> > 2.0V).
- b. For sinusoidal voltages, (|dV<sub>CM</sub> | / dt)<sub>max</sub> = ηf<sub>CM</sub>V<sub>CM</sub> (p-p).
- c. No external pull-up is required for a high logic state on the enable input. If the V<sub>E</sub> pin is not used, tying V<sub>E</sub> to V<sub>CC</sub> will result in improved CM<sub>R</sub> performance. For single-channel products only. See [Application Information](#).
- d. CM<sub>L</sub> is the maximum tolerable rate of fall of the common mode voltage to assure that the output will remain in a low logic state (that is, V<sub>O</sub> < 0.8V).

## Package Characteristics

All Typicals at  $T_A = 25^\circ\text{C}$ .

Parameter	Sym.	Package	Min.	Typ.	Max.	Units	Test Conditions	Figure	Note
Input-Output Insulation	$I_{I-O}^a$	Single 8-Pin DIP Single SO-8	—	—	1	$\mu\text{A}$	45% RH, $t = 5\text{s}$ , $V_{I-O} = 3\text{ kV DC}$ , $T_A = 25^\circ\text{C}$		b, c
Input-Output Momentary Withstand Voltage <sup>d</sup>	$V_{ISO}$	8-Pin DIP SO-8	3750	—	—	Vrms	RH $\leq 50\%$ , $T_A = 25^\circ\text{C}$ , $t = 1\text{ minute}$		b, c
Input-Output Resistance	$R_{I-O}$	8-Pin SO-8	—	$10^{12}$	—	$\Omega$	$V_{I-O} = 500\text{ Vdc}$		b, e, f
Input-Output Capacitance	$C_{I-O}$	8-Pin DIP SO-8	—	0.6	—	pF	$f = 1\text{ MHz}$ , $T_A = 25^\circ\text{C}$		b, e, f
Input-Input Insulation Leakage Current	$I_{I-I}$	Dual Channel	—	0.005	—	$\mu\text{A}$	RH $\leq 45\%$ , $t = 5\text{s}$ , $V_{I-I} = 500\text{V}$		g
Resistance (Input-Input)	$R_{I-I}$	Dual Channel	—	$10^{11}$	—	$\Omega$			g
Capacitance (Input-Input)	$C_{I-I}$	Dual 8-Pin Dip	—	0.03	—	pG	$f = 1\text{ MHz}$		g
		Dual SO-8	—	0.25	—				

- a. The JEDEC Registration specifies  $0^\circ\text{C}$  to  $+70^\circ\text{C}$ . Broadcom specifies  $-40^\circ\text{C}$  to  $+85^\circ\text{C}$ .
- b. The device is considered a two-terminal device: pins 1, 2, 3, and 4 shorted together, and pins 5, 6, 7, and 8 shorted together.
- c. In accordance with UL 1577, each optocoupler is proof tested by applying an insulation test voltage  $\geq 4500\text{ Vrms}$  for 1 second (leakage detection current limit,  $I_{I-O} \leq 5\ \mu\text{A}$ ). This test is performed before the 100% production test for partial discharge (Method b) shown in the IEC/EN/DIN EN 60747-5-5 Insulation Characteristics Table, if applicable.
- d. The Input-Output Momentary Withstand Voltage is a dielectric voltage rating that should not be interpreted as an input-output continuous voltage rating. For the continuous voltage rating, refer to the IEC/EN/DIN EN 60747-5-5 Insulation Characteristics Table (if applicable), your equipment level safety specification, or Broadcom Application Note 1074, *Optocoupler Input-Output Endurance Voltage*.
- e. Each channel.
- f. Measured between the LED anode and cathode shorted together and pins 5 through 8 shorted together. For dual-channel products only.
- g. Measured between pins 1 and 2 shorted together, and pins 3 and 4 shorted together. For dual-channel products only.

