

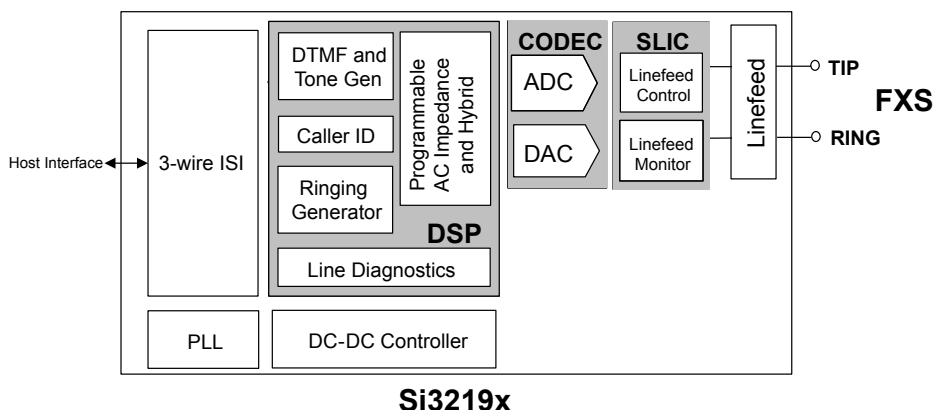
Si3219x: ProSLIC® Single FXS Solution

ProSLIC® Single-Channel FXS Solutions

The Si3219x Single FXS ProSLIC® devices implement a complete foreign exchange station (FXS) telephony interface solution in a single package that conforms to all relevant global specifications. The Si3219x ProSLIC ICs operate from a 3.3 V supply and have a 3-wire ISI digital interface with 1.8 to 3.3 V I/O. The Si3219x devices are designed to operate with a capacitive boost tracking battery supply for lower power, cost, and footprint than other tracking supplies in the industry. Self-testing and metallic loop testing (MLT) are facilitated by the built-in DSP, monitor ADC, and test load. The devices are available with wideband audio for better-than-PSTN voice quality and DTMF detection. The Si3219x devices are available in a 4 x 6 mm, 38-pin QFN package.

Applications:

- VoIP Gateways and Routers
- xDSL IADs
- Optical Network Terminals/Units (ONT/U)
- Analog Terminal Adapters (ATA)
- Cable eMTA
- Wireless Fixed Terminals (WFT)
- Wireless Local Loop (WLL)
- WiMAX CPE



KEY FEATURES

- Complete FXS solution in a single 4 x 6 mm package
- Performs all BORSCHT functions
- Ideal for short- or medium-loop applications
- Ultra-low power consumption
- Patented low-power ringing
- Adaptive ringing
- Dynamic overhead control
- Ringing current limiting
- Simplified configuration and diagnostics
 - Supported by ProSLIC API
 - Audio diagnostics with loopback
 - Integrated test load
- Global programmability
- Wideband voice support
- On-hook transmission
- Loop or ground start operation
- Smooth polarity reversal
- A-Law/ μ -Law companding, linear PCM
- Flexible integrated tracking dc-dc controller supporting patent-pending low-cost capacitive boost configuration
- Software-programmable parameters:
 - Ringing frequency, amplitude, cadence, and waveshape
 - Two-wire ac impedance
 - Transhybrid balance
 - DC current loop feed (10–45 mA)
 - Loop closure and ring trip thresholds
 - Ground key detect threshold
- Pulse metering
- DTMF generation
- DTMF detection (Si32193)
- 3.3 V operation
- Support for 1.8 V I/O
- Maximum battery up to –100 V
- Pb-free/RoHS-compliant packaging
- 3-wire Integrated Serial Interface (ISI)

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1. Ordering Guide

Table 1.1. Si3219x Ordering Guide

| P/N | Description | Package Type ² | Max V _{BAT} | Temperature |
|---------------|--|---------------------------|----------------------|--------------|
| Si32192-A-FM1 | Wideband FXS, ISI interface, parametric MLT | QFN38 | -100 V | 0 to 70 °C |
| Si32192-A-GM1 | Wideband FXS, ISI interface, parametric MLT | QFN38 | -100 V | -40 to 85 °C |
| Si32193-A-FM1 | Wideband FXS, ISI interface, parametric MLT, DTMF detection | QFN38 | -100 V | 0 to 70 °C |
| Si32193-A-GM1 | Wideband FXS, ISI interface, parametric MLT, DTMF detection | QFN38 | -100 V | -40 to 85 °C |
| Si32192-A-ZM1 | Wideband FXS, ISI interface, parametric MLT, customer-specific | QFN38 | -100 V | 0 to 70 °C |
| Si32192-A-ZM2 | Wideband FXS, ISI interface, parametric MLT, customer-specific | QFN38 | -100 V | 0 to 70 °C |
| Si32193-A-ZM1 | Wideband FXS, ISI interface, parametric MLT, DTMF detection, customer-specific | QFN38 | -100 V | 0 to 70 °C |
| Si32193-A-ZM2 | Wideband FXS, ISI interface, parametric MLT, DTMF detection, customer-specific | QFN38 | -100 V | 0 to 70 °C |

Note:

1. Adding the suffix "R" to the part number (e.g. Si3219x-A-FMR) denotes tape and reel.
2. QFN - Quad-Flat No-leads.

Table 1.2. Ordering GuideSi3219x Evaluation Kits

| Part Number | Description | V _{BAT} Max ¹ |
|--------------------|--|-----------------------------------|
| Si32193ACB10SL0EVB | ISI Wideband FXS with DTMF detection and low-cost capacitive boost dc-dc converter EVB | -100 V |

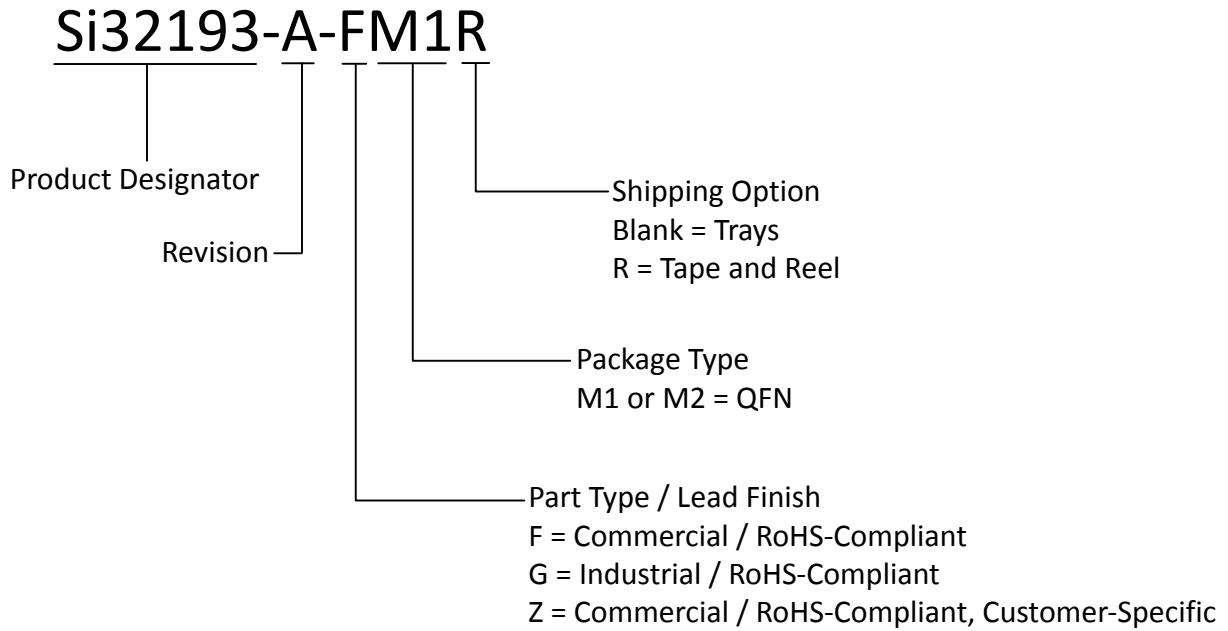
Note:

1. EVB V_{BAT} max may be limited by BOM option.

1.1 Product Identification

The product identification number is a finished goods part number or is specified by a finished goods part number, such as a special customer part number.