Figure 1: Typical High Level Output Current vs. Temperature

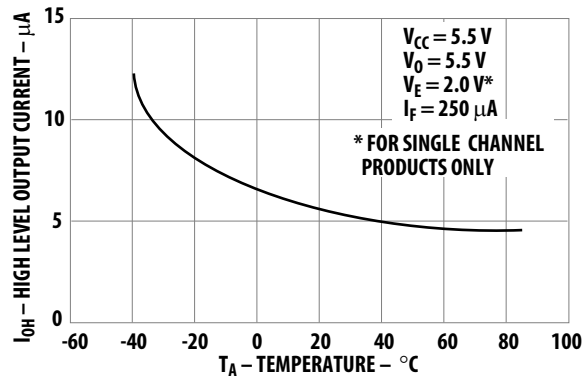
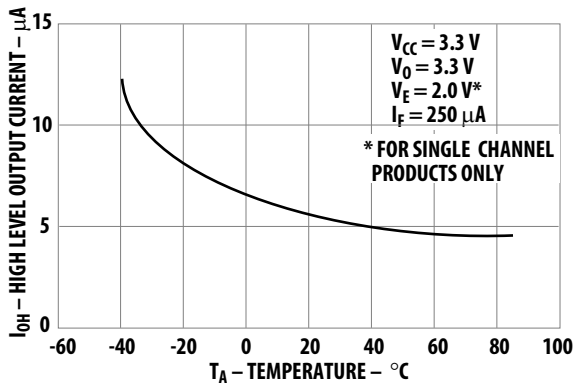


Figure 2: Typical Output Voltage vs. Forward Input Current

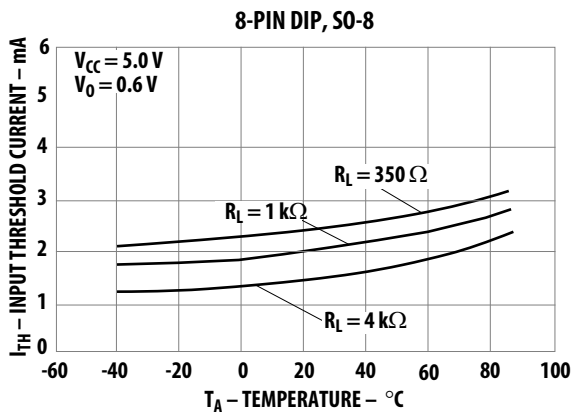
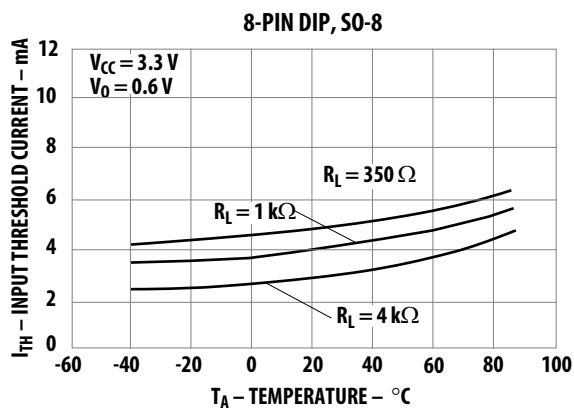


Figure 3: Typical Low Level Output Voltage vs. Temperature

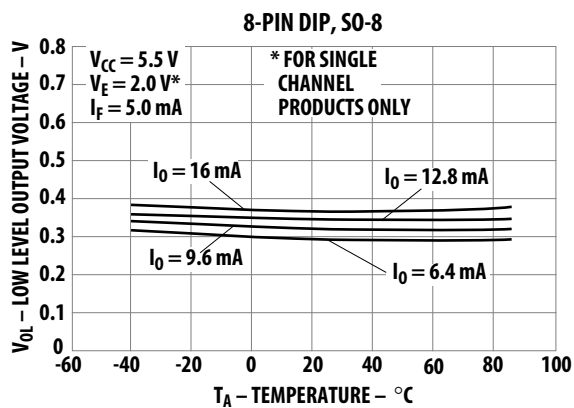
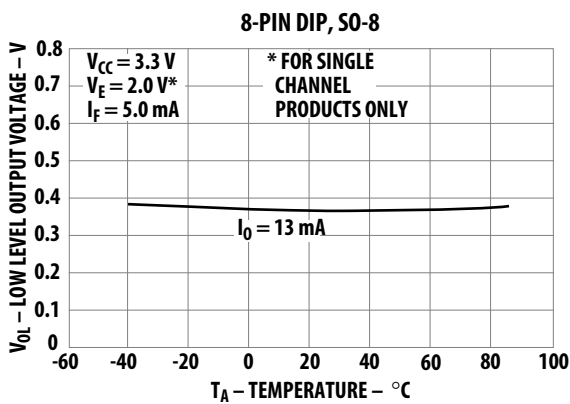


Figure 4: Typical Low Level Output Current vs. Temperature

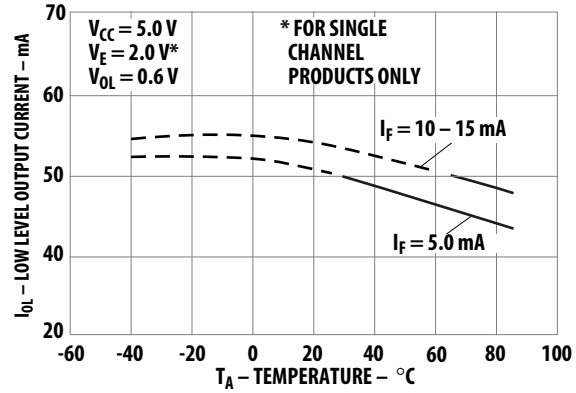
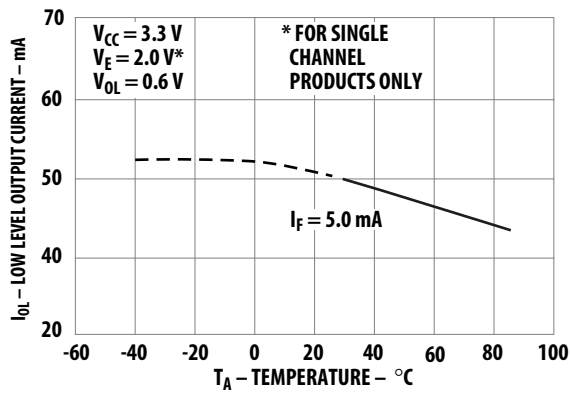