Example:



2. Functional Description

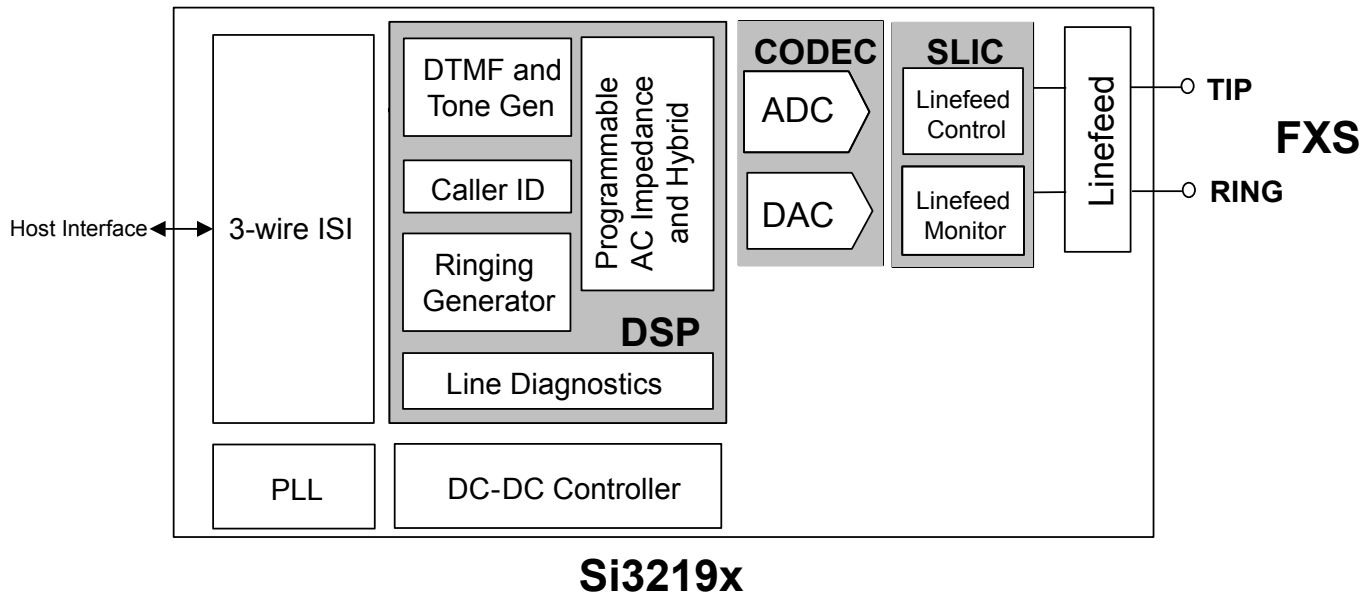


Figure 2.1. Si3219x Functional Block Diagram

The Si3219x series provides all SLIC, codec, DTMF detection, and signal generation functions needed for one complete analog telephone interface. The Si3219x performs all battery, over-voltage, ringing, supervision, codec, hybrid, and test (BORSCHT) functions; it also supports extensive metallic loop and self-test capabilities.

The Si3219x supports wideband audio (150 Hz–6.8 kHz) compliant with PKT-SP-HDV-104-120823, and is configurable to support the full ITU-T-G.722-201209 bandwidth (50 Hz–7 kHz). The wideband mode provides an expanded audio band with a 16 kHz sample rate for enhanced audio quality.

The Si3219x series supports a 3-wire ISI digital interface.

The Si3219x incorporates a programmable dc-dc converter controller that reacts to line conditions to provide the optimal battery voltage required for each line-state. Si3219x ICs are available with voltage ratings of –100 V.

Programmable on-hook voltage, programmable off-hook loop current, reverse polarity operation, loop or ground start operation, and on-hook transmission are supported. Loop current and voltage are continuously monitored by an integrated monitoring ADC.

The Si3219x single ProSLIC devices support ringing with or without a programmable dc offset and can operate in low-power ringing and adaptive-ringing modes. The available voltage offset, frequency, waveshape, and cadence options are designed to ring the widest variety of terminal devices and to reduce external controller requirements.

A complete audio transmit and receive path is integrated, including ac impedance and hybrid gain. These features are software-programmable, allowing a single hardware design to meet global requirements.

3. Electrical Specifications

Table 3.1. Recommended Operating Conditions

| Parameter | Symbol | Test Condition | Min | Typ | Max | Unit |
|--|--------------------|---------------------|------|-----|--------------------|------|
| Ambient Temperature | T_A | F-grade, Z-grade | 0 | 25 | 70 | °C |
| | | G-grade | -40 | 25 | 85 | °C |
| Silicon Junction Temperature, High Voltage Die | T_{JHV} | Continuous | — | — | Internally Limited | °C |
| Silicon Junction Temperature, Low Voltage Die | T_{JLV} | Continuous | — | — | 125 | °C |
| Supply Voltages | V_{DDD}, V_{DDA} | | 3.13 | 3.3 | 3.47 | V |
| Battery Voltage ² | V_{BAT} | | -100 | — | -15 | V |
| IO Supply Voltage | V_{DDIO} | | 1.71 | — | 3.47 | V |

Note:

1. All minimum and maximum specifications apply across the recommended operating conditions. Typical values apply at nominal supply voltages and an operating temperature of 25 °C unless otherwise stated.
2. Minimum and maximum battery voltage limits are dependent upon loop conditions and dc-dc converter configuration.

Table 3.2. Power Supply Characteristics

| Parameter | Symbol | Test Condition | Min | Typ | Max | Unit |
|----------------------------------|------------|--|-----|-------|-----|------|
| Supply Currents: | I_{DD} | V_T and $V_R = \text{Hi-Z}$, $R_{STB} = 0$ | — | 5.13 | — | mA |
| Reset | I_{VBAT} | | — | — | — | mA |
| Supply Currents: | I_{DD} | V_T and $V_R = \text{Hi-Z}$ | — | 11.15 | — | mA |
| High Impedance, Open | I_{VBAT} | | — | 0.6 | — | mA |
| Supply Currents: | I_{DD} | $V_{TR} = -48 \text{ V}$, Automatic Power Save Mode Enabled | — | 11.6 | — | mA |
| Forward/Reverse, On-hook | I_{VBAT} | | — | 0.6 | — | mA |
| Supply Currents: | I_{DD} | $V_{TR} = -48 \text{ V}$, Automatic Power Save Mode Disabled | — | 30.44 | — | mA |
| Forward/Reverse, On-hook | I_{VBAT} | | — | 2.1 | — | mA |
| Supply Currents: | I_{DD} | V_T or $V_R = -48 \text{ V}$ V_R or $V_T = \text{Hi-Z}$, Automatic Power Save Mode Enabled | — | 11.5 | — | mA |
| Tip/Ring Open, On-hook | I_{VBAT} | | — | 0.4 | — | mA |
| Supply Currents: | I_{DD} | V_T or $V_R = -48 \text{ V}$ V_R or $V_T = \text{Hi-Z}$, Automatic Power Save Mode Disabled | — | 30.6 | — | mA |
| Tip/Ring Open, On-hook | I_{VBAT} | | — | 1.3 | — | mA |
| Supply Currents: | I_{DD} | $V_{TR} = 48 \text{ V}$ | — | 44.7 | — | mA |
| Forward/Reverse OHT, On-hook | I_{VBAT} | | — | 2.9 | — | mA |
| Supply Currents: | I_{DD} | $I_{LOOP} = 20 \text{ mA}$, $R_{LOAD} = 200 \Omega$ | — | 44.4 | — | mA |
| Forward/Reverse Active, Off-hook | I_{VBAT} | | — | 21.65 | — | mA |
| Supply Currents: | I_{DD} | $V_{TR} = 55 V_{RMS} + 0 V_{DC}$, low power ringing, sinusoidal, $f = 20 \text{ Hz}$, $R_{LOAD} = 3 \text{ REN} = 2333 \Omega$ | — | 33.8 | — | mA |
| Ringing | I_{VBAT} | | — | 24.4 | — | mA |

Note:

1. All specifications are for a tracking low-cost capacitive boost dc-dc converter at 25 °C. V_{DDD} , $V_{DDA} = 3.3 \text{ V}$; $V_{DC} = 12 \text{ V}$.
2. I_{DD} includes $I_{DDIO} + I_{DDD} + I_{DDA}$.

Table 3.3. AC Characteristics

| Parameter | Test Condition | Min | Typ | Max | Unit |
|--|---|---|-----|------|---------|
| TX/RX Performance | | | | | |
| Overload Compression | 2-Wire - PCM | See Figure 3.4 Overload Compression Performance on page 14. | — | — | |
| Single Frequency Distortion | 200 Hz to 3.4 kHz (μ -law/A-law) | — | — | -40 | dB |
| | 200 Hz to 3.4 kHz (16-bit linear) | — | — | -63 | dB |
| Signal-to-(Noise + Distortion) Ratio ¹ | 200 Hz to 3.4 kHz transmit or receive path Active off-hook, and OHT, any Z _T | See Figure 3.3 Transmit and Receive Path SNDR on page 13. | — | — | |
| Audio Tone Generator Signal-to-Distortion Ratio ¹ | 0 dBm0, Active off-hook, and OHT, any Z _T | 46 | — | — | dB |
| Intermodulation Distortion | | — | — | -41 | dB |
| Gain Accuracy ¹ | 2-Wire to PCM or PCM to 2-Wire 1014 Hz, Any gain setting | -0.2 | — | 0.2 | dB |
| Attenuation Distortion vs. Frequency | 0 dBm 0 ⁵ | See Figure 3.5 Receive Path Frequency Response on page 14 and Figure 3.6 Transmit Path Frequency Response on page 15. | | | |
| Group Delay vs. Frequency | | See Figure 3.7 Transmit Group Delay Distortion on page 15 and Figure 3.8 Receive Group Delay Distortion on page 16. | | | |
| Gain Tracking ² | 1014 Hz sine wave, reference level -10 dBm, Signal level: | — | — | — | — |
| | 3 dB to -37 dB | — | — | 0.25 | dB |
| | -37 dB to -50 dB | — | — | 0.5 | dB |
| | -50 dB to -60 dB | — | — | 1.0 | dB |
| Round-Trip Group Delay | 1014 Hz, Within same time-slot | — | 450 | 500 | μ s |
| 2-Wire Return Loss ³ | 200 Hz to 3.4 kHz | 26 | 30 | — | dB |
| Transhybrid Balance ³ | 300 Hz to 3.4 kHz | 26 | 30 | — | dB |
| Noise Performance | | | | | |
| Idle Channel Noise ⁴ | C-Message weighted | — | 8 | 14 | dBrnC |
| | Psophometric weighted | — | -82 | -76 | dBmP |
| PSRR from V _{DDD} , V _{DDIO} @ 3.3 V | RX and TX, 200 Hz to 3.4 kHz | — | 55 | — | dB |
| Longitudinal Performance | | | | | |

| Parameter | Test Condition | Min | Typ | Max | Unit |
|---|--|-----|-----|-----|----------|
| Longitudinal to Metallic/PCM Balance (forward or reverse) | 200 Hz to 1 kHz | — | 60 | — | dB |
| | 1 kHz to 3.4 kHz | — | 58 | — | dB |
| Metallic/PCM to Longitudinal Balance | 200 Hz to 3.4 kHz | — | — | — | dB |
| Longitudinal Impedance | 200 Hz to 3.4 kHz at TIP or RING | — | 50 | — | Ω |
| Longitudinal Current Capability | Active off-hook 60 Hz Reg 73 = 0x0B | — | 25 | — | mA |

Note:

1. Analog signal measured as $V_{TIP} - V_{RING}$. Assumes ideal line impedance matching.
2. The quantization errors inherent in the μ/A -law companding process can generate slightly worse gain tracking performance in the signal range of 3 to -37 dB for signal frequencies that are integer divisors of the 8 kHz PCM sampling rate.
3. $V_{DD} = 3.3$ V, $V_{BAT} = -52$ V, no fuse resistors; $R_L = 600 \Omega$, $Z_S = 600 \Omega$ synthesized using RS register coefficients.
4. The level of any unwanted tones within the bandwidth of 0 to 4 kHz does not exceed -55 dBm.
5. 0 dBm 0 is equal to 0 dBm into 600Ω .