Figure 5: Typical Input Diode Forward Characteristic

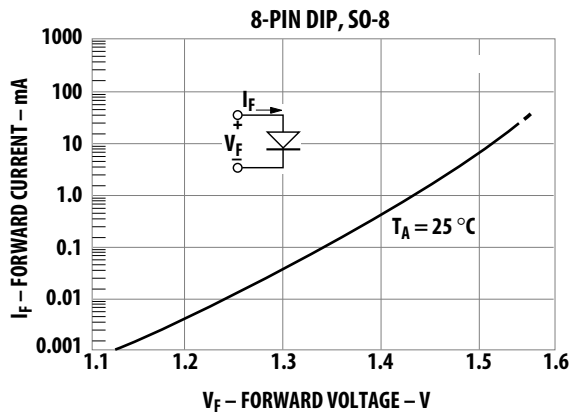


Figure 6: Test Circuit for  $t_{PHL}$  and  $t_{PLH}$

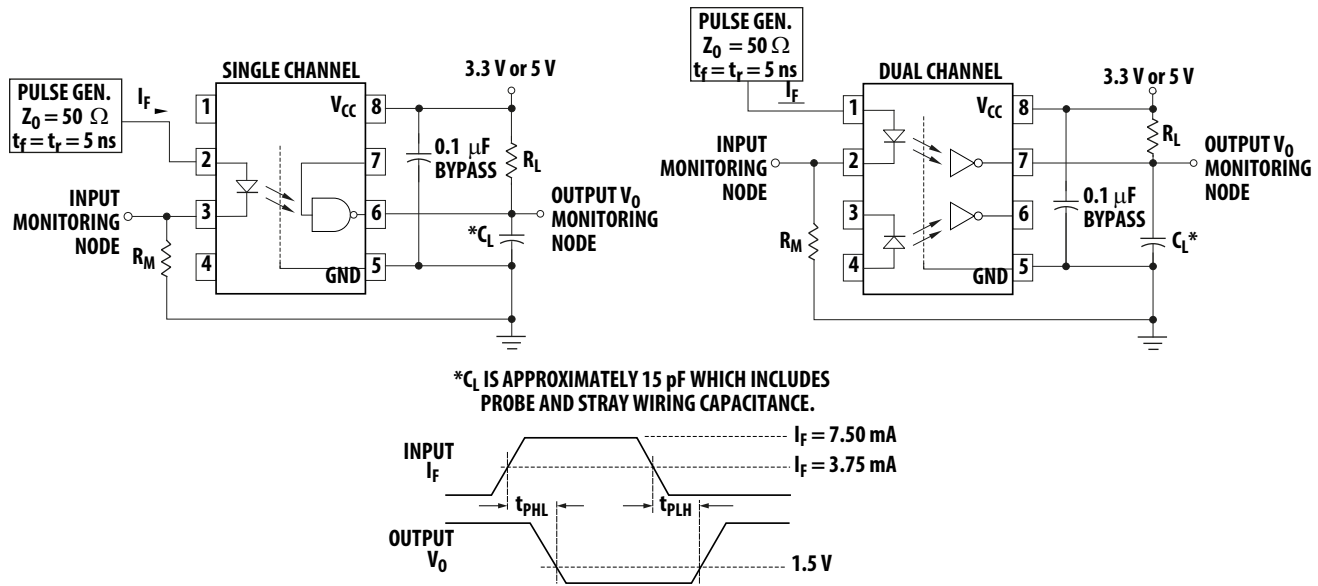


Figure 7: Typical Propagation Delay vs. Temperature

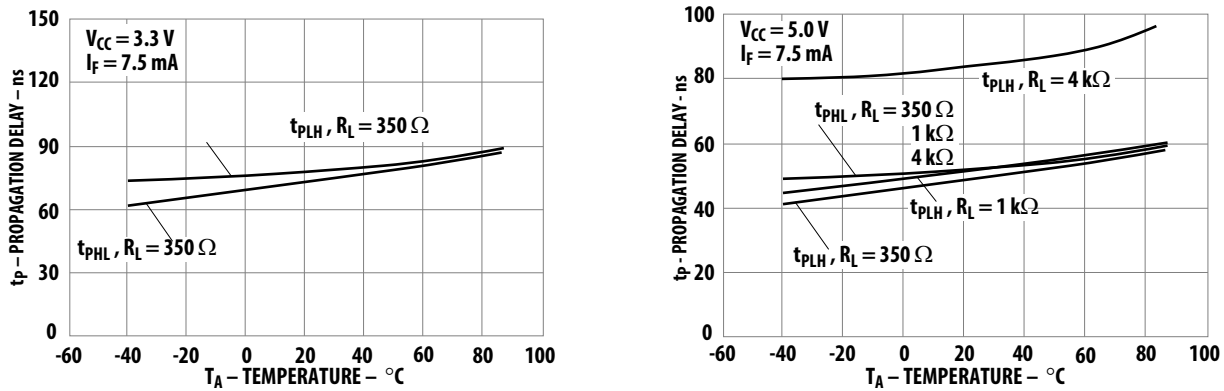