Table 3.4. Linefeed Characteristics

| Parameter | Symbol | Test Condition | Min | Typ | Max | Unit |
|---|---------------|--|------|-----|-----|------------|
| DC Feed Current | | Differential | — | — | 45 | mA |
| | | Common Mode | — | — | 30 | mA |
| | | Differential + Common Mode | — | — | 45 | mA |
| DC Loop Current Accuracy | | $I_{LIM} = 18 \text{ mA}$ | — | — | 10 | % |
| DC Open Circuit Voltage Accuracy | | Active Mode; $V_{OC} = 48 \text{ V}$, $V_{TIP} - V_{RING}$ | — | — | 4 | V |
| DC Differential Output Resistance | R_{DO} | $I_{LOOP} < I_{LIM}$ | 160 | — | 640 | Ω |
| DC On-Hook Voltage Accuracy-Ground Start (TIP Open) | V_{OHTO} | $I_{RING} < I_{LIM}$; V_{RING} wrt ground, $V_{RING} = -51 \text{ V}$ | — | — | 4 | V |
| DC Output Resistance-Ground Start (TIP Open) | R_{ROTO} | $I_{RING} < I_{LIM}$; RING to ground | 160 | — | 640 | Ω |
| DC Output Resistance-Ground Start (TIP Open) | R_{TOTO} | TIP to ground | 400 | — | — | k Ω |
| Loop Closure Detect Threshold Accuracy | | $I_{THR} = 13 \text{ mA}$ | — | — | 10 | % |
| Ground Key Detect Threshold Accuracy | | $I_{THR} = 13 \text{ mA}$ | — | — | 10 | % |
| Ring Trip Threshold Accuracy | | AC detection, $V_{RING} = 70 \text{ Vpk}$, no offset, $I_{TH} = 80 \text{ mA}$ | — | — | 4 | mA |
| | | DC detection, 20 V dc offset, $I_{TH} = 13 \text{ mA}$ | — | — | 1 | mA |
| Ringing Amplitude | $V_{RINGING}$ | Open circuit, $V_{BAT} = -100 \text{ V}$ | -100 | — | — | V_{PK} |
| Sinusoidal Ringing Total Harmonic Distortion | R_{THD} | 50 V_{RMS} , 0 V_{OFFSET} , 0–5 REN | — | 1 | — | % |
| Ringing Frequency Accuracy | | $f = 16 \text{ Hz to } 60 \text{ Hz}$ | — | — | 1 | % |
| Ringing Cadence Accuracy | | Accuracy of ON/OFF times | — | — | 50 | ms |
| Loop Voltage Sense Accuracy | | $V_{TIP} - V_{RING} = 48 \text{ V}$ | — | 2 | 4 | % |
| Loop Current Sense Accuracy | | $I_{LOOP} = 18 \text{ mA}$ | — | 7 | 10 | % |

Table 3.5. Digital I/O Characteristics

| Parameter | Symbol | Test Condition | Min | Typ | Max | Unit |
|---------------------------|----------|---------------------------------|-----------------------|-----|-----------------------|---------------|
| High-Level Input Voltage | V_{IH} | PSCLK, RSTB, ISI_MOSI | $0.7 \times V_{DDIO}$ | — | V_{DDIO} | V |
| Low-Level Input Voltage | V_{IL} | PSCLK, RSTB, ISI_MOSI | — | — | $0.3 \times V_{DDIO}$ | V |
| High-Level Output Voltage | V_{OH} | ISI_MISO, $I_O = -4 \text{ mA}$ | $V_{DDIO} - 0.6$ | — | — | V |
| Low-Level Output Voltage | V_{OL} | ISI_MISO, $I_O = 4 \text{ mA}$ | — | — | 0.4 | V |
| Input Leakage Current | I_L | | — | — | 10 | μA |

Table 3.6. Charge Pump Characteristics

| Parameter | Symbol | Test Condition | Min | Typ | Max | Unit |
|------------------------------|----------|----------------|-----------------------|-----|-------------------|------|
| Output Voltage (DCDRV, DCFF) | V_{CP} | | $2 \times V_{DD} - 1$ | — | $2 \times V_{DD}$ | V |
| Output Current | I_{CP} | | — | — | 3 ¹ | mA |

Note:

1. Peak drive current capability is >60 mA.

Table 3.7. Switching Characteristics (General Inputs)

| Parameter | Symbol | Min | Typ | Max | Unit |
|--|-----------|-----|-----|-----|---------|
| RSTB Pulse Width | t_{RST} | 200 | — | — | μ s |
| RSTB High to First Register/RAM Access | t_{RCS} | 5 | — | — | ms |

Note:

1. All timing is referenced to the 50% level of the waveform. Input test levels are $V_{IH} = V_{DDIO} - 0.4$ V, $V_{IL} = 0.4$ V. Rise and Fall times are referenced to the 20% and 80% levels of the waveform.
2. MSIF_LOCK signal on SoC must be active before valid SPI communication.

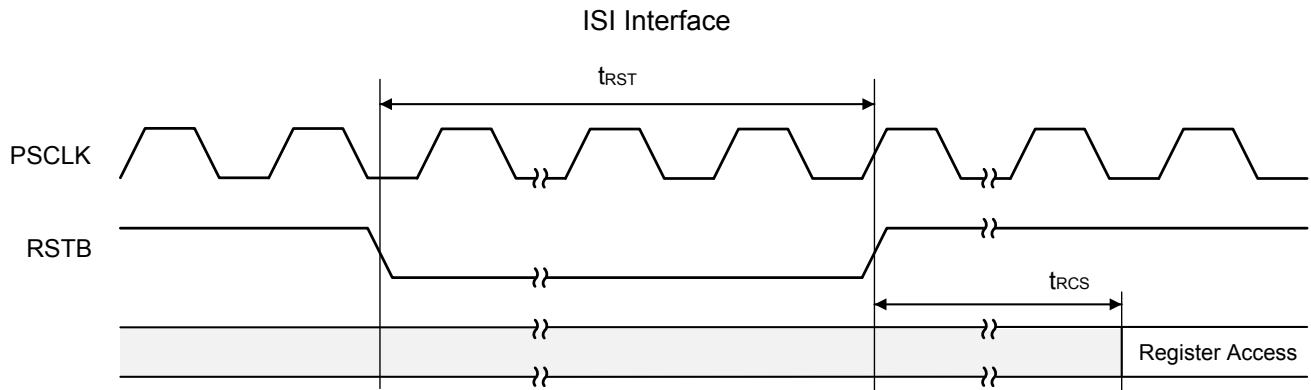
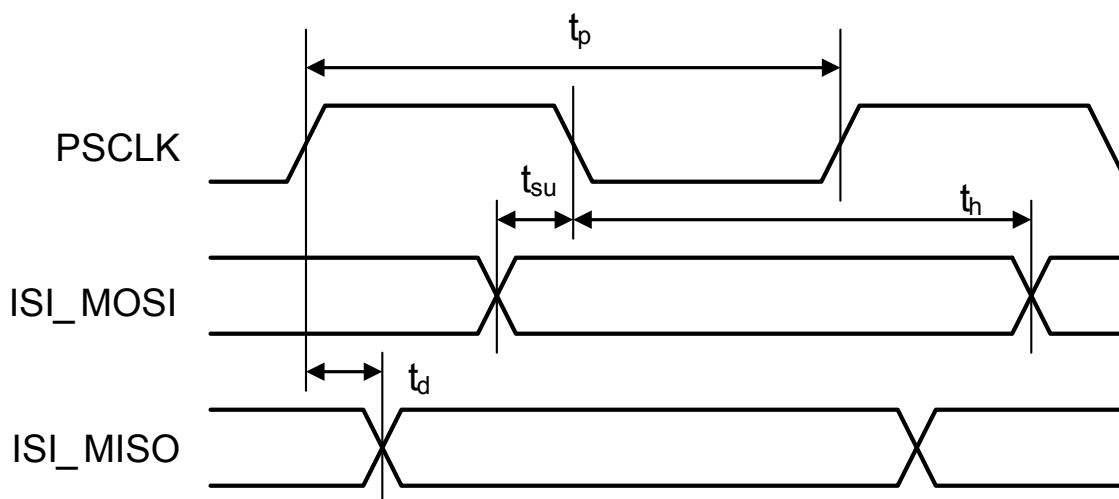
**Figure 3.1. Si3219x Reset Timing Diagram**

Table 3.8. Switching Characteristics (ISI)

| Parameter | Symbol | Min | Typ | Max | Unit |
|------------------------------------|----------|-----|-------|-----|------|
| Setup Time, ISI_MOSI to PSCLK Fall | t_{su} | 7.5 | — | — | ns |
| Hold Time, ISI_MOSI to PSCLK Fall | t_h | 5 | — | — | ns |
| Delay Time, PSCLK Rise to ISI_MISO | t_d | — | — | 16 | ns |
| PSCLK Period | t_p | — | 40.69 | — | ns |
| PSCLK Duty Cycle | | 40 | 50 | 60 | % |

Note:

1. Timing should be guaranteed by ISI-enabled host SoC.

**Figure 3.2. ISI Timing Diagram****Table 3.9. Thermal Conditions**

| Parameter | Symbol | Test Condition | Value | Unit |
|---|---------------|----------------|--------------------|------|
| Storage Temperature Range | T_{STG} | | -55 to 150 | °C |
| Thermal Resistance, Typical QFN-38 ¹ | θ_{JA} | | 35 | °C/W |
| | θ_{JC} | | 15 | °C/W |
| Maximum Junction Temperature, (High Voltage Die) ³ | T_{JHV} | Continuous | Internally Limited | °C |
| Maximum Junction Temperature (Low Voltage Die) ² | T_{JLV} | | 125 | °C |

Note:

1. The thermal resistance of an exposed pad package is assured when the recommended printed circuit board layout guidelines are followed correctly. The specified performance requires that the exposed pad be soldered to an exposed copper surface of at least equal size and that multiple vias are added to enable heat transfer between the top-side copper surface and a large internal/bottom copper plane. Thermal resistance values are empirical measurements taken from 2-Layer EVBs.
2. Operation of the Si3219x above 125 °C junction temperature may degrade device reliability.
3. Si3219x linefeed is equipped with on-chip thermal limiting circuitry that shuts down the circuit when the junction temperature exceeds the thermal shutdown threshold. The thermal shutdown threshold is normally set to 145 °C; when in the ringing state the thermal shutdown is set to 200 °C.

Table 3.10. Absolute Maximum Ratings

| Parameter | Symbol | Value | Unit |
|------------------------|---------------------|--------------------------|------|
| Supply Voltage | V_{DD}, V_{DDIO} | -0.5 to 4.0 | V |
| Digital Input Voltage | V_{IND} | -0.3 to $V_{DDIO} + 0.5$ | V |
| Battery Supply Voltage | V_{BAT} | +0.4 to 102 | V |
| Tip or Ring Voltage | V_{TIP}, V_{RING} | $V_{BAT} - 0.4$ | V |
| TIP, RING Current | I_{TIP}, I_{RING} | ± 100 | mA |

Note:

- Permanent device damage may occur or the reliability of the device may be affected if the device is operated at or above the absolute maximum ratings.