Figure 8: Typical Pulse Width Distortion vs. Temperature

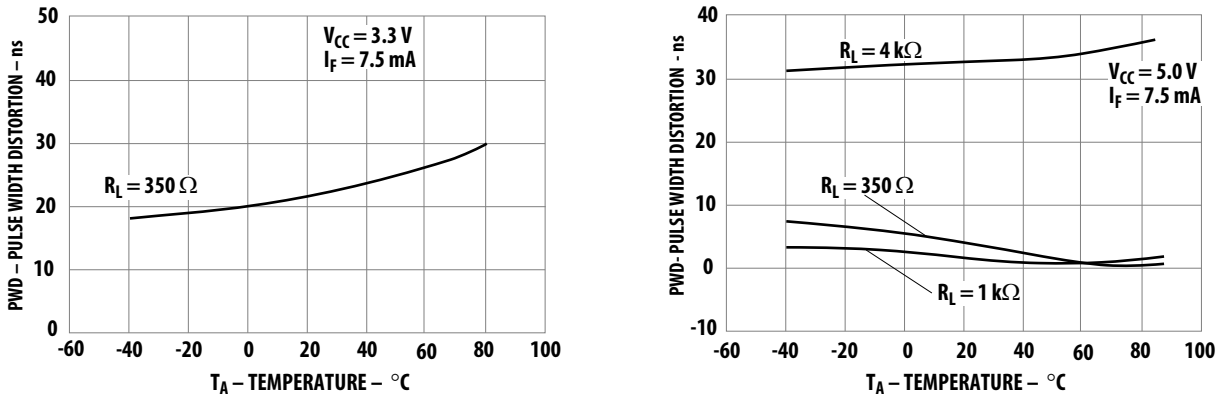


Figure 9: Test Circuit for  $t_{EHL}$  and  $t_{ELH}$

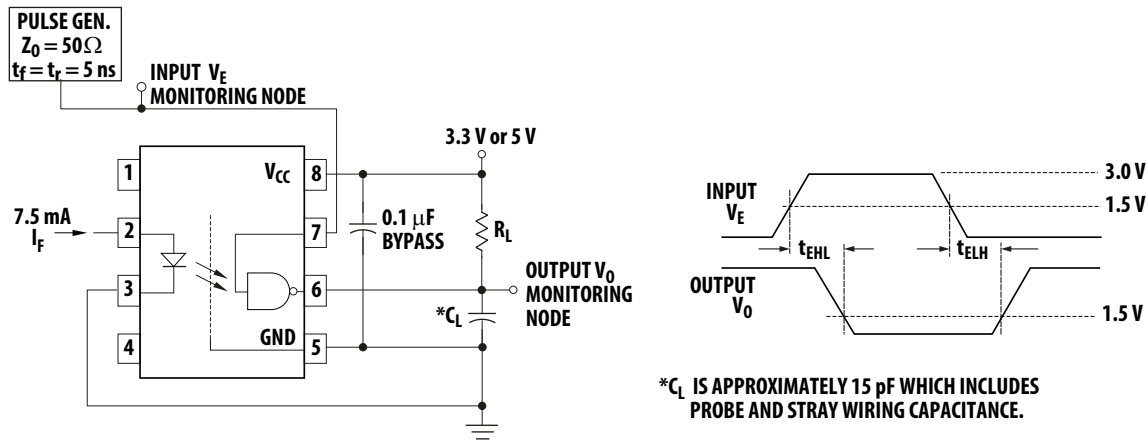


Figure 10: Test Circuit for Common Mode Transient Immunity and Typical Waveforms

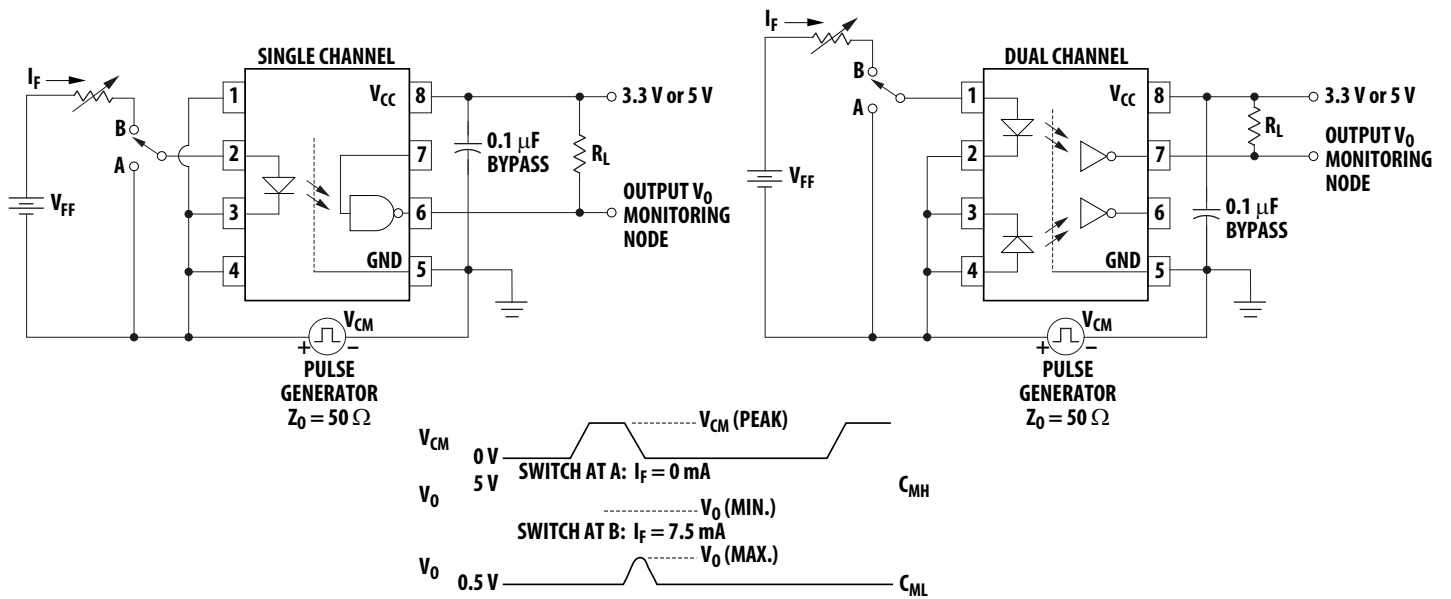