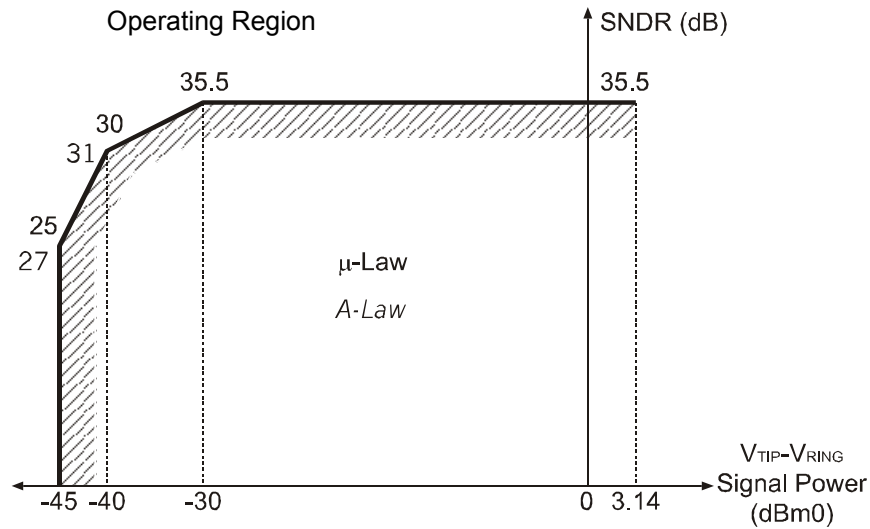


Figure 3.3. Transmit and Receive Path SNDR

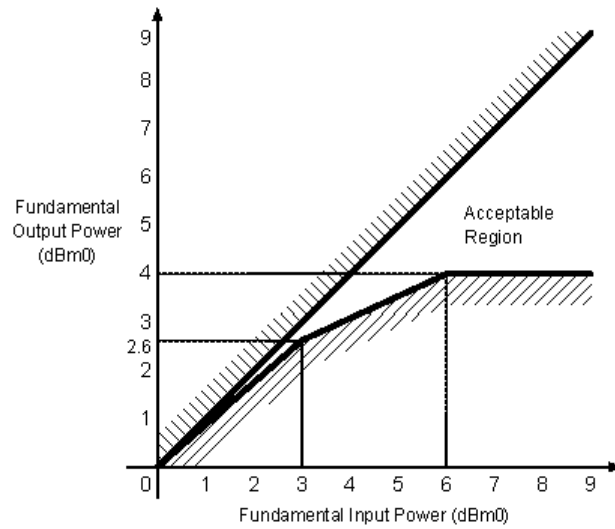


Figure 3.4. Overload Compression Performance

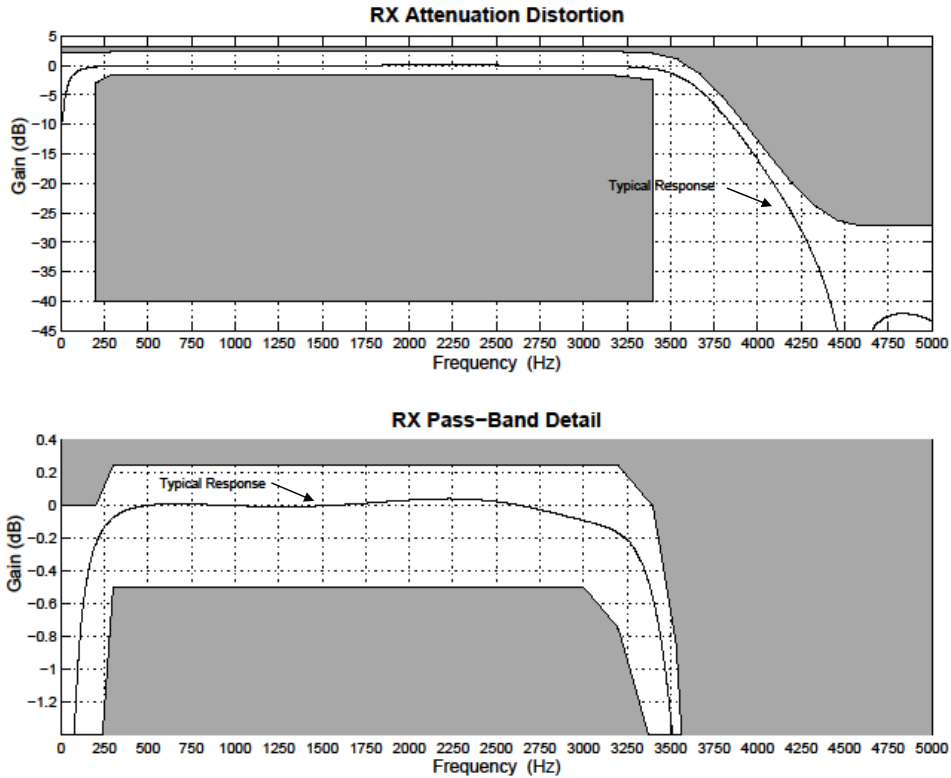


Figure 3.5. Receive Path Frequency Response

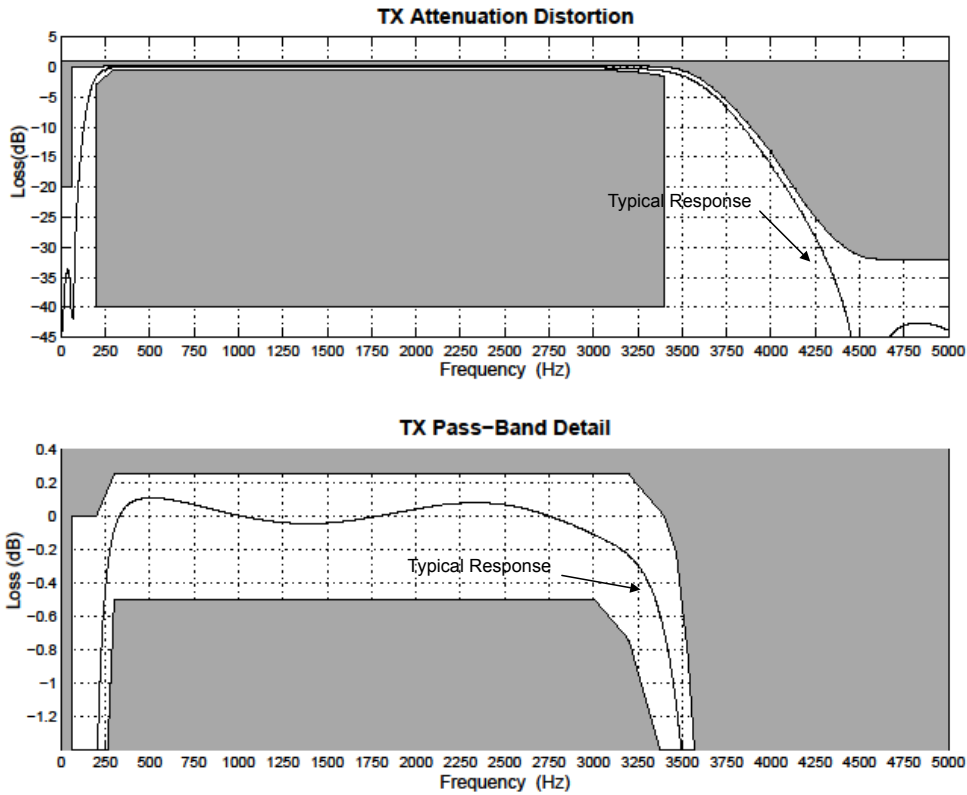


Figure 3.6. Transmit Path Frequency Response

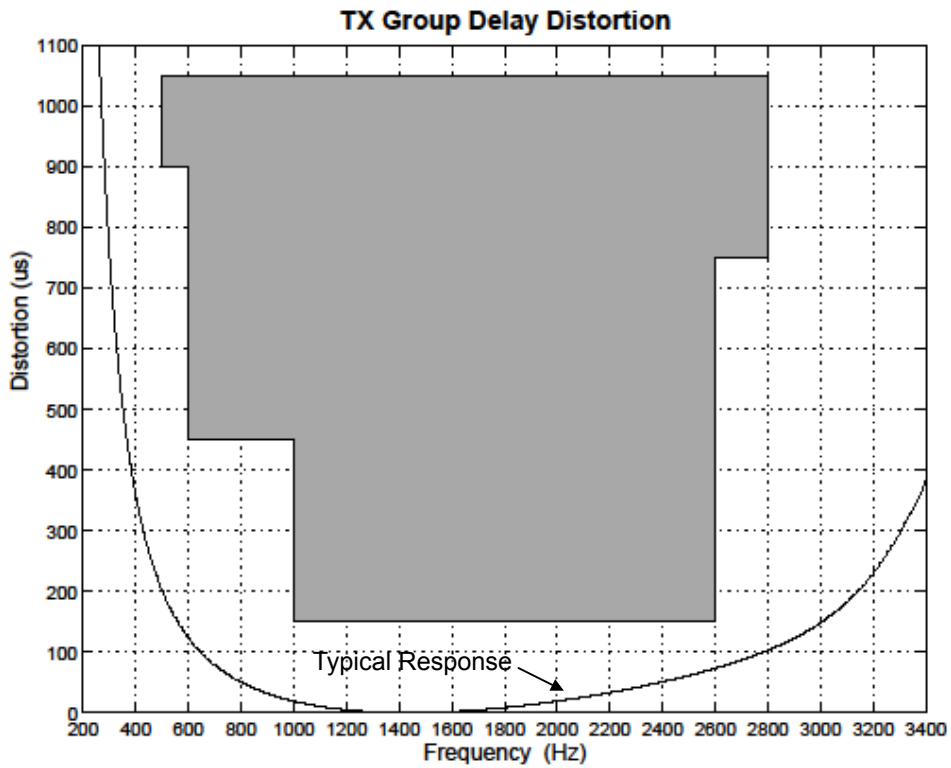


Figure 3.7. Transmit Group Delay Distortion

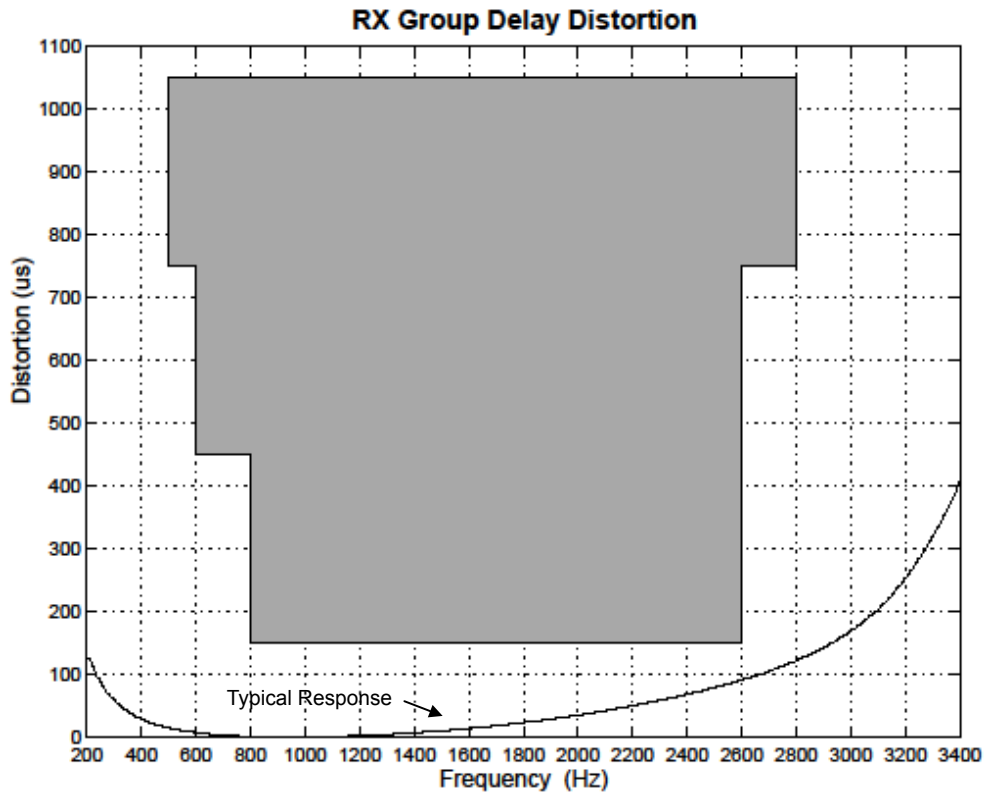


Figure 3.8. Receive Group Delay Distortion

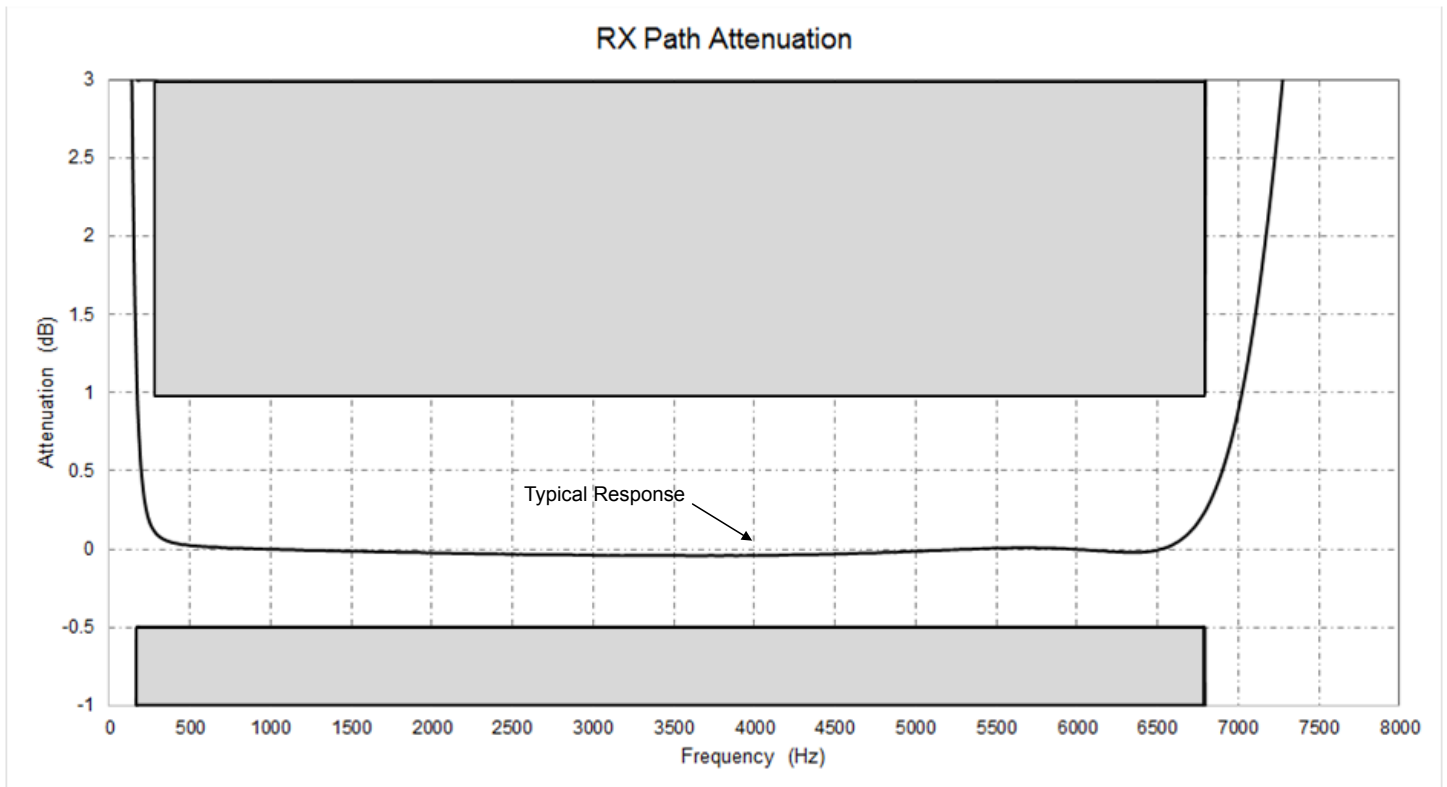


Figure 3.9. Receive Wideband Frequency Response

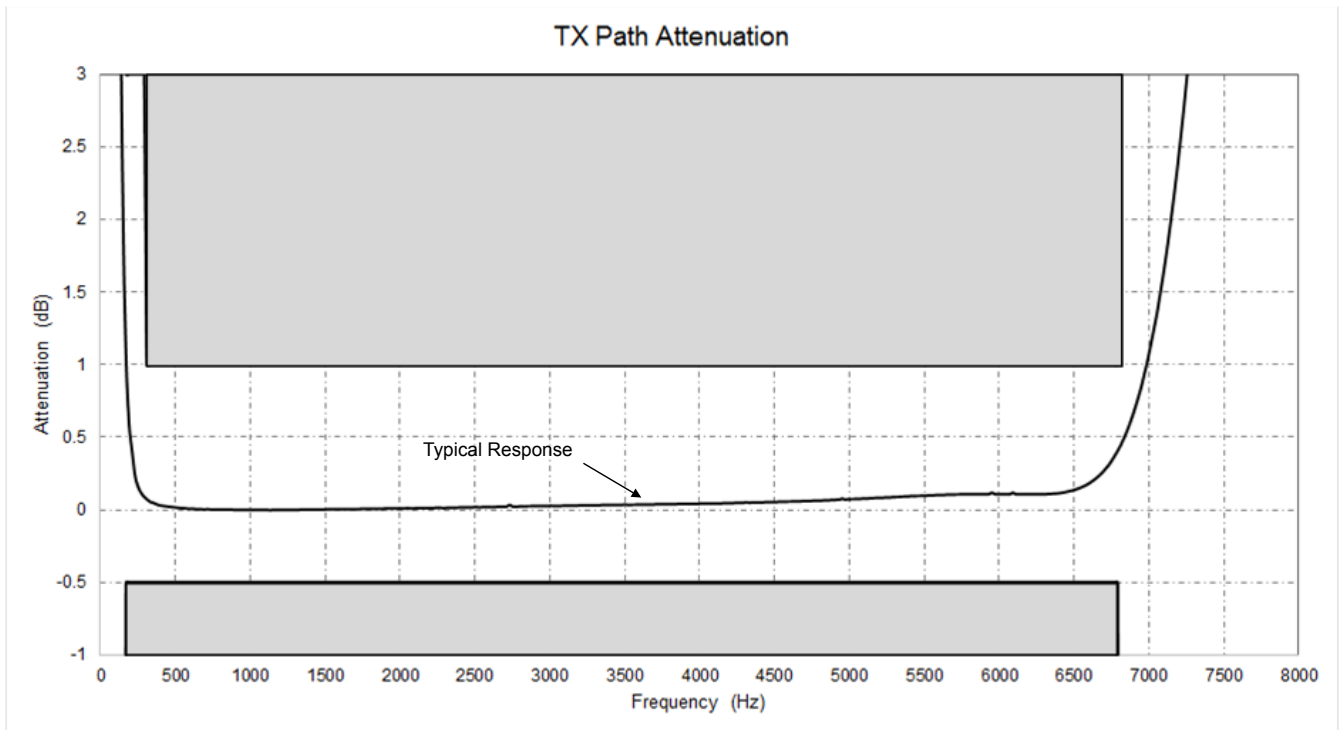


Figure 3.10. Transmit Wideband Frequency Response

4. FXS Features

4.1 DC Feed Characteristics

ProSLIC internal linefeed circuitry provides completely programmable dc feed characteristics.

When in the active state, the ProSLIC operates in one of three dc linefeed operating regions: a constant-voltage region, a constant-current region, or a resistive region, as shown in the figure below. The constant-voltage region has a low resistance, typically 160 Ω . The constant-current region approximates infinite resistance.

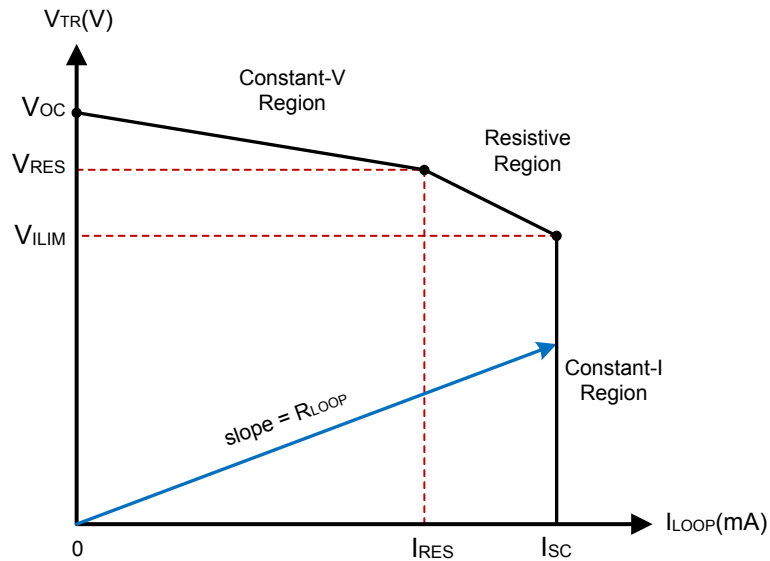


Figure 4.1. Dual ProSLIC DC Feed Characteristics

4.2 Linefeed Operation States

The linefeed interface includes eight different register-programmable operating states as listed in [Table 4.1 Linefeed Operating States on page 19](#). The Open state is the default condition in the absence of any preloaded register settings. The device may also automatically enter the open state in the event of a linefeed fault condition.

4.3 Line Voltage and Current Monitoring

The ProSLIC continuously monitors the TIP, RING, and battery voltages and currents via an on-chip Monitor ADC and stores the resulting values in individual RAM locations. Additionally, VTIP, VRING, loop current, and longitudinal current values are calculated based on the differential and common mode voltage measurements. The ADC updates all registers at a rate of 2 kHz or greater.

4.4 Power Monitoring and Power Fault Detection

The Si3219x line monitoring functions are used to continuously protect against excessive power conditions.

If the Si3219x detects an overpower condition, it automatically sets that device to the open state and generates a "power alarm" interrupt.

The interrupt can be masked, but masking the automatic transition to open is not recommended since it is used to protect the Si3219x HVIC under excessive power conditions.

4.5 Thermal Overload Shutdown

If the die temperature exceeds the maximum junction temperature threshold, T_{Jmax} , of 145 °C or 200 °C (depending on the operating state), the device has the ability to shut itself down to a low-power state without user intervention. A thermal alarm interrupt is generated to notify host that the device has been switched to open state.