Figure 11: Recommended Printed Circuit Board Layout

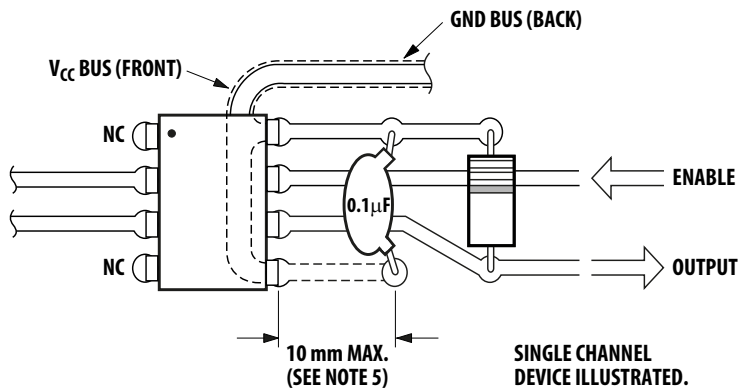
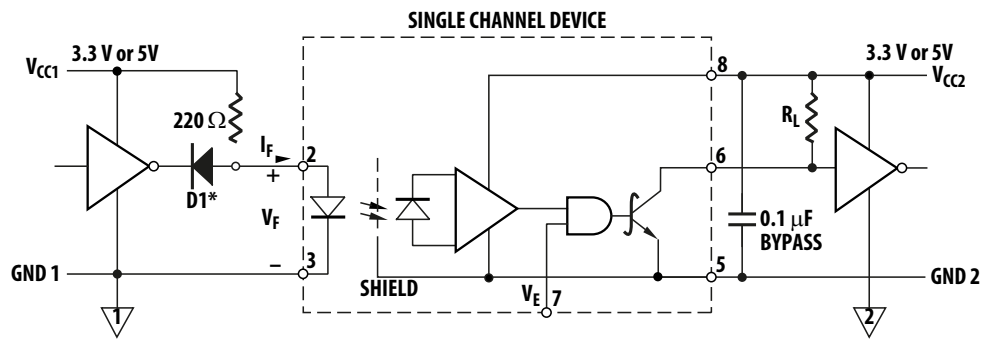
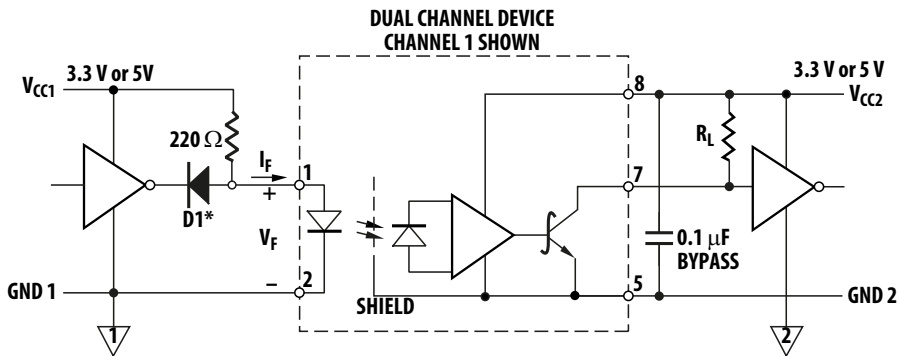


Figure 12: Recommended LVTTTL Interface Circuit



\*DIODE D1 (1N916 OR EQUIVALENT) IS NOT REQUIRED FOR UNITS WITH OPEN COLLECTOR OUTPUT.



## Application Information

### Common-Mode Rejection for HCPL-260L Families

Figure 13 shows the recommended drive circuit for optimal common-mode rejection performance. Note the following two points:

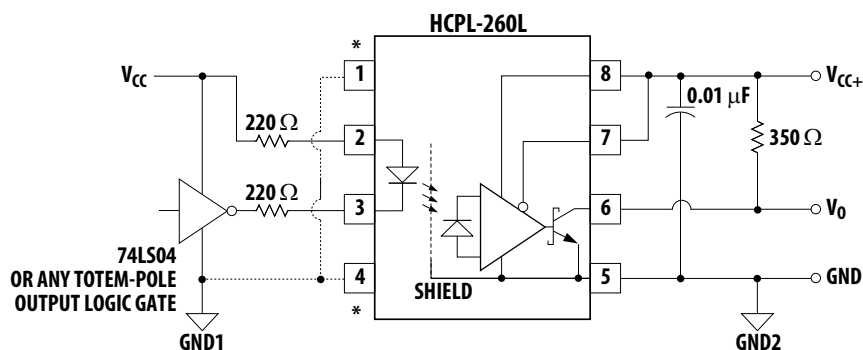
- The enable pin is tied to  $V_{CC}$  rather than floating (this applies to single-channel parts only).
- Two LED-current setting resistors are used instead of one. This is to balance  $I_{LED}$  variation during common-mode transients.

If the enable pin is left floating, it is possible for common-mode transients to couple to the enable pin, resulting in common-mode failure. This failure mechanism only occurs when the LED is on and the output is in the Low state. It is identified as occurring when the transient output voltage rises above 0.8V. Therefore, the enable pin should be connected to either  $V_{CC}$  or logic-level high for best common-mode performance with the output low ( $CMR_L$ ). This failure mechanism is only present in single-channel parts that have the enable function.

Also, common-mode transients can capacitively couple from the LED anode (or cathode) to the output-side ground causing current to be shunted away from the LED (which can be bad if the LED is on) or conversely cause current to be injected into the LED (bad if the LED is meant to be off). Figure 14 shows the parasitic capacitances that exist between LED anode/cathode and output ground ( $C_{LA}$  and  $C_{LC}$ ). Also shown in Figure 14 on the input side is an AC-equivalent circuit.

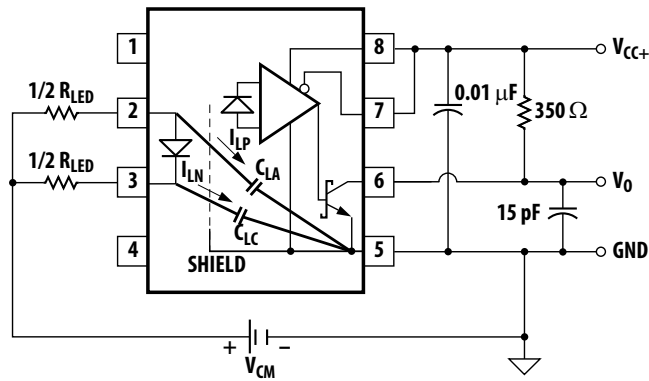
For transients occurring when the LED is on, common-mode rejection ( $CMR_L$ , since the output is in the “low” state) depends upon the amount of LED current drive ( $I_F$ ). For conditions where  $I_F$  is close to the switching threshold ( $I_{TH}$ ),  $CMR_L$  also depends on the extent which  $I_{LP}$  and  $I_{LN}$  balance each other. In other words, any condition where common-mode transients cause a momentary decrease in  $I_F$  will cause common-mode failure for transients that are fast enough.

Figure 13: Recommended Drive Circuit for High-CMR



\* HIGHER CMR MAY BE OBTAINABLE BY CONNECTING PINS 1, 4 TO INPUT GROUND (GND1).