Table 4.1. Linefeed Operating States

| Linefeed State | Description |
|----------------------------------|--|
| Open | Output is high-impedance and audio is not transmitted. This is the default state after powerup or following a hardware reset. This state can also be used in the presence of line fault conditions and to generate open switch intervals (OSIs). This state is used in line diagnostics mode as a high impedance state during linefeed testing. A power fault condition may also force the device into the open state. |
| Forward Active Reverse Active | Linefeed circuitry and audio are active. In Forward Active state, the TIP lead is more positive than the RING lead; in Reverse Active state, the RING lead is more positive than the TIP lead. |
| Forward OHT Reverse OHT | Provides data transmission during an on-hook loop condition (e.g., transmitting caller ID data between ringing bursts). Linefeed circuitry and audio are active. In Forward OHT state, the TIP lead is more positive than the RING lead; in Reverse OHT state, the RING lead is more positive than the TIP lead. |
| TIP Open | Provides an active linefeed on the RING lead and sets the TIP lead to high impedance ($\geq 400\text{ k}\Omega$) for ground start operation in forward polarity. Loop closure and ground key detect circuitry are active. |
| RING Open | Provides an active linefeed on the TIP lead and sets the RING lead to high impedance ($\geq 400\text{ k}\Omega$) for ground start operation in reverse polarity. Loop closure and ground key detect circuitry are active. |
| Ringing | Drives programmable ringing signal onto TIP and RING leads with or without dc offset. |

4.6 Loop Closure Detection

The Si3219x provides a completely programmable loop closure detection mechanism. The loop closure detection scheme provides two unique thresholds to allow hysteresis, and also includes a programmable debounce filter to eliminate false detection. A loop closure detect status bit provides continuous status, and a maskable interrupt bit is also provided.

4.7 Ground Key Detection

The Si3219x provides a ground key detect mechanism using a programmable architecture similar to the loop closure scheme. The ground key detect scheme provides two unique thresholds to allow hysteresis and also includes a programmable debounce filter to eliminate false detection. A ground key detect status bit provides continuous status, and a maskable interrupt bit is also provided.

4.8 Ringing Generation

The Si3219x supports the patented Low-Power Ringing (LPR) method exclusively, which when used with a tracking battery scheme, maximizes the ringing power transferred to the load and reduces overall power consumption. Ringing is fully programmable including frequency, amplitude, dc offset, wave shape and crest factor. The Si3219x also supports automatic ring cadencing and ringtrip detection (AC and DC).