Figure 14: AC Equivalent Circuit



Likewise for common-mode transients that occur when the LED is off (that is,  $CMR_H$ , since the output is “high”), if an imbalance between  $I_{LP}$  and  $I_{LN}$  results in a transient  $I_F$  equal to or greater than the switching threshold of the optocoupler, the transient “signal” may cause the output to spike below 2V (which constitutes a  $CMR_H$  failure).

By using the circuit in Figure 13, good CMR can be achieved. The balanced  $I_{LED}$ -setting resistors help equalize  $I_{LP}$  and  $I_{LN}$  to reduce the amount by which  $I_{LED}$  is modulated from transient coupling through  $C_{LA}$  and  $C_{LC}$ .

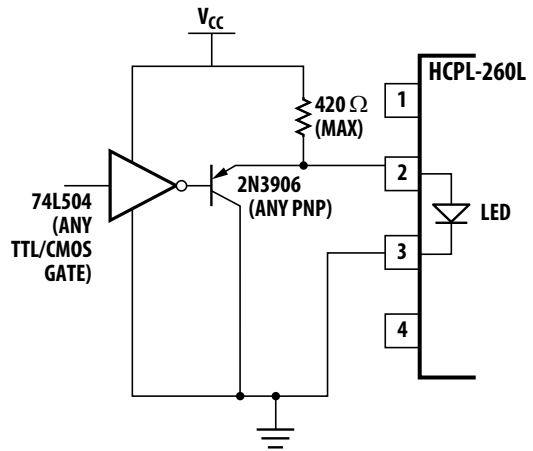
### CMR with Other Drive Circuits

CMR performance with drive circuits other than that shown in Figure 13 may be enhanced by following these guidelines:

- Use of drive circuits where current is shunted from the LED in the LED “off” state (as shown in Figure 15 and Figure 16). This is beneficial for good  $CMR_H$ .
- Use of  $I_{FH} > 3.5$  mA. This is good for high  $CMR_L$ .

Figure 15 shows a circuit that can be used with any totem-pole-output TTL/LSTTL/HCMOS logic gate. The buffer PNP transistor allows the circuit to be used with logic devices that have low current-sinking capability. It also helps maintain the driving-gate power-supply current at a constant level to minimize ground shifting for other devices connected to the input-supply ground.

Figure 15: TTL Interface Circuit



When using an open-collector TTL or open-drain CMOS logic gate, the circuit in Figure 16 may be used. When using a CMOS gate to drive the optocoupler, the circuit shown in Figure 17 may be used. The diode in parallel with the  $R_{LED}$  speeds the turn-off of the optocoupler LED.

Figure 16: TTL Open-Collector/Open-Drain Gate Drive Circuit

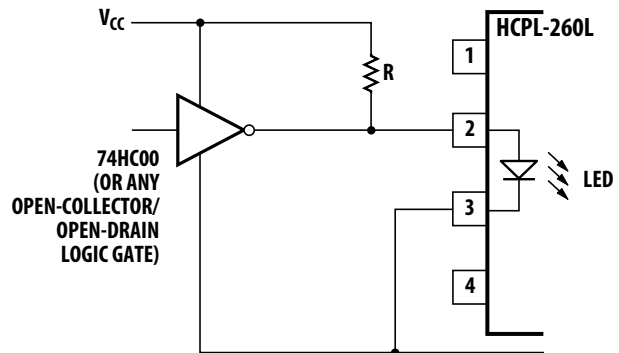
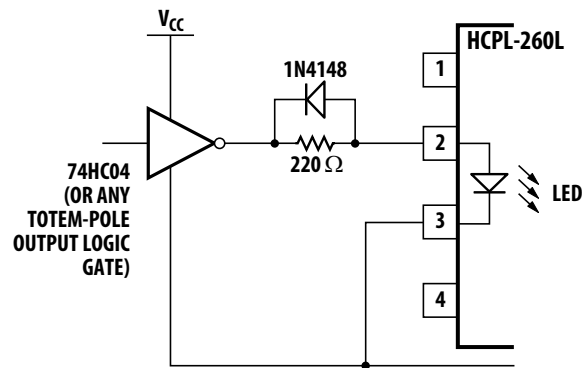


Figure 17: CMOS Gate Drive Circuit



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