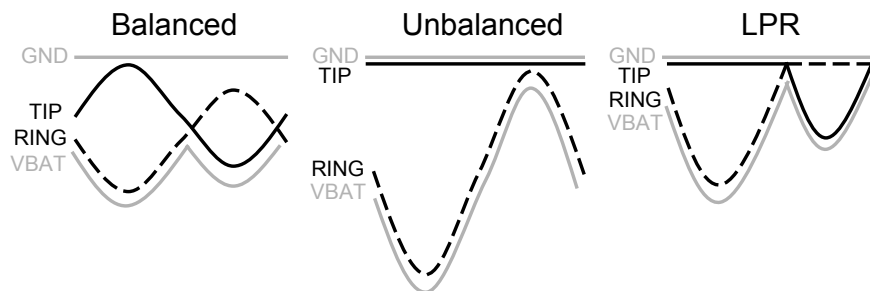


Figure 4.2. Ringing Mode

4.9 Polarity Reversal

The Si3219x supports polarity reversal for message waiting and various other signaling modes. The ramp rate can be programmed for a smooth or abrupt transition to accommodate different application requirements.

4.10 Two-wire Impedance Synthesis

The ac two-wire impedance synthesis is generated on-chip using a DSP-based scheme to optimally match the output impedance of the Si3219x to the reference impedance. Most real or complex two-wire impedances can be generated with appropriate register coefficients.

4.11 Transhybrid Balance Figure

The trans-hybrid balance function is implemented on-chip using a DSP-based scheme to effectively cancel the reflected receive path signal from the transmit path.

4.12 Tone Generators

The Si3219x includes two digital tone generators that allow a wide variety of single- or dual-tone frequency and amplitude combinations. Each tone generator has its own set of registers that hold the desired frequency, amplitude, and cadence to allow generation of DTMF and call progress tones for different requirements. The tones can be directed to either receive or transmit paths.

4.13 DTMF Detection (Si32193 Only)

In DTMF, two tones generate a DTMF digit. One tone is chosen from four possible row tones, and one tone is chosen from four possible column tones. The sum of these tones constitutes one of 16 possible DTMF digits.

4.14 Pulse Metering

The pulse metering system for the Si3219x is designed to inject a 12 or 16 kHz billing tone into the audio path with maximum amplitude of 0.85 VRMS at TIP and RING into a 200 Ω ac load impedance. The tone is generated in the DSP via a table lookup that guarantees spectral purity by not allowing drift. The tone will ramp up until it reaches a host-programmed threshold, at which point it will maintain that level until instructed to ramp down, thus creating a trapezoidal envelope.

The amplitude is controlled by an automatic gain control circuit (AGC). While the tone is ramping up, the AGC takes the feedback audio and applies it to a band pass filter, which is programmed for the 12 or 16 kHz frequency of interest. When the peak is detected, the ramp is stopped.

4.15 DC-DC Controller

The Si3219x-C devices integrate a dc-dc controller to control an external tracking dc-dc converter which generates the high voltage supply (VBAT) to the SLIC. The tracking VBAT voltage generated from a single positive dc input is optimized to minimize power consumption by closely tracking the SLIC state, even tracking the ringing waveforms.

The dc-dc controller output DCDRV is driven by an internal charge pump which allows it to connect directly to the gate of the MOSFET switch of the dc-dc converter. This eliminates the need for the MOSFET predrive circuit that is typically required when other SLICs are used with a MOSFET with VTH greater than V_{DD}. See [Table 3.6 Charge Pump Characteristics on page 11](#).

4.16 Wideband Audio

The Si3219x supports wideband audio (150 Hz–6.8 kHz) compliant with PKT-SP-HDV-104-120823, and is configurable to support the full ITU-T-G.722-201209 bandwidth (50 Hz–7 kHz). The wideband provides an expanded audio band at a 16-bit, 16 kHz sample rate for enhanced audio quality while maintaining standard telephony audio compatibility. Wideband audio samples are transmitted and received at an effective 16 kHz rate by using multiple timeslots in a single 8 kHz frame.

4.17 Test Facilities

The Si3219x supports a rich set of metallic loop tests to diagnose external faults, as well as a set of inward self-tests to support diagnostics of the Si3219x-based voice port. Implementation of metallic loop tests required the ProSLIC® MLT API, while the inward self-tests are included in the standard ProSLIC® API.

Table 4.2. Supported Tests

| Test | Description |
|----------------------------|---|
| Metallic Loop Tests | |
| Voltages | Measures ac and dc voltages from T-R, T-G, and R-G. |
| Receiver Off-hook | Discriminates between resistive fault and off-hook terminating device. |
| REN | Measures Ringer Equivalence Number (REN). |
| REN Capacitance | Measures T-R capacitance of on-hook load. |
| Capacitance | Measures 3-Terminal Capacitance. |
| Resistance | Measures resistance from T-R, T-G, or R-G. |
| Inward Self-Tests | |
| PCM Loopback | Configures Si3219x for 8- or 16-bit PCM loopback. |
| DC Feed | Verify dc Feed I/V and loop closure using integrated test load. |
| Ring and Ringtrip | Verify ringing voltage (ac and dc). Optional ringtrip check to support system level signaling verification. |
| Battery | Verify VBAT. |
| Audio Gain | Measure gain of RX (host to line) and TX (line to host) paths without using an external load, test equipment, or requiring the host to provide audio samples. |

In addition to these specific test suites, the user is free to use the general test facilities listed in the following table:

Table 4.3. General Test Facilities

| Test | Description |
|--------------------------|---|
| Monitor ADC | Provides TIP/RING voltages (inside and outside overcurrent protection), TIP/RING currents and VBAT voltage. |
| Audio Diagnostic Filters | Three cascaded second-order Biquad filters with peak hold and averaging capabilities. |
| Loopback Modes | Digital and analog loopback modes to isolate portions of the audio path. |
| Tone Generators | The dual-tone generators may be used as general-purpose test signal generators. |

5. System Interfaces

5.1 Integrated Serial Interface

The ISI interface supported on select Si3219x devices is a three-wire proprietary interface which serializes SPI and PCM communications and interrupts, reducing the SoC interface from nine wires to three (PCLK, ISI_MISO, ISI_MOSI). SPI communications and PCM data transfers are embedded in the serial data. The host side of the ISI is integrated onto selected SoCs from several vendors.

ISI is a point to point interface; therefore, it is not possible to daisy-chain more than one ISI ProSLIC device.

Both μ -255 Law (μ -Law) and A-law companding formats are supported in addition to 16-bit linear data mode with no companding.

5.2 Input/Output Voltage Selection

The digital host interface I/O (ISI) on the Si3219x may directly interface to 1.8 V to 3.3 V devices. The I/O voltage selection is made by supplying the VDDIO pin with the appropriate I/O supply voltage.

To avoid power supply sequencing issues, VDDIO should be connected to the same supply as VDD in 3.3 V interface designs. Other voltages between 1.8 V and 3.3 V can also be used for VDDIO (for example 2.5 V), but steps must be taken to ensure that the VDDIO supply comes up after the VDD supply if VDDIO is not connected to VDD.

6. Pin Descriptions

6.1 Pin Descriptions: 38 Pin QFN (Si32192/3)

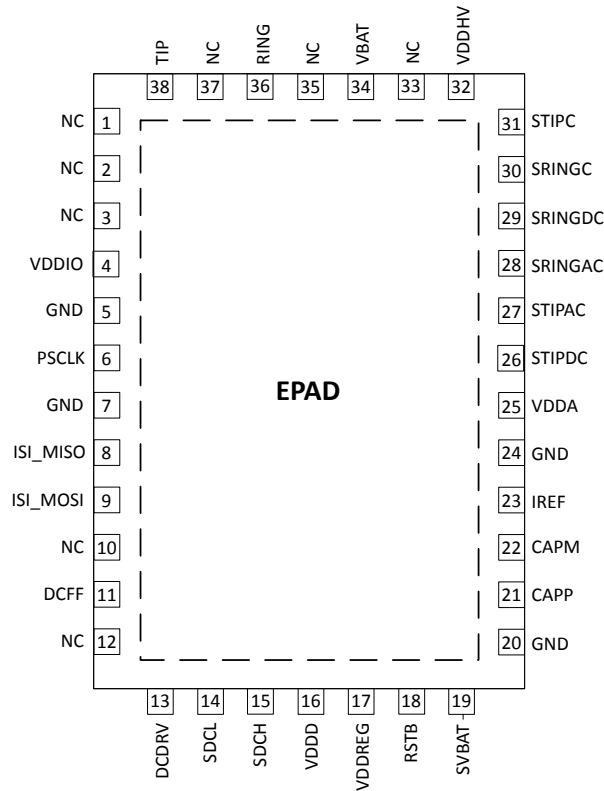


Table 6.1. Si3219x Pin Descriptions

| Pin # | Pin Name | Description |
|-------|----------|--|
| 1 | NC | No connection. This pin should be left unconnected. |
| 2 | NC | No connection. This pin should be left unconnected. |
| 3 | NC | No connection. This pin should be left unconnected. |
| 4 | VDDIO | Digital IO Supply Voltage. 3.3 V or 1.8 V digital power supply for internal circuitry. |
| 5 | GND | Ground. Connect to ground. |
| 6 | PSCLK | ISI Bus Clock Input. Clock input for ISI bus timing. |
| 7 | GND | Ground. Connect to ground. |

| Pin # | Pin Name | Description |
|-------|----------|--|
| 8 | ISI_MISO | Transmit ISI Output. ISI Master Input, Slave Output. Output data to ISI bus. |
| 9 | ISI_MOSI | Transmit ISI Input. ISI Master Output, Slave input. Input data from ISI bus. |
| 10 | NC | No connection. This pin should be left unconnected. |
| 11 | DCFF | DC-DC Charge Pump Output. |
| 12 | NC | No connection. This pin should be left unconnected. |
| 13 | DCDRV | DC Drive. DC-DC converter control signal output which drives external transistor. |
| 14 | SDCL | DC Monitor. DC-DC converter monitor input used to detect overcurrent situations in the converter. |
| 15 | SDCH | DC Monitor. DC-DC converter monitor input used to detect overcurrent situations in the converter. |
| 16 | VDDD | IC Voltage Supply. 3.3 V digital power supply for internal circuitry. |
| 17 | VDDREG | Regulated Core Power Supply. |
| 18 | RSTB | Reset Input. Active low input. Hardware reset used to place all control registers in the default state. |
| 19 | SVBAT | VBAT Sense. Input used to sense voltage on DC-DC converter output voltage lead. |
| 20 | GND | Ground. Connect to ground. |
| 21 | CAPP | SLIC Stabilization Capacitor. Capacitor used in dc feed low-pass filter. |
| 22 | CAPM | SLIC Stabilization Capacitor. Capacitor used in dc feed low-pass filter. |
| 23 | IREF | Current Reference Input. Connects to an external resistor used to provide a high accuracy reference current. |
| 24 | GND | Ground. Connect to ground. |
| 25 | VDDA | Analog Supply Voltage. Analog 3.3 V power supply for internal analog circuitry. |
| 26 | STIPDC | TIP DC Sense. Analog DC input used to sense voltage on TIP lead. |

| Pin # | Pin Name | Description |
|-------|----------|---|
| 27 | STIPAC | TIP AC Sense. Analog AC input used to sense voltage on TIP lead. |
| 28 | SRINGAC | RING AC Sense. Analog AC input used to sense voltage on RING lead. |
| 29 | SRINGDC | RING DC Sense. Analog DC input used to sense voltage on RING lead. |
| 30 | SRINGC | RING Coarse Sense Input. Voltage sensing outside protection circuit. |
| 31 | STIPC | TIP Coarse Sense Input. Voltage sensing outside protection circuit. |
| 32 | VDDHV | Analog Supply Voltage. Analog 3.3 V power supply for internal analog circuitry. |
| 33 | NC | No connection. This pin should be left unconnected. |
| 34 | VBAT | Battery Voltage Supply. Connect to battery supply from DC-DC converter. |
| 35 | NC | No connection. This pin should be left unconnected. |
| 36 | RING | RING Terminal. Connect to the RING lead of the subscriber loop. |
| 37 | NC | No connection. This pin should be left unconnected. |
| 38 | TIP | TIP Terminal. Connect to the TIP lead of the subscriber loop. |
| — | EPAD | Exposed paddle. Connect to ground. |

Table 7.1. Package Dimensions

| Dimension | Min | Nom | Max |
|-----------|-----------|------|------|
| A | 0.80 | 0.85 | 0.90 |
| A1 | 0.00 | 0.02 | 0.05 |
| A3 | 0.20REF | | |
| b | 0.18 | 0.20 | 0.25 |
| D | 3.90 | 4.00 | 4.10 |
| D2 | 2.50 | 2.60 | 2.70 |
| e | 0.40 BSC. | | |
| E | 5.90 | 6.00 | 6.10 |
| E2 | 4.50 | 4.60 | 4.70 |
| L | 0.30 | 0.40 | 0.50 |
| aaa | — | — | 0.10 |
| bbb | — | — | 0.07 |
| ccc | — | — | 0.10 |
| ddd | — | — | 0.05 |
| eee | — | — | 0.08 |
| fff | — | — | 0.10 |

Note:

1. All dimensions shown are in millimeters (mm) unless otherwise noted.
2. Dimensioning and Tolerancing per ANSI Y14.5M-1982.
3. Recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

8. Land Pattern

8.1 Land Pattern: 38-Pin QFN

The figure below shows the recommended land pattern details for the 38-Pin QFN package. The table below lists the values for the dimensions shown in the illustration.

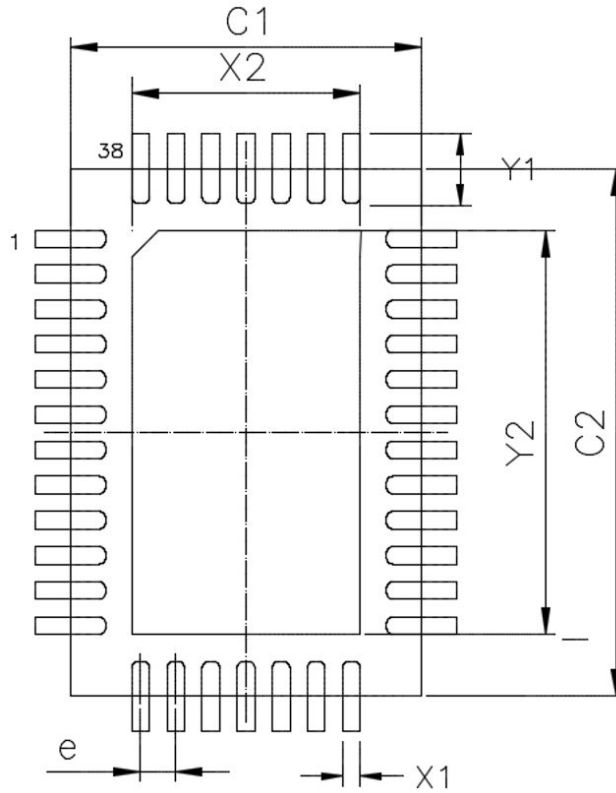


Figure 8.1. 38-Pin QFN Land Pattern

Table 8.1. PCB Land Pattern

| Dimension | mm |
|-----------|------|
| C1 | 4.00 |
| C2 | 6.00 |
| e | 0.40 |
| X1 | 0.20 |
| Y1 | 0.80 |
| X2 | 2.70 |
| Y2 | 4.70 |

Note:**General**

1. All dimensions shown are at Maximum Material Condition (MMC). Least Material Condition (LMC) is calculated based on a Fabrication Allowance of 0.05mm.
2. This Land Pattern Design is based on the IPC-7351 guidelines.

Solder Mask Design

1. All metal pads are to be non-solder mask defined (NSMD). Clearance between the solder mask and the metal pad is to be 60 µm minimum, all the way around the pad.

Stencil Design

1. A stainless steel, laser-cut and electro-polished stencil with trapezoidal walls should be used to assure good solder paste release.
2. The stencil thickness should be 0.125mm (5 mils).
3. The ratio of stencil aperture to land pad size should be 1:1 for all perimeter pins.
4. A 4x3 array of 0.85 mm square openings on 1.2 mm pitch should be used for the center ground pad to achieve a target solder coverage of ~50%.

Card Assembly

1. A No-Clean, Type-3 solder paste is recommended.
2. The recommended card reflow profile is per the JEDEC/IPC J-STD-020 specification for Small Body Components.

9. Top Marking

9.1 Top Marking 38-pin QFN

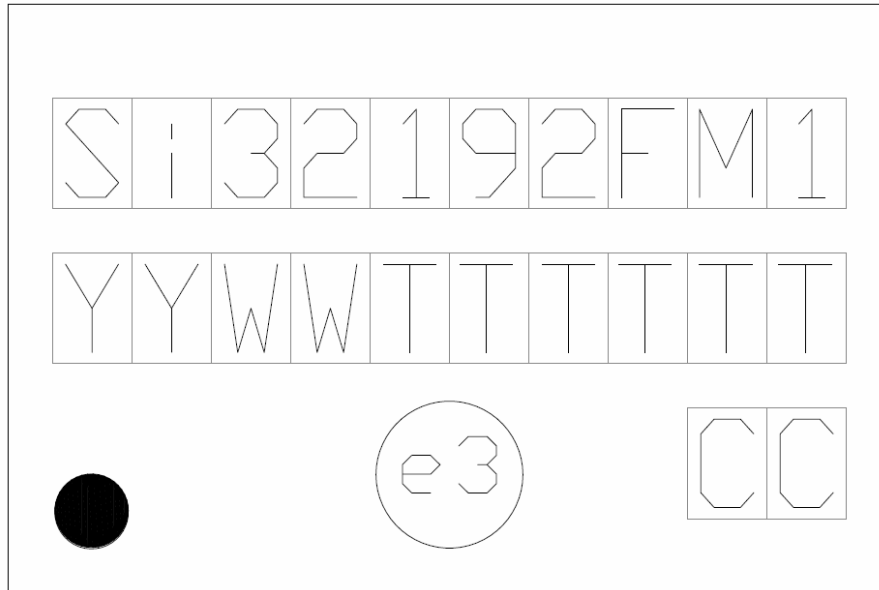


Figure 9.1. 38-Pin QFN Top Marking

Table 9.1. Top Marking Explanation

| | | |
|---------------------|--|---|
| Mark Method: | Laser | |
| Pin 1 Mark: | Circle = 0.50 mm Diameter (Bottom-Left-Justified) | |
| Font Size | 0.60 mm Right-Justified | |
| Line 1 Mark Format: | Device Part Number | Si32192FM1 |
| Line 2 Mark Format: | YY = Year WW = Work Week TTTTTT = Mfg Code | Manufacturing Code from the Assembly Purchase Order form. Assigned by the Assembly House. Corresponds to the year and work week of the assembly release. |
| Line 3 Mark Format: | Circle = 1.0 mm Diameter Center-Justified CC = Country of Origin ISO Code Abbreviation | "e3" Pb-Free Symbol TH, TW, or CN |

10. Revision History

Revision 1.0

June, 2018

- Updated [Table 3.2 Power Supply Characteristics on page 7](#).
- Updated [Table 3.3 AC Characteristics on page 8](#).
 - Updated Idle Channel Noise specification.
- Numerous clarifications throughout.

Revision 0.9

April, 2017

- Initial revision.



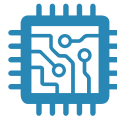
